

## EE120 Fall 2016: PS6 Solutions

GSI: Phil Sandborn

Problem 1.

Given:  $\mathcal{F}\{x[n]\} = X(e^{j\omega})$

$$\text{a) } \mathcal{F}\{\text{Im}\{x[n]\}\} = \mathcal{F}\left\{\frac{x[n]-x^*[n]}{2j}\right\}$$

$$\mathcal{F}\{\text{Im}\{x[n]\}\} = \frac{\mathcal{F}\{x[n]\}}{2j} - \frac{\mathcal{F}\{x^*[n]\}}{2j} = \frac{1}{2j}X(e^{j\omega}) - \frac{1}{2j}\sum_{n=-\infty}^{\infty} x^*[n]e^{-j\omega n}$$

$$\mathcal{F}\{\text{Im}\{x[n]\}\} = \frac{1}{2j}X(e^{j\omega}) - \frac{1}{2j}\left(\sum_{n=-\infty}^{\infty} x[n]e^{j\omega n}\right)^*$$

$$\mathcal{F}\{\text{Im}\{x[n]\}\} = \frac{1}{2j}X(e^{j\omega}) - \frac{1}{2j}X^*(e^{-j\omega})$$

$$\text{b) } \mathcal{F}\{x[-n]\} = \sum_{n=-\infty}^{\infty} x[-n]e^{-j\omega n}$$

Using the substitution:  $n' = -n$ ,

$$\mathcal{F}\{x[-n]\} = \sum_{n'=-\infty}^{\infty} x[n']e^{j\omega n'}$$

$$\mathcal{F}\{x[-n]\} = X(e^{-j\omega})$$

$$\text{c) } \mathcal{F}\{\text{Odd}\{x[n]\}\} = \mathcal{F}\left\{\frac{x[n]}{2} - \frac{x[-n]}{2}\right\}$$

$$\mathcal{F}\{\text{Odd}\{x[n]\}\} = \frac{1}{2}X(e^{j\omega}) - \frac{1}{2}\mathcal{F}\{x[-n]\}$$

$$\mathcal{F}\{\text{Odd}\{x[n]\}\} = \frac{1}{2}X(e^{j\omega}) - \frac{1}{2}X(e^{-j\omega})$$

## Problem 2.

Given this LDE:

$$y[n] + \frac{1}{2}y[n-1] = x[n]$$

a) Frequency Response?

$$Y(e^{j\omega}) + \frac{1}{2}Y(e^{j\omega})e^{-j\omega} = X(e^{j\omega})$$

$$Y(e^{j\omega})\left(1 + \frac{1}{2}e^{-j\omega}\right) = X(e^{j\omega})$$

$$H(e^{j\omega}) = \frac{Y(e^{j\omega})}{X(e^{j\omega})} = \frac{1}{\left(1 + \frac{1}{2}e^{-j\omega}\right)}$$

b) Output for  $x_1[n] = \left(\frac{1}{2}\right)^n u[n]$  ?

$$X_1(e^{j\omega}) = \frac{1}{\left(1 - \frac{1}{2}e^{-j\omega}\right)}$$

$$Y(e^{j\omega}) = X_1(e^{j\omega})H(e^{j\omega}) = \frac{1}{\left(1 - \frac{1}{2}e^{-j\omega}\right)\left(1 + \frac{1}{2}e^{-j\omega}\right)}$$

When taking IDTFT, we have a formula of the form of  $G(e^{j\omega}) = \frac{1}{\left(1 - \frac{1}{4}e^{-j\omega}\right)}$ , with a frequency-scaling. First we take the inverse transform of  $G(e^{j\omega})$  to get a discrete time signal  $g[n]$  then apply the frequency scaling rule to get  $y[n]$ .

$$G(e^{j2\omega}) = \frac{1}{\left(1 - \frac{1}{4}e^{-j[2\omega]}\right)}$$

$$G(e^{j\omega}) = \frac{1}{\left(1 - \frac{1}{4}e^{-j\omega}\right)}$$

$$g[n] = \left(\frac{1}{4}\right)^n u[n]$$

$$y[n] = \left(\frac{1}{4}\right)^{n/2} u[n/2] \text{ for } n \text{ even, and zero for } n \text{ odd.}$$

c) Output for  $x_2[n] = \left(-\frac{1}{2}\right)^n u[n]$  ?

$$X_2(e^{j\omega}) = \frac{1}{\left(1 + \frac{1}{2}e^{-j\omega}\right)}$$

$$Y(e^{j\omega}) = X_2(e^{j\omega})H(e^{j\omega}) = \frac{1}{\left(1 + \frac{1}{2}e^{-j\omega}\right)\left(1 + \frac{1}{2}e^{-j\omega}\right)}$$

$$Y(e^{j\omega}) = \frac{1}{\left(1 + \frac{1}{2}e^{-j\omega}\right)^2}$$

This has a form from the DTFT pair table, so we can find the output discrete time signal:

$$y[n] = (n+1)\left(-\frac{1}{2}\right)^n u[n]$$

- d) Output for  $x_3[n] = \delta[n] + \frac{1}{2}\delta[n - 1]$  ?

