

# Systems

- system Transform input into output

$$T \left[ \underset{\substack{\uparrow \\ \text{input}}}{x(n_1, n_2)} \right] = \underset{\substack{\uparrow \\ \text{output}}}{y(n_1, n_2)}$$

- Linear :  $T \left\{ a x_1(n_1, n_2) + b x_2(n_1, n_2) \right\}$   
 $= a T \left\{ x_1(n_1, n_2) \right\} + b T \left\{ x_2(n_1, n_2) \right\}$

Shift Invariance :

- If  $T \left\{ x(n_1, n_2) \right\} = y(n_1, n_2)$   
Then  $T \left[ x(n_1 - k_1, n_2 - k_2) \right] = y(n_1 - k_1, n_2 - k_2)$

2D-LSI System.

LSI  $\longrightarrow$  Impulse response.

$h(n_1, n_2)$

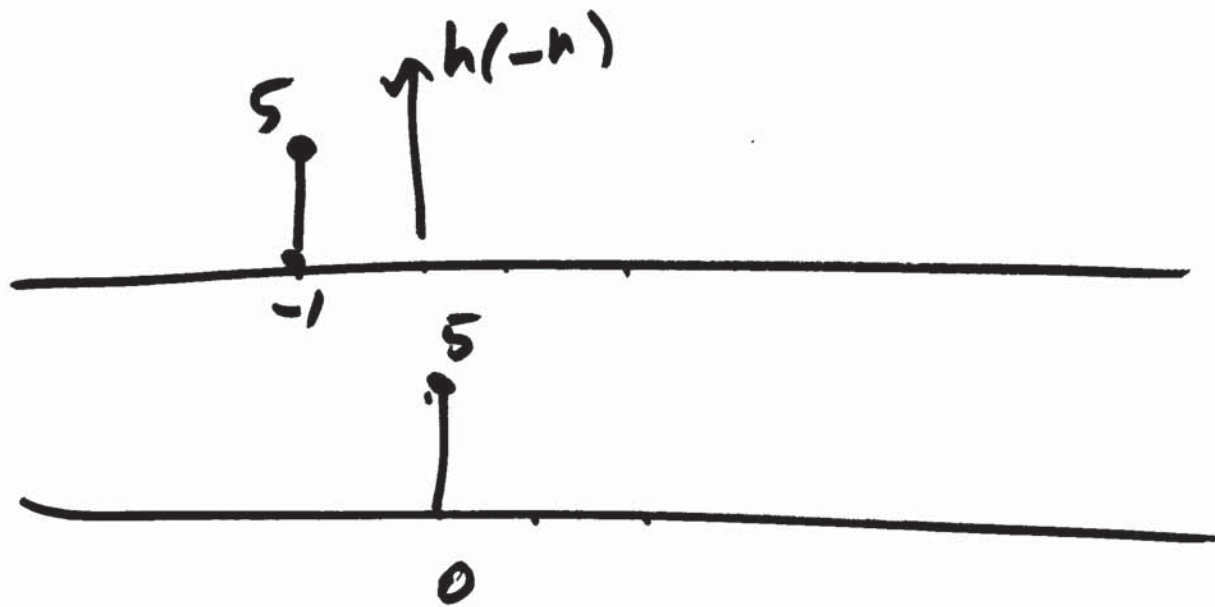
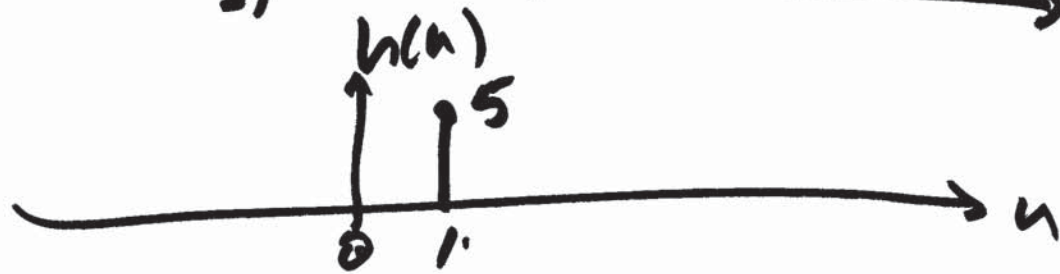
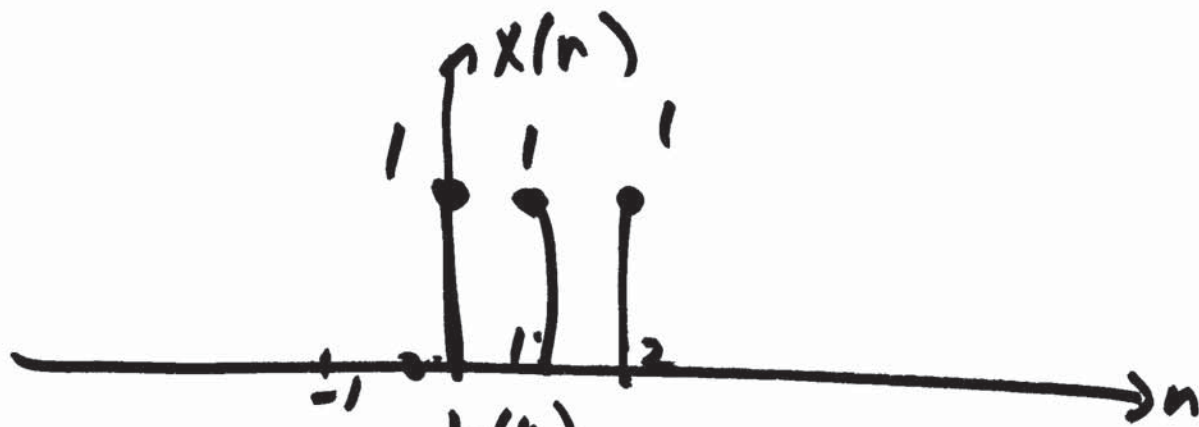
- enables to entirely characterize an LSI system.

