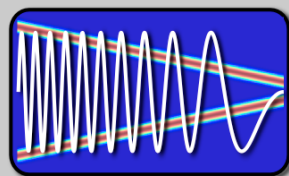


EE123



Digital Signal Processing

Lecture 15

Announcements

- Last Time
 - Today:
 - Resampling
 - Interchanging operations
 - multi-rate processing
- Today
 - Multi-Rate
 - Polyphase
 - Practical A/D
 - Quantization noise

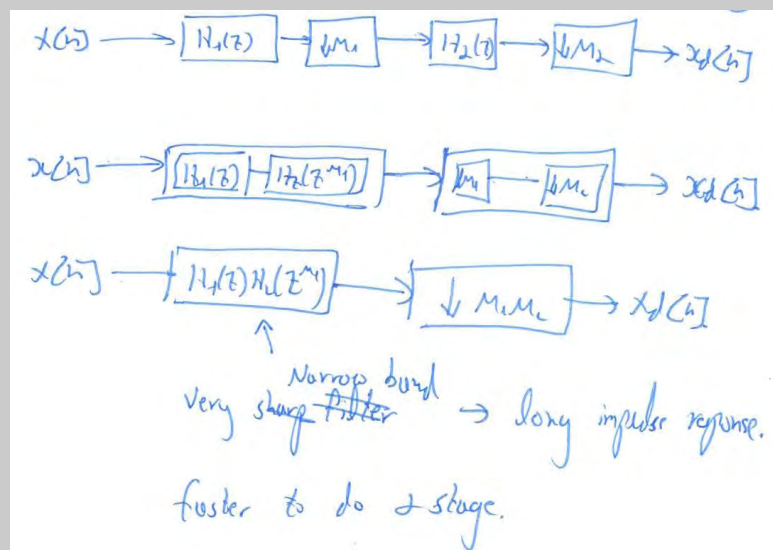
Interchanging Operations

- Last time, showed:

$$x[n] \rightarrow H(z) \rightarrow \uparrow L \rightarrow y[n] \quad \equiv \quad x[n] \rightarrow \uparrow L \rightarrow H(z^L) \rightarrow y[n]$$

$$x[n] \rightarrow \downarrow M \rightarrow H(z) \rightarrow y[n] \quad \equiv \quad x[n] \rightarrow H(z^M) \rightarrow \downarrow M \rightarrow y[n]$$

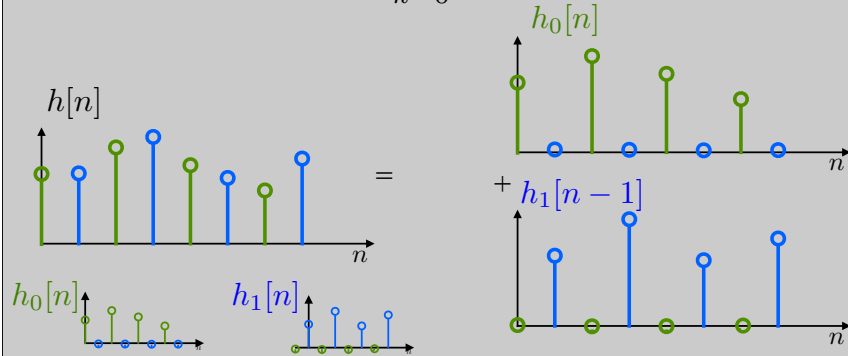
Multi-Rate Filtering



Polyphase Decomposition

- We can decomposed an impulse response to:

$$h[n] = \sum_{k=0}^{M-1} h_k[n - k]$$



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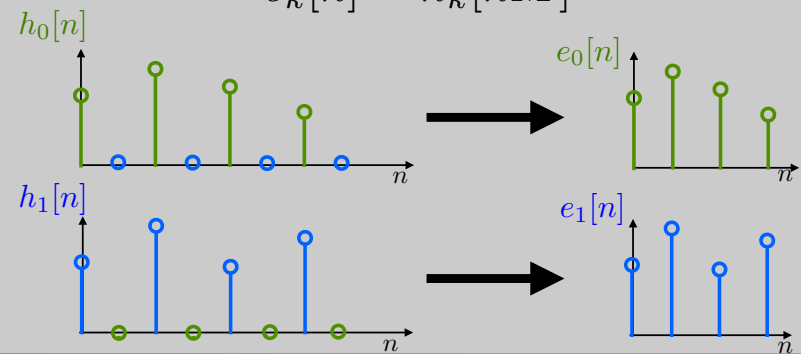
5