$$X_3(e^{j\omega}) = 1 + \frac{1}{2}e^{-j\omega}$$

$$Y(e^{j\omega}) = X_3(e^{j\omega}) H(e^{j\omega}) = \left(1 + \frac{1}{2}e^{-j\omega}\right) \frac{1}{\left(1 + \frac{1}{2}e^{-j\omega}\right)} = 1$$

A constant DTFT has a corresponding pair with a Kronecker delta function, so the output discrete time signal:

$$y[n] = \delta[n]$$

- e) Output for  $x_4[n] = \delta[n] - \frac{1}{2}\delta[n - 1]$  ?

$$X_4(e^{j\omega}) = 1 - \frac{1}{2}e^{-j\omega}$$

$$Y(e^{j\omega}) = X_4(e^{j\omega}) H(e^{j\omega}) = \frac{\left(1 - \frac{1}{2}e^{-j\omega}\right)}{\left(1 + \frac{1}{2}e^{-j\omega}\right)} = \frac{1}{\left(1 + \frac{1}{2}e^{-j\omega}\right)} - \frac{1}{2} \frac{e^{-j\omega}}{\left(1 + \frac{1}{2}e^{-j\omega}\right)}$$

Using DTFT pairs and properties, we can directly write output discrete time signal and simplify:

$$y[n] = \left(-\frac{1}{2}\right)^n u[n] - \frac{1}{2}\delta[n - 1] * \left(\left(-\frac{1}{2}\right)^n u[n]\right)$$

$$y[n] = \left(-\frac{1}{2}\right)^n u[n] + \left(-\frac{1}{2}\right)^n u[n - 1]$$

$$y[n] = \left(-\frac{1}{2}\right)^n (u[n] + u[n - 1])$$

- f) Output for  $X_5(e^{j\omega}) = \frac{\left(1 - \frac{1}{4}e^{-j\omega}\right)}{\left(1 + \frac{1}{2}e^{-j\omega}\right)}$  ?

$$Y(e^{j\omega}) = X_5(e^{j\omega}) H(e^{j\omega}) = \frac{\left(1 - \frac{1}{4}e^{-j\omega}\right)}{\left(1 + \frac{1}{2}e^{-j\omega}\right)\left(1 + \frac{1}{2}e^{-j\omega}\right)}$$

$$Y(e^{j\omega}) = \frac{1}{\left(1 + \frac{1}{2}e^{-j\omega}\right)^2} - \frac{1}{4} \frac{e^{-j\omega}}{\left(1 + \frac{1}{2}e^{-j\omega}\right)^2}$$

Using DTFT pairs and properties, we can directly write output discrete time signal and simplify:

$$y[n] = (n + 1) \left(-\frac{1}{2}\right)^n u[n] - \frac{1}{4}\delta[n - 1] * \left((n + 1) \left(-\frac{1}{2}\right)^n u[n]\right)$$

$$y[n] = (n + 1) \left(-\frac{1}{2}\right)^n u[n] - \frac{1}{4} \left((n) \left(-\frac{1}{2}\right)^{n-1} u[n - 1]\right)$$

$$y[n] = \left(-\frac{1}{2}\right)^n \left\{ (n + 1)u[n] + \frac{1}{2}n \cdot u[n - 1] \right\}$$

- g) Output for  $X_6(e^{j\omega}) = 1 + 2e^{-3j\omega}$  ?

$$Y(e^{j\omega}) = X_6(e^{j\omega}) H(e^{j\omega}) = \frac{\left(1 + 2e^{-3j\omega}\right)}{\left(1 + \frac{1}{2}e^{-j\omega}\right)}$$

$$Y(e^{j\omega}) = \frac{1}{\left(1 + \frac{1}{2}e^{-j\omega}\right)} + 2 \frac{e^{-3j\omega}}{\left(1 + \frac{1}{2}e^{-j\omega}\right)}$$

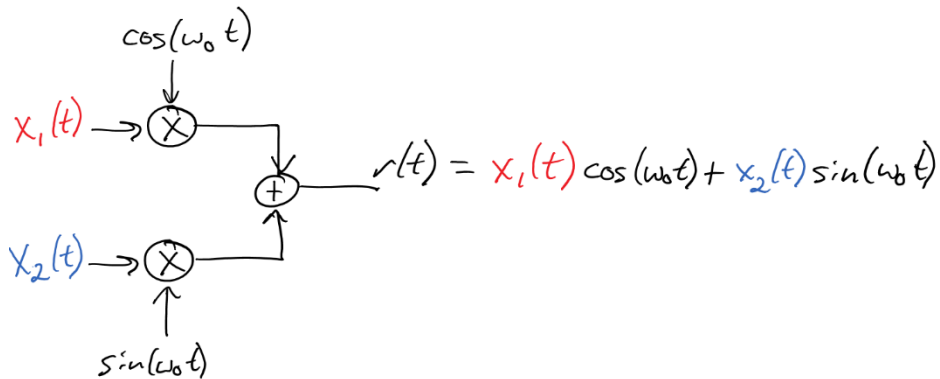
$$y[n] = \left(-\frac{1}{2}\right)^n u[n] + 2\delta[n - 3] * \left\{ \left(-\frac{1}{2}\right)^n u[n] \right\}$$

$$y[n] = \left(-\frac{1}{2}\right)^n u[n] + 2 \left(-\frac{1}{2}\right)^{n-3} u[n - 3]$$

$$y[n] = \left(-\frac{1}{2}\right)^n (u[n] - 16u[n - 3])$$

### Problem 3.

Part a.