- tells you output given input.

$$Y(n_1, n_2) = X * h \xrightarrow{\text{Convolution}}$$

$$y(n_1, n_2) = \sum_{k_1} \sum_{k_2} x(k_1, k_2) h(n_1 - k_1, n_2 - k_2) \quad 2D$$

$$Y(n) = \sum_k X(k) h(n - k) \quad 1D$$



$$y(0) = 0$$

$$y(1) = 5$$

# Separable LSI system

$$h(n_1, n_2) = h_1(n_1) h_2(n_2)$$

$$X(n_1, n_2) \quad N \times N$$

$$L: \quad M \times M.$$



general:

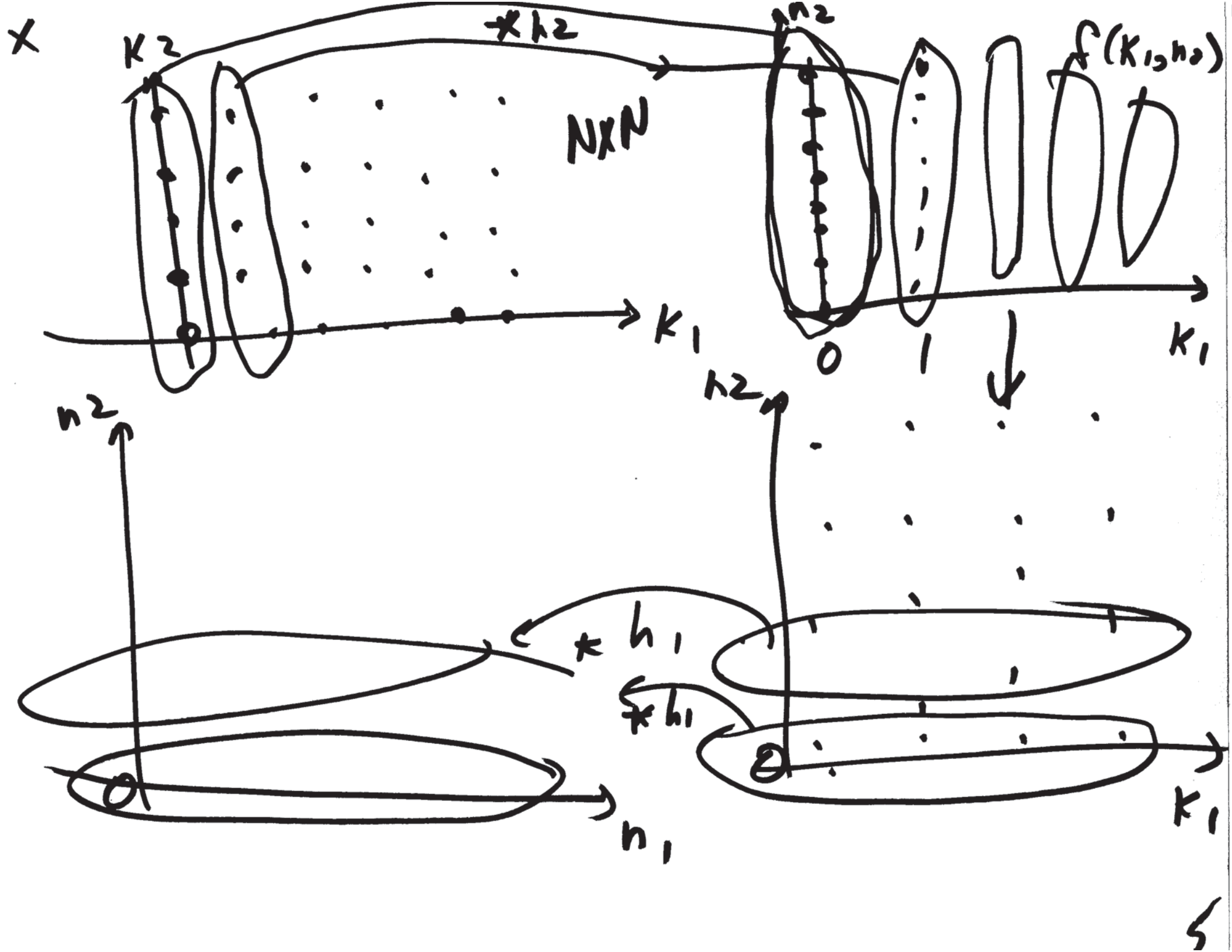
$$y(n_1, n_2) = \sum_{k_1} \sum_{k_2} x(k_1, k_2) h(n_1 - k_1, n_2 - k_2)$$

