

# EECS 128 LECTURE NOTES 5

## GOALS:

- design of lead-lag compensators using Bode techniques
- examples

## REFS

FPE § 6.7

The best reference for this material

is :

DiStefano, Stubbleud, Williams

"Feedback : Control Systems"

(Schaum's Outline Series)

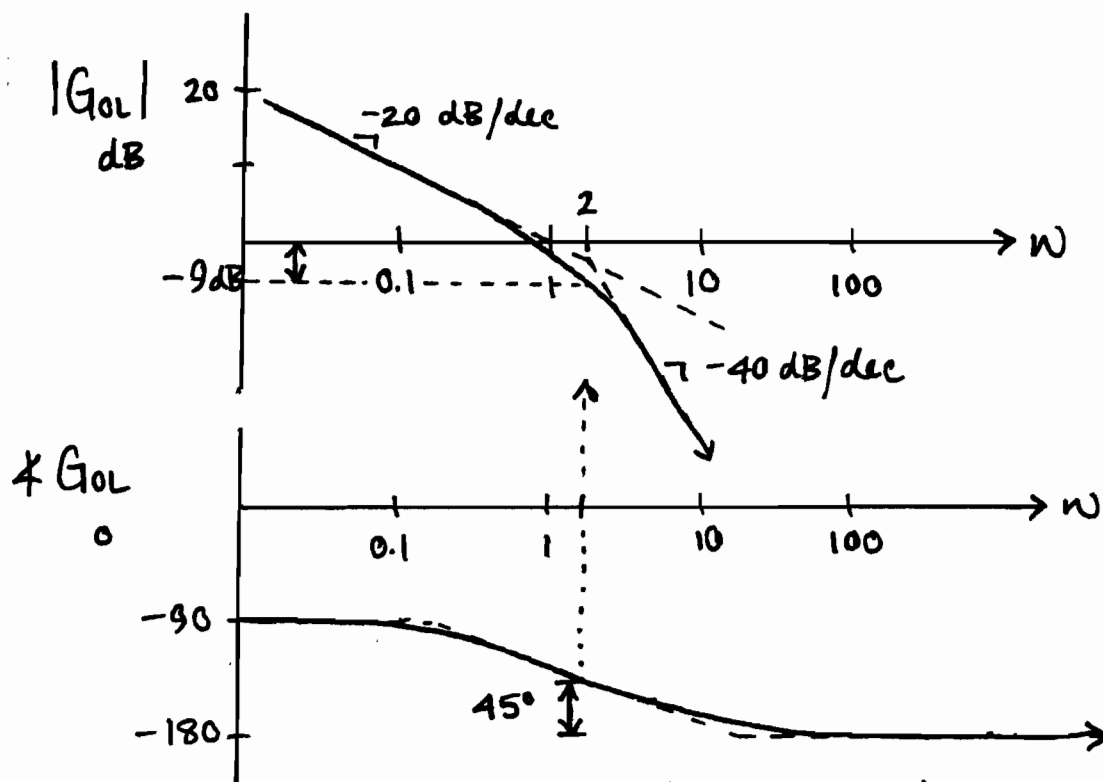
# Compensator Design in The Frequency Domain

