

Lecture 25: Feedback I

• Announcements:

- HW#10 due Wednesday at 8 a.m.
- No homework over Thanksgiving break
- HW#11 will be online next week
- Lab#3 (Design Project) due Friday, Dec. 11, at 11:59 p.m.

↳ For 240A: use resistor temperature coefficients previously given in lecture

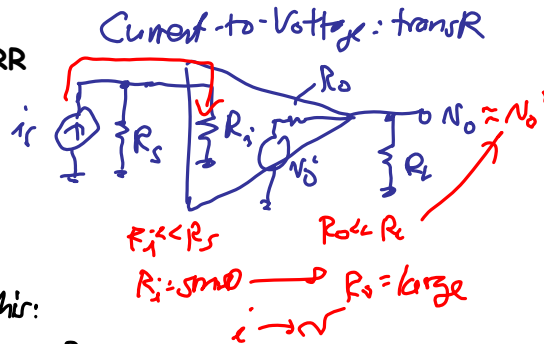
- Pre-Lecture Feedback Loading Handout online

• Lecture Topics:

- ↳ Recognizing Feedback Configurations
- ↳ Effect of FB on Z_i and Z_o .
- ↳ Feedback Loading

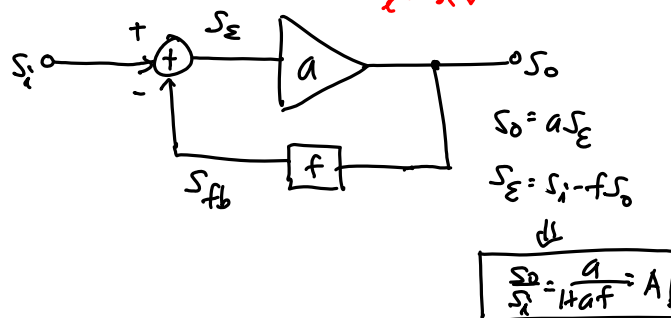
• Last Time:

- Finished PSRR



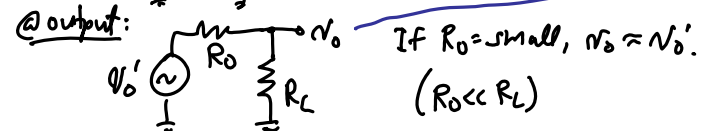
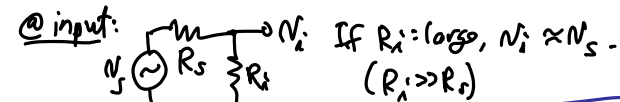
Feedback

⇒ we know this:



Benefit of Negative FB

- ① Stabilizes the gain of the amp against parameter changes & active device variations
- ② Modifies R_i and $R_o \rightarrow$ basically improves their values according to the type of amplifier implemented
e.g., voltage amp: $R_i = \text{large}, R_o = \text{small}$



current-to-voltage amp: $R_i = \text{small}, R_o = \text{small}$

voltage-to-current amp: $R_i = \text{large}, R_o = \text{large}$

current-to-current amp: $R_i = \text{small}, R_o = \text{large}$

$V_o = V_o' \left(\frac{R_L}{R_L + R_o} \right)$

- ③ Reduces distortion; improves linearity. V_o (when $R_o \ll R_L$)
- ④ Increases bandwidth (w. 3dB).

Disadvantages of Neg. FB

- ① Gain is reduced \rightarrow reduction factor \sim equal to the amount of gain stabilization, distortion reduction, etc...
Solution: Add more stages of gain \rightarrow but this adds cost & power...
- ② Feedback causes stability problems (if not compensated properly)

Gain Sensitivity Reduction Via FB

$$A = \frac{a}{1+af} \rightarrow \frac{dA}{da} = \frac{(1+af) - af}{(1+af)^2} = \frac{1}{(1+af)^2}$$

} small
↑ weak sig!

For a Δ in op amp gain: S_a

$$\frac{SA}{SA} = \frac{1}{(1+af)^2} \rightarrow SA = \frac{S_a}{(1+af)^2}$$

... and the fractional change:

$$\frac{SA}{A} = \frac{1+af}{a} \frac{S_g}{(1+af)^2} \Rightarrow \frac{SA}{A} = \frac{S_g}{1+af}$$

Distortion Reduction via FB

Slope = a_2
Slope = a_1
 $a_3 = 0$

Now, close the loop: (add neg. FB)

$0 < S_0 < S_{01}$: $A_1 = \frac{a_1}{1+af} \approx \frac{1}{f}$ for a_1 large $\approx 20K$

$S_{01} < S_0 < S_{02}$: $A_2 = \frac{a_2}{1+a_2f} \approx \frac{1}{f}$ $a_2 = 1000$

Power vs ω graph showing signal and noise. Filter and LNA blocks are shown.