sep. sys.  $\downarrow$

$$= \sum_{k_1} \sum_{k_2} x(k_1, k_2) h_1(n_1 - k_1) h_2(n_2 - k_2)$$

$$= \sum_{k_1} h_1(n_1 - k_1) \underbrace{\sum_{k_2} x(k_1, k_2) h_2(n_2 - k_2)}_{f(k_1, n_2)}$$

$$f(k_1, n_2) = x(k_1) * h_2$$



Summary

1. Series of 1D convoliter To get.

$f(k_1, n_2)$  :

N columns  $\rightarrow$

$\left\{ \begin{array}{l} N \text{ convoliter.} \\ NM \text{ op / convol} \end{array} \right.$

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$N^2 M$  op To build  $f$

(2) Series of 1-D conv To get  $g(n, n_1)$

N rows  $\rightarrow$

$\left\{ \begin{array}{l} N \text{ convoliter.} \\ NM \text{ op / convol} \end{array} \right.$

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$N^2 M$

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Total  $2 \cdot N^2 M$ .



# LSI system

\* impulse response.  $\rightarrow h(n_1, n_2)$

\* stability of LSI system.

\* BIBO stability  $\rightarrow$

$\nearrow$  bounded input, bounded output.

\* Can show, necessary + suff. cond.  
 $h(n_1, n_2)$  to be BIBO stable.

$$\sum_{n_1=-\infty}^{+\infty} \sum_{n_2=-\infty}^{+\infty} |h(n_1, n_2)| < \infty$$

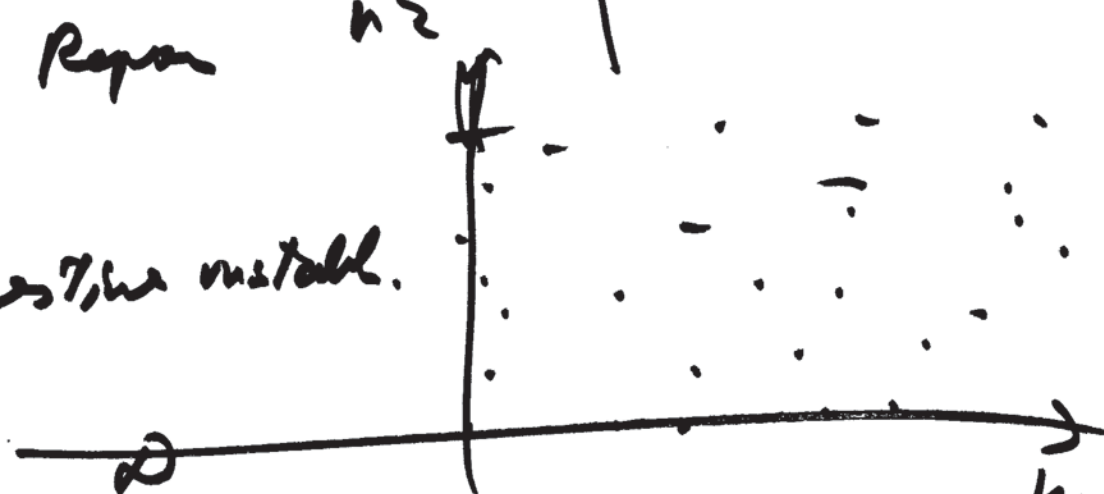
Finite impulse response  
(FIR).

