

EE 140 Homework #3 Fall 2009

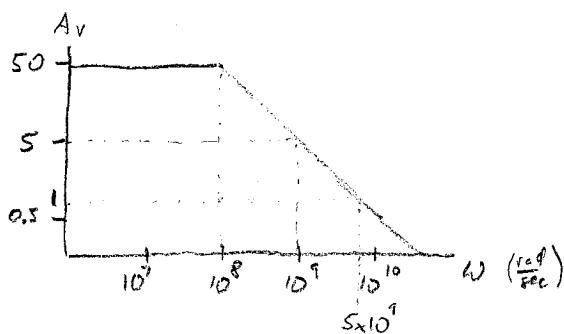
1A)  $A_v = 100 \quad f_p = 5 \text{ MHz} \quad C_L = 1 \mu\text{F}$

$$f_u = A_v f_p = 500 \text{ MHz}$$

$$R_o = \frac{1}{2\pi f_p C_L} = (10\pi \cdot 10^6 \cdot 10^{-12})^{-1} = \frac{1}{\pi} \cdot 10^5 \Omega$$

$$G_m = \frac{A_v}{R_o} = \frac{100}{\frac{1}{\pi} \cdot 10^5} = \pi \text{ mS}$$

1B)



$A_v$  decreases by  $10x$  /decade after  $\omega_p$ , so  $\omega_p$  must be exactly one decade before  $10^9 \frac{\text{rad}}{\text{sec}}$  where  $A_v = 5$ , given that  $A_{v,DC} = 50$ .

$$\omega_p = 10^8 \frac{\text{rad}}{\text{sec}}$$

$$\omega_a = A_{v,DC} \cdot \omega_p = 5 \cdot 10^7 \frac{\text{rad}}{\text{sec}}$$

1C)  $R_o = 10^7 \quad G_m = 10 \text{ mS} \quad \omega_a = 10^9 \frac{\text{rad}}{\text{sec}}$

$$A_v = G_m R_o = 10^5$$

$$\omega_p = \frac{\omega_a}{A_v} = 10^4 \frac{\text{rad}}{\text{sec}}$$

$$C_L = \frac{1}{\omega_p R_o} = 10^{-11} \text{ F} = 10 \text{ pF}$$

2)	$G_m [\text{S}]$	$R_o [\Omega]$	$C_L [\text{F}]$	$A_v$	$\omega_p [\frac{\text{rad}}{\text{sec}}]$	$\omega_a [\frac{\text{rad}}{\text{sec}}]$
a)	$1 \text{ m}$	$50 \text{ k}$	$2 \text{ p}$	$50$	$10 \text{ M}$	$500 \text{ M}$
b)	$5 \text{ m}$	$20 \text{ k}$	$5 \text{ p}$	$100$	$10 \text{ M}$	$1 \text{ G}$
c)	$25 \text{ m}$	$4 \text{ k}$	$5 \text{ p}$	$100$	$50 \text{ M}$	$5 \text{ G}$
d)	$2 \text{ m}$	$15 \text{ k}$	$13.33 \text{ p}$	$30$	$5 \text{ M}$	$150 \text{ M}$
e)	$50 \text{ n}$	$100$	$2.5 \text{ p}$	$5$	$4 \text{ G}$	$20 \text{ G}$

Useful relations:  $A_v = G_m R_o$ ,  $\omega_p = \frac{1}{R_o C_L}$ ,  $\omega_a = A_v \omega_p$

3) First, a few things to note:

- (1) If I can achieve the required  $g_m$ ,  $R_o$ , and  $C_L$ , then I can also achieve the required  $A_V$ ,  $w_p$ , and  $w_n$ .
- (2) L presents a tradeoff between  $R_o$  (via 2) and  $w_p$  (via  $C_L$ )
- (3) W presents a tradeoff between  $g_m$  and  $w_p$  (via  $C_L$ )
- (4)  $V_{DSAT}$  presents a tradeoff between  $g_m$  and  $I_D$  (via  $I_D$ )
- (5) This problem has no unique solution, so we need to make an intelligent choice about which parameter to choose first.
  - a) Minimum  $V_{DSAT}$  gives the greatest  $g_m$  efficiency ( $\frac{g_m}{I_D}$ )
  - b) Minimum length gives the minimum capacitance / area

$\Rightarrow$  A good place to start is to minimize either  $V_{DSAT}$  or L.  
Because I claim the math is easier, I'll start with minimum L.

