

② a) I decided to use the following design flow:

Synthesize w/ Matlab in  
z-domain



Break up into Biquads  
using "tf2sos"



estimate Biquad  $Q$ s,  
use inverse bilin.  
transform

$$s_p = \frac{2}{T} \left( \frac{z_p - 1}{z_p + 1} \right)$$

$$\Rightarrow \left. \begin{array}{l} Q_1 = 0.61 \\ Q_2 = 0.16 \\ Q_3 = 6.3 \end{array} \right\} \begin{array}{l} \text{"Low Q"} \\ \text{"Hi Q"} \end{array} \quad (>3)$$

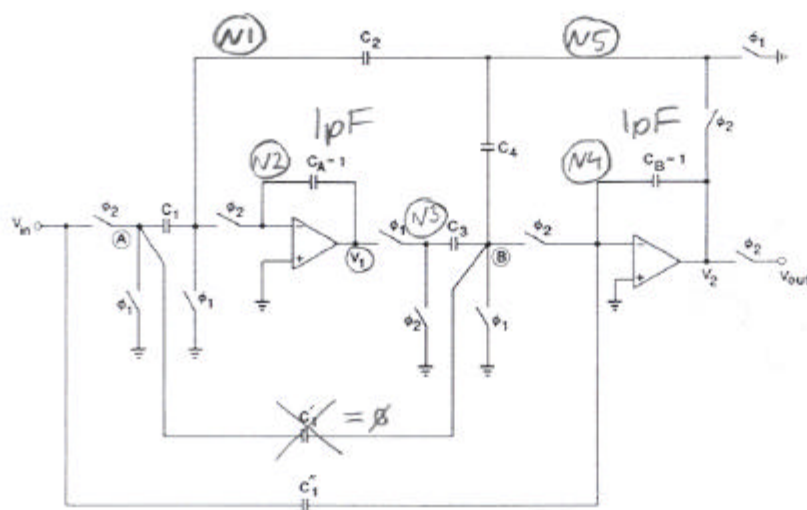
Implement using Hi- and  
Lo Q circuits as given in  
Gregorian & Temes pp. 280.

→ see attached schematics  
with design equations

$$F_{pass} = 2 \cdot \frac{1 \text{ MHz}}{10 \text{ MHz}} = 0.2$$

$$F_{stop} = 2 \cdot \frac{1.5 \text{ MHz}}{10 \text{ MHz}} = 0.3$$

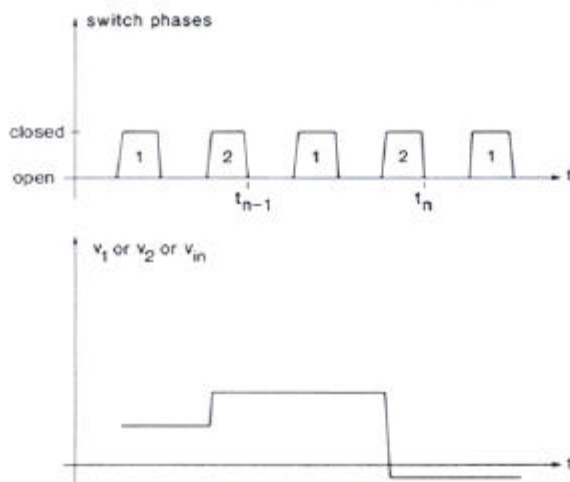
$$\Rightarrow \text{ellipord}(\dots) = \underline{\underline{6}} \quad (\text{filter order})$$



(c)

LoQ

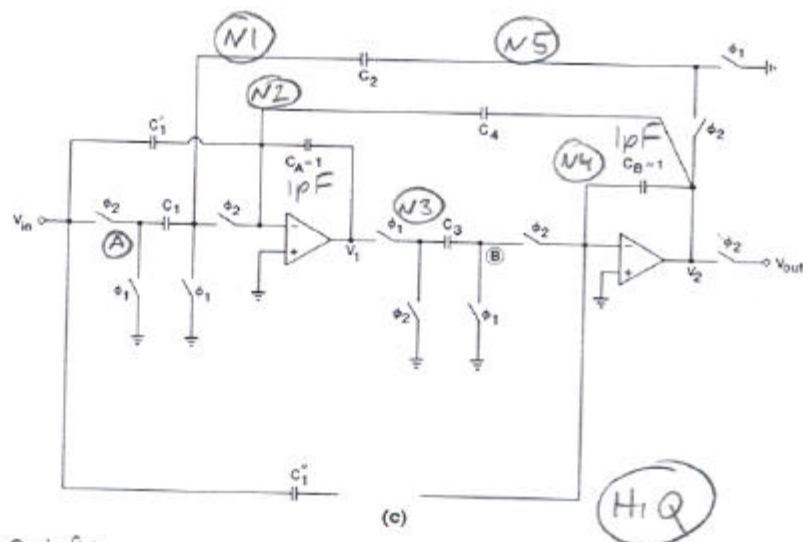
$$H(z) = \frac{a_2 z^2 + a_1 z + a_0}{b_2 z^2 + b_1 z + b_0}$$