→ Always stable.



Infinite impulse response  
(IIR)

Some stable, some unstable.

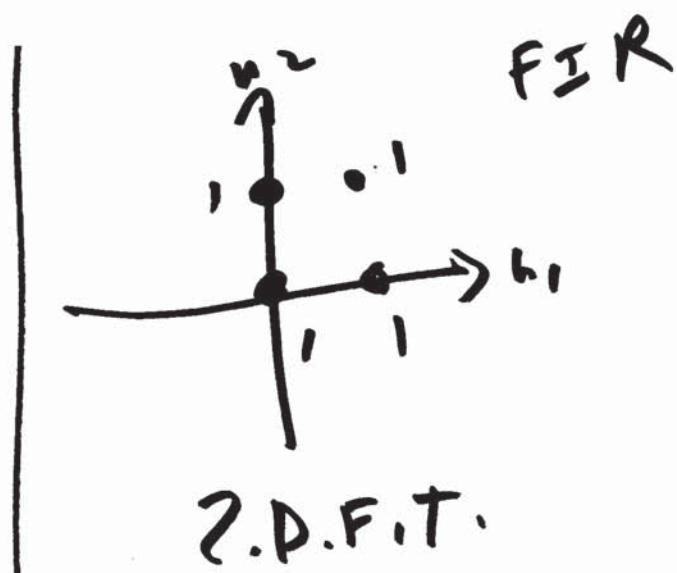


$\sum_{n=-\infty}^{\infty} \frac{1}{n}$  blows up  $\sum_{n=0}^{\infty} \frac{1}{n^2}$  does not blow up.

$h(n_1, n_2) \xrightarrow{\text{z.T.}} \sum_{n_1} \sum_{n_2} h(n_1, n_2) z_1^{-n_1} z_2^{-n_2}$   
 $= H(z_1, z_2) = \text{2D polynomial in } z_1, z_2$

FIR

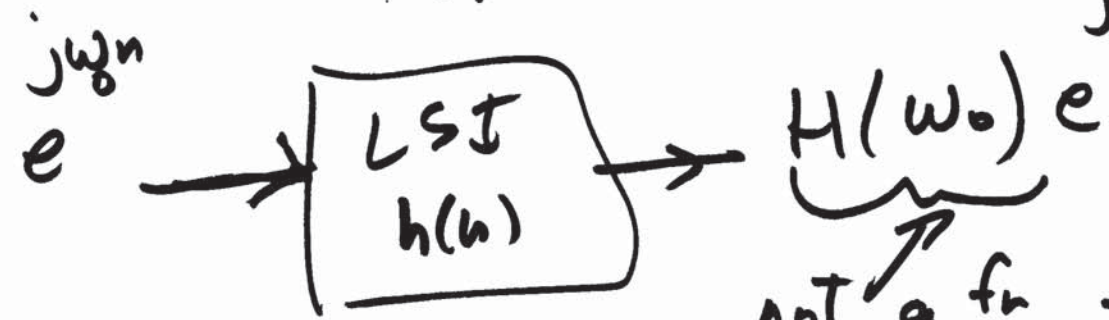




$$H(z_1, z_2) = 1 + z_1^{-1} + z_2^{-1} + z_1^{-1} z_2^{-1}$$

2D polynomial in  $z_1$  and in  $z_2$

$$H(\omega_1, \omega_2) = \sum_{n_1} \sum_{n_2} h(n_1, n_2) e^{j\omega_1 n_1} e^{j\omega_2 n_2}$$