"Inspection" Analysis of FB Ckts.

↓ starts with...

Identification of FB Connection Types

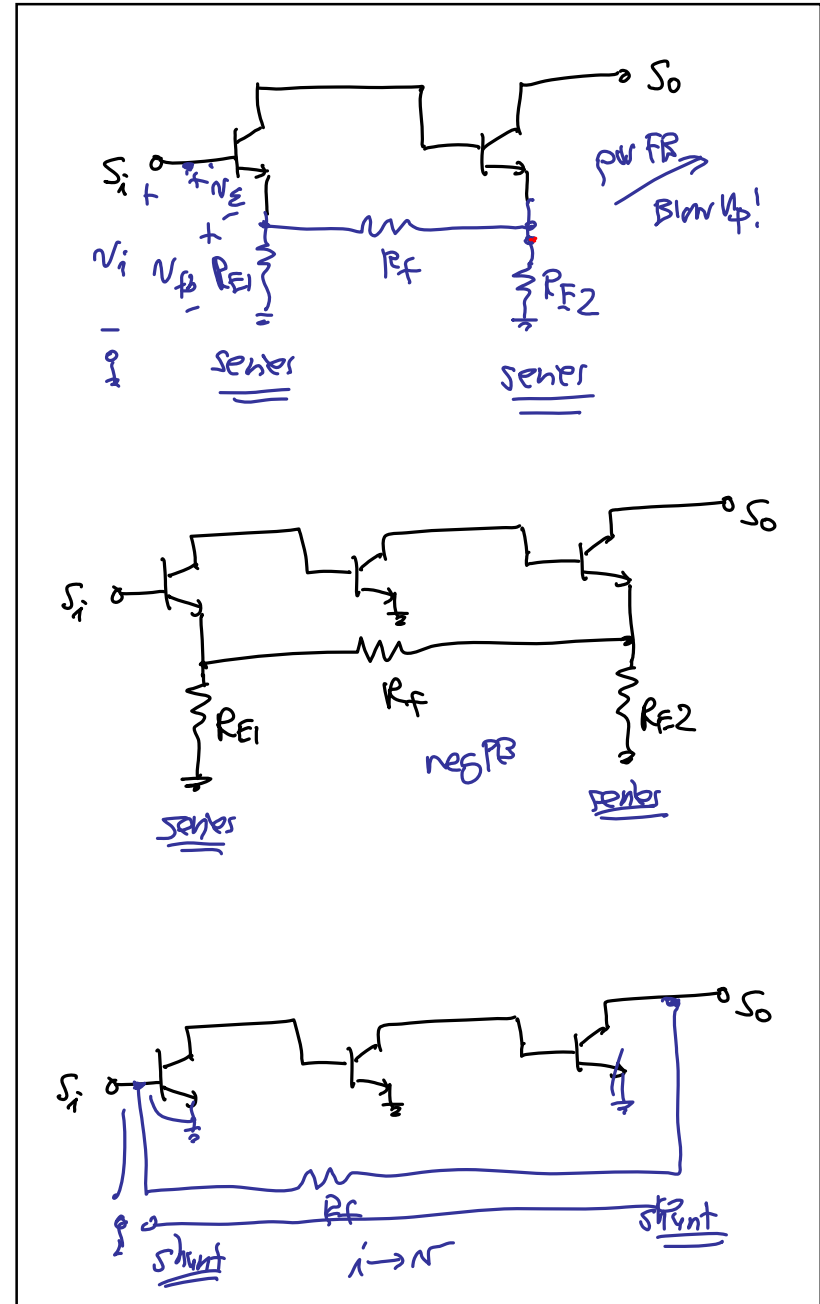
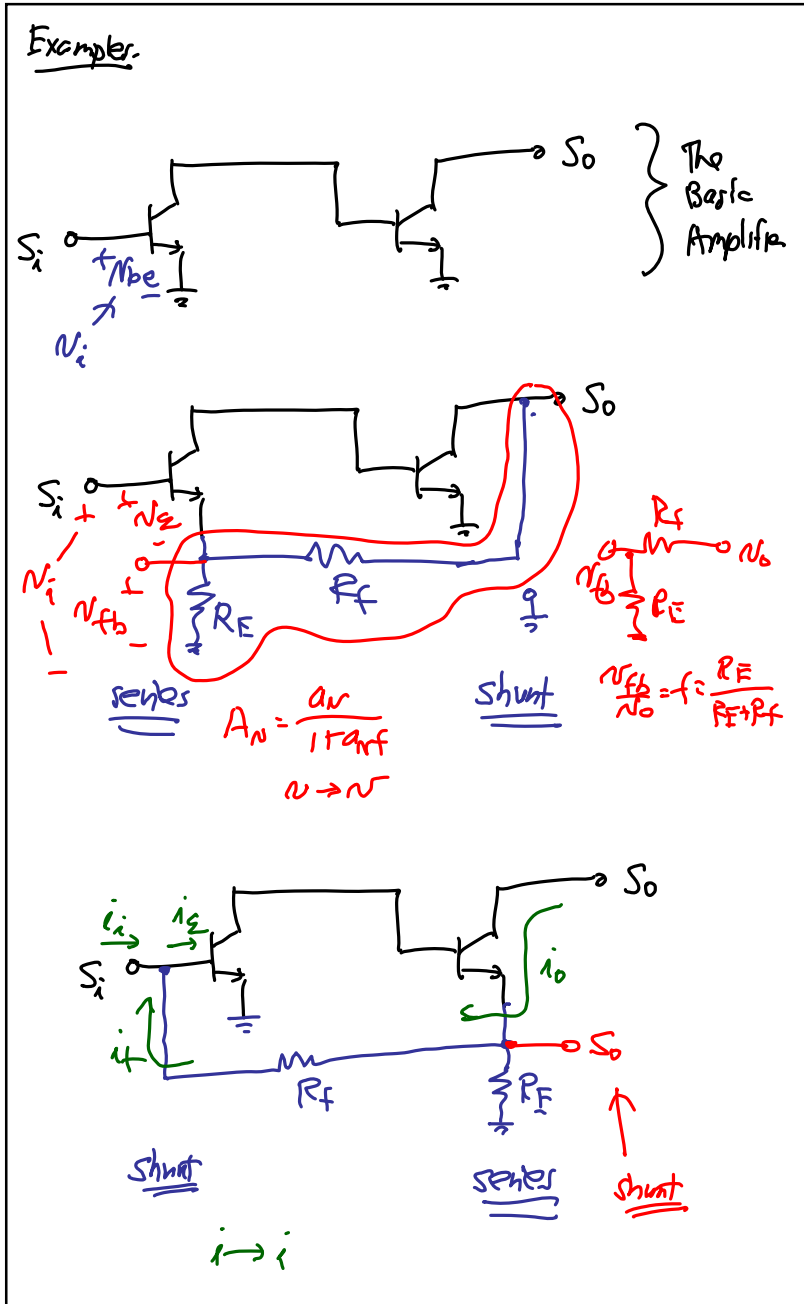
Series Connection - FB network part in series w/ amplifier part