Polyphase Decomposition

- Define:

$$h_k[n] \rightarrow \downarrow M \rightarrow e_k[n]$$

$$e_k[n] = h_k[nM]$$



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Polyphase Decomposition

$$e_k[n] \rightarrow \uparrow M \rightarrow h_k[n]$$

recall upsampling \Rightarrow scaling

$$H_k[z] = E_k[z^M]$$

Also, recall:

$$h[n] = \sum_{k=0}^{M-1} h_k[n - k]$$

So,

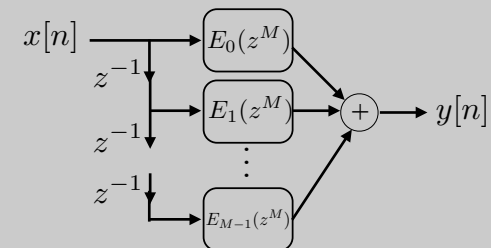
$$H(z) = \sum_{k=0}^{M-1} E_k(z^M)z^{-k}$$

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Polyphase Decomposition

$$H(z) = \sum_{k=0}^{M-1} E_k(z^M)z^{-k}$$



Why should you care?

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Polyphase Implementation of Decimation

$$x[n] \rightarrow H(z) \rightarrow y[n] \rightarrow \downarrow M \rightarrow w[n] = y[nM]$$

• Problem:

–Compute all $y[n]$ and then throw away --
wasted computation!

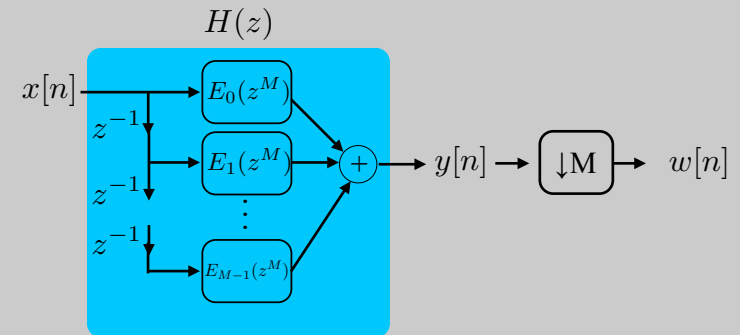
- For FIR length $N \Rightarrow N$ mults/unit time

–Can interchange Filter with compressor?

- Not in general!

Polyphase Implementation of Decimation

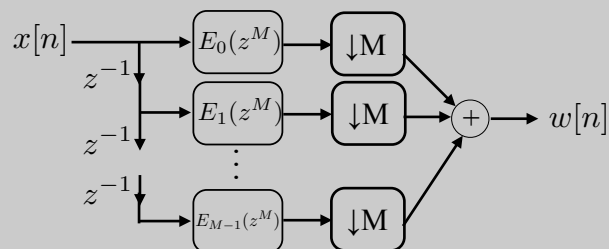
$$x[n] \rightarrow H(z) \rightarrow y[n] \rightarrow \downarrow M \rightarrow w[n] = y[nM]$$



Polyphase Implementation of Decimation

$$x[n] \rightarrow H(z) \rightarrow y[n] \rightarrow \downarrow M \rightarrow w[n] = y[nM]$$

Interchange sum with decimation

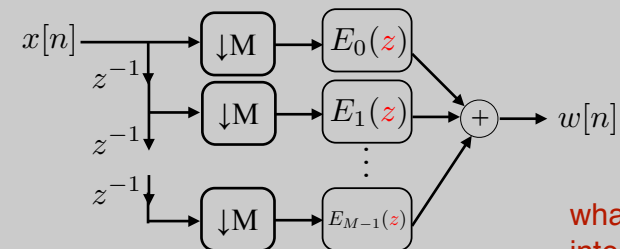


now, what can we do?

Polyphase Implementation of Decimation

$$x[n] \rightarrow H(z) \rightarrow y[n] \rightarrow \downarrow M \rightarrow w[n] = y[nM]$$

Interchange filter with decimation



what about
interpolation?