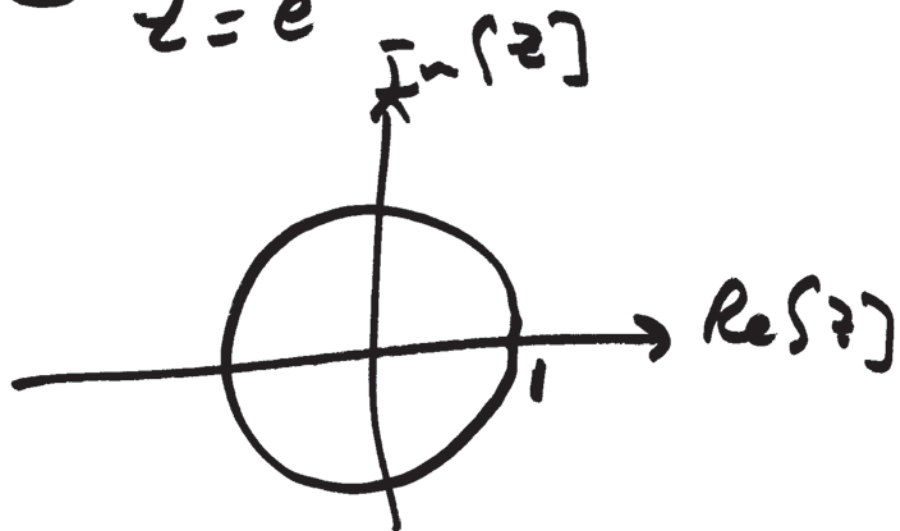
$$H(\omega_0) = \left[ H(\omega) \right]_{\omega=\omega_0} = \left[ \sum_{n=-\infty}^{+\infty} h(n) e^{-j\omega n} \right]_{\omega=\omega_0}$$

not a fn of  $n$ .

$$H(\omega) = \sum_{n=-\infty}^{+\infty} h(n) e^{-j\omega n}$$

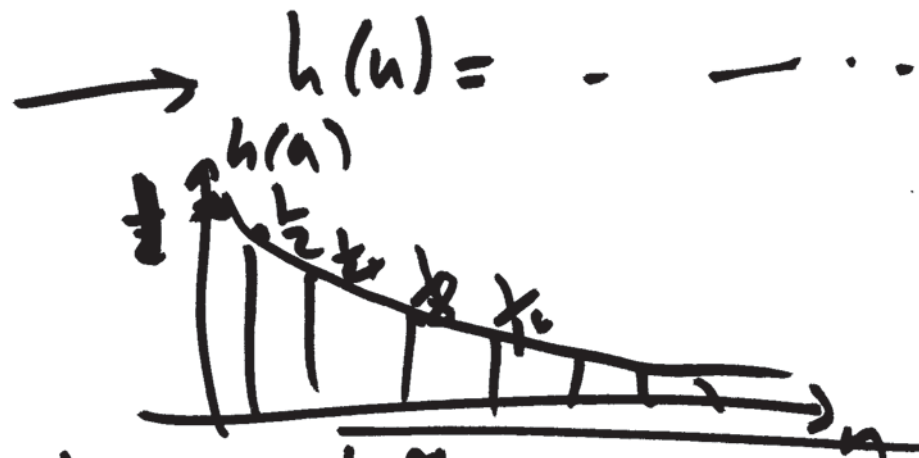
$$H(z) = \sum_{n=-\infty}^{+\infty} h(n) z^{-n}$$