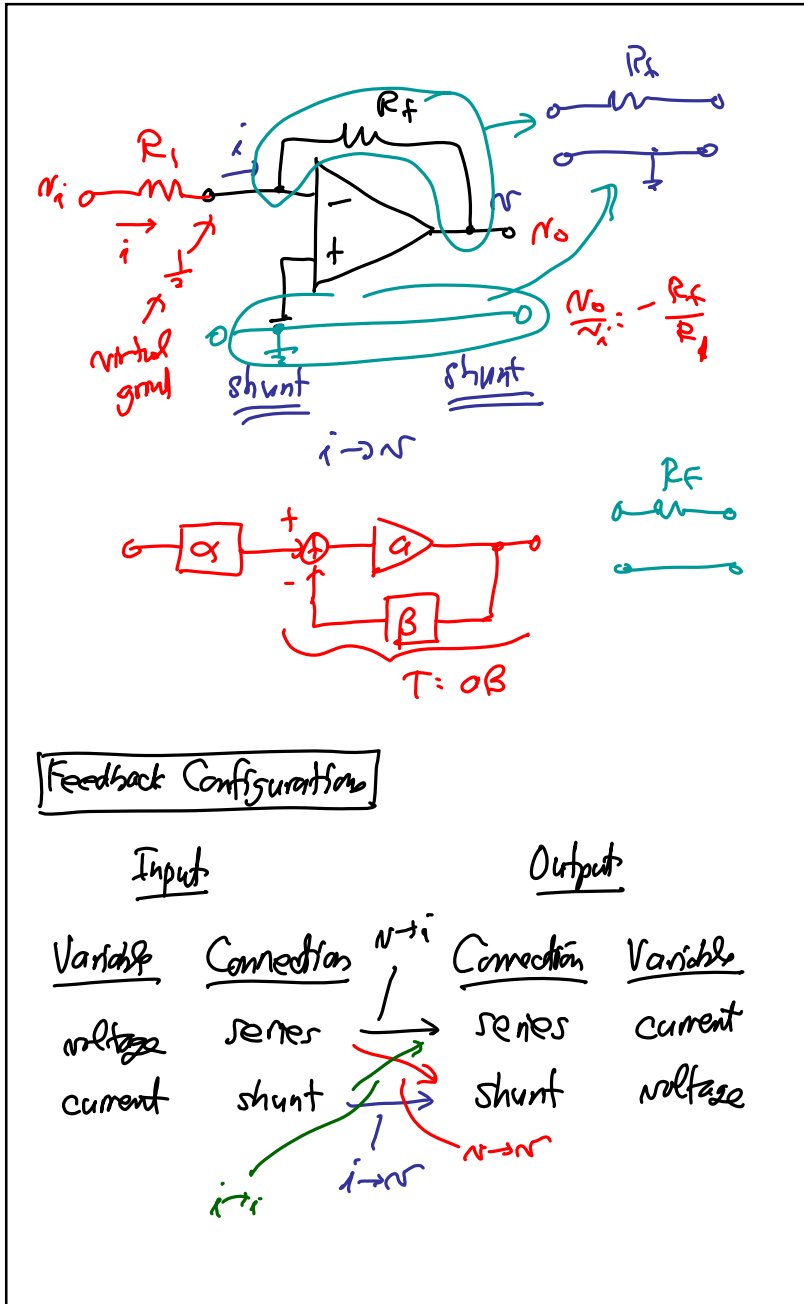
must go thru both the FB part & the amplifier part to get from the node of interest to ground