(d)

FIGURE 5.10. continued.

$$\begin{aligned} C_1'' &= a_0 \\ C_1' &= a_2 - a_0 \\ C_1 &= \frac{a_0 + a_1 + a_2}{C_3} \\ C_4 &= b_2 - 1 \\ C_2 C_3 &= b_1 + b_2 + 1 \end{aligned}$$



$$H(z) = \frac{a_2 z^2 + a_1 z + a_0}{b_2 z^2 + b_1 z + b_0}$$

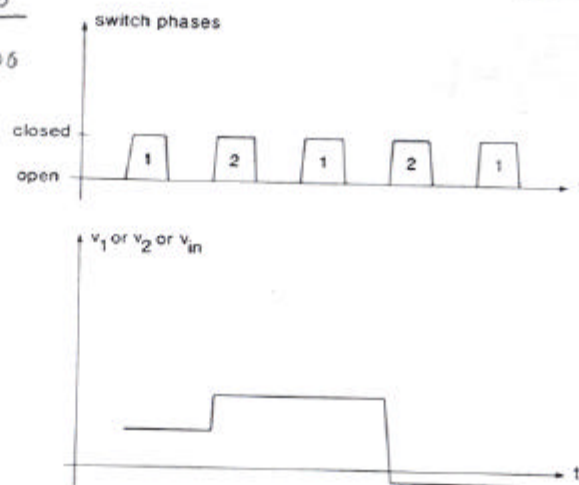
$$C_1^{II} = \frac{a_2}{b_2}$$

$$C_1^I = \frac{a_2 - a_0}{b_2 C_3}$$

$$C_1 = \frac{a_0 + a_1 + a_2}{b_2 C_3}$$

$$C_4 = \frac{1 - 1/b_2}{C_3}$$

$$C_2 C_3 = \frac{1 + b_1 + b_2}{b_2}$$



(d)

FIGURE 5.13. continued.

2) b) Poles and Zeros

-----  
Biquad 1:

Z1 =  
-4.959483412658553e-001 +8.683520270003670e-001i  
-4.959483412658553e-001 -8.683520270003670e-001i  
P1 =  
7.127856685302593e-001 +1.653876720637911e-001i  
7.127856685302593e-001 -1.653876720637911e-001i

Biquad 2:

Z2 =  
3.825418155817085e-001 +9.239381793883470e-001i  
3.825418155817085e-001 -9.239381793883470e-001i  
P2 =  
7.246789558726074e-001 +4.320114746276476e-001i  
7.246789558726074e-001 -4.320114746276476e-001i

Biquad 3:

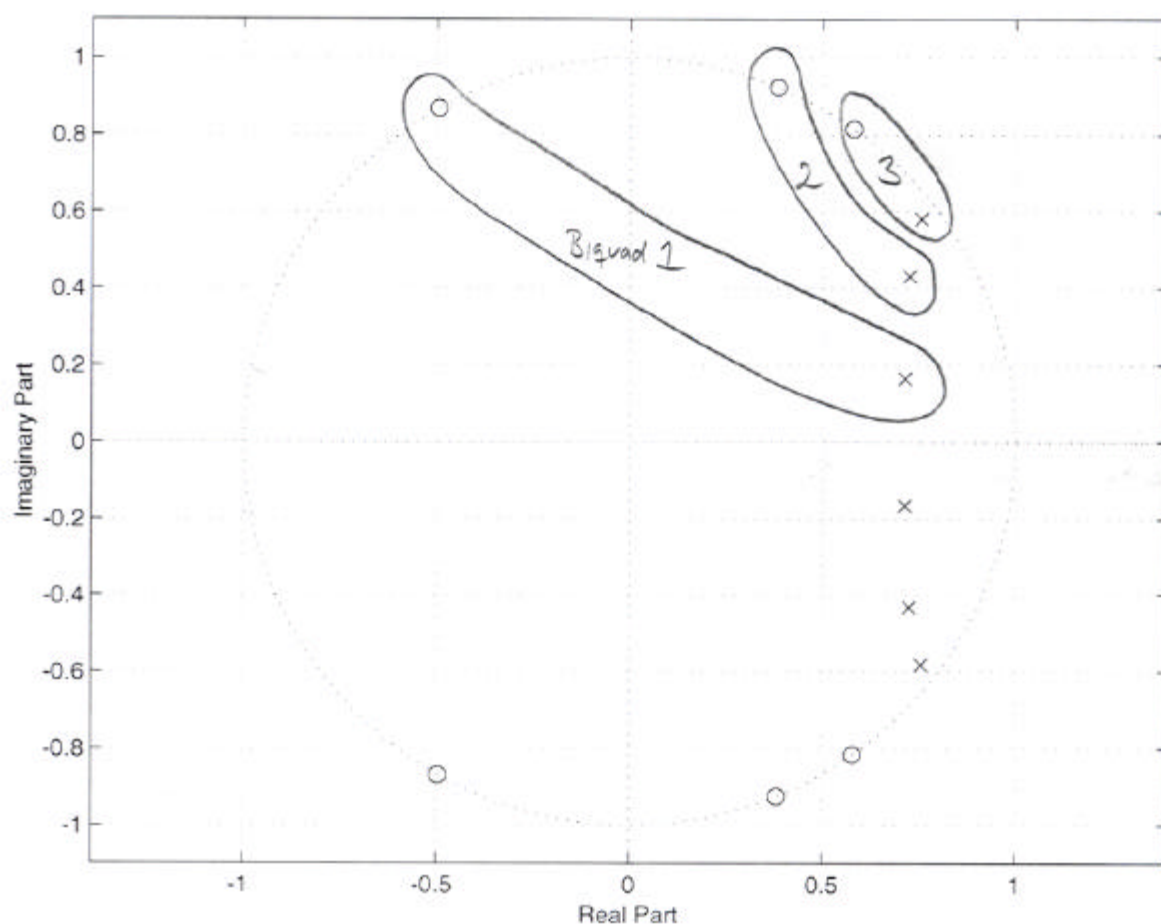
Z3 =  
5.785179590177820e-001 +8.156696458088302e-001i  
5.785179590177820e-001 -8.156696458088302e-001i  
P3 =  
7.553705109318878e-001 +5.808504547226241e-001i  
7.553705109318878e-001 -5.808504547226241e-001i

(Grouping done by Matlab "tf2sos"  
→ see next page)

② b

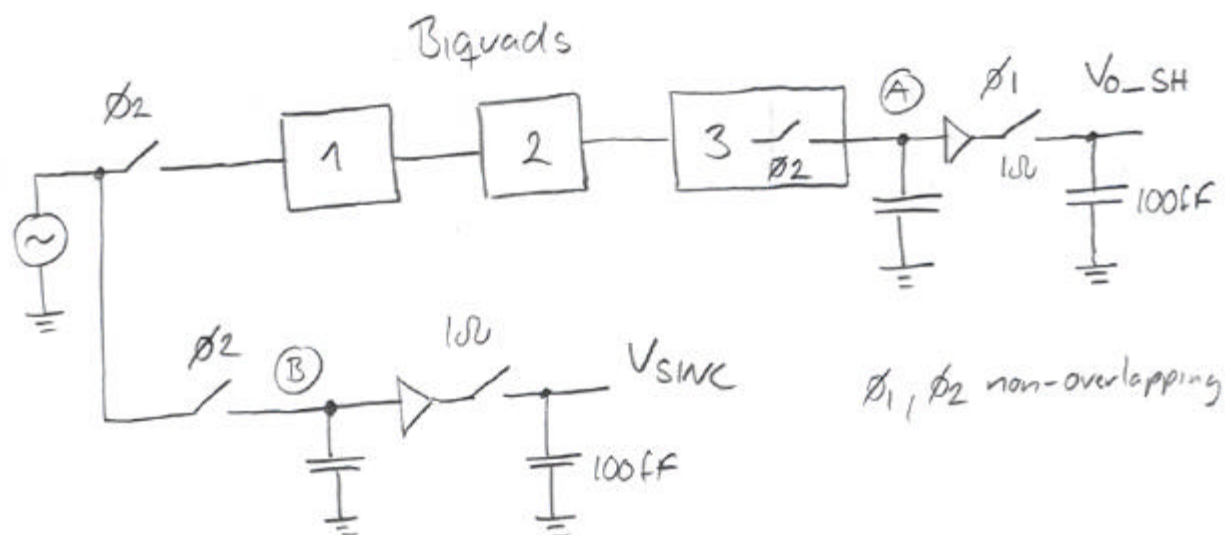
p/z grouping by Matlab:

(<sup>u</sup>Minimum sensitivity to coefficient tolerances,  
see e.g. Oppenheim & Schaffer)





② ③ Top Schematic:



→ (A), (B) are "track and hold nodes", the held value is transferred with  $\phi_1$  to form an "ideal" staircase voltage  $\equiv$  Dirac sampling with falling edge of  $\phi_2$

→  $V_{SINC}$  will be used in the postprocessor to eliminate sinc distortion.

$$V_{O-SH}[dB] - V_{SINC}[dB] \Rightarrow \text{actual discrete time response}$$

→ Biquads scaled in Matlab using:

$$\frac{\sum a}{\sum b} \stackrel{!}{=} 1 \quad \text{in} \quad \frac{a_2 z^2 + a_1 z + a_0}{b_2 z^2 + b_1 z + b_0} \quad z \rightarrow 1 @ DC$$

(see Matlab script)

→ Good agreement w/ Spectre simulation, see Plots 1, 2

② d) found:  $A_{\min} \approx \underline{\underline{1200}}$  for  $R_p = 0.12 \text{ dB}$ ,  
assuming  $f_{\text{corner}}$  is allowed to shift  
 $\approx \frac{8 \text{ kHz}}{1 \text{ MHz}} = 0.8 \%$

→ Stopband attenuation appears to be  
insensitive to low gain / gain variations.

→ See plots 3, 4

---

h: (wvew12s11() - wvew12s112())

h: harmonic="0";dB20((v "vsinc" ?result "pac2-pac" ?resultsD)

300m

↑  $|H(f)|$  [dB]

PASSBAND RESPONSE

(Gain =  $10^6$ )

Plot 1

SPEC  $\leq 0.1$ dB

sinc corrected response

sinc distorted response

Sinc

(dB)

300m  
200m  
100m  
0.00  
-100m  
-200m  
-300m

0.00

200K

400K

600K

800K

1.00M

1.20M

freq (Hz)

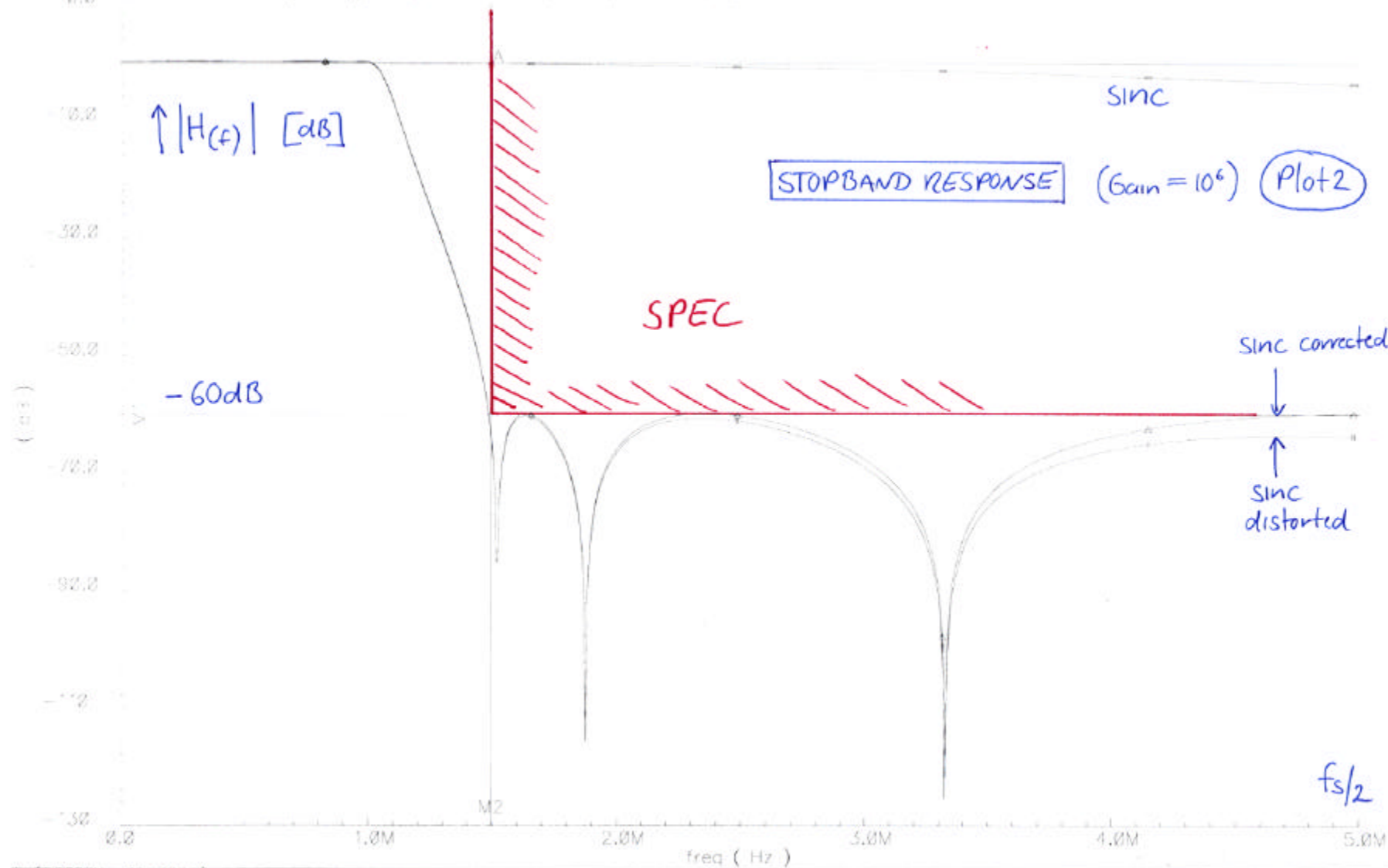
A: (\* 2.37584)