$$H(\omega) = [H(z)]_{z=e^{j\omega}}$$



$$H(z) = \frac{1}{1 - \frac{1}{2}z^{-1}}$$

ROC:  $\left\} \text{outside circle}$



$$H(z) = \frac{Y(z)}{X(z)} = \frac{1}{1 - \frac{1}{2}z^{-1}}$$

$$\sum_{n=-\infty}^{\infty} |h(n)| < \infty$$

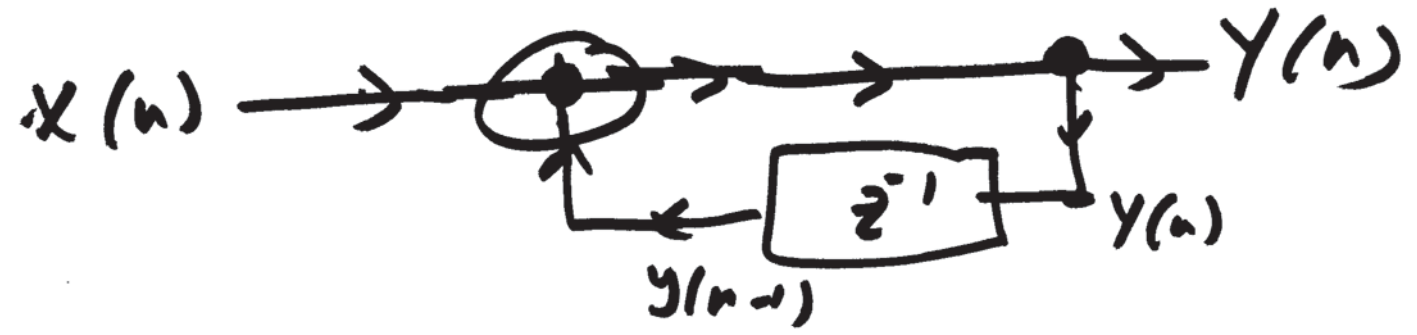
BIBO

$$Y(z) \left[ 1 - \frac{1}{2}z^{-1} \right] = X(z)$$

~~$y(n) - \frac{1}{2}y(n-1) = x(n)$~~

$$y(n) - \frac{1}{2}y(n-1) = x(n)$$

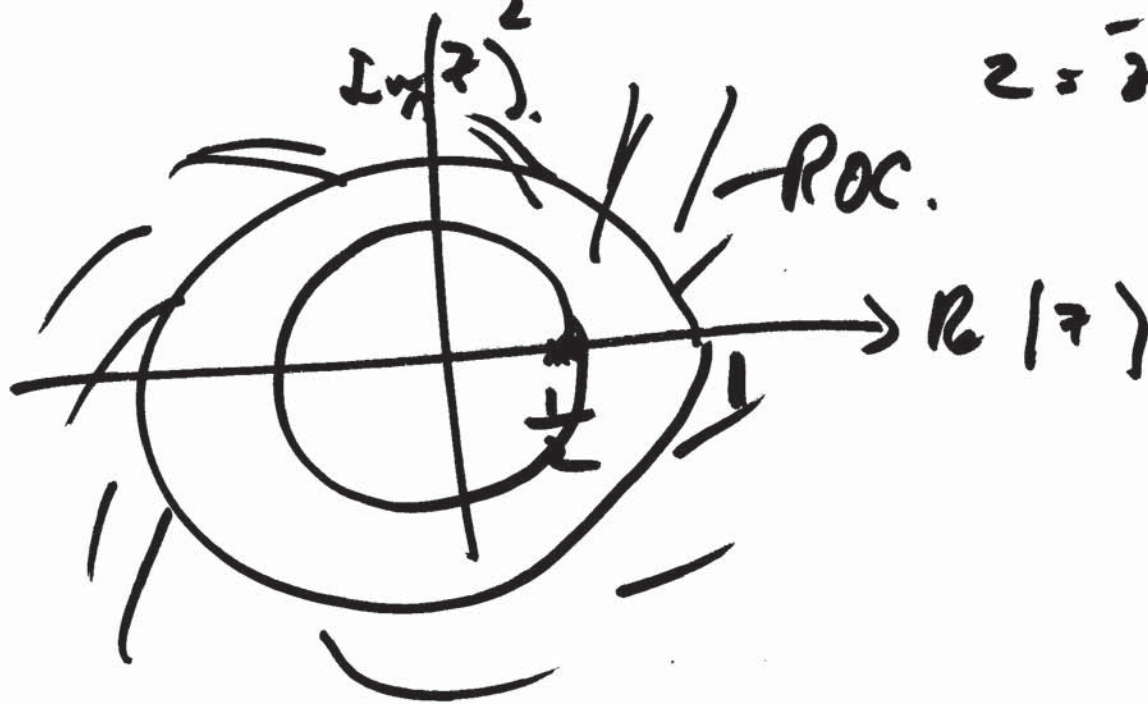
$$y(n) = x(n) + \frac{1}{2}y(n-1)$$



pole

$$1 - \frac{1}{2} z^{-1} = 0 \Rightarrow 1 = \frac{1}{2} z^{-1}$$

$$z = \frac{1}{2} \Rightarrow z = \frac{1}{2}$$

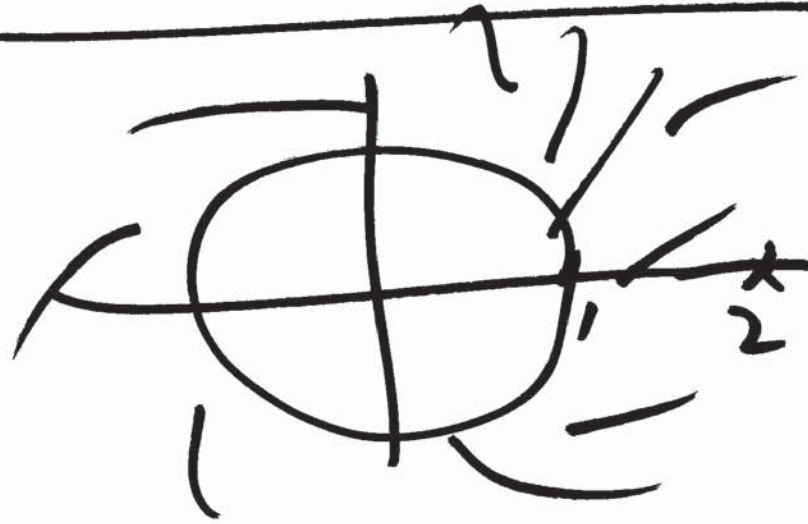


ROC.

$$|z| > \frac{1}{2}$$