## Using Bode:

### 1. GAIN FACTOR COMPENSATION

EXAMPLE:  $K \cdot G_{OL}(j\omega) = \frac{K}{j\omega (1 + \frac{j\omega}{2})}$

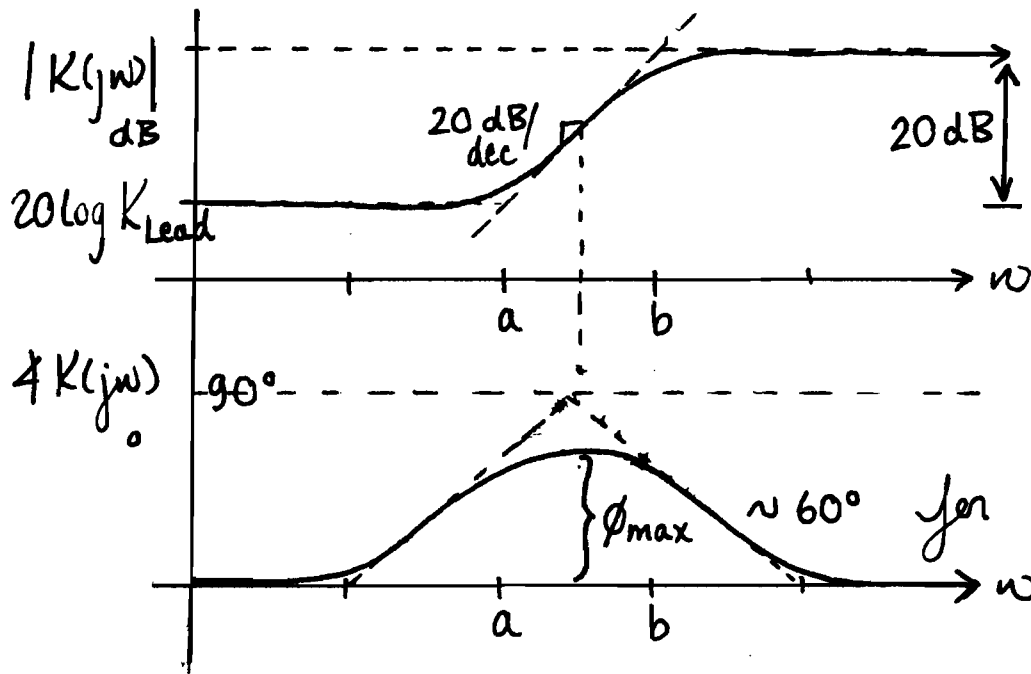
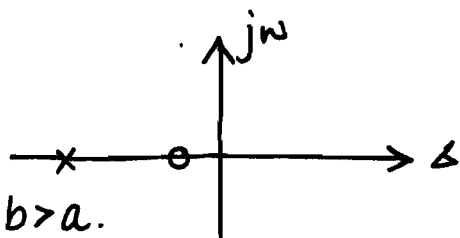
Q: I'd like a PM of  $45^\circ$ . What is the maximum BW I can achieve with just gain compensation?



A: Pure gain compensation doesn't change OL phase, so find  $\omega$  for which  $\phi_{G_{OL}} = -180 + 45 = -135^\circ$   
 ( $\omega = 2$ ) and compute  $K$  to  
 make  $\omega = 2$  the new crossover freq.  $\omega_c$ :  
 $20 \log K = 9 \Rightarrow K = 2.82 \leftarrow$

2. LEAD COMPENSATION:

$$K(j\omega) = K_{\text{lead}} \frac{j\omega/a + 1}{j\omega/b + 1} \quad \text{where } b > a.$$



here,  $\frac{a}{b} = 0.1$   
 $\therefore$  total gain is 20 dB.

$$\angle K(j\omega) \triangleq \phi(\omega) = \tan^{-1}\left(\frac{\omega}{a}\right) - \tan^{-1}\left(\frac{\omega}{b}\right)$$

maximize  $\phi$  w.r.t  $\omega$ :

$$\frac{\partial \phi}{\partial \omega} = \frac{1}{1 + \left(\frac{\omega}{a}\right)^2} \cdot \frac{1}{a} - \frac{1}{1 + \left(\frac{\omega}{b}\right)^2} \cdot \frac{1}{b} = 0$$

$\therefore$  solve

$$\omega_{\text{max}} = \sqrt{ab}$$

$$\begin{aligned} \therefore \log \omega_{\text{max}} &= \log \sqrt{a} + \log \sqrt{b} \\ &= \frac{1}{2} [\log(a) + \log(b)] \end{aligned}$$