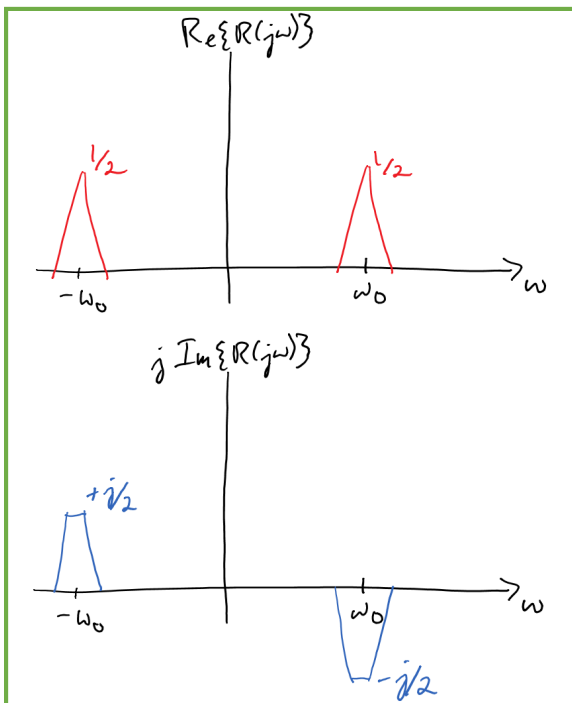


$$R(j\omega) = \frac{1}{2\pi} \underline{X}_1(j\omega) * \pi [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$$

$$+ \frac{1}{2\pi} \underline{X}_2(j\omega) * \pi [\delta(\omega - \omega_0) - \delta(\omega + \omega_0)]$$

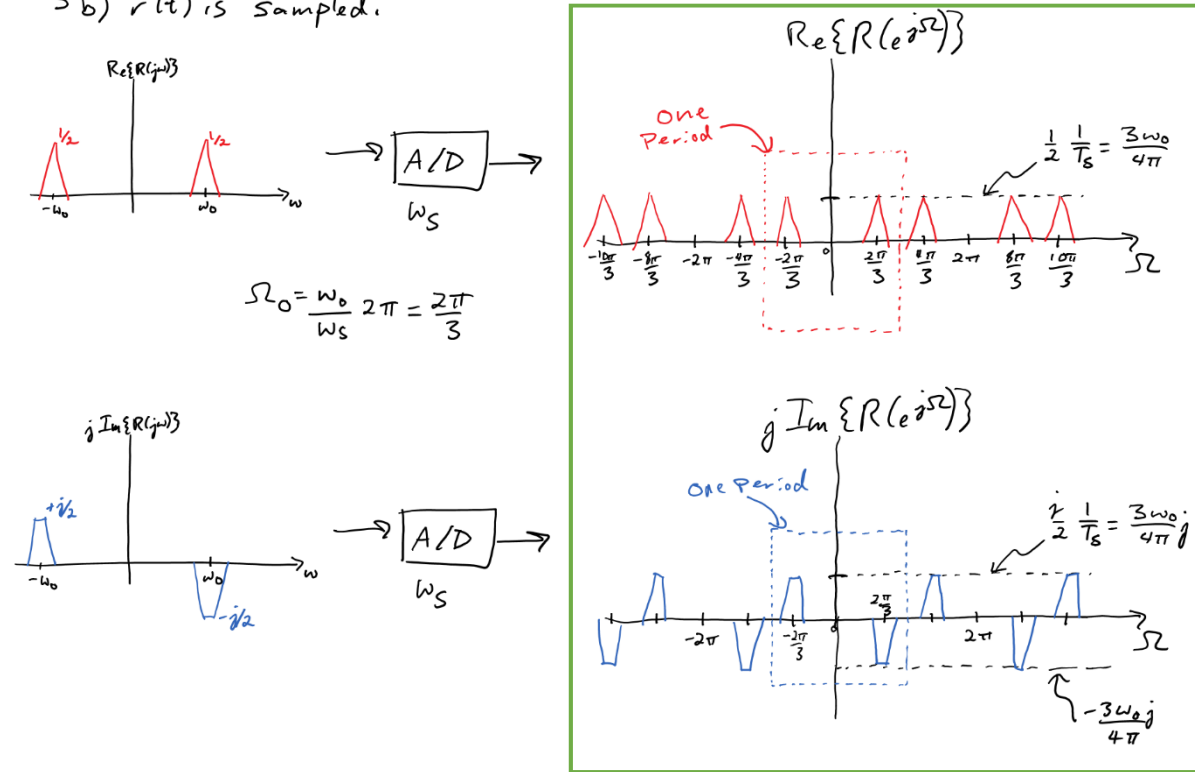
$\underline{X}_1(j\omega)$  and  $\underline{X}_2(j\omega)$  are real, so  $\text{Re}\{R(j\omega)\} = \frac{1}{2\pi} \underline{X}_1(j\omega) * \pi [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$

$$\text{Im}\{R(j\omega)\} = \frac{1}{2\pi} \underline{X}_2(j\omega) * (-1)\pi [\delta(\omega - \omega_0) - \delta(\omega + \omega_0)]$$

