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EE C247B - ME C218 Introduction to MEMS Design Spring 2015

Prof. Clark T.-C. Nguyen

Dept. of Electrical Engineering & Computer Sciences
 University of California at Berkeley
 Berkeley, CA 94720

Lecture Module 13: Equivalent Circuits II

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Lecture Outline

- Reading: Senturia, Chpt. 6, Chpt. 14
- Lecture Topics:
 - ↳ Input Modeling
 - Force-to-Velocity Equiv. Ckt.
 - Input Equivalent Ckt.
 - ↳ Current Modeling
 - Output Current Into Ground
 - Input Current
 - Complete Electrical-Port Equiv. Ckt.
 - ↳ Impedance & Transfer Functions

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Input Modeling

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Electromechanical Analogies

$F(t) = F \cos(\omega t) \rightarrow x(t) = X \cos \omega t$
 $N(t) = N \cos \omega t \rightarrow i(t) = I \cos \omega t$

Equation of Motion:
 $m_{eq} \ddot{x} + c_{eq} \dot{x} + k_{eq} x = F(t)$
 \Rightarrow using phasor concepts:
 $F = j\omega m_{eq} \dot{x} + \frac{k_{eq}}{j\omega} x + c_{eq} \dot{x}$
 \Rightarrow by analogy:

$F \rightarrow N$	$m_{eq} \rightarrow l_x$	$c_{eq} \rightarrow r_x$
$\dot{x} \rightarrow i$	$k_{eq} \rightarrow \frac{1}{c_x}$	

[Parameter Relationships in the Current Analogy]

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Bandpass Biquad Transfer Function

$$F = j\omega m_{eq} \dot{x} + \frac{k_{eq}}{j\omega} x + c_{eq} \dot{x}$$

$$\Rightarrow \text{Converting to full phasor form:}$$

$$F = (j\omega)(j\omega x) m_{eq} + \frac{k_{eq}}{j\omega} (j\omega x) + c_{eq} (j\omega x)$$

$$\frac{X}{F}(j\omega) = \frac{1}{k_{eq}} \left[-\omega^2 \frac{m_{eq}}{k_{eq}} + 1 + j \frac{c_{eq}\omega}{k_{eq}} \right]^{-1} = \frac{1}{k_{eq}} \left[-\left(\frac{\omega}{\omega_0}\right)^2 + 1 + j \frac{\omega}{Q\omega_0} \right]^{-1}$$

$$\left[\frac{k_{eq}}{m_{eq}} = \omega_0^2, Q = \frac{m_{eq}\omega_0}{c_{eq}} = \frac{k_{eq}}{\omega_0 c_{eq}} \rightarrow \frac{k_{eq}}{c_{eq}} = Q\omega_0 \right]$$

$$\left| \frac{X}{F}(j\omega) \right|$$

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Force-to-Velocity Relationship

- The relationship between input voltage v_1 and force F_{d1} :

$$F_{d1} \approx -V_P \frac{\partial C_1}{\partial x} v_1$$
- When displacement x is the mechanical output variable:

$$\frac{X(s)}{F_{d1}(s)} = \frac{1}{k} \frac{\omega_o^2}{s^2 + (\omega_o/Q)s + \omega_o^2}$$
- When velocity v is the mechanical output variable:

$$\frac{v(s)}{F_{d1}(s)} = \frac{sX(s)}{F_{d1}(s)} = \frac{1}{k} \frac{\omega_o^2 s}{s^2 + (\omega_o/Q)s + \omega_o^2}$$

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Force-to-Velocity Equiv. Ckt.

- Combine the previous lumped LCR mechanical equivalent circuit with a circuit modeling the capacitive transducer → circuit model for voltage-to-velocity
- $$U = -\dot{x} \quad |_{x=m} \quad |_{x=b} \quad r_x = b$$
- $$c_x = 1/k$$

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Equiv. Circuit for a Linear Transducer

- A transducer ...
 - converts energy from one domain (e.g., electrical) to another (e.g., mechanical)
 - has at least two ports
 - is not generally linear, but is virtually linear when operated with small signals (i.e., small displacements)

Electrical: Voltage V , Current I
 Mechanical: Velocity $U = -\dot{x}$, Force F

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Equiv. Circuit for a Linear Transducer

Electrical | Mechanical

- For physical consistency, use a transformer equivalent circuit to model the energy conversion from the electrical domain to mechanical domain

$$\begin{bmatrix} e_2 \\ f_2 \end{bmatrix} = \begin{bmatrix} \eta & 0 \\ 0 & -\frac{1}{\eta} \end{bmatrix} \begin{bmatrix} e_1 \\ f_1 \end{bmatrix}$$

Describing Matrix

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Electromechanical Equivalent Circuit

- $e_2 = F_{d1}$, $e_1 = v_1$, just need η_1 :
- From the matrix: $e_2 = \eta e_1$

$$F_{d1} \approx -V_P \frac{\partial C_1}{\partial x} v_1 \rightarrow \eta_1 = \left| V_P \frac{\partial C_1}{\partial x} \right|$$

Electrical | Mechanical

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Complete Electrical-Port Equiv. Circuit

Static electrode-to-mass overlap capacitance

$$\eta_{e1} = V_P \frac{\partial C_1}{\partial x} = V_P \frac{C_{o1}}{d_1}$$

$$\eta_{e2} = V_P \frac{\partial C_2}{\partial x} = V_P \frac{C_{o2}}{d_2}$$

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