wt: (wvew12s1i1() - wvew12s1i2())

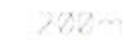
wt: harmonic="0";dB20((v "vsinc" ?result "pac2-pac" ?results))

wt: harmonic="0";dB20((v "vo\_sh" ?result "pac2-pac" ?results))



At (-50027V - 324.187m)

3227



2007

2.92

227

–200m

-322m

—400m

A: (299.272K - 59.3953m)      delta: (501.808K - 121.808m)  
B: (900.82K - 62.4126m)      slope: -202.403m

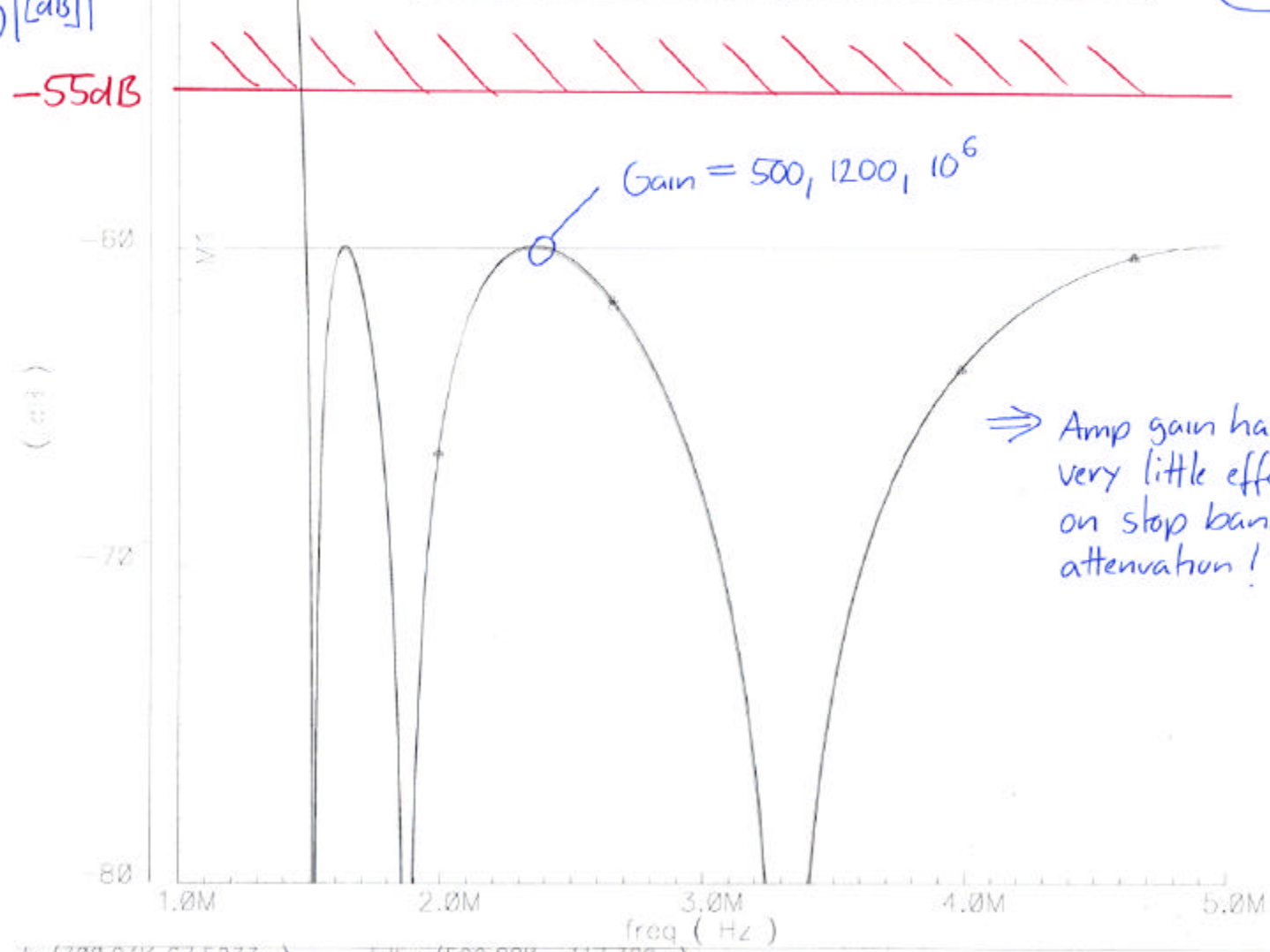
Δ: harmonic="0";(dB20((v "vo\_sh" ?r-: harmonic="0";(dB20((v "vo\_sh" ?r  
 #: harmonic="0";(dB20((v "vo\_sh" ?r