Shunt Connection - FB network part in shunt w/ amplifier part

can get from the node of interest to ground via either FB network part or the amplifier part

over

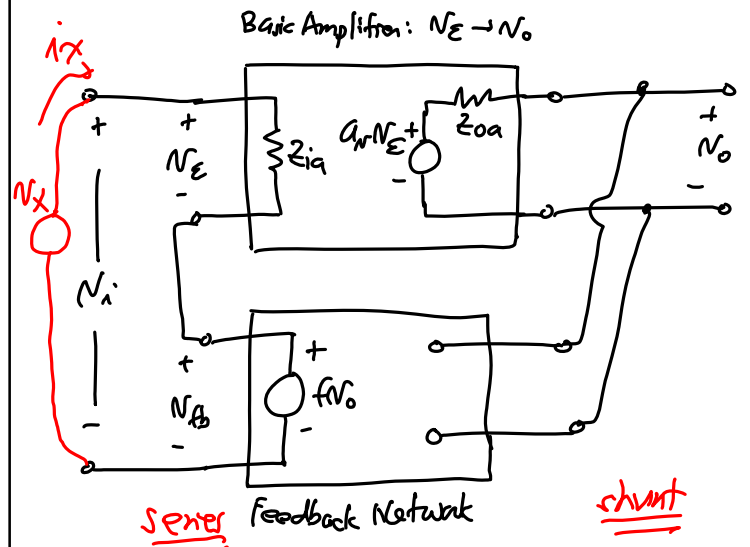




Effect of FB on Z_i & Z_o

Ex. Series-Shunt FB

Assumption: FB network has ideal impedances
i.e., it does not load the basic amplifier



Find the T.F.:

$$\left. \begin{aligned} N_E &= V_i - N_{fb} \\ N_o &= a_n N_E \\ N_{fb} &= f_v N_o \end{aligned} \right\} \Rightarrow \frac{V_o}{V_i} = \frac{a_n}{1 + a_n f_v} \quad \checkmark$$

(as expected)

Find $Z_i = \frac{V_x}{i_x}$:

$$V_x = V_\Sigma + V_{fb}$$

$$= V_\Sigma + fV_o = V_\Sigma + fA_{OL}V_\Sigma = V_\Sigma(1 + A_{OL}f)$$

$i_x = \frac{V_\Sigma}{Z_{io}}$

$$Z_i = \frac{V_x}{i_x} = \frac{V_\Sigma(1 + A_{OL}f)}{\frac{V_\Sigma}{Z_{io}}} = Z_{io}(1 + A_{OL}f) = Z_i$$

closed loop input impedance

original open-loop impedance of op amp

loop gain

loop gain!

When the series connection @ input } Closed loop input impedance raised by (1 + A_{OL}f)!

↓

If $Z_i \uparrow \rightarrow$ better voltage amplifier!

↓

at least it accepts a voltage input better