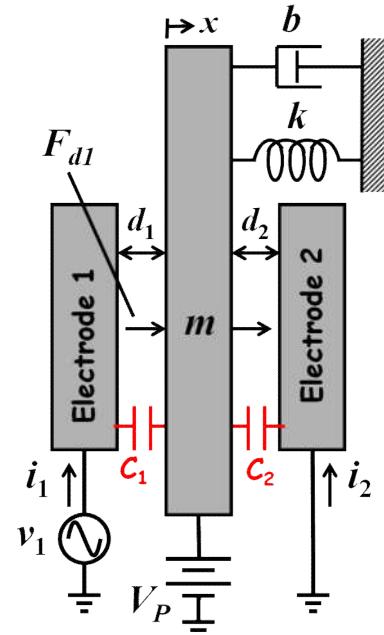


Lecture 23: Mechanical Circuit Analysis & Gyros

- Announcements:
  - Module 13 on Equivalent Circuits II online
  - Module 14 on Sensing Circuits online
  - Module 15 on Gyros, Noise, & MDS
  - HW#6 online and due Tuesday, April 17
  - Project Slide Set #1 due Friday, April 13
  - You do need to be in groups of three
    - ↳ Not 2, not 4
    - ↳ Part of the learning experience in any project entails how to work with others
  - Next Tuesday, April 17, Eta Kappa Nu will show up at the beginning of class to guide you through course evaluations
    - ↳ Bring your computer
- -----
- Reading: Senturia, Chpt. 6, Chpt. 14
- Lecture Topics:
  - ↳ Input Modeling
    - Force-to-Velocity Equiv. Ckt.
    - Input Equivalent Ckt.
  - ↳ Current Modeling
    - Output Current Into Ground
    - Input Current
  - Complete Electrical-Port Equiv. Ckt.
  - ↳ Impedance & Transfer Functions
- -----

- -----
- Reading: Senturia, Chpt. 14, Chpt. 16, Chpt. 21
- Lecture Topics:
  - ↳ **Gyroscopes**
- Reading: Senturia, Chpt. 14
- Lecture Topics:
  - ↳ **Detection Circuits**
    - Velocity Sensing
    - Position Sensing
- -----
- Last Time:
  - In the midst of generating the full equivalent circuit

[Input Current Expression]  $\Rightarrow$  then yields complete equivalent input ckt



Get  $I_1(j\omega)$ :

$$i_1(t) = C_1(x, t) \frac{dv_1(t)}{dt} + V_1(t) \frac{dC_1(x, t)}{dx}$$

$$(V_1(t) = N_1 - V_p) \Rightarrow i_1 = C_1 \frac{dv_1}{dt} + (N_1 - V_p) \frac{\partial C_1}{\partial x} \frac{\partial x}{\partial t}$$

↑  
f(t)                      magnitude of  $N_1 = v_1$  for wt

$$\therefore I_1(j\omega) = j\omega C_1 N_1 + j\omega V_p \frac{\partial C_1}{\partial x} X - j\omega V_p \frac{\partial C_1}{\partial x} X$$

Feedthrough Current                  Motional Current due to mass motion

$$@DC: x = \frac{F_{d1}}{k} = -\frac{1}{k} V_p \frac{\partial C_1}{\partial x} N_1$$

$$@resonance: x = \frac{(Q F_d)}{k} = -\frac{Q}{jk} V_p \frac{\partial C_1}{\partial x} N_1 = X$$

Then: ( $\omega_0$   $\rightarrow$  resonance)  $\uparrow$   
 $90^\circ$  phase lag  $\uparrow$   
 $\downarrow$   $\text{ne}_1$   $\downarrow$  plug in to  $*'$

$$I_1(j\omega) = j\omega_0 C_1 |N_1| + j\omega_0 (V_p \frac{\partial C_1}{\partial x})^2 \frac{Q}{jk} |N_1|$$

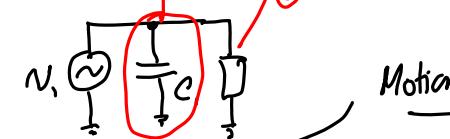
$$= j\omega_0 C_1 |N_1| + \frac{Q}{k} \omega_0^2 \text{ne}_1 |N_1|$$

$90^\circ$  phase-shifted  
from  $N_1$   
 $\downarrow$

This is a capacitor  
in shunt w/ input!

@resonance

In phase  
w/  $N_1$ !  
 $\uparrow$   
This is an effective resistance @  $\omega_0$   
seen looking into the electrode!