(6) Assume  $|V_{DS}| = |V_{DSS}| = V_{out} = \frac{V_{DD}}{2} = 1.5V$ . This means  $w_p = 2w_n$

A)  $L = 0.5 \mu m$        $\lambda = 0.1 \cdot \frac{1 \mu m}{L} \Rightarrow 2 = 0.2$

$$R_o = \frac{1}{2} \frac{1 + 2V_{DS}}{\lambda I_D} \Rightarrow I_D = \frac{1 + 2V_{DS}}{2 \lambda R_o}$$

so for  $R_o = 50 k\Omega$ ,  $I_D = 65 \mu A$ .

$$g_m = \frac{2I_D}{V_{DSAT}} \Rightarrow V_{DSAT} = \frac{2I_D}{g_m}$$

so for  $g_m = 1mS$ ,  $V_{DSAT} = 130 mV$

$$W = \frac{I_D}{I_D w_p} = \frac{I_D}{\frac{1}{2L} k_n V_{DSAT}^2 (1+2V_{DS})} = \frac{65 \cdot 10^{-6}}{\frac{1}{2 \cdot 0.5} 200 \cdot 10^{-6} \cdot 0.13^2 \cdot (1 + 0.2 \cdot 1.5)} \\ = 14.79 \mu m$$

$$C_L = C_{gd(nmos)} + C_{gd(pmos)} + C_{db(nmos)} + C_{db(pmos)}$$

$$= w_p C_{ol} + (w_n + w_p) \frac{k_n C_{DSO}}{\sqrt{1 + \frac{|V_{DS}|}{4}}} \quad C_{ol} = 0.5 fF/\mu m$$

$$= 2w_n C_{ol} + 3w_n C_* = 2.5 w_n [fF]$$

$$= 37 fF$$

$$\frac{k_n C_{DSO}}{\sqrt{1 + \frac{|V_{DS}|}{4}}} = 0.5 fF/\mu m \\ = C_*$$

B)  $L = 0.5 \text{ mm} \Rightarrow \lambda = 0.2$

$$I_D = \frac{1+2V_{DS}}{2\lambda R_o}, \text{ so for } R_o = 20 \text{ k}\Omega \quad I_D = 162.5 \text{ mA.}$$

$$V_{DSAT} = \frac{2I_D}{g_m}, \text{ so for } g_m = 5 \text{ mS} \quad V_{DSAT} = 65 \text{ mV.}$$

This is less than 100 mV, the minimum allowable  $V_{DSAT}$ , so we need a new approach.

$$V_{DSAT} = V_{DSAT,min} = 100 \text{ mV}$$

$$I_D = \frac{1}{2} g_m V_{DSAT}, \text{ where } g_m \text{ is taken from the spec.}$$

$$R_o = \frac{1}{2} \frac{1+2V_{DS}}{2I_D} \Rightarrow \lambda = \frac{1}{2R_o I_D - V_{DS}}, \text{ where } R_o \text{ is from the spec.}$$

$$\lambda = 0.1 \cdot \frac{W}{L} \Rightarrow L = \frac{0.1}{\lambda}$$

$$W = \frac{I_D}{I_D/W} = \frac{I_D}{\frac{1}{2} \lambda W^2 V_{DSAT}^2 (1+2V_{DS})}$$

Plugging in using the values for B ( $V_{DSAT} = 100 \text{ mV}$ ,  $g_m = 5 \text{ mS}$ ,  $R_o = 20 \text{ k}\Omega$ ),

$$\text{we find } I_D = 250 \text{ mA}, \quad L = 0.85 \text{ mm} \text{ and } W = 180.625 \text{ mm}$$

$$\text{Using } C_L = 2W_h C_{ox} + 3W_n C \text{ from part A, } \quad C_L = 452 \text{ fF}$$

C) Using the approach from A, we would again get  $V_{DSAT} = 65 \text{ mV} < V_{DSAT,min}$ .

Instead, fix  $V_{DSAT}$  as in part B.

