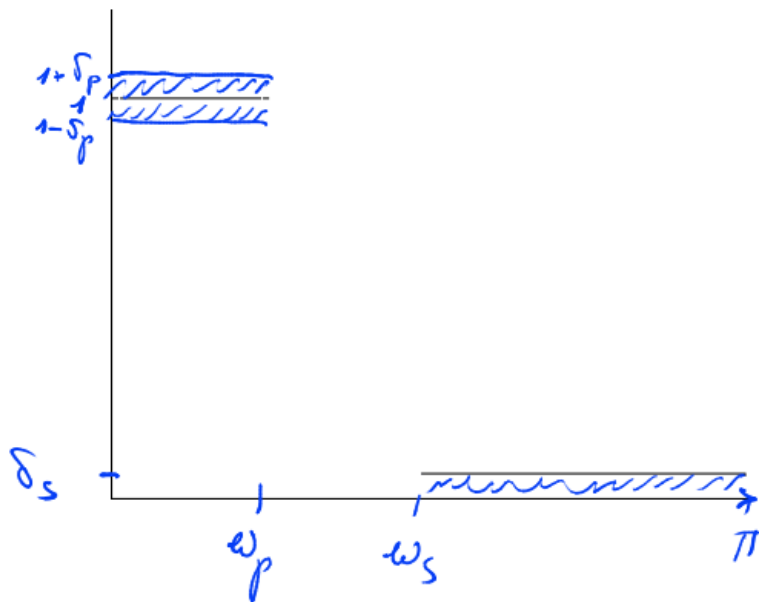


Lecture 11/09/11

Optimal min-max design through optimization:

SPECS:



(a) Maximum pass-band ripple:

$$1 - \delta_p \leq |H(e^{j\omega})| \leq 1 + \delta_p, \quad 0 \leq \omega \leq \omega_p$$

(b) Minimum stop-band attenuation

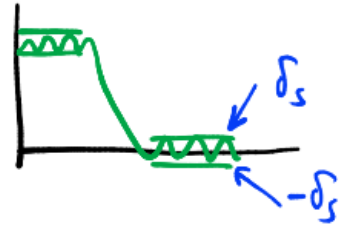
$$|H(e^{j\omega})| \leq \delta_s, \quad \omega_s \leq \omega \leq \pi$$

maximum stop-band attenuation (linear phase) ⊕

minimize δ_s

Recall $\tilde{H}(e^{j\omega})$ is Re

subject to: $1 - \delta_p \leq \tilde{H}(e^{j\omega_k}) \leq 1 + \delta_p, 0 \leq \omega_k \leq \omega_p$
 $-\delta_s \leq \tilde{H}(e^{j\omega_k}) \leq \delta_s, \omega_s \leq \omega_k \leq \pi$



$\delta_s > 0$

$$\tilde{H}(e^{j\omega_k}) \Big|_{0 \leq \omega_k \leq \omega_p} = \begin{bmatrix} 1 & 2\cos(\omega_k) & \dots & 2\cos(\omega_k \frac{M}{2}) \\ \vdots & \vdots & \ddots & \vdots \\ 1 & 2\cos(\omega_p) & \dots & 2\cos(\omega_p \frac{M}{2}) \end{bmatrix}$$