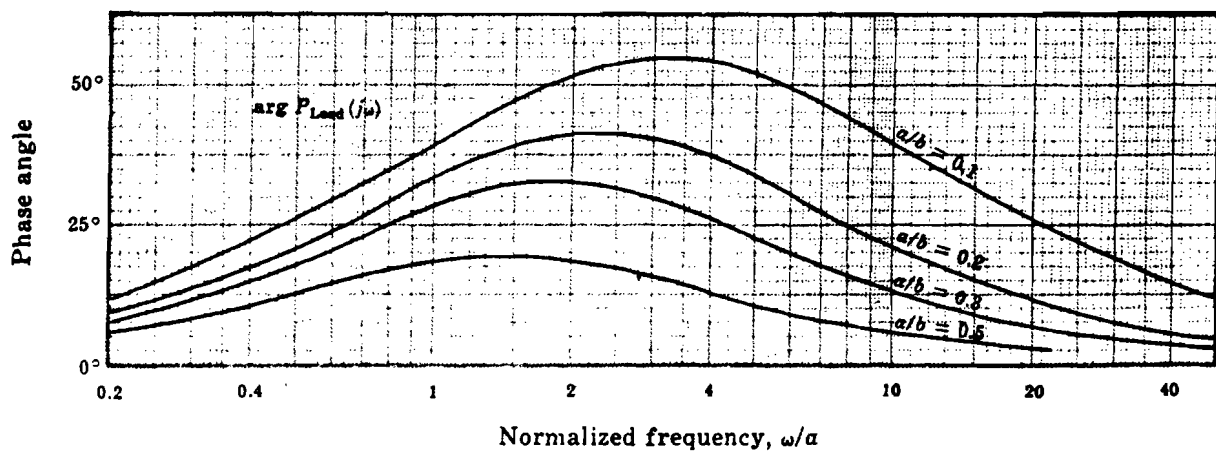
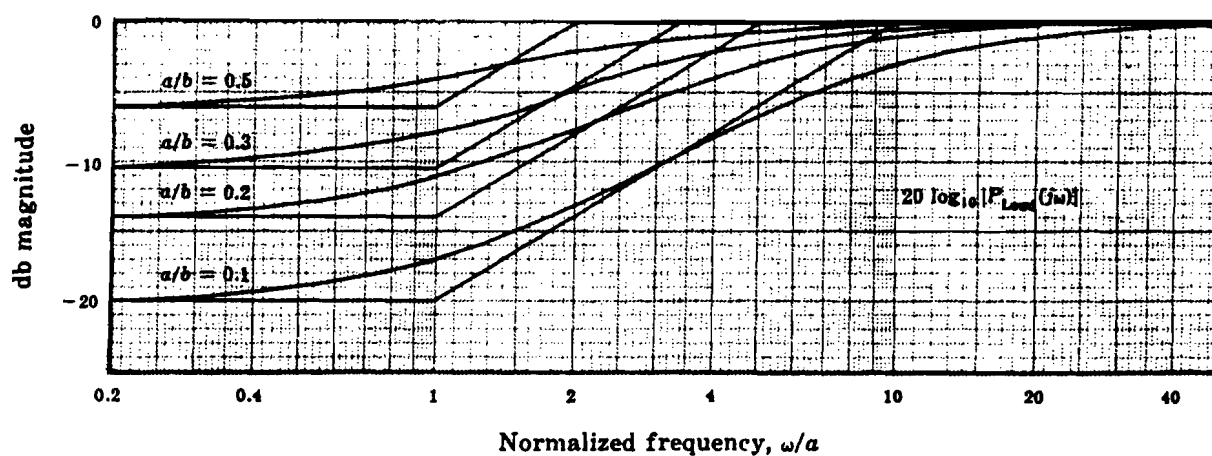
$$\text{and } \phi_{\text{max}} = \sin^{-1} \frac{1 - \frac{a}{b}}{1 + \frac{a}{b}}$$

$$\Rightarrow \frac{a}{b} = \frac{1 - \sin \phi_{\text{max}}}{1 + \sin \phi_{\text{max}}}$$

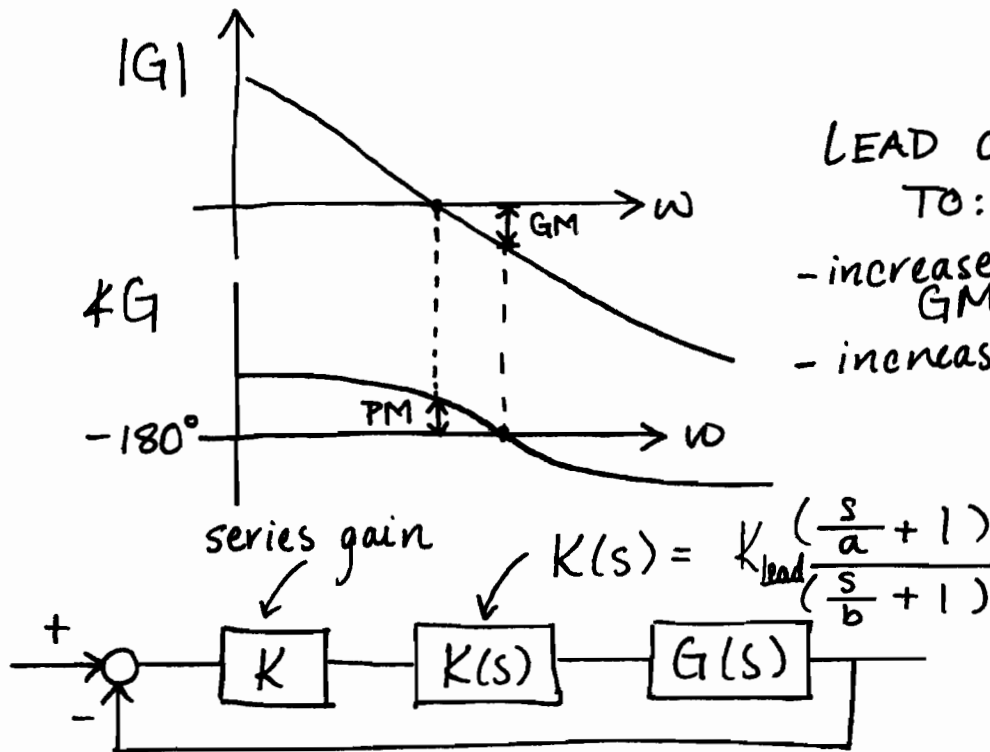
Below are the Bode plots for a lead compensator:

$$P_{\text{Lead}}(j\omega) = \frac{\frac{a}{b} (1 + j\omega/a)}{1 + j\omega/b}$$

for various lead ratios  $\frac{a}{b}$ .