$|H(f)| [dB] \uparrow$

-55dB

STOPBAND RESPONSE W/ DIFFERENT AMP GAIN

Plot 4



A: (300.94K 67.5233m) delta: (599.88K -117.389m)  
 B: (900.82K -49.8657m) slope: -195.687n

% EE247 Homework #4 Boris Murmann

Matlab Script

①

% filter specification

```
clear;
Wpass = 2*pi*1e6;      % passband edge [rad]
Wstop = 2*pi*1.5e6;    % start of stopband [rad]
Rp = 0.1;              % passband ripple (dB)
Rs = 60;               % stopband attenuation (dB)
```

% sampling frequency and normalized band specifications

```
Wsamp = 2*pi*10e6;
Wp = 2*Wpass/Wsamp;
Ws = 2*Wstop/Wsamp;
```

% discrete time filter synthesis

```
[N,Wn] = ellipord(Wp, Ws, Rp, Rs);
[num,den] = ellip(N, Rp, Rs, Wn);
freqz(num,den);
```

% break up into biquads - matlab provides optimized pole zero grouping

```
[SOS,G] = tf2sos(num,den)
```

```
num1 = SOS(1,1:3);
num2 = SOS(2,1:3);
num3 = SOS(3,1:3);
den1 = SOS(1,4:6);
den2 = SOS(2,4:6);
den3 = SOS(3,4:6);
```

% calculate Q of continous time equivalent sections

```
[Z1, P1, K1] = tf2zp( num1, den1 );
[Z2, P2, K2] = tf2zp( num2, den2 );
[Z3, P3, K3] = tf2zp( num3, den3 );
s1 = (Wsamp/pi) * (P1(1)-1) / (P1(1)+1);
s2 = (Wsamp/pi) * (P2(1)-1) / (P2(1)+1);
s3 = (Wsamp/pi) * (P3(1)-1) / (P3(1)+1);
Q1 = 0.5*sqrt(1+(imag(s1)/real(s1))^2);
Q2 = 0.5*sqrt(1+(imag(s2)/real(s2))^2);
Q3 = 0.5*sqrt(1+(imag(s3)/real(s3))^2);
```

% normalize to positive powers of z and divide by "a2" to match (5.27) Gregorian

```
NUM1 = num1 / den1(3);
NUM2 = num2 / den2(3);
NUM3 = num3 / den3(3);
DEN1 = den1 / den1(3);
DEN2 = den2 / den2(3);
DEN3 = den3 / den3(3);
```

% scale for unity gain at DC (z=1) for each biquad

```
k1 = sum(DEN1) / sum(NUM1);
k2 = sum(DEN2) / sum(NUM2);
k3 = sum(DEN3) / sum(NUM3);
NUM1 = k1 * NUM1;
NUM2 = k2 * NUM2;
NUM3 = k3 * NUM3;
```