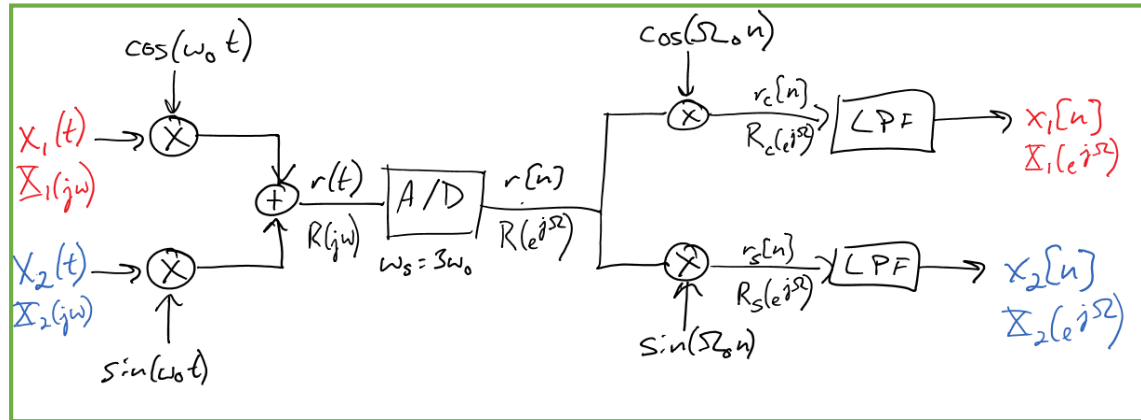


Part 3b.

3 b)  $r(t)$  is sampled.



Part 3c.



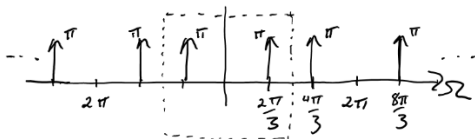
Part 3d.

$$r_c[n] = r[n] \cdot \cos\left(\frac{2\pi n}{3}\right) \xleftrightarrow{\text{DTFT}} \frac{1}{2\pi} R(e^{j\Omega}) \circledast \pi \sum_l \left[ \delta\left(\Omega - \frac{2\pi}{3} - 2\pi l\right) + \delta\left(\Omega + \frac{2\pi}{3} - 2\pi l\right) \right] = R_c(e^{j\Omega})$$

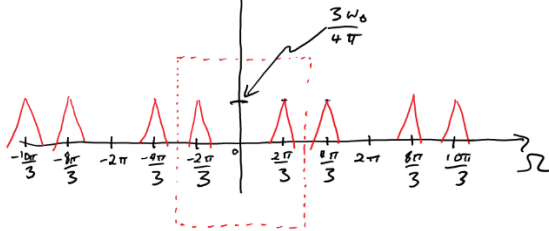
convolve both real and imaginary parts separately:

Real part

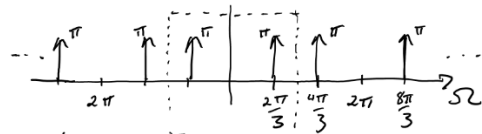
$$\text{DTFT}\{\cos(\Omega_0 n)\}$$



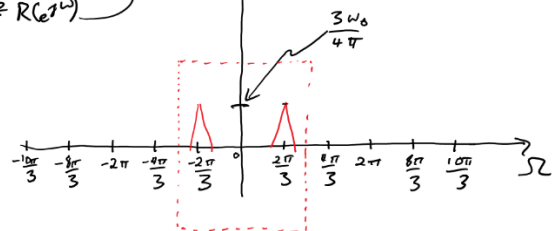
$$\circledast \text{Re}\{R(e^{j\Omega})\}$$



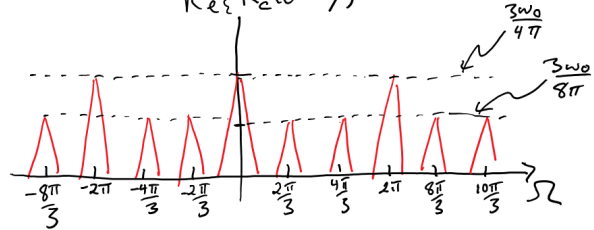
$$\text{DTFT}\{\cos(\Omega_0 n)\}$$



$$\text{Define } \hat{R}(e^{j\Omega}) = \text{one period of } R(e^{j\Omega}) \rightarrow \text{Regular Convolution} \rightarrow \text{Re}\{\hat{R}(e^{j\Omega})\}$$