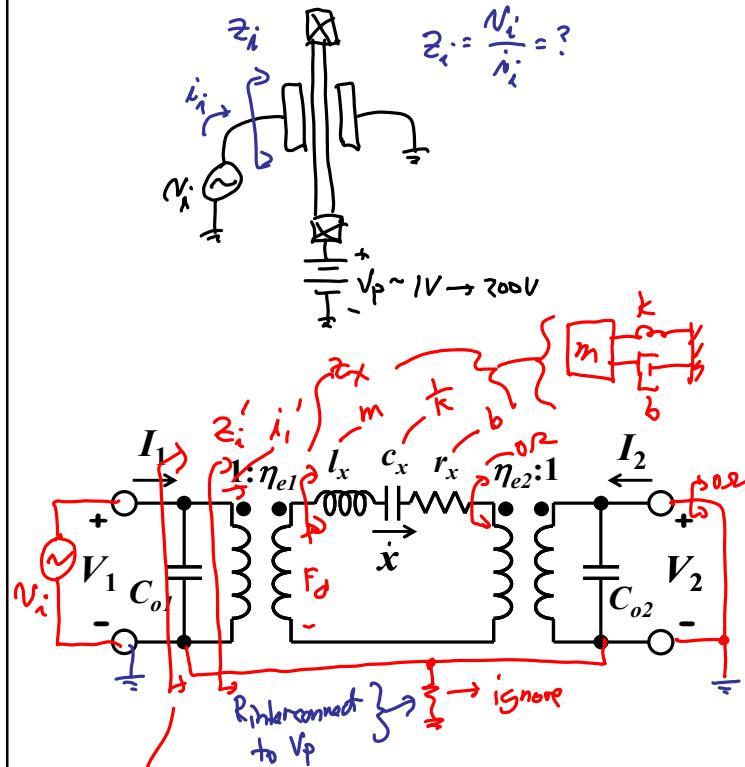


Motional Resistance:

$$R_{x1} = \frac{V_1}{I_1} = \frac{k}{\omega_0 Q \text{ne}_1} = \frac{m \omega_0}{Q \text{ne}_1} = \frac{b}{\eta^2 \text{ne}_1} : R_{x1}$$

The equivalent ckt. better  
got this right!

Input Impedance Into Port 1



$$z_i = \frac{V_i}{i_i} = C_{o1} \parallel z_i'$$

$$z_i' = \frac{V_i}{i_i'} = \frac{e_i}{f_i}$$

$$\begin{bmatrix} e_2 \\ f_2 \end{bmatrix} = \begin{bmatrix} \eta & 0 \\ 0 & -\frac{1}{\eta} \end{bmatrix} \begin{bmatrix} e_1 \\ f_1 \end{bmatrix} \Rightarrow \begin{cases} e_2 = \eta e_1 \rightarrow e_1 = \frac{e_2}{\eta} \\ f_2 = -\frac{1}{\eta} f_1 \rightarrow f_1 = -\eta f_2 \end{cases}$$

$$\frac{e_1}{f_1} = z_i' = \frac{e_2}{\eta} \left( \frac{1}{-\eta f_2} \right) = -\frac{1}{\eta^2} \frac{e_2}{f_2} \rightarrow \frac{V_i}{i_i'} = z_i' = \frac{1}{\eta^2} \frac{e_2}{f_2}$$

$$z_i' = \frac{1}{\eta^2} z_x$$

$$z_i' = \frac{1}{h_{el}^2} \left( j\omega h_x + \frac{1}{j\omega C_x} + R_x \right)$$

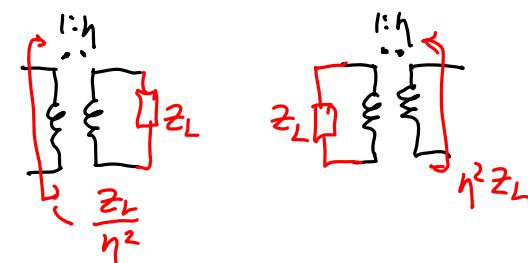
$$= j\omega \left( \frac{h_x}{h_{el}^2} \right) + \frac{1}{j\omega (h_{el}^2 C_x)} + \frac{R_x}{h_{el}^2}$$

$$\left. \begin{array}{l} L_{x1} \\ C_{x1} \\ R_{x1} \end{array} \right\} \text{model noise!}$$

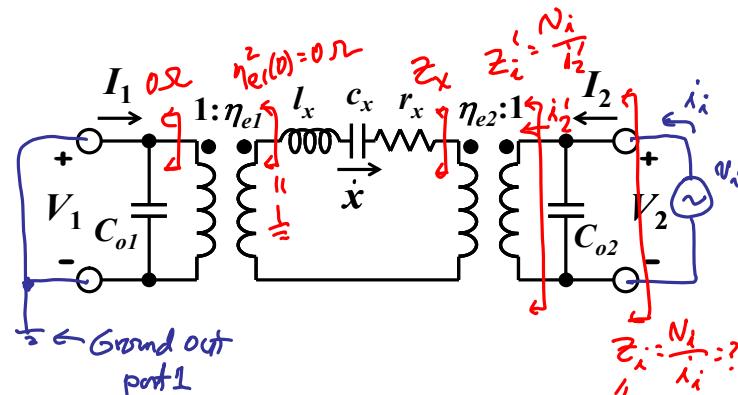
$$h_{el}^2 = 4KTR_x$$

Purely Electrical Equiv. Ckt.