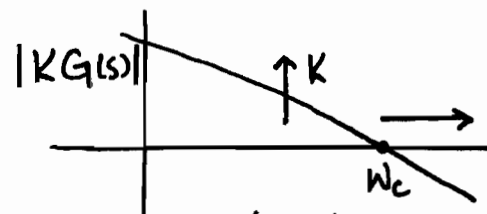


## HOW LEAD COMPENSATION IS GENERALLY USED IN DESIGN:



### Procedure:

1. Use the series gain  $K$  to achieve a desired  $\omega_c$ .

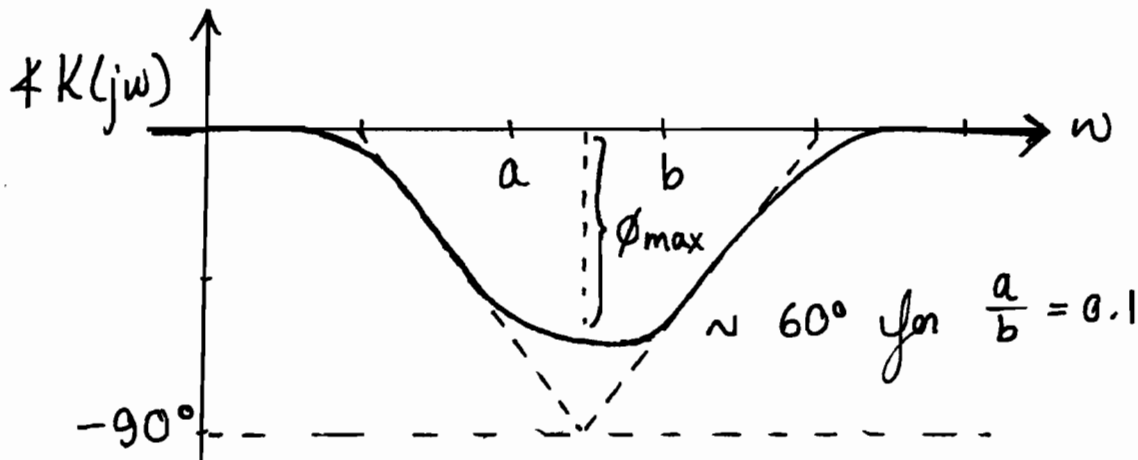
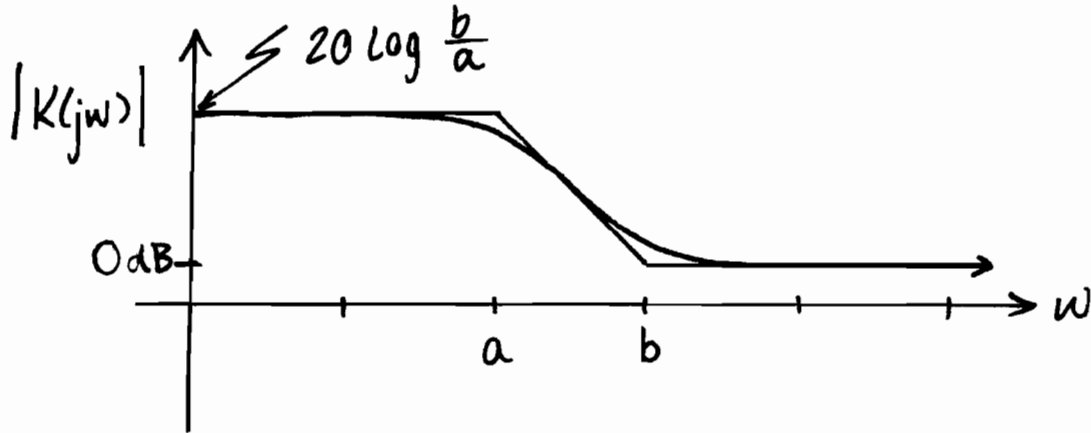
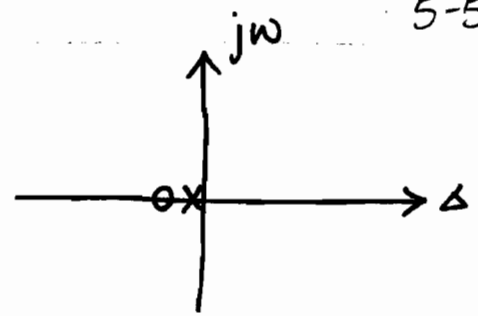


2. Compute what  $\phi_{max}$  has to be to get desired phase margin @  $\omega_c$ , and choose  $\frac{a}{b}$  ratio accordingly.
3. Choose  $K_{lead}$  such that  $\left|K(j\omega)\right|_{\omega=\sqrt{ab}} = 1$ .