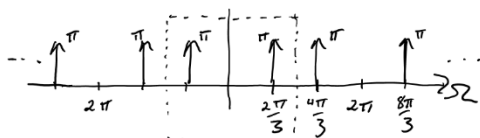


$$\text{Re}\{R_c(e^{j\Omega})\}$$

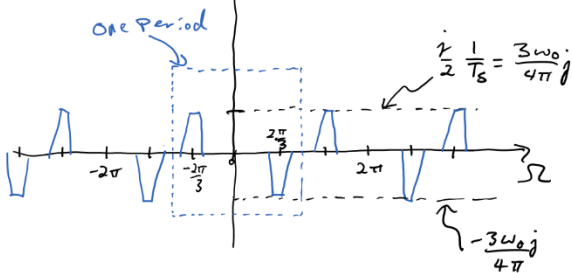


Imaginary part

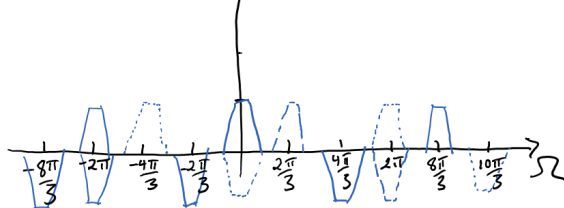
$$\text{DTFT}\{\cos(\Omega_0 n)\}$$



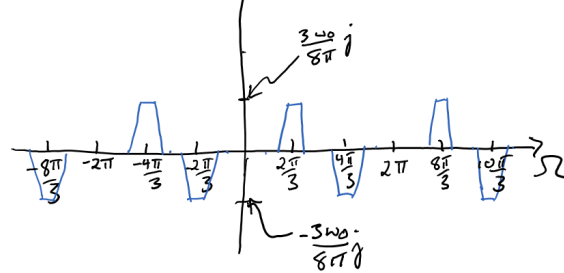
$$j \text{Im}\{R(e^{j\Omega})\}$$

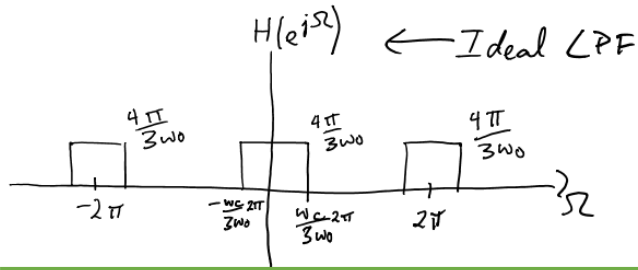


$$\text{Im}\{R_c(e^{j\Omega})\}$$



$$\text{Im}\{R_c(e^{j\Omega})\}$$



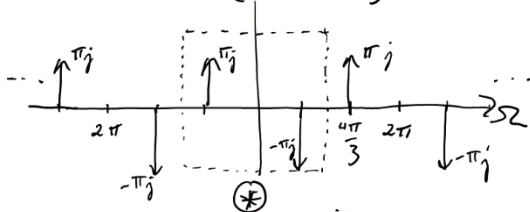


$$r_s[n] = r[n] \cdot \sin\left(\frac{2\pi}{3}n\right) \xleftrightarrow{\text{DTFT}} \frac{1}{2\pi} R(e^{j\Omega}) \circledast \int_{-\pi}^{\pi} \delta\left(\Omega - \frac{2\pi}{3} - 2\pi l\right) - \delta\left(\Omega + \frac{2\pi}{3} - 2\pi l\right) = R_s(e^{j\Omega})$$

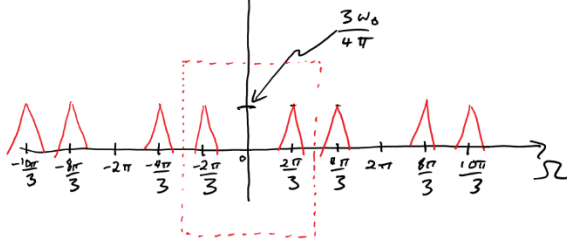
convolve both real and imaginary parts separately:

Real part

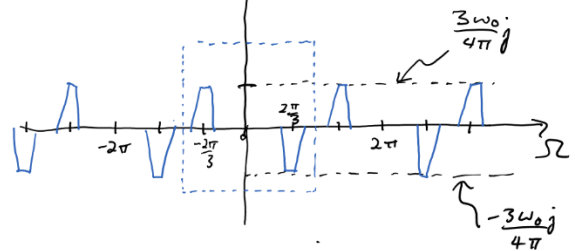
$$\text{DTFT}\{\sin(\Omega_0 n)\}$$

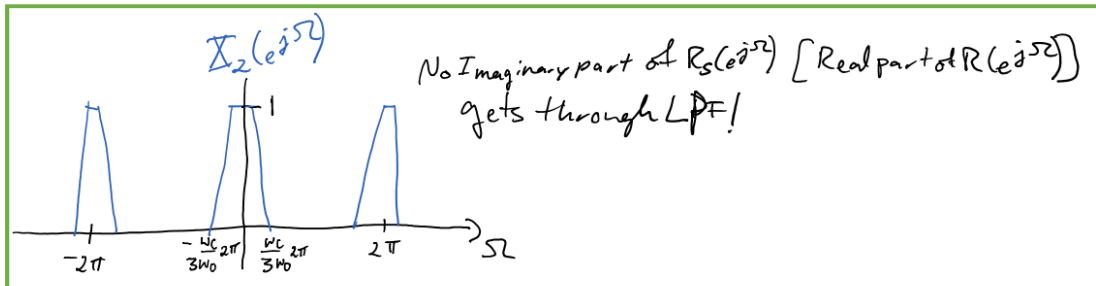
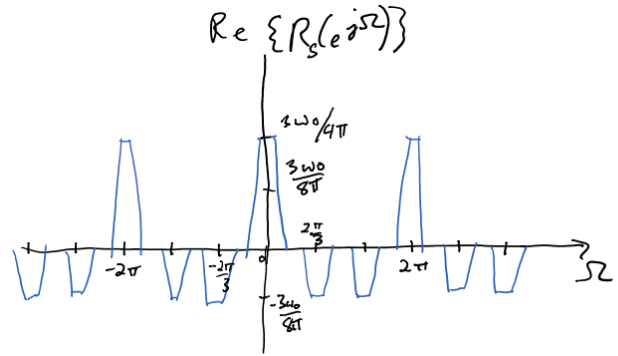
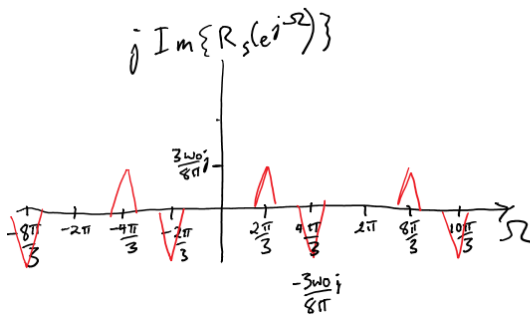
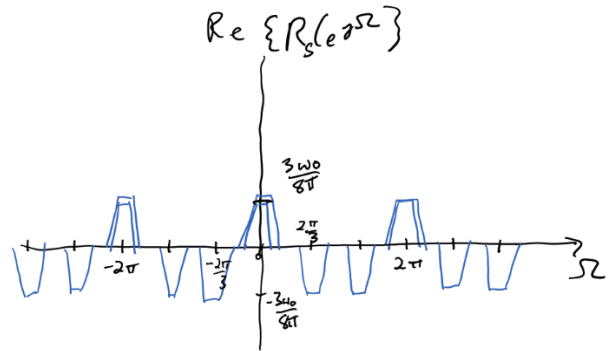
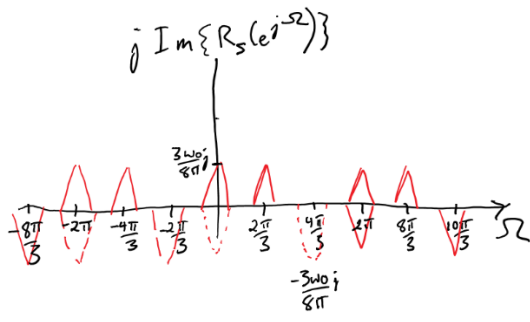


$$\text{Re}\{R(e^{j\Omega})\}$$



$$j \text{Im}\{R(e^{j\Omega})\}$$







Problem 4.

$$a) H(e^{j\omega}) = e^{-j\omega} \frac{1 - \frac{1}{2}e^{j\omega}}{1 - \frac{1}{2}e^{-j\omega}}$$

$$|H(e^{j\omega})|^2 = \left| e^{-j\omega} \frac{1 - \frac{1}{2}e^{j\omega}}{1 - \frac{1}{2}e^{-j\omega}} \right|^2$$

$$|H(e^{j\omega})|^2 = \left| \frac{1 - \frac{1}{2}e^{j\omega}}{1 - \frac{1}{2}e^{-j\omega}} \right|^2$$

$$|H(e^{j\omega})|^2 = \left[ \frac{1 - \frac{1}{2}e^{j\omega}}{1 - \frac{1}{2}e^{-j\omega}} \right] \cdot \left[ \frac{1 - \frac{1}{2}e^{-j\omega}}{1 - \frac{1}{2}e^{j\omega}} \right] = 1$$

$$\boxed{|H(e^{j\omega})| = 1}$$

QED.

We have used the fact that for any complex number,  $z$ ,  $|z|^2 = zz^*$ .

b) If we let:

$$z_1 = e^{-j\omega} = r_1 e^{j\theta_1}; z_2 = 1 - \frac{1}{2}e^{j\omega} = r_2 e^{j\theta_2}; \text{ and } z_3 = 1 - \frac{1}{2}e^{-j\omega} = r_3 e^{j\theta_3},$$

We can write  $H(e^{j\omega})$ :

$$H(e^{j\omega}) = \frac{z_1 z_2}{z_3} = \frac{r_1 r_2}{r_3} \exp(\theta_1 + \theta_2 - \theta_3)$$

We can write  $\angle H(e^{j\omega}) = \theta_1 + \theta_2 - \theta_3$

$$\theta_1 = -\omega$$

$$z_2 = 1 - \frac{1}{2}e^{j\omega} = 1 - \frac{1}{2}\cos(\omega) - \frac{j}{2}\sin(\omega);$$

$$\theta_2 = \text{atan} \left( -\frac{\sin(\omega)}{2} \frac{1}{1 - \frac{1}{2}\cos(\omega)} \right)$$

$$z_3 = 1 - \frac{1}{2}e^{-j\omega} = 1 - \frac{1}{2}\cos(-\omega) - \frac{j}{2}\sin(-\omega);$$

$$\theta_3 = \text{atan} \left( -\frac{\sin(-\omega)}{2} \frac{1}{1 - \frac{1}{2}\cos(-\omega)} \right)$$

$$\theta_3 = \text{atan} \left( \frac{\sin(\omega)}{2} \frac{1}{1 - \frac{1}{2}\cos(\omega)} \right)$$