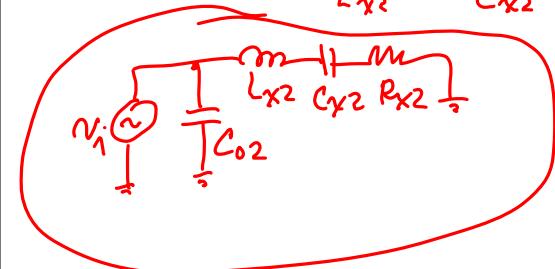
Xformer Inspection Analysis



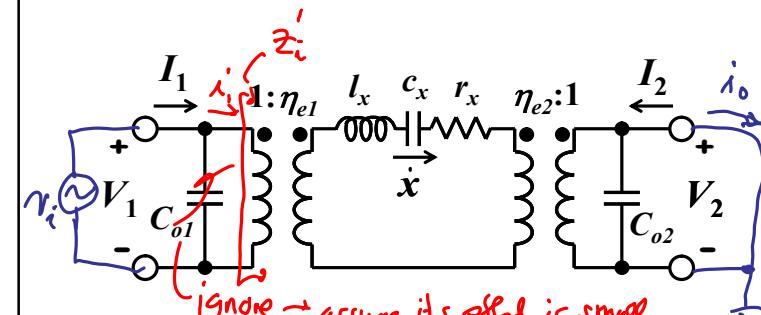
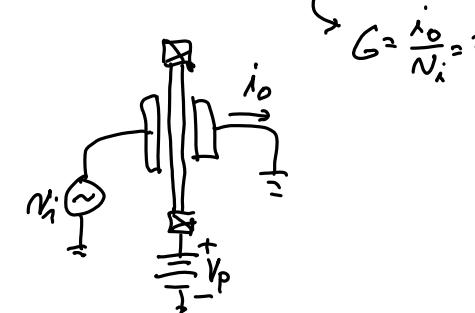
Input Impedance Into Port 2



$$Z_i' = \frac{N_i}{i_2} = \frac{Z_x}{\eta_{e2}^2} = j\omega \left( \frac{l_x}{\eta_{e2}} \right) + \frac{1}{j\omega (\eta_{e2}^2 C_x)} + \frac{r_x}{\eta_{e2}^2}$$



Port 1-to- Port 2 Transconductance



$$\begin{aligned} \dot{x} &= \frac{1}{\eta_{e1}} \dot{i}_1 \\ \dot{i}_o &= \eta_{e2} \dot{x} \end{aligned} \quad \left. \begin{aligned} \dot{i}_o &= \frac{\eta_{e2}}{\eta_{e1}} \dot{i}_1 = \frac{\eta_{e2}}{\eta_{e1}} \left( \frac{N_i}{Z_i'} \right) \end{aligned} \right\}$$

$$\dot{i}_o = \frac{\eta_{e2}}{\eta_{e1}} \left[ \eta_{e1}^2 \frac{1}{j\omega l_x + \frac{1}{j\omega C_x + r_x}} \right] N_i$$

$$\therefore \frac{i_o}{N_i}(j\omega) = \frac{N_{e1}N_{e2}}{j\omega L_x + \frac{1}{j\omega C_x} + R_x}$$

$$\frac{i_o}{N_i}(j\omega) = \left[ j\omega L_{x12} + \frac{1}{j\omega C_{x12}} + R_{x12} \right]^{-1}$$

where  $L_{x12} = \frac{R_x}{N_{e1}N_{e2}}$ ,  $C_{x12} = N_{e1}N_{e2}C_x$

$$R_{x12} = \frac{R_x}{N_{e1}N_{e2}}$$

→ separate into magnitude & freq. response components

$$\frac{i_o}{N_i}(s) = \frac{1}{sL_x + \frac{1}{sC_x} + R_x} = \frac{s(\frac{1}{L_x})}{s^2 + \frac{1}{L_x C_x} + s(\frac{R_x}{L_x})}$$

$$\left( \frac{1}{L_x C_x} = \omega_0^2, Q = \frac{\omega_0 L_x}{R_x} \rightarrow \frac{R_x}{L_x} = \frac{\omega_0}{Q} \right)$$

$$\frac{i_o}{N_i}(s) = \frac{1}{R_x} \frac{s(\frac{\omega_0}{Q})}{s^2 + s(\frac{\omega_0}{Q}) + \omega_0^2} = \frac{1}{R_x} H(s)$$

Gain Term      Frequency Shaping Term      Resonance magnitude

