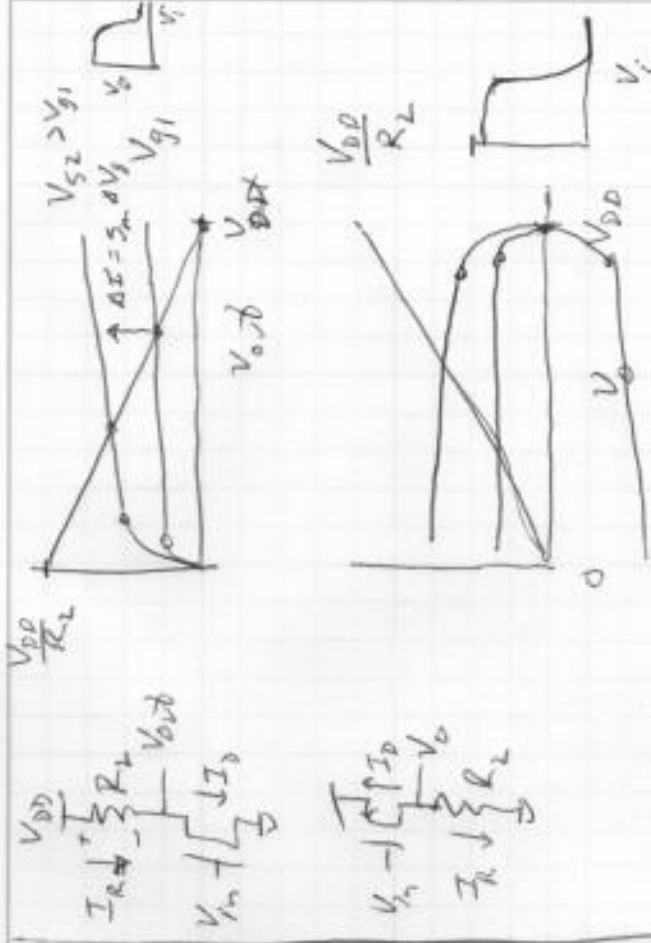
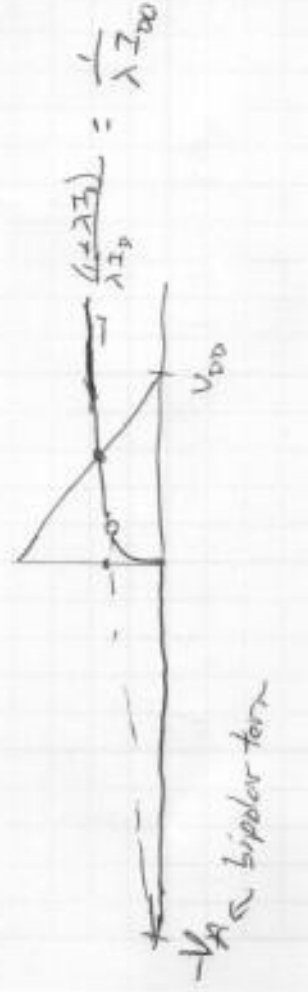


Midterm 2 weeks from today  
 in class  
 1 page, 2 sides notes  
 no calculators  
 HW 1-4  
 Lab 1+(2?)

CS amp.  
 Load line  
 small signal model (W3L1P 4<sup>th</sup>)

sticker  
 Usually  $r_o \gg R_L$  } why?  
 slope of  $I_D$  vs  $V_{DS} \ll$  slope of  $\frac{1}{R_L}$  }

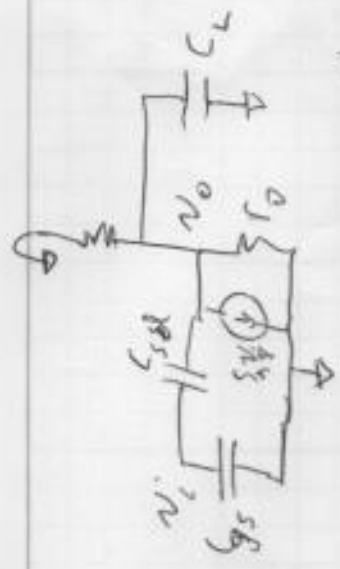
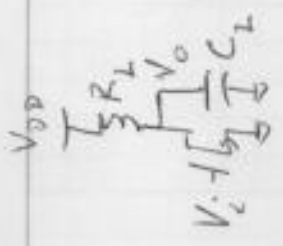