Computation:

Each Filter: $N/M * (1/M)$ mult/unit time

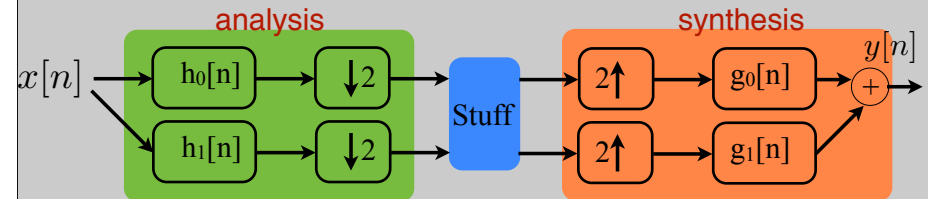
Total: N/M mult/unit time

Multirate Filter Bank

- For many types of applications we would like to process bands in frequency separately
 - Analysis/synthesis of speech, Audio, more
- Idea:
 - Split frequency
 - Decimate to reduce computation
 - (Much like wavelets!)

Multirate FilterBank

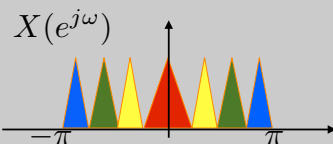
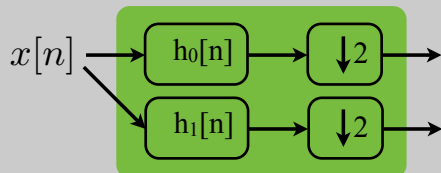
- $h_0[n]$ is low-pass, $h_1[n]$ is high-pass
- Often $h_1[n] = e^{j\omega n} h_0[n]$ or $H_1(e^{j\omega}) = H_0(e^{j(\omega-\pi)})$



Note: notation different than in lecture 9,10 for Haar

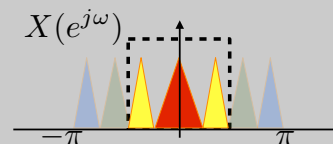
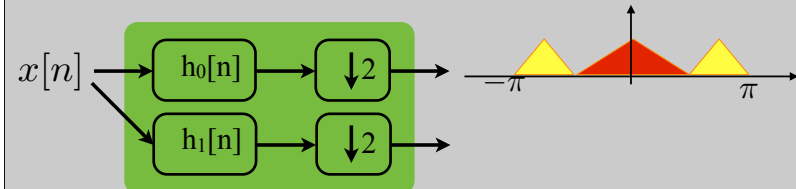
Subtleties in Time-Freq Tiling

- Assume h_0, h_1 are ideal low,high pass filters



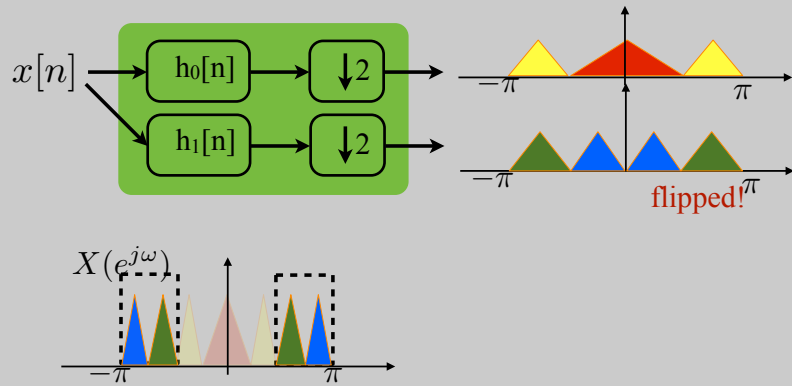
Subtleties in Time-Freq Tiling

- Assume h_0, h_1 are ideal low,high pass filters



Subtleties in Time-Freq Tiling

- Assume h_0, h_1 are ideal low, high pass filters

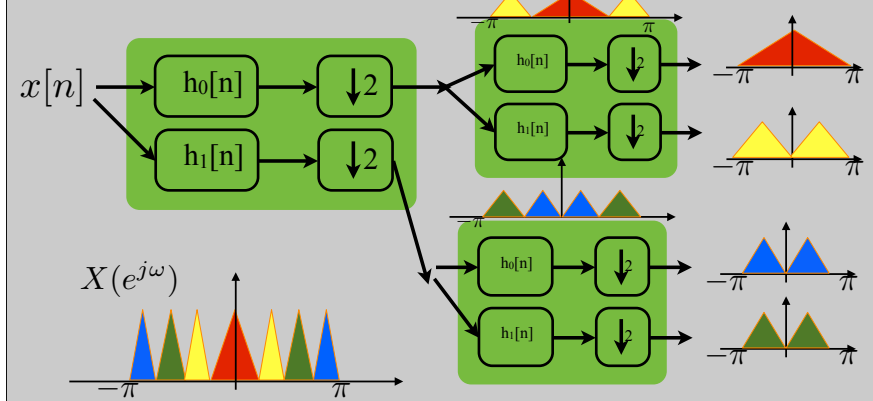


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Subtleties in Time-Freq Tiling