### 3. LAG compensation:

$$K(j\omega) = \frac{b}{a} \frac{1 + \frac{j\omega}{b}}{1 + \frac{j\omega}{a}}$$

where  $b > a$



How LAG compensation is generally used in design:

- choose  $b$  to be  $\sim \frac{\omega_c}{50}$  so the phase lag does not significantly affect PM

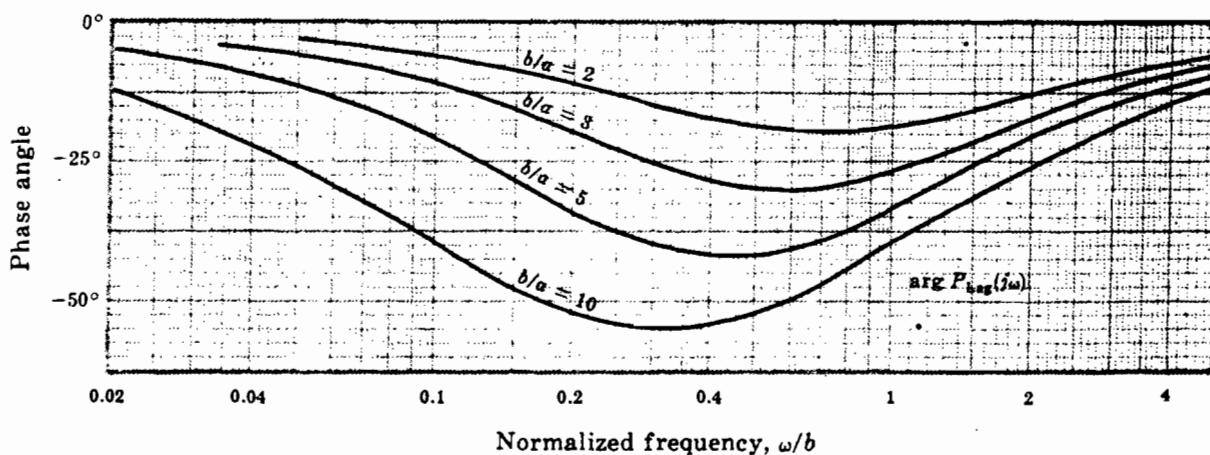
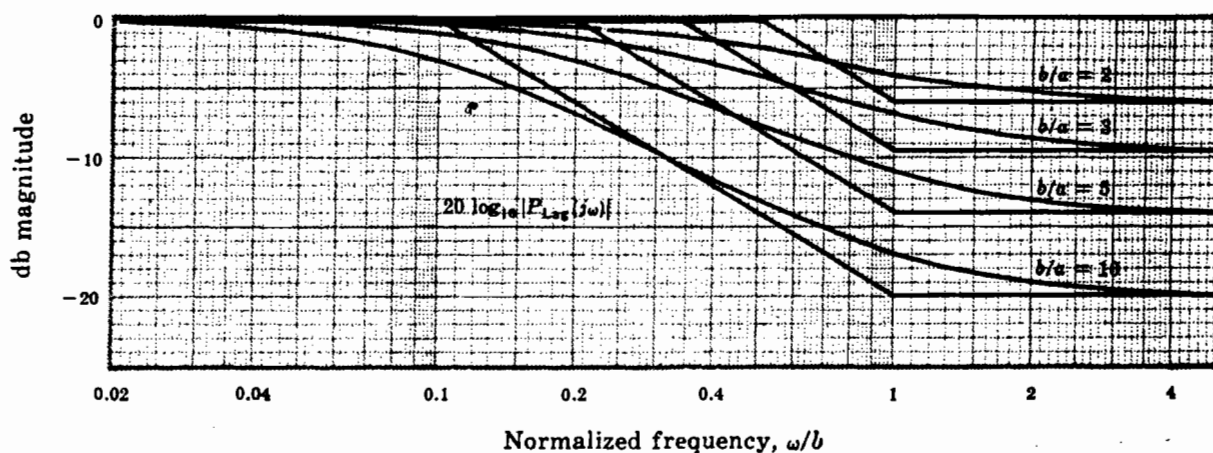
- choose  $\frac{b}{a}$  ratio to achieve low freq. gain to satisfy ss error specs:

$$e_{ss} = \lim_{s \rightarrow 0} s \cdot \frac{1}{\left[ 1 + \frac{b(1+\frac{s}{b})}{a(1+\frac{s}{a})} G(s) \right]} R(s) \leq \text{spec.}$$