$$h(n) = \frac{1}{2}^n u(n)$$

unstable



$$h(n) = 2^n u(n)$$

$$= 1 + 2 + 4 + 8 + \dots$$

$$H(z) = \frac{1}{1 - 2z^{-1}}$$

$M(z)$ :

$$1 + 5z^{-1} + 3z^{-2} + 9z^{-3} + 12z^{-4} + 18z^{-5}$$

↓  
2D polynomial can always  
Be factored.

→ Fundamental Thm of Algebra.

Any 1-D polynomial ~~of~~ of deg.  $n$   
can be factored into  $n$   
polynomials of deg. 1.

$$= \prod_{i=1}^n (1 - \alpha_i z^{-1})$$

# Story in 2D

IIR :

$$H(z_1, z_2) = \frac{1}{P(z_1^{-1}, z_2^{-1})}$$

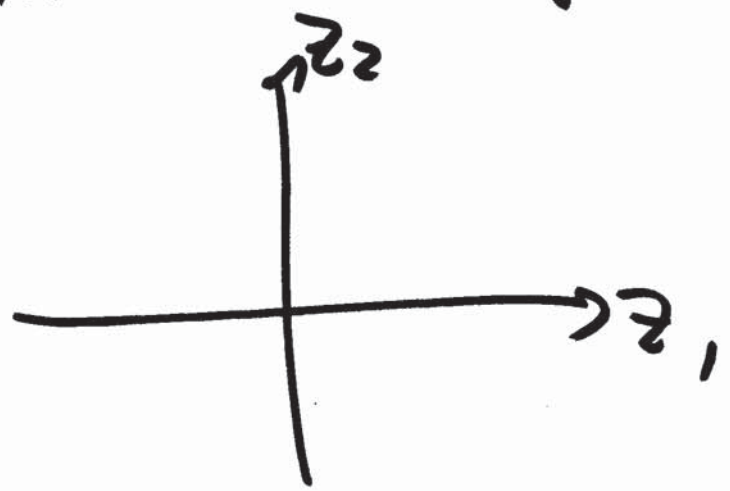
$$P(z_1^{-1}, z_2^{-1}) = (1 - z_1^{-1})(1 - z_2^{-1})$$

Thm:

Hayes 1980's :

Set of reducible 2D polynomials is of measure 0 in the set of all 2D polynomials.