Arctangent is an odd function, so  $-\theta_2 = \theta_3$

$$\boxed{\angle H(e^{j\omega}) = \theta_1 + \theta_2 - \theta_3 = -\omega + 2 \text{atan} \left( \frac{\sin(\omega)}{\cos(\omega) - 2} \right)}$$

c) The input function,  $x[n] = \cos\left(\frac{\pi}{3}n\right)$  can be expressed in basis functions:

$$x[n] = \frac{1}{2}e^{j\frac{\pi}{3}n} + \frac{1}{2}e^{-j\frac{\pi}{3}n}$$

Therefore, we can write the output of the system,

$$y[n] = \frac{1}{2}e^{j\frac{\pi}{3}n} |H(e^{j\frac{\pi}{3}})| e^{j\angle H(e^{j\frac{\pi}{3}})} + \frac{1}{2}e^{-j\frac{\pi}{3}n} |H(e^{-j\frac{\pi}{3}})| e^{j\angle H(e^{-j\frac{\pi}{3}})}$$

$$y[n] = \frac{1}{2}e^{j\frac{\pi}{3}n} e^{j\angle H(e^{j\frac{\pi}{3}})} + \frac{1}{2}e^{-j\frac{\pi}{3}n} e^{j\angle H(e^{-j\frac{\pi}{3}})}$$

$$\angle H(e^{j\frac{\pi}{3}}) = -\frac{\pi}{3} + 2 \text{atan} \left( \frac{\sin(\frac{\pi}{3})}{\cos(\frac{\pi}{3}) - 2} \right) = -\frac{2\pi}{3}$$

$$\angle H(e^{-j\frac{\pi}{3}}) = \frac{\pi}{3} - 2 \operatorname{atan}\left(\frac{\sin(\frac{\pi}{3})}{\cos(\frac{\pi}{3}) - 2}\right) = -\angle H(e^{j\frac{\pi}{3}}) = \frac{2\pi}{3}$$

$$y[n] = \frac{1}{2} e^{j\frac{\pi}{3}n} e^{j2\pi/3} + \frac{1}{2} e^{-j\frac{\pi}{3}n} e^{-j2\pi/3}$$

$$y[n] = \frac{1}{2} e^{j(\frac{\pi}{3}n + \frac{2\pi}{3})} + \frac{1}{2} e^{-j(\frac{\pi}{3}n + \frac{2\pi}{3})}$$

$$y[n] = \cos\left(\frac{\pi}{3}n + \frac{2\pi}{3}\right)$$

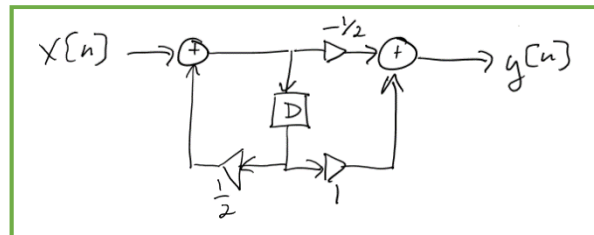
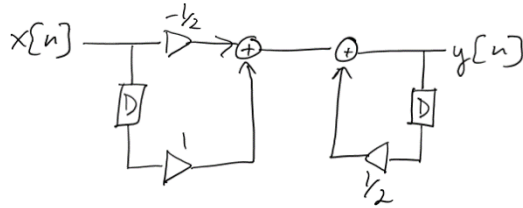
d)

$$H(e^{j\omega}) = \frac{Y(e^{j\omega})}{X(e^{j\omega})}$$

$$Y(e^{j\omega}) - \frac{1}{2} Y(e^{j\omega}) e^{-j\omega} = X(e^{j\omega}) e^{-j\omega} - \frac{1}{2} X(e^{j\omega})$$

$$y[n] - \frac{1}{2} y[n-1] = x[n-1] - \frac{1}{2} x[n]$$

$$y[n] = -\frac{1}{2} x[n] + x[n-1] + \frac{1}{2} y[n-1]$$



Problem 5.

a)  $h(t) = e^{-2t}u(t)$

Take CTFT:

$$\mathcal{F}\{h(t)\} = H_1(j\omega) = \frac{1}{2+j\omega}$$

Key points on the magnitude

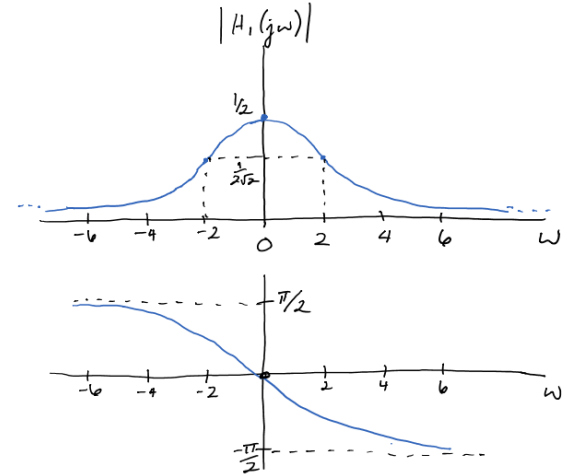
graph:

$$|H_1(j0)| = 1/2$$

$$|H_1(\pm j2)| = \frac{1}{2\sqrt{2}} \text{ (3dB point)}$$

$$|H_1(j\omega)| = \frac{1}{\sqrt{4+\omega^2}}$$

$$\angle H_1(j\omega) = -\arctan\left(\frac{\omega}{2}\right)$$



b)  $h_2[n] = (1 - e^{-0.1})e^{-0.1n}u[n]$

Take DTFT:

$$DTFT\{h_2[n]\} = H_2(e^{j\Omega}) = \frac{(1-e^{-0.1})}{1-e^{-0.1-j\Omega}}$$

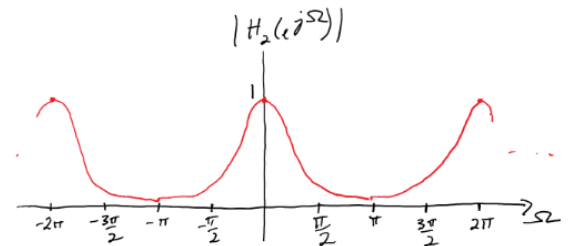
Key points on the magnitude graph:

$$|H_2(e^{j0})| = 1$$

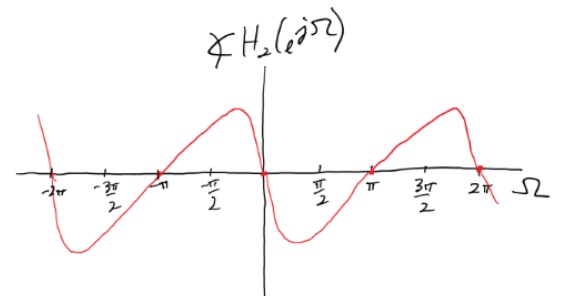
$$|H_2(e^{j\pi})| \cong 0.05$$

Periodic in  $\Omega_0 = 2\pi$

$$|H_2(e^{j\Omega})| = \sqrt{1+a^2 - 2a \cos(\omega)}$$



$$\angle H_2(e^{j\Omega}) = \arctan\left(\frac{\sin(\omega)}{\cos(\omega) - a}\right)$$



c) We can make the filters match by selecting a sampling frequency, since  $\Omega = \frac{\omega}{\omega_s} 2\pi$

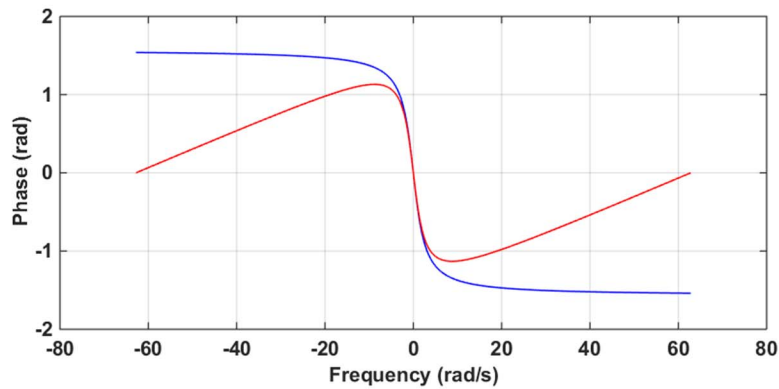
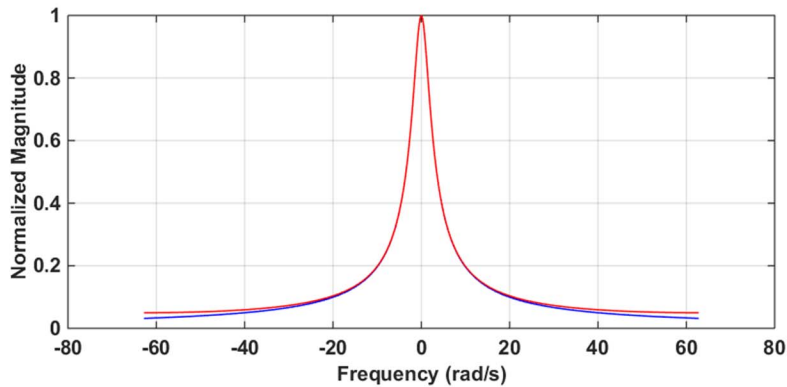
First, normalize  $H'_1(j\omega) = 2H_1(j\omega)$

$\Omega_{3dB} \cong 0.1$  by inspection. We know  $\Omega_{3dB} = \frac{\omega_{3dB}}{\omega_s} 2\pi$ , and  $\omega_{3dB} = 2$ . We can solve for  $\omega_s = 125.6$  rad/s.

The discrete filter behaves like the continuous time filter inside one period of  $H_2(e^{j\Omega})$ . The period of  $H_2$  in CTFT variable  $\omega$  corresponds to

$$\Omega = \frac{\omega_{filter}}{\omega_s} 2\pi = \pi$$

So  $\omega_{filter}$  is 63 rad/s, or  $\sim 10$ Hz.



However, it can be seen that the filter only has a monotonic phase response inside  $\pm 3\pi$  rad/s, so it may be relevant to restrict the bandwidth further if the application called for no dispersion (linear phase).