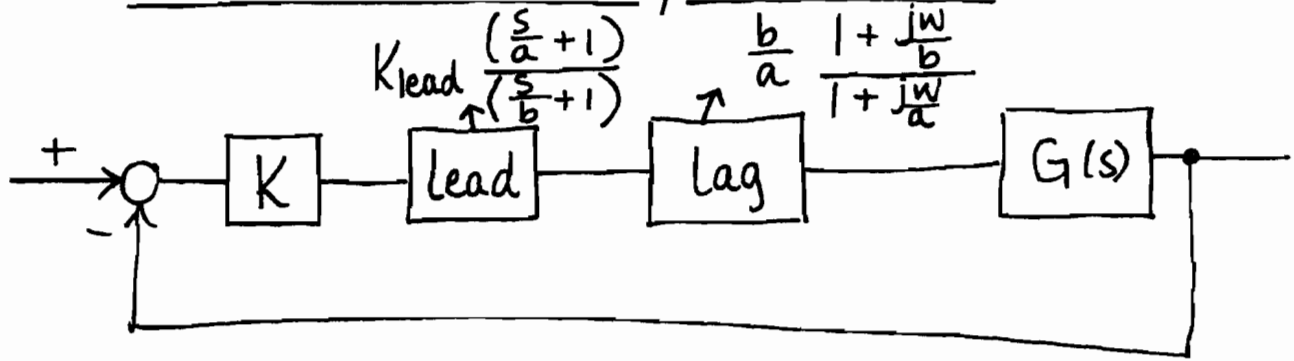
Below are the Bode plots for a lag compensator:

$$P_{\text{lag}}(j\omega) = \frac{1 + j\omega/b}{1 + j\omega/a}$$

for various lag ratios  $\frac{b}{a}$ .



#### 4. LEAD-LAG compensation



$$\therefore K(s) = K \cdot K_{\text{lead}} \frac{(\frac{s}{a} + 1)}{(\frac{s}{b} + 1)} \cdot \frac{b}{a} \frac{1 + \frac{s}{b}}{1 + \frac{s}{a}}$$

#### Procedure:

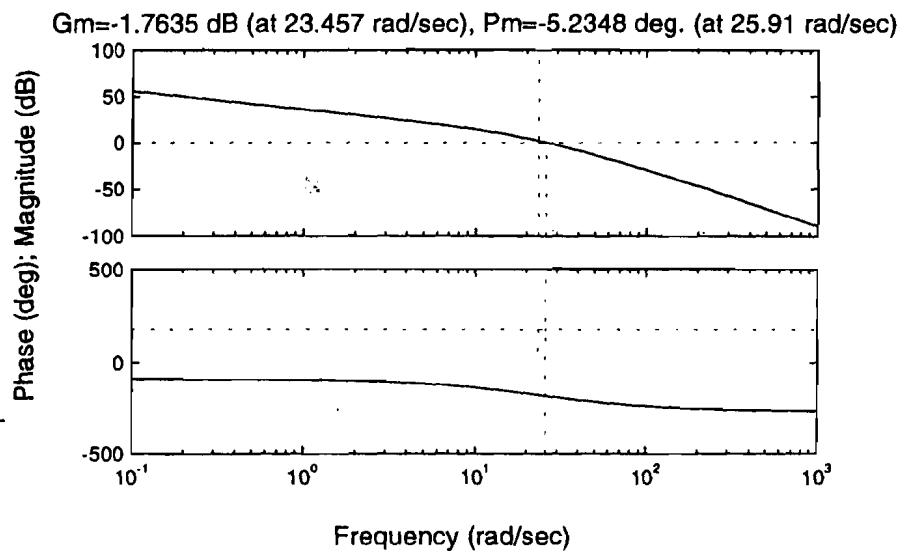
1. Use lead compensator : series gain  $K$  to achieve desired  $\omega_c$  and PM.
2. Use lag compensation to achieve desired low frequency gain to satisfy steady state error specs.

WORKED PROBLEM

$$G(s) = \frac{34381}{s(s+15.5)(s+35.5)} \quad D(s)$$

Design a lead compensator so that  $PM = 52^\circ$  and  $GM = 4.4$ . Verify that your design works using MATLAB.

uncompensated system  $\rightarrow G(s)$



Solution: use MATLAB command MARGIN to find the Gain and Phase margins

$$\left. \begin{array}{l} PM = -5.235^\circ \\ GM = -1.76 \text{ dB} \end{array} \right\} \text{unstable closed loop system}$$

We can design the compensator and leave the crossover frequency  $\omega_c = 25.91$  unchanged.