Note that  $g_m$  is changing w/ bias point

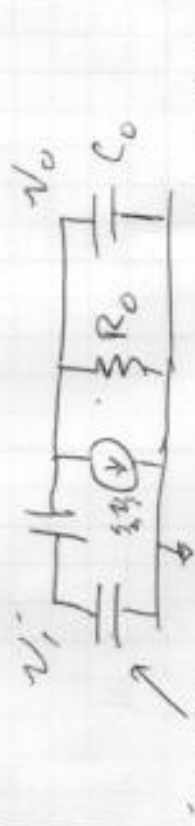
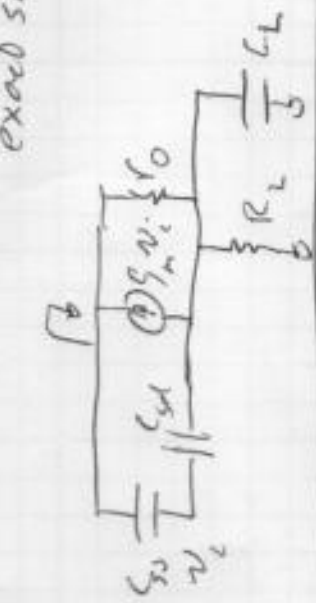
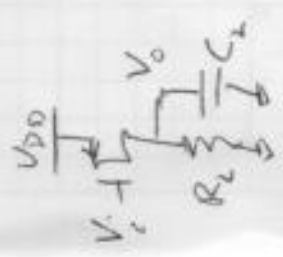
$$g_m = \left\{ \begin{array}{l} \frac{2I_{D0}}{V_{GS}} \\ \mu_{eff} \frac{C_{ox}}{2} (V_{GS}) (1 + \lambda V_{DS}) \end{array} \right.$$

$$I_{D0} = 3\mu_{eff} C_{ox} \frac{W}{L} (1 + \lambda V_{DS})$$

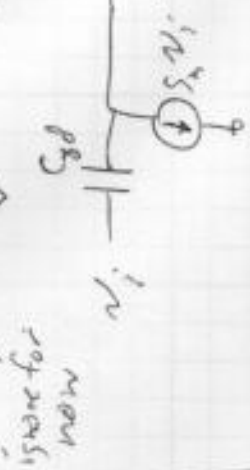




exact same!



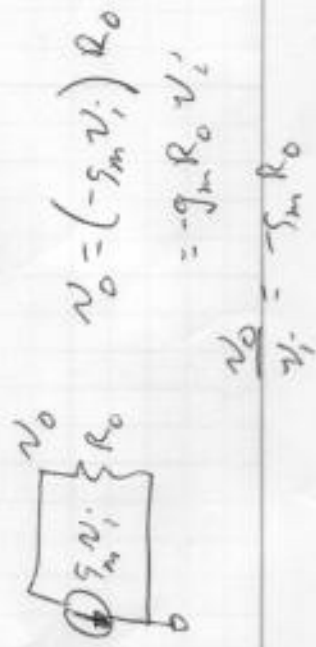
$$R_o = r_o \parallel R_L$$



$C_o \text{ typ} \gg C_{gs}$

"inspection" - really low freq } goal: eliminate  
 " high freq } component  
 in between

low freq:  $Z_c$  huge - ignore caps  
 compared to:  $R_o \approx \frac{1}{g_m}$