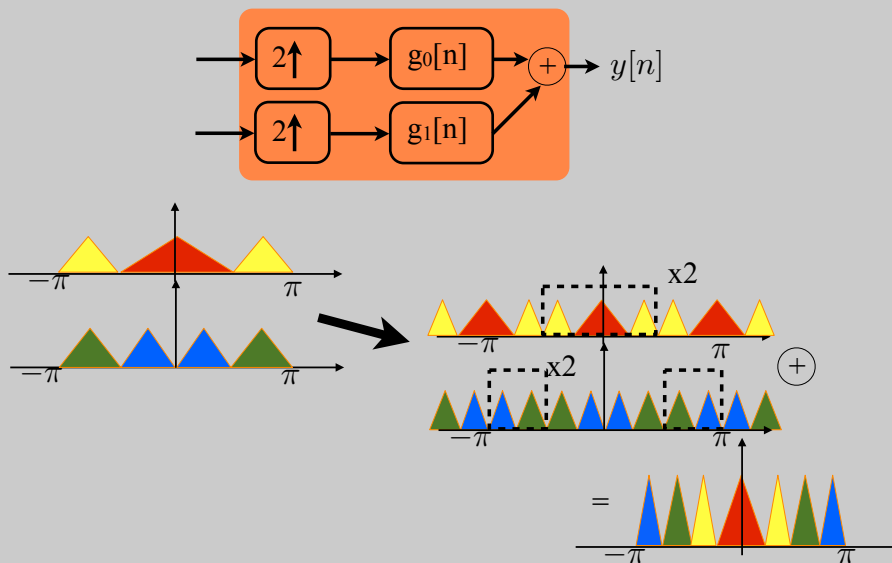
- Assume h_0, h_1 are ideal low, high pass filters



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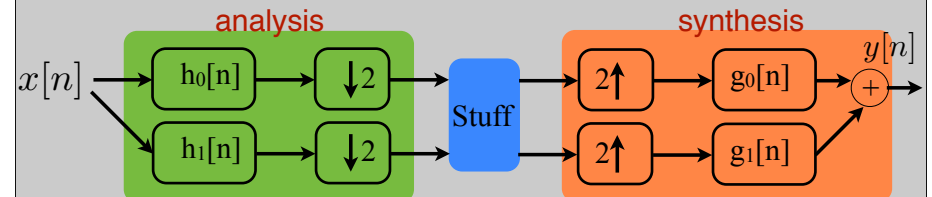
Perfect Reconstruction Ideal Filters



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Perfect Reconstruction non-Ideal Filters



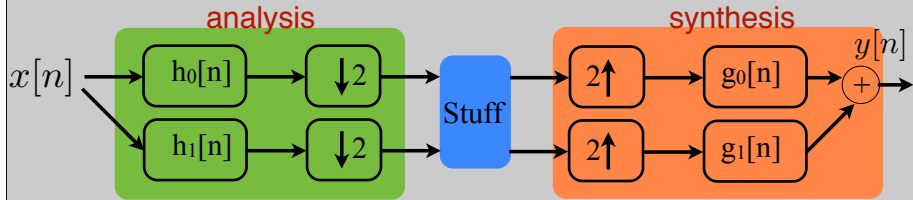
$$Y(e^{j\omega}) = \frac{1}{2} [G_0(e^{j\omega})H_0(e^{j\omega}) + G_1(e^{j\omega})H_1(e^{j\omega})] X(e^{j\omega}) + \frac{1}{2} [G_0(e^{j\omega})H_0(e^{j(\omega-\pi)}) + G_1(e^{j\omega})H_1(e^{j(\omega-\pi)})] X(e^{j(\omega-\pi)})$$

need to cancel! aliasing

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Quadrature Mirror Filters - perfect recon