To achieve  $PM = 52^\circ$ , need a phase increase of  $52 - (-5.235) = 57.2^\circ$

$$\text{Let } D(s) = K_{\text{lead}} \frac{\left(\frac{s}{a} + 1\right)}{\left(\frac{s}{b} + 1\right)}$$

$$\text{where (we know) } \frac{a}{b} = \frac{1 - \sin \phi_{\text{max}}}{1 + \sin \phi_{\text{max}}}$$

To keep  $\omega_c = 25.91$ , let  $\omega_c = \sqrt{ab} = 25.91$

$$\text{and } |D(s)|_{s=j\sqrt{ab}} = 1.$$

This gives

$$D(s) = \sqrt{\frac{b}{a}} \frac{s + \omega_c \sqrt{\frac{a}{b}}}{s + \omega_c \sqrt{\frac{b}{a}}}$$

$$\Rightarrow D(s) = 3.4 \frac{s + 7.62}{s + 88.1}$$

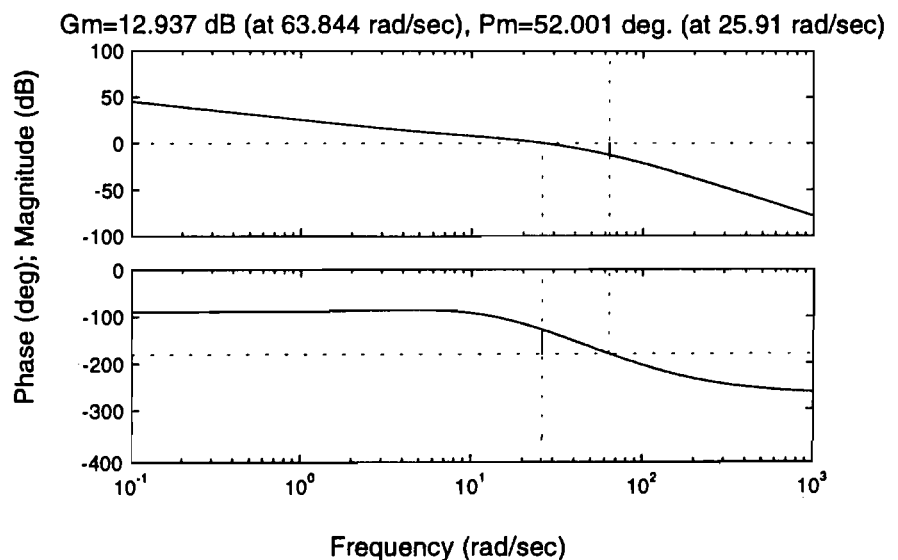
Check: compensated system has

$$PM = 52^\circ, GM = 12.94 \text{ dB} = 4.44 \leftarrow$$

```
% Open-loop Transfer function
Gnum = 34381;
Gden = conv([1 15.5 0],[1 35.5]);
% Draw Open-loop Bode plot
figure(1)
subplot(2,1,1)
W = logspace(-1, 3, 500);
bode(Gnum, Gden, W);
% Find the Gain and Phase Margin
[Mag, Phase] = bode(Gnum, Gden, W);
margin(Mag, Phase, W);
title('uncompensated system')
[Gm, Pm, Wcg, Wcp] = margin(Mag, Phase, W);
```

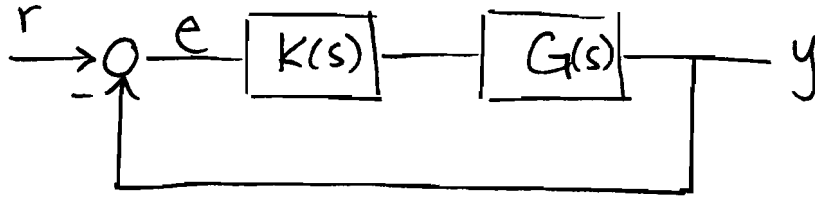
```
%Phase lead needed if keep Wcp same
Plead = (52.001 - Pm) * pi / 180;
%The ratio of pole and zero of lead
b_a = (1 + sin(Plead)) / (1 - sin(Plead));
%The lead compensator
Wc = Wcp;
Dnum = sqrt(b_a) * [1 Wc/sqrt(b_a)];
Dden = [1 Wc*sqrt(b_a)];
%Compensated system
num = conv(Gnum, Dnum);
den = conv(Gden, Dden);
subplot(2,1,2);
bode(num, den, W);
% Find the Gain and Phase Margin
[Mag, Phase] = bode(num, den, W);
margin(Mag, Phase, W);
title('lead compensated system')
```