% calculate component values, for Q>3 use hi Q circuit

```
C0=1e-12;
if Q1>3
    Caps1 = HiQ(NUM1, DEN1, C0)
else
    Caps1 = LoQ(NUM1, DEN1, C0)
end
if Q2>3
    Caps2 = HiQ(NUM2, DEN2, C0)
else
```

```
    Caps2 = LoQ(NUM2, DEN2, C0)
end
if Q3>3
    Caps3 = HiQ(NUM3, DEN3, C0)
else
    Caps3 = LoQ(NUM3, DEN3, C0)
end
```

②



```
% Lo Q (Q<3) Gregorian p.286
function Caps = LoQ(NUM, DEN, C0)
a2 = NUM(1);
a1 = NUM(2);
a0 = NUM(3);
b2 = DEN(1);
b1 = DEN(2);
b0 = DEN(3);
C2 = sqrt(b1+b2+1);
C3 = sqrt(b1+b2+1);
C1x = a2-a0;
C1xx = a0;
C1 = (a0+a1+a2)/C3;
C4 = b2-1;
Caps = C0*[C1,C1x,C1xx,C2,C3,C4];
```

```
% Hi Q (Q>3) Gregorian p.291
function Caps = HiQ(NUM, DEN, C0)
a2 = NUM(1);
a1 = NUM(2);
a0 = NUM(3);
b2 = DEN(1);
b1 = DEN(2);
b0 = DEN(3);
C2 = sqrt( (1+b1+b2)/b2 );
C3 = sqrt( (1+b1+b2)/b2 );
C1x = (a2-a0) / (b2*C3);
C1xx = a2/b2;
C1 = (a0+a1+a2)/ (b2*C3);
C4 = (1 - (1/b2)) / C3;
Caps = C0*[C1,C1x,C1xx,C2,C3,C4];
```

Spectre Netlist

①

\*\*\* EE247 HW #4 Boris Murmann

```
simulator lang=spectre
parameters fs=10Meg c0=1p Ron=1 AmpGain=1e6
//
*** first low Q section
+c11= 4.529440715720658e-013
+c1x1= 4.163336342344337e-029 // =zero
+c1xx1= 6.857132907366235e-014
+c21= 4.529440715720658e-013
+c31= 4.529440715720658e-013
+c41= 8.677048917763188e-013
//
*** second low Q section
+c12= 6.072036269121990e-013
+c1x2= -8.881784197001252e-028 // =zero
+c1xx2= 2.985596869872886e-013
+c22= 6.072036269121991e-013
+c32= 6.072036269121991e-013
+c42= 4.049018365178336e-013
//
*** hi Q section
+c13= 6.302625148882051e-013
+c1x3= -2.079239384327061e-027 // =zero
+c1xx3= 4.712310360217251e-013
+c23= 6.302625148882049e-013
+c33= 6.302625148882049e-013
+c43= 1.460155701618134e-013

*** stimulus
vin1 (vin 0) vsource pacmag=1
//vin1 (vin 0) vsource type=sine ampl=1 freq=fs/10
vphi1 (phi1 0) vsource type=pulse val0=0 val1=1 period=1/fs
width=0.45/fs rise=.1n fall=.1n delay=0.05/fs
vphi2 (phi2 0) vsource type=pulse val0=0 val1=1 period=1/fs
width=0.45/fs rise=.1n fall=.1n delay=0.55/fs

*** circuit
b1 (vin vo1 phi1 phi2) loq pc0=c0 pc1=c11 pc1xx=c1xx1 pc2=c21
pc3=c31 pc4=c41
b2 (vo1 vo2 phi1 phi2) loq pc0=c0 pc1=c12 pc1xx=c1xx2 pc2=c22
pc3=c32 pc4=c42
b3 (vo2 vout phi1 phi2) hiq pc0=c0 pc1=c13 pc1xx=c1xx3 pc2=c23
pc3=c33 pc4=c43

*** sample and hold output
c1 (vout 0) capacitor c=100f
buf1 (voutbuf 0 vout 0) vcvs gain=1
s2 (voutbuf vo_sh phi1 0) relay vt1=0 vt2=1 rclosed=1
ropen=100e9
c2 (vo_sh 0) capacitor c=100f

*** sample and hold input (-> sinc response for correction)
s11 (vin vth phi2 0) relay vt1=0 vt2=1 rclosed=1
ropen=100e9
c11 (vth 0) capacitor c=100f
buf11 (vthbuf 0 vth 0) vcvs gain=1
s21 (vthbuf vsinc phi1 0) relay vt1=0 vt2=1 rclosed=1
```

