

EECS 42 Introduction Digital Electronics

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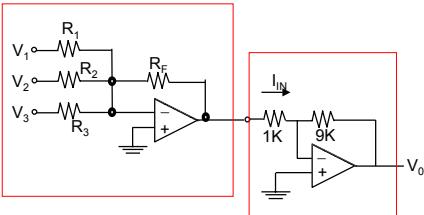
Lecture # 15 Op-Amp Circuits and Comparators 4.3-4.4 (light on non-ideal)

- A) Cascade Op-Amps
- B) Integration/Differentiation Op-Amps
- C) I vs. V of Op-Amps – Source Limits
- D) Comparator Circuits
- E) D to A Converters

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CASCADE OP-AMP CIRCUITS



How do you get started on finding V_0 ?

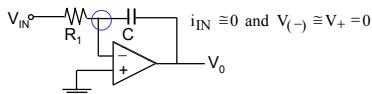
Hint: Identify Stages

Hint: I_{IN} does not affect V_{O1}

See the further examples of op-amp circuits in the reader

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INTEGRATING OP-AMP



How do you get started on finding V_0 ?

Hint: $i_{IN} \approx 0$ and $V_{(-)} \approx V_+ = 0$

Hint: KCL at V_- node with $I_{IN} = 0$

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NEGATIVE FEEDBACK

Familiar examples of negative feedback:

Thermostat controlling room temperature

Driver controlling direction of automobile

Photochromic lenses in eyeglasses

Fundamentally pushes toward stability

Familiar examples of positive feedback:

Microphone "squawk" in room sound system

Mechanical bi-stability in light switches

Thermonuclear reaction in H-bomb

Fundamentally pushes toward instability or bi-stability

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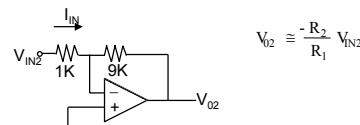
CASCADE OP-AMP SOLUTION

FIRST STAGE IS "SUMMING JUNCTION" AMPLIFIER

Solution:

$$\begin{aligned} i_{IN} &\approx 0 \text{ and } V_{(-)} \approx V_+ = 0 \\ \text{KCL: } \frac{V_1}{R_1} + \frac{V_2}{R_2} + \frac{V_3}{R_3} - \frac{V_0}{R_F} &= 0 \\ \Rightarrow V_{01} &= -\frac{R_F}{R_1} V_1 - \frac{R_F}{R_2} V_2 - \frac{R_F}{R_3} V_3 \end{aligned}$$

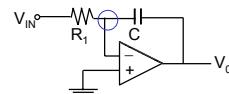
SECOND STAGE IS "INVERTING" AMPLIFIER



$$V_{02} \approx -\frac{R_2}{R_1} V_{IN2}$$

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INTEGRATING OP-AMP



$$\frac{0 - V_{IN}}{R_1} + C \frac{\partial(0 - V_O)}{\partial t} = 0$$

Integrate from t_0 to t to get $V_O(t)$

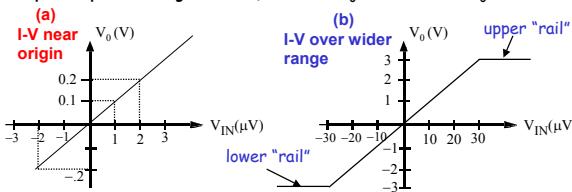
$$V_O(t) = \frac{-1}{R_1 C} \int_{t_0}^t V_{IN}(t) dt'$$

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OP-AMP I-V CHARACTERISTICS WITH RAILS

- Circuit model (ideal op-amp) gives the essential linear part
- But V_0 cannot rise above some physical voltage related to the positive power supply V_{CC} ("upper rail") $V_0 < V_{+RAIL}$
- And V_0 cannot go below most negative power supply, V_{EE} i.e., limited by lower "rail" $V_0 > V_{-RAIL}$

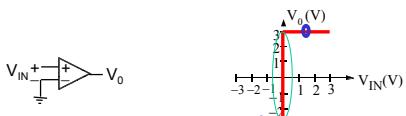
Example: Amplifier with gain of 10^5 , with max V_0 of 3V and min V_0 of -3V.



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SIMPLE A/D CONVERTER

I-V with equal X and Y axes



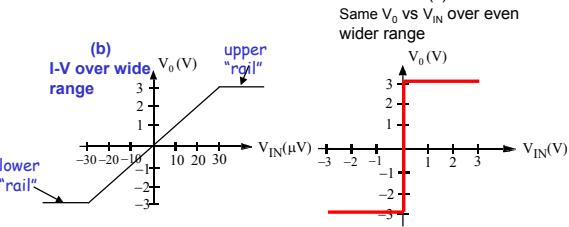
Note:

- (a) displays **linear amplifier behavior** ($|V_{IN}| < 30 \mu V$) and stops at rails
- (b) shows **comparator decision function** (1 bit A/D converter centered at $V_{IN} = 0$) where lower rail = logic "0" and upper rail = logic "1"

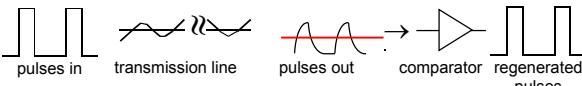
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OP-AMP I-V CHARACTERISTICS WITH RAILS (cont.)

Example: Amplifier with gain of 10^5 , with upper rail of 3V and lower rail of -3V. We plot the V_0 vs V_{IN} characteristics on two different scales

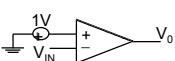


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ONE-BIT A/D CONVERSION REQUIRED IN DIGITAL SYSTEMS

As we saw, we set comparator **threshold** at a suitable value (e.g., halfway between rails) and comparator output goes to +rail if $V_{IN} > V_{THRESHOLD}$ and to -rail if $V_{IN} < V_{THRESHOLD}$.

What would this circuit do?

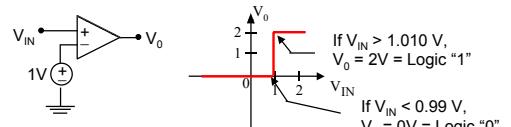


The **inverse** pulse shaped function is generated by applying the input voltage to V- and setting V+ to the threshold voltage.

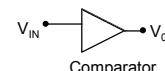
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OP-AMP USE AS COMPARATOR (A/D) MODE

Simple comparator with threshold at 1V. Design lower rail at 0V and upper rail at 2V (logic "1"). A = large (e.g. 10^2 to 10^5)



NOTE: The actual diagram of a comparator would not show an amplifier with "offset" power supply as above. It would be a simple triangle, perhaps with the threshold level (here 1V) specified.

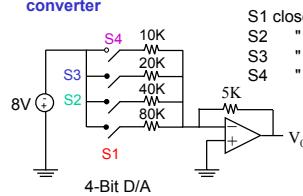


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D/A CONVERSION

Example: Digital representation of sound to analog (so you can hear it!) → D/A conversion

The summing junction op-amp provides a simple means of D/A conversion via **weighted-adder D/A converter**



4-Bit D/A

Another way (not shown) is to sum charges instead of current with capacitor networks

Binary number	Analog output (volts)
0 0 0 0	0
0 0 0 1	1
0 0 1 1	1.5
0 1 0 0	2
0 1 0 1	2.5
0 1 1 0	3
0 1 1 1	3.5
1 0 0 0	4
1 0 0 1	4.5
1 0 1 0	5
1 0 1 1	5.5
1 1 0 0	6
1 1 0 1	6.5
1 1 1 0	7
1 1 1 1	7.5

MSB LSB

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CHARACTERISTIC OF A 4-BIT DAC