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## EE C247B - ME C218 Introduction to MEMS Design Spring 2017

**Prof. Clark T.-C. Nguyen**

Dept. of Electrical Engineering & Computer Sciences  
 University of California at Berkeley  
 Berkeley, CA 94720

**Lecture Module 13: Equivalent Circuits II**

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## Lecture Outline

- Reading: Senturia, Chpt. 6, Chpt. 14
- Lecture Topics:
  - ↪ Input Modeling
    - Force-to-Velocity Equiv. Ckt.
    - Input Equivalent Ckt.
  - ↪ Current Modeling
    - Output Current Into Ground
    - Input Current
    - Complete Electrical-Port Equiv. Ckt.
  - ↪ Impedance & Transfer Functions

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## Input Modeling

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## Electromechanical Analogies

Equation of Motion:

$$m_{eq} \ddot{x} + c_{eq} \dot{x} + k_{eq} x = F(t)$$

⇒ using phasor concepts:

$$F = j\omega m_{eq} \dot{x} + \frac{k_{eq}}{j\omega} x + c_{eq} \dot{x}$$

⇒ by analogy:

$F \rightarrow N$	$m_{eq} \rightarrow l_x$	$c_{eq} \rightarrow r_x$
$\dot{x} \rightarrow i$	$k_{eq} \rightarrow \frac{1}{c_x}$	

Parameter Relationships in the Current Analogy

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### Bandpass Biquad Transfer Function

$$F = j\omega m_{eq} \dot{x} + \frac{k_{eq}}{j\omega} x + c_{eq} \dot{x}$$

$$\Rightarrow \text{Converting to full phasor form:}$$

$$F = (j\omega)(j\omega x) m_{eq} + \frac{k_{eq}}{j\omega} (j\omega x) + c_{eq} (j\omega x)$$

$$\frac{X}{F}(j\omega) = \frac{1}{k_{eq}} \left[ -\omega^2 \frac{m_{eq}}{k_{eq}} + 1 + j \frac{c_{eq}\omega}{k_{eq}} \right]^{-1} = \frac{1}{k_{eq}} \left[ -\left(\frac{\omega}{\omega_0}\right)^2 + 1 + j \frac{\omega}{Q\omega_0} \right]^{-1}$$

$$\left[ \frac{k_{eq}}{m_{eq}} = \omega_0^2, Q = \frac{m_{eq}\omega_0}{c_{eq}} = \frac{k_{eq}}{\omega_0 c_{eq}} \rightarrow \frac{k_{eq}}{c_{eq}} = Q\omega_0 \right]$$

$$\left| \frac{X}{F}(j\omega) \right|$$

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### Force-to-Velocity Relationship

- The relationship between input voltage  $v_1$  and force  $F_{d1}$ :
 
$$F_{d1} \approx -V_P \frac{\partial C_1}{\partial x} v_1$$
- When displacement  $x$  is the mechanical output variable:
 
$$\frac{X(s)}{F_{d1}(s)} = \frac{1}{k} \frac{s^2 + \omega_0^2}{s^2 + (\omega_0/Q)s + \omega_0^2}$$
- When velocity  $v$  is the mechanical output variable:
 
$$\frac{v(s)}{F_{d1}(s)} = \frac{sX(s)}{F_{d1}(s)} = \frac{1}{k} \frac{\omega_0^2 s}{s^2 + (\omega_0/Q)s + \omega_0^2}$$

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### Force-to-Velocity Equiv. Ckt.

- Combine the previous lumped LCR mechanical equivalent circuit with a circuit modeling the capacitive transducer  $\rightarrow$  circuit model for voltage-to-velocity

$$U = -\dot{x} \quad I_x = m \quad r_x = b$$

$$c_x = 1/k$$

Electrical | Mechanical

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### Equiv. Circuit for a Linear Transducer

- A transducer ...
  - converts energy from one domain (e.g., electrical) to another (e.g., mechanical)
  - has at least two ports
  - is not generally linear, but is virtually linear when operated with small signals (i.e., small displacements)

Electrical | Mechanical

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### Equiv. Circuit for a Linear Transducer

Electrical | Mechanical

- For physical consistency, use a transformer equivalent circuit to model the energy conversion from the electrical domain to mechanical domain

$$\begin{bmatrix} e_2 \\ f_2 \end{bmatrix} = \begin{bmatrix} \eta & 0 \\ 0 & -\frac{1}{\eta} \end{bmatrix} \begin{bmatrix} e_1 \\ f_1 \end{bmatrix}$$

Describing Matrix

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### Complete Electrical-Port Equiv. Circuit

Static electrode-to-mass overlap capacitance

$$l_x = m \quad c_x = \frac{1}{k} \quad r_x = b$$

$$\eta_{e1} = V_P \frac{\partial C_1}{\partial x} = V_P \frac{C_{o1}}{d_1} \quad \eta_{e2} = V_P \frac{\partial C_2}{\partial x} = V_P \frac{C_{o2}}{d_2}$$

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### Condensed Equiv. Circuit (Symmetrical)

If  $\eta_{e1} = \eta_{e2}$ , then ...

where

$$\begin{cases} L_x = \frac{m}{\eta_e^2} \\ C_x = \frac{\eta_e^2}{k} \\ R_x = \frac{b}{\eta_e^2} \end{cases}$$

Holds for the symmetrical case, where port 1 and port 2 are identical

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### Phasings of Signals

- Below: plots of resonance electrical and mechanical signals vs. time, showing the phasings between them

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