⊕ Similarly for $\omega_s \leq \omega_k \leq \pi$

⊕ $\omega_p, \omega_s, M, \delta_p$ given

⊕ solution is a linear program in \tilde{h}, δ_s

⊕ know δ used since 60's

⊕ can add constraints e.g. $|h_i| \leq \kappa$

* Based on notes by Stephen Boyd

```

M = 16;
wp = 0.5*pi;
ws = 0.6*pi;
dp = 0.05;

MM = M*15;
w = linspace(0,pi,MM);

idxp = find(w <=wp);
idxs = find(w >=ws);

Ap = [ones(length(idxp),1) 2*cos(kron(w(idxp)', [1:M/2]))];
As = [ones(length(idxs),1) 2*cos(kron(w(idxs)', [1:M/2]))];

% optimization

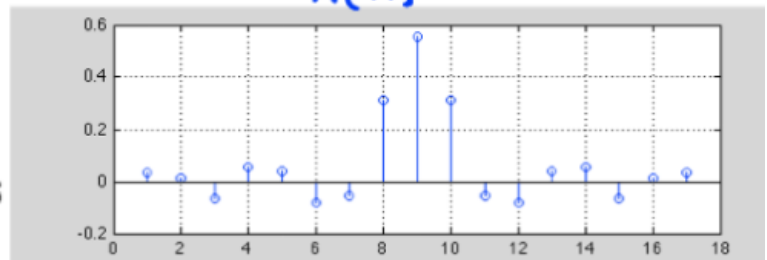
cvx_begin
    variable hh(M/2+1,1);
    variable ds(1,1);

    minimize(ds)
    subject to
        Ap*hh <=1+dp;
        Ap*hh >=1-dp;
        As*hh < ds;
        As*hh > -ds;
        ds>0;
cvx_end

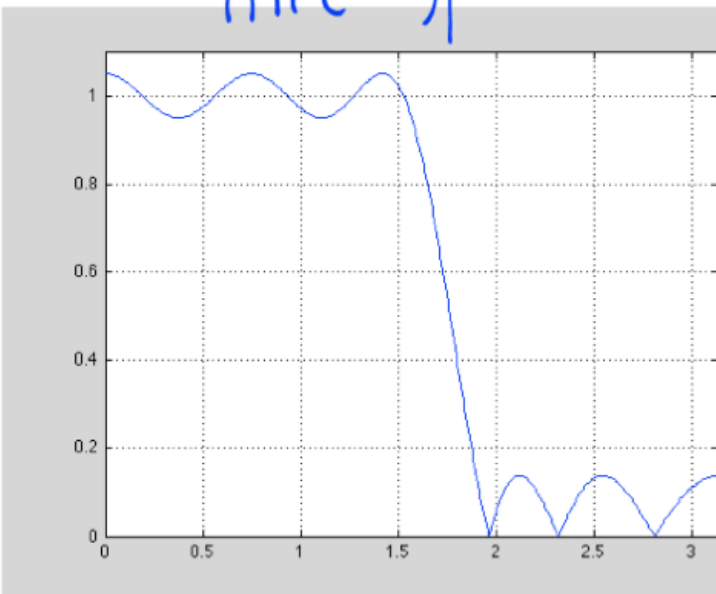
h = [hh(end:-1:1) ; hh(2:end)];

```

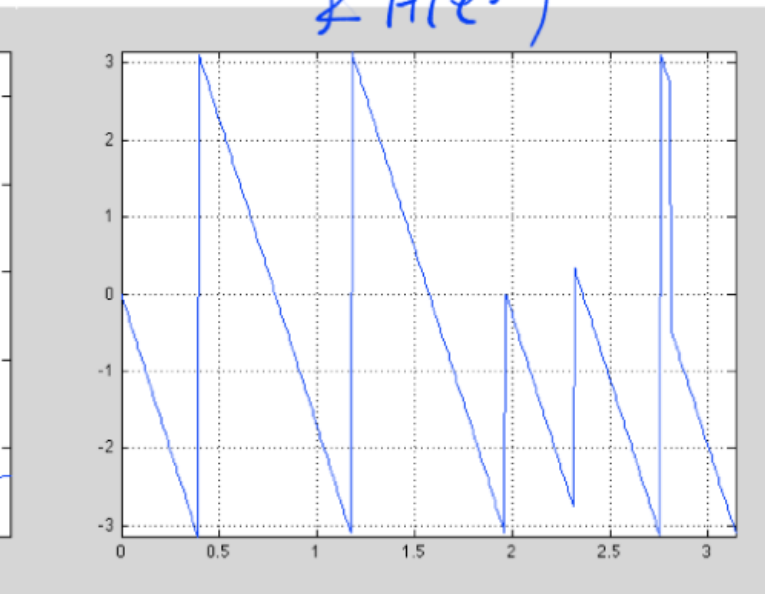
$h[n]$



$|H(e^{j\omega})|$



$\angle H(e^{j\omega})$



Variations

(-) fix δ_s , minimize δ_p

(-) fix δ_s and δ_p , M min. $\omega_s - \omega_p = \Delta\omega$

use bisection: (-) fix $\Delta\omega, \delta_p, M$, solve \tilde{h}, δ_s
 (-) if feasible decrease $\Delta\omega$
 if not increase

(-) fix $\delta_p, \delta_s, \omega_p, \omega_s$ minimize M
bisection!