Plugging in to the same expressions in part B using the values from part C ( $V_{DSAT} = 100 \text{ mV}$ ,  $g_m = 25 \text{ mS}$ ,  $R_o = 4 \text{ k}\Omega$ ), we find

$$I_D = 1.25 \text{ mA}, \quad L = 0.85 \text{ mm, and } W = 903.125 \text{ mm. Then } C_L = 2.26 \text{ pF}$$

D) Using the approach from A,  $L = 0.5 \text{ mm} \Rightarrow \lambda = 0.2$

$$I_D = \frac{1+2V_{DS}}{2\lambda R_o}, \text{ so for } R_o = 15 \text{ k}\Omega \quad I_D = 216.67 \text{ mA}$$

$$V_{DSAT} = \frac{2I_D}{g_m}, \text{ so for } g_m = 5 \text{ mS} \quad V_{DSAT} = 216.7 \text{ mV}$$

$$W = \frac{I_D}{I_D/W}, \text{ so } W = 17.75 \text{ mm. Then } C_L = 44 \text{ fF}$$

$$E) \quad L = 0.5 \text{ mm} \Rightarrow \lambda = 0.2$$

$$\text{for } R_o = 100, \quad I_D = \frac{1 + \lambda V_{DS}}{2 \lambda R_o} = 32.5 \text{ mA}$$

$$\text{for } g_m = 50 \text{ mS} \quad V_{DSAT} = \frac{2I_D}{g_m} = 1.3 \text{ V}$$

$$W = \frac{I_D}{I_D/\lambda} = 74 \text{ um}$$

$$C_L = 185 \text{ fF}$$

Caveat: Typically values of W and L in an IC process are limited to multiples of  $\frac{1}{2} L_{min}$ , or in this process 0.25 um. I neglected this fact in solving this problem, but since we have plenty of margin in our bandwidth, we could certainly resize the device slightly and still meet the spec.

Summary of results:

	$g_m [\text{S}]$	$R_o [\Omega]$	$C_L [\text{F}]$	$I_D [\text{A}]$	$W [\text{um}]$	$L [\text{um}]$	$V_{DSAT} [\text{V}]$
a)	1m	50k	37f	65 $\mu$	14.79	0.5	130m
b)	5m	20k	452f	250 $\mu$	180.625	0.85	100m
c)	25m	4k	2.26p	1.25m	903.125	0.85	100m
d)	2m	15k	44f	216.67 $\mu$	17.75	0.5	216.7m
e)	50m	100	185f	32.5m	74	0.5	1.3

Note: Had I approached A, D, and E by fixing  $V_{DSAT} = V_{DSAT,min} = 100 \text{ mV}$ , the value of L found to meet the spec would be less than  $L_{min}$ .

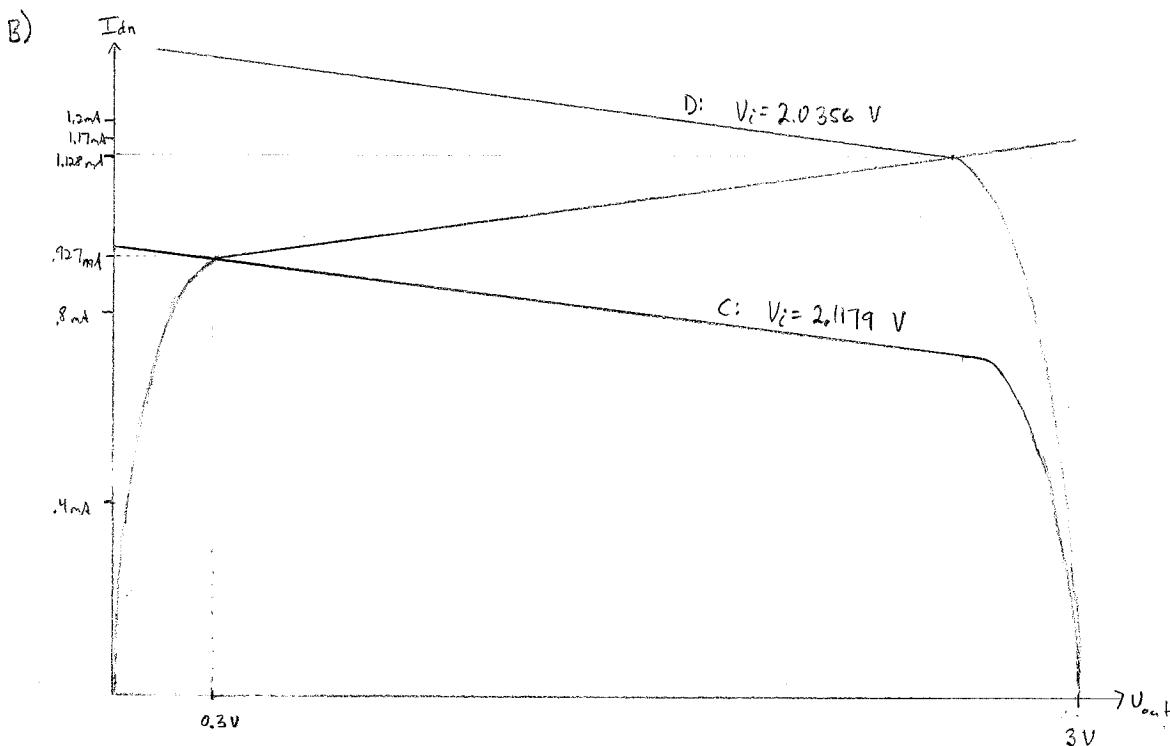
4)