②

```

ropen=100e9
c21 (vsinc 0) capacitor c=100f

*****
* Lo Q biquad
subckt loq (vi vos phi1 phi2)
parameters pc0 pc1 pclxx pc2 pc3 pc4
rsim1 (vi n2 phi2 phi1 a n1) rsim_series
Csamp=pc1
rsim2 (v1 n4 phi1 phi2 n3 b) rsim_series_inv
Csamp=pc3
clxx (vi n4) capacitor c=pclxx
c2 (n1 n5) capacitor c=pc2
c4 (n5 b) capacitor c=pc4
switch1 (n5 0 phi1 0) relay vt1=0 vt2=1 rclosed=Ron
ropen=100e9
switch2 (n5 vo phi2 0) relay vt1=0 vt2=1 rclosed=Ron
ropen=100e9
switch3 (vos vo phi2 0) relay vt1=0 vt2=1
rclosed=Ron ropen=100e9
amp1 (v1 0 0 n2) vcvs gain=AmpGain
amp2 (vo 0 0 n4) vcvs gain=AmpGain
ca (n2 v1) capacitor c=pc0
cb (n4 vo) capacitor c=pc0
ends loq

* Hi Q biquad
subckt hiq (vi vos phi1 phi2)
parameters pc0 pc1 pclxx pc2 pc3 pc4
rsim1 (vi n2 phi2 phi1 a n1) rsim_series
Csamp=pc1
rsim2 (v1 n4 phi1 phi2 n3 b) rsim_series_inv
Csamp=pc3
clxx (vi n4) capacitor c=pclxx
c2 (n1 n5) capacitor c=pc2
c4 (n2 vo) capacitor c=pc4
switch1 (n5 0 phi1 0) relay vt1=0 vt2=1 rclosed=Ron
ropen=100e9
switch2 (n5 vo phi2 0) relay vt1=0 vt2=1 rclosed=Ron
ropen=100e9
switch3 (vos vo phi2 0) relay vt1=0 vt2=1
rclosed=Ron ropen=100e9
amp1 (v1 0 0 n2) vcvs gain=AmpGain
amp2 (vo 0 0 n4) vcvs gain=AmpGain
ca (n2 v1) capacitor c=pc0
cb (n4 vo) capacitor c=pc0
ends hiq

* simulated resistor with capacitor in series
subckt rsim_series (v1 vr phi1 phi2 n1 n2)
parameters Csamp
cs (n1 n2) capacitor c=Csamp
s1 (v1 n1 phi1 0) relay vt1=0 vt2=1 rclosed=Ron ropen=100e9
s2 (n1 0 phi2 0) relay vt1=0 vt2=1 rclosed=Ron ropen=100e9
s3 (n2 0 phi2 0) relay vt1=0 vt2=1 rclosed=Ron ropen=100e9
s4 (n2 vr phi1 0) relay vt1=0 vt2=1 rclosed=Ron ropen=100e9
cp1 (n1 0) capacitor c=1f
cp2 (n2 0) capacitor c=1f

```

3

ends rsim\_series

```
* simulated resistor with capacitor in series and charge inversion
subckt rsim_series_inv (v1 vr phi1 phi2 n1 n2)
  parameters Csamp
  cs      (n1 n2)          capacitor c=Csamp
  s1      (v1 n1 phi1 0 ) relay vt1=0 vt2=1 rclosed=Ron ropen=100e9
  s2      (n1 0 phi2 0 ) relay vt1=0 vt2=1 rclosed=Ron ropen=100e9
  s3      (n2 0 phi1 0 ) relay vt1=0 vt2=1 rclosed=Ron ropen=100e9
  s4      (n2 vr phi2 0 ) relay vt1=0 vt2=1 rclosed=Ron ropen=100e9
  cp1     (n1 0)           capacitor c=1f
  cp2     (n2 0)           capacitor c=1f
ends rsim_series_inv
*****
```

```
pss2 pss fund=fs maxacfreq=fs/2 tstab=1/fs/4
pac2 pac start=1k stop=fs/2 lin=1000
*tran1 tran stop=10/fs method=gear2only
```

```
simOptions options
+      rawfmt=psfbin
+      reltol=1e-5
+      vabstol=1n
+      iabstol=1p
```

```
alter1 alter param=AmpGain value=1.2e3
pss3 pss fund=fs maxacfreq=fs/2 tstab=1/fs/4
pac3 pac start=1k stop=fs/2 lin=1000
```

```
alter2 alter param=AmpGain value=0.5e3
pss4 pss fund=fs maxacfreq=fs/2 tstab=1/fs/4
pac4 pac start=1k stop=fs/2 lin=1000
```