IIR Design

(6)

Historically \rightarrow Continuous IIR design was highly advanced
use results from C.T to D.T

- C.T IIR designs have closed form \rightarrow easy to use.
- With these designs, easy to control amp. response but not phase response.

Common types:

- (1) Butterworth \rightarrow monotonic, no ripple
- (2) Chebyshev \rightarrow type I pass-band ripple
type II stop band ripple
- (3) Elliptic \rightarrow Ripples in both s.b and p.b
Has zeroes.

Design of DT IIR Filters from Analog.

(7)

Discretize by using one of many techniques.

$$H_c(s) \rightarrow H(z)$$

must satisfy:

(1) Imaginary axis \rightarrow mapped to unit circle

$$s = j\Omega \Rightarrow z = e^{j\omega}$$

(2) stability of $H_c(s)$ should result in stable $H(z)$

Two methods:

(1) Impulse invariance \rightarrow match impulse

$$h[n] = T h_c(nT)$$

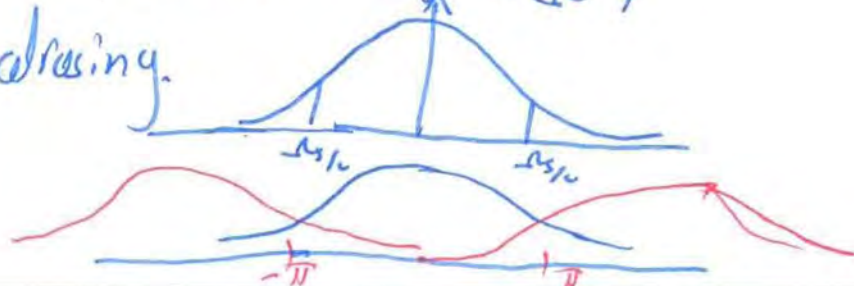
\Downarrow

$$H(e^{j\omega}) = \sum_{k=-\infty}^{\infty} H_c(j\Omega + j\Omega_s k) \Big|_{\Omega = \frac{\omega}{T}}$$

if $H_c(j\Omega)$ band limited $|\Omega| \leq \Omega_s/2$ then:

$$H(e^{j\omega}) = H_c(j\Omega) \Big|_{\Omega = \frac{\omega}{T}}$$

otherwise \rightarrow aliasing.



(-) Bilinear transform \rightarrow basic idea.

(8)

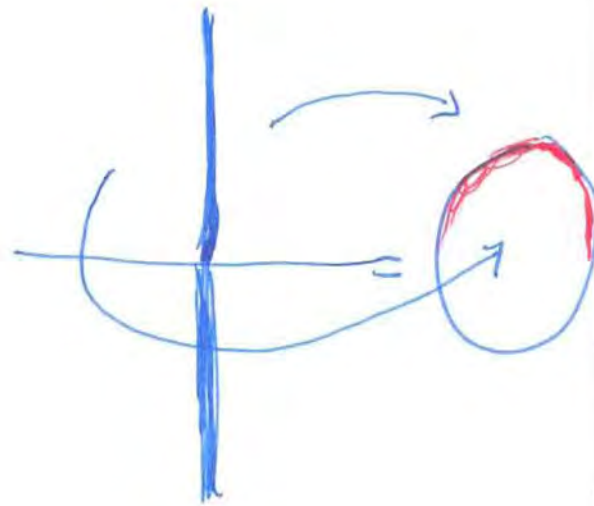
(-) Distorts $H_c(j\Omega)$ such that it is band limited.

$$s = \frac{2}{T} \frac{1-z^{-1}}{1+z^{-1}}$$

$$z = \frac{1 + \frac{T}{2}s}{1 - \frac{T}{2}s}$$

for $s=0 \Rightarrow z=1$

$s=\infty \Rightarrow z=-1$



Matlab:

butter, cheby1, cheby2, ellip

Next time FIR ...