A)  $V_{Dsatn} = V_{as} - V_T = V_{BN} - V_T = 0.8 \text{ V} - 0.5 \text{ V} = [0.3 \text{ V}]$

$$I_{dn} = \frac{1}{2} k_n \frac{W}{L} V_{Dsatn}^2 (1 + 2V_{BS}) \quad \text{where } V_{BS} = V_{out} = V_{Dsatn}$$

$$\lambda = 0.1 \frac{W}{L} = 0.1 \quad W = 100 \mu\text{m} \quad L = 1 \mu\text{m} \quad k_n = 200 \cdot 10^{-6} \frac{\text{A}}{\text{V}^2}$$

$$I_{dn} = \frac{1}{2} \cdot 200 \cdot 10^{-6} \cdot \frac{100}{1} \cdot 0.3^2 \cdot (1 + 0.1 \cdot 0.3) = [92.7 \mu\text{A}]$$



Minimum  $I_{dn}$  for NMOS saturation: Found in part A to be  $92.7 \mu\text{A}$ .

Maximum  $I_{dn}$  for NMOS saturation: Occurs at  $V_{DS} = 3 \text{ V}$  ( $V_{DD}$ )

Note: PMOS is out of saturation, but that isn't what the problem asked.

$$I_{dpmax} = \frac{1}{2} \frac{W}{L} k_p V_{Dsat}^2 (1 + \lambda(3)) = 9 \cdot 10^{-4} (1.3) = [1.17 \text{ mA}]$$

C)  $|I_{dp}| = I_{dn}$ , so  $|I_{dp}| = 92.7 \mu\text{A}$  at this point

$$|I_{dp}| = \frac{1}{2} \frac{W}{L} k_p (|V_i - V_{DD}| - |V_{tp}|)^2 (1 + 1/2(V_{out} - V_{SD}))$$

$$92.7 \cdot 10^{-6} = \frac{1}{2} \cdot \frac{100}{1} \cdot 100 \cdot 10^{-6} ((3 - V_i) - 0.5)^2 (1 + 0.1(3 - 0.3))$$

$$\Rightarrow [V_i = 2.1179 \text{ V}]$$

D) The PMOS leaves saturation when  $|V_{out} - V_{DD}| = |V_i - V_{DD}| - |V_{TP}|$

$$I_{Dp} = \frac{1}{2} \frac{W}{L} k_p^* (V_{DD} - V_{out})^2 (1 + \lambda (V_{DD} - V_{out}))$$

$$= I_{Dn} = \frac{1}{2} \frac{W}{L} k_n^* (V_{SN} - V_{in})^2 (1 + \lambda V_{out})$$

$$k_p^* (V_{DD} - V_{out})^2 (1 + \lambda (V_{DD} - V_{out})) = k_n^* (V_{SN} - V_{in})^2 (1 + \lambda V_{out})$$

$$(V_{DD} - V_{out})^2 (1 + \lambda (V_{DD} - V_{out})) = 0.18 (1 + \lambda V_{out})$$

$$(V_{DD} - V_{out})^2 (1 + \lambda (V_{DD} - V_{out})) - 0.18 (1 + \lambda V_{out}) = 0$$