lead compensated system —  $D(s)G(s)$



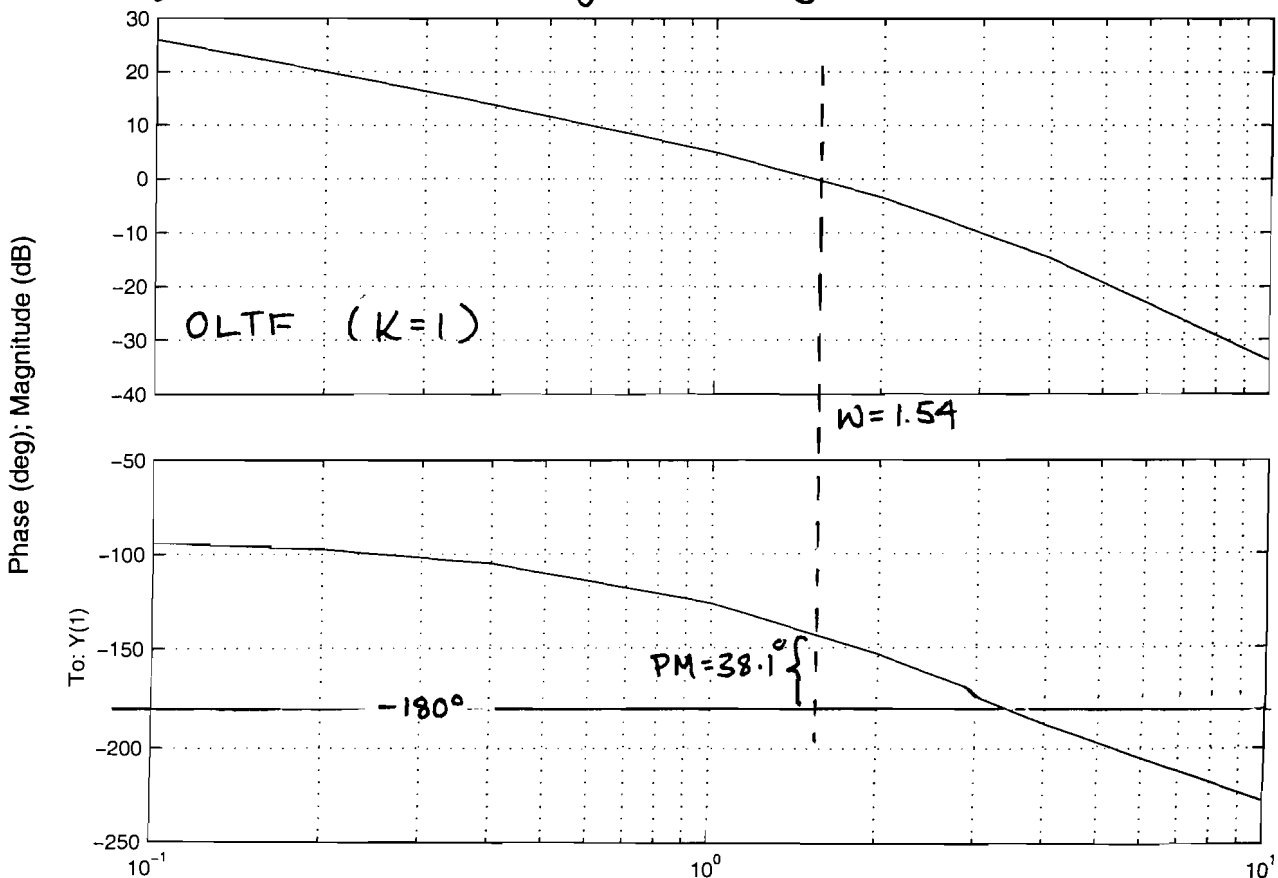
# DESIGN EXAMPLE USING BODE

$$G(j\omega) = \frac{2}{j\omega \left(\frac{j\omega}{2} + 1\right) \left(\frac{j\omega}{6} + 1\right)}$$



Design  $K(s)$  for:

1. SS error to ramp inputs  $\leq 0.05\%$
2.  $PM = 45^\circ \pm 5^\circ$
3. gain crossover frequency  $\omega_c \geq 1$  rad/s.



## A. PURE GAIN COMPENSATION?

$$\text{Recall } E(s) = \frac{1}{1+G(s)} R(s)$$

$$\therefore e_{ss} = \lim_{s \rightarrow 0} s E(s)$$

$$= \lim_{s \rightarrow 0} s \frac{\left(\frac{1}{s^2}\right) \leftarrow (\text{ramp})}{1+G(s)}$$

$$= \frac{1}{\lim_{s \rightarrow 0} s G(s)} \quad \left[ \text{we denote } K_v \equiv \lim_{s \rightarrow 0} s G(s) \right]$$

$\lim_{s \rightarrow 0} s G(s) = 2 \leftarrow$  large ss error to ramp inputs

we require (Spec. 1)  $\lim_{s \rightarrow 0} s K G(s) = 20$  [for 0.05%]

$\therefore$  try  $K(s) = K = 10 \leftarrow$   
[or  $20 \log_{10} 10 = 20 \text{ dB} \leftarrow$ ]