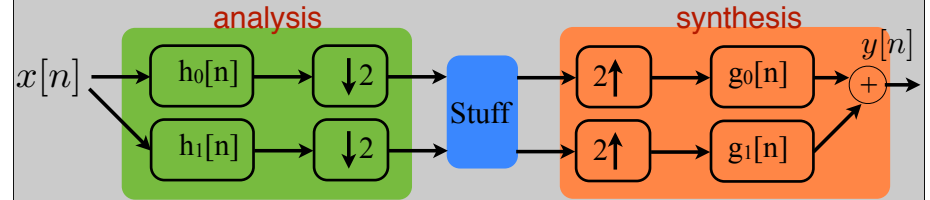
QMF - mirror around $\pi/2$

$$H_1(e^{j\omega}) = H_0(e^{j(\omega-\pi)})$$

$$G_0(e^{j\omega}) = 2H_0(e^{j\omega})$$

$$G_1(e^{j\omega}) = -2H_1(e^{j\omega})$$

Quadrature Mirror Filters - perfect recon



QMF - mirror around $\pi/2$

$$H_1(e^{j\omega}) = H_0(e^{j(\omega-\pi)})$$

$$G_0(e^{j\omega}) = 2H_0(e^{j\omega})$$

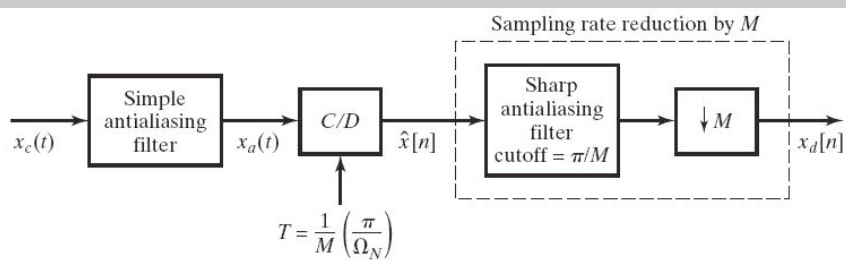
$$G_1(e^{j\omega}) = -2H_1(e^{j\omega})$$

Example Haar:

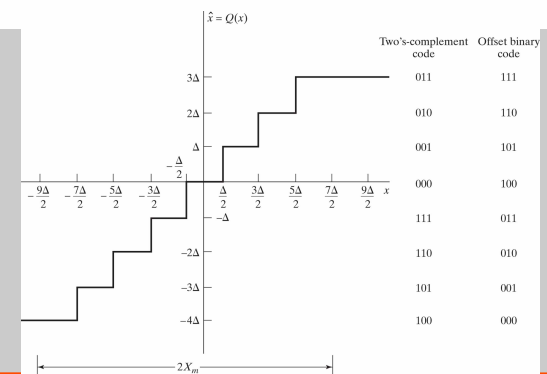
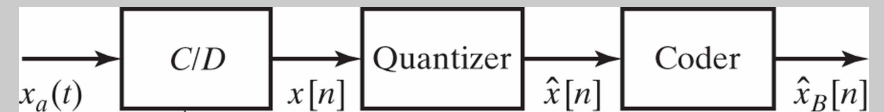