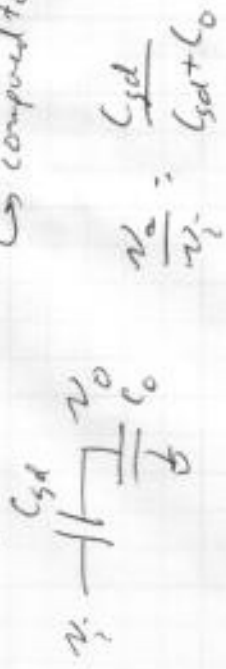


$$V_o = (-g_m V_i) R_o$$

$$= -g_m R_o V_i$$

$$\frac{V_o}{V_i} = -g_m R_o$$

high freq:  $Z_c$  small  
 compared to  $R_o, \frac{1}{g_m}$

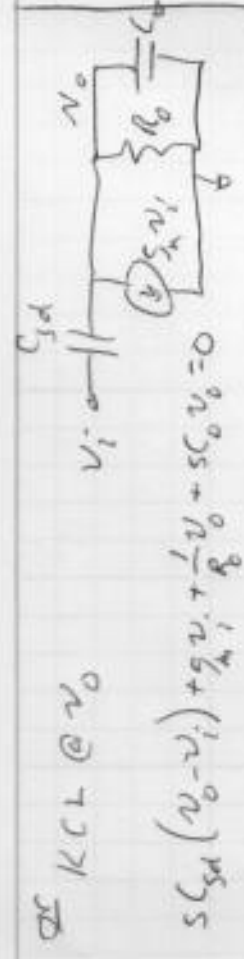


mid  $V_i = 0$   
 ignore  $r_o$   $C_{gs}$   
 ignore  $r_o$   $C_{gs}$   
 $r_o \gg \frac{1}{g_m}$   
 $g_m \gg \omega C_{gs}$



$$V_o = (-g_m V_i) Z_c$$

$$\frac{dV_o}{dV_i} = -\frac{g_m}{s C_o}$$



$g_m = 100 \mu S$     $R_o = 1 M$     $C_L = 1 pF$     $C_{sd} = 10 pF$   
 $A_{v_0} = 100$   
 $\omega_p = 10^6 \text{ rad/s}$   
 $\omega_z = \frac{10^{-10}}{10^{-14}} = 10^{10} \frac{\text{rad}}{s}$   
 $H(s) = \begin{cases} -100 & \omega \ll 10^6 \text{ rad/s} \\ -\frac{10^3}{s} & \text{in between} \\ \frac{1}{100} & \omega \gg 10^{10} \text{ rad/s} \end{cases}$

$v_o(s C_{sd} + s C_o + \frac{1}{R_o}) = (g_m - s C_{sd}) v_i$

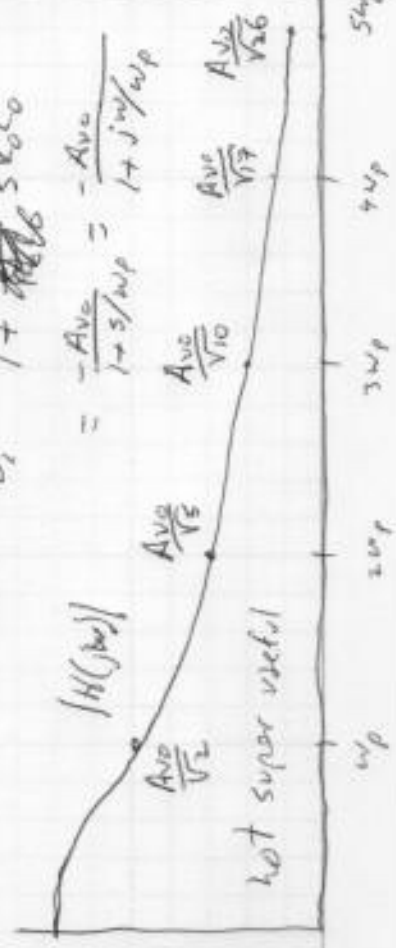
$H(s) = \frac{v_o}{v_i} = \frac{-g_m (1 - s C_{sd})}{\frac{1}{R_o} (1 + s(C_o + C_{sd}) R_o)}$   
 $= -g_m R_o \frac{1 + \frac{s}{\omega_z}}{1 + \frac{s}{\omega_p}}$

$\omega_z = \frac{g_m}{C_{sd}}$  RHP    $\omega_p = \frac{1}{R_o(C_o + C_{sd})}$

Single pole amplifiers - ignore  $C_{sd}$  for now



$KCL @ v_o \Rightarrow H(s) = \frac{v_o}{v_i} = \frac{-g_m R_o}{1 + s R_o C_o} = \frac{-A_{v_0}}{1 + s/\omega_p} = \frac{-A_{v_0}}{1 + j\omega/\omega_p}$