2D  $\rightarrow$  unit bisurface

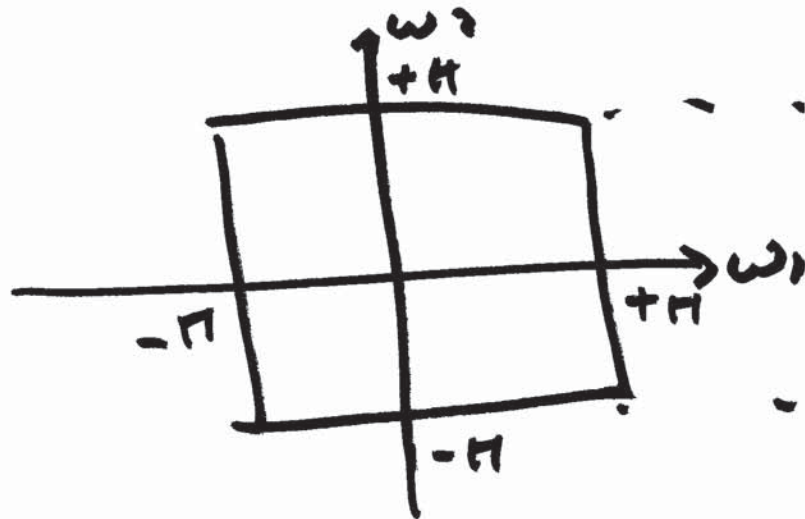


~~$P(z_1, z_2)$~~



Do the poly surface cross the  
unit Bi-surface?  $\Rightarrow$  some things  
covered later.

Proving stability in 2D is hard



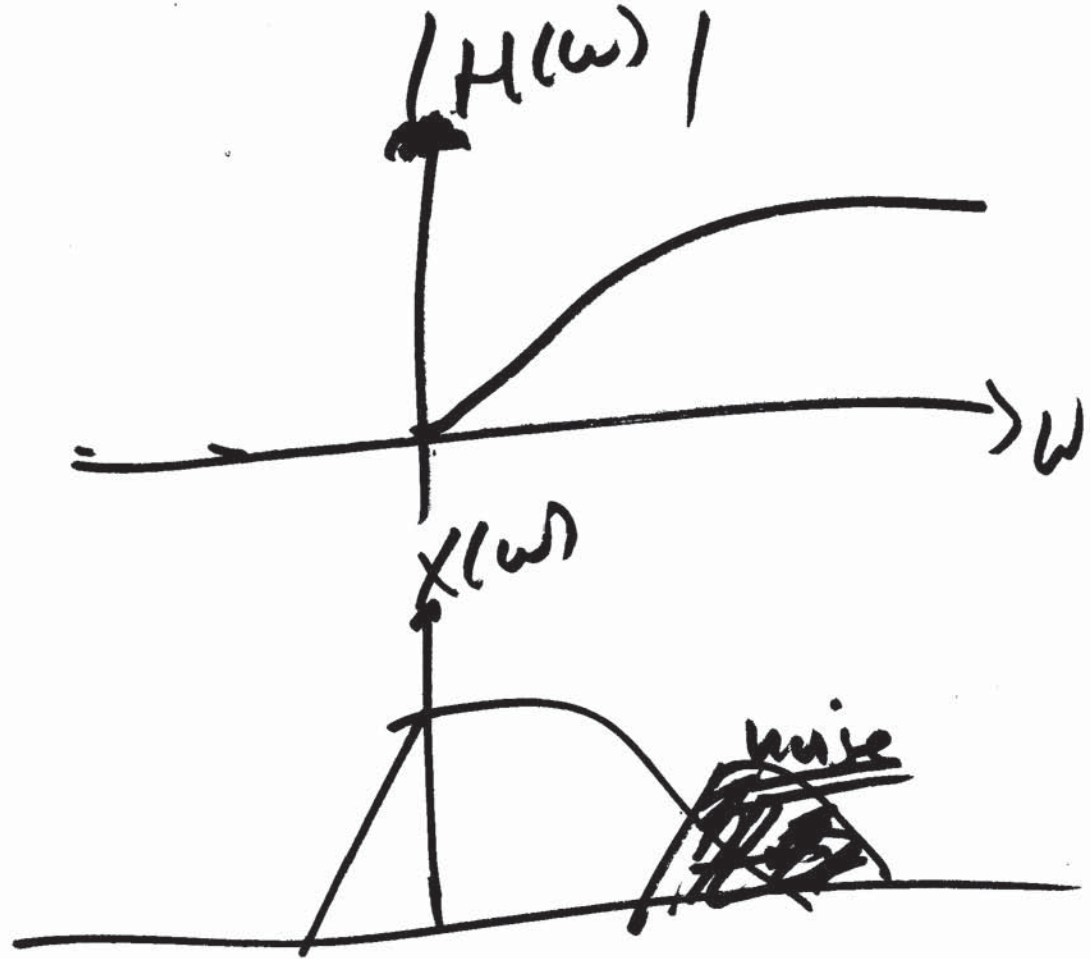
$X(n_1, n_2)$

discrete  
integers



$X(\omega_1, \omega_2)$

Real variables.



Signal.