A Few Notes on Practical A/D

Oversampling A/D to simplify C.T. Anti-Aliasing Filter

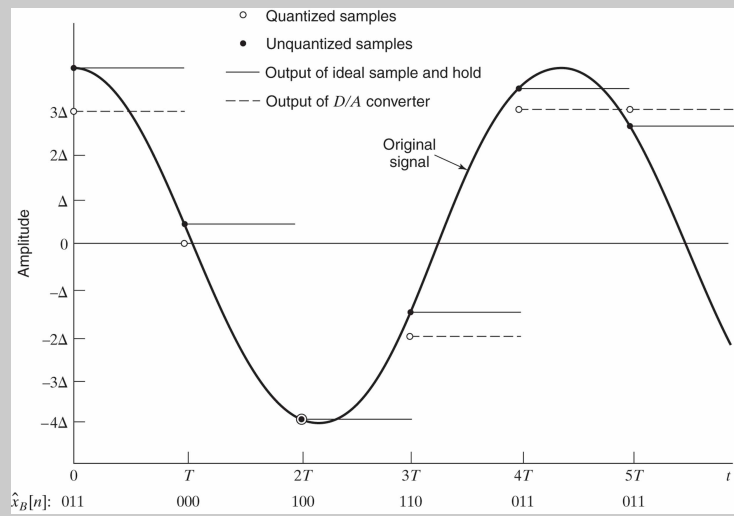


Sampling and Quantization



Sampling and Quantization

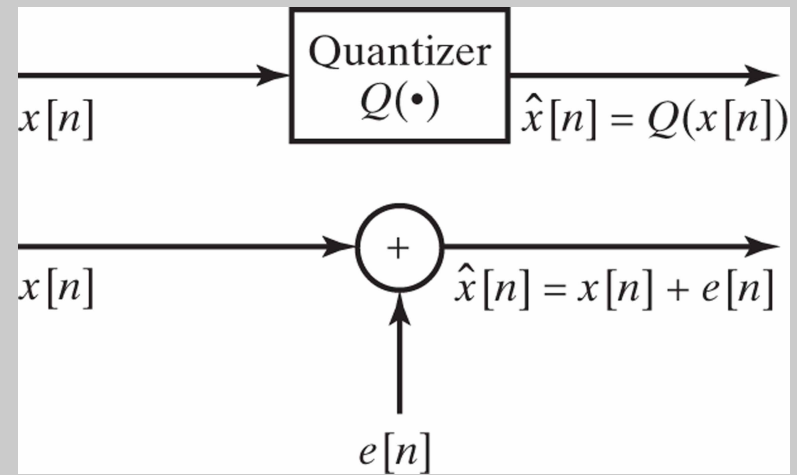
Figure 4.55 Sampling, quantization, coding, and D/A conversion with a 3-bit quantizer.



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Noise Model For Quantization

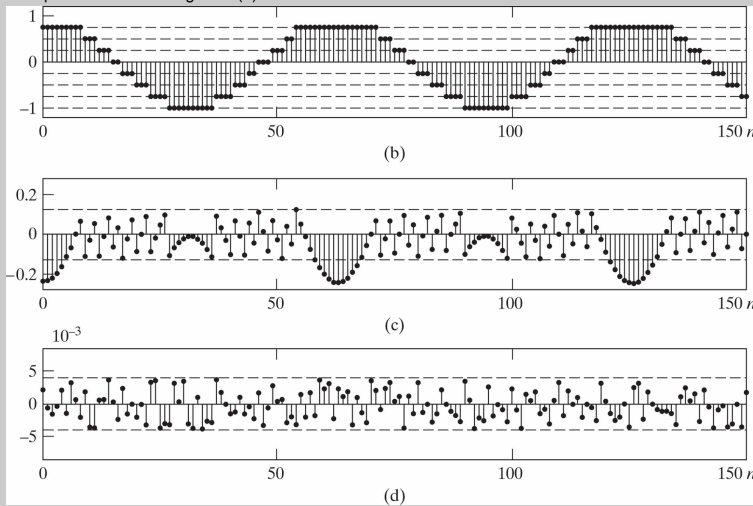


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Quantization Noise

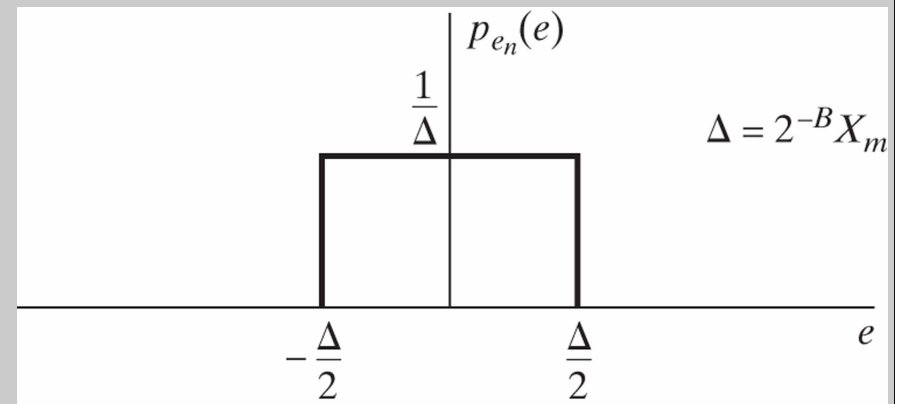
Figure 4.57 (continued) (b) Quantized samples of the cosine waveform in part (a) with a 3-bit quantizer. (c) Quantization error sequence for 3-bit quantization of the signal in (a). (d) Quantization error sequence for 8-bit quantization of the signal in (a).



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PDF of Quantization Noise



$$\text{SNR}_Q = 6.02B + 10.8 - 20 \log_{10} \left(\frac{X_m}{\sigma_x} \right) \begin{matrix} \text{Signal amp} \\ \text{rms of amp} \end{matrix}$$

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