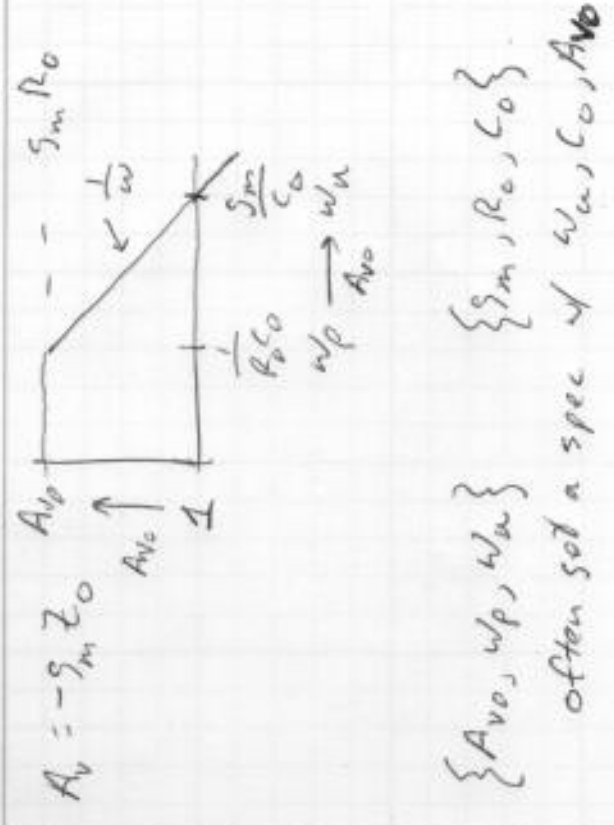
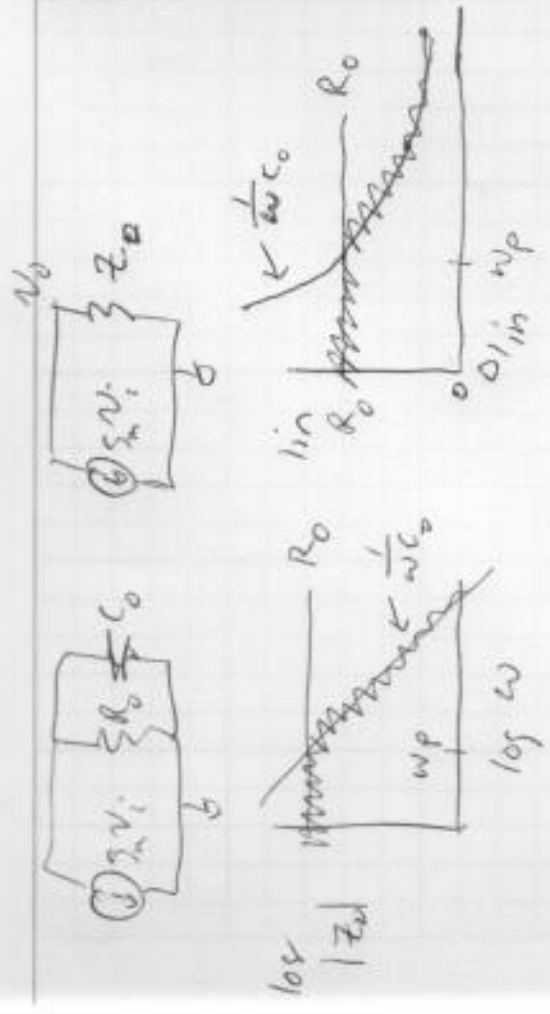
when is  $|H(j\omega)| = 1$ ?

$|H(j\omega)| = 1 = \frac{g_m R_o}{|1 + j\omega R_o C_o|} \approx \frac{g_m R_o}{\omega R_o C_o} = \frac{g_m}{\omega C_o}$

$\omega_u = \frac{g_m}{C_o}$

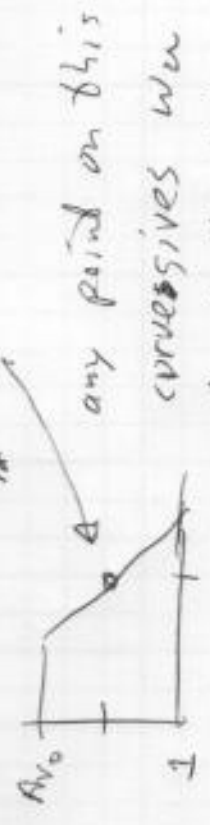
-or-

$\omega_u = A_{v_0} \omega_p = (g_m R_o) \frac{1}{R_o C_o} = \frac{g_m}{C_o}$



$\{A_{v0}, w_p, w_a\}$   
 $\{g_m, R_o, C_o\}$   
 often set a spec w/  $w_a, C_o, A_{v0}$

need a gain of 100 at 1 MHz, 1 pF load



any point on this curve gives  $w_a$   
 • bounds  $A_{v0}$  between

increase  $R_o$  by  $\alpha$

increase  $C_o$  by  $\alpha$

increase  $g_m$  by  $\alpha$