$$(3 - V_{out})^2 (1 + 0.1 (3 - V_{out})) - 0.18 (1 + 0.1 V_{out}) = 0$$

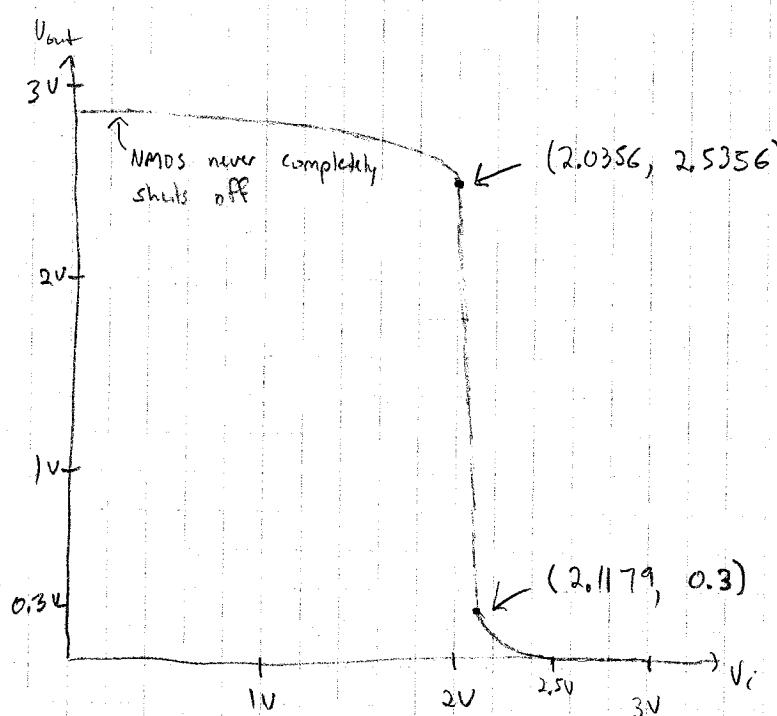
Solving in MATLAB,  $V_{out} = 2.5356 \text{ V}$

$$V_{DD} - V_{out} = V_{DD} - V_i - |V_{TP}|$$

$$3 - 2.5356 = 3 - V_i - 0.5 \Rightarrow V_i = 2.0356$$

At this point,  $I_o = \frac{1}{2} \frac{W}{L} k_n^* (V_{SN} - V_{in})^2 (1 + \lambda V_{out}) = 1.1282 \text{ mA}$

E)



F) Assuming approx constant slope in high-gain region,

$$A_v = -\frac{\partial V_o}{\partial V_i} = -\frac{0.3 - 2.5356}{2.1179 - 2.0356} = -27.164$$

input range: 2.0356 V to 2.1179 V

output range: 0.3 V to 2.5356 V

G)

Left edge: (2.0356, 2.5356)

$$\begin{aligned} A_v &= -g_{mp} (r_{op} \| r_{on}) \\ &= -\frac{2 \cdot 1.1282 \text{ mA}}{0.4649 \text{ V}} \cdot \left[ \frac{(1 + 0.1 \cdot 4.649)}{0.1 \cdot 1.1282 \times 10^{-3}} \parallel \frac{(1 + 0.1 \cdot 2.5356)}{0.1 \cdot 1.1282 \times 10^{-3}} \right] \\ &= -4.8587 \text{ mS} \cdot (9.275 \text{ k}\Omega \parallel 11.111 \text{ k}\Omega) \\ &= -4.8587 \text{ mS} \cdot 5.055 \text{ k}\Omega \\ &= -24.5116 \end{aligned}$$

Right edge: (2.1179, 0.3)

$$\begin{aligned} A_v &= -g_{mp} (r_{op} \| r_{on}) = \frac{2 \cdot 9.27 \times 10^{-3}}{0.3821} \cdot \left[ \frac{1 + 0.1 \cdot 2.7}{0.1 \cdot 9.27 \times 10^{-3}} \parallel \frac{1 + 0.1 \cdot 0.3}{0.1 \cdot 9.27 \times 10^{-3}} \right] \\ &= -4.8521 \text{ mS} \cdot (13.7 \text{ k}\Omega \parallel 11.111 \text{ k}\Omega) \\ &= -29.7689 \end{aligned}$$

Middle: (2.0768, 1.4178)  $I_D = \frac{1}{2} Q_K (0.3)^2 (1 + 2 \cdot 1.4178) = 1.0278 \text{ mA}$

$$\begin{aligned} A_v &= -g_{mp} (r_{op} \| r_{on}) = \frac{2 \cdot 1.0278 \times 10^{-3}}{0.4232} \cdot \left[ \frac{1 + 0.1 \cdot 1.5822}{0.1 \cdot 1.0278 \times 10^{-3}} \parallel \frac{1 + 0.1 \cdot 1.4178}{0.1 \cdot 1.0278 \times 10^{-3}} \right] \\ &= -4.8573 \text{ mS} \cdot (14.269 \text{ k}\Omega \parallel 11.109 \text{ k}\Omega) = -4.8573 \text{ mS} \cdot 5.5942 \text{ k}\Omega \\ &= -27.1727 \end{aligned}$$

Note: Actual  $R_o$  (and  $A_v$ ) at edge points is less than predicted here because we are at the edge of saturation and triode. I used the saturation equations above, but I just as well might have used the triode equations and gotten a much smaller  $R_o$ . The truth lies somewhere between the two.