

# Optimal FIR Filter Design

## Type I Generalized Linear Phase Filter.

LPF  $H(\omega) = H_m(\omega) e^{j\beta} e^{-j\alpha\omega}$

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$N = \# \text{ of taps is odd.}$   
 $= 2M + 1$

Assume Type I  $\implies \beta = 0$

$$h(n) = h(N - n - 1)$$

$$H_m(\omega) = \sum_{n=0}^M \overline{a(n)} \cos(\omega n) \triangleq G(\omega)$$

$a(0) = h(0)$   
 $a(n) = 2h(M-n)$   
 all other n.

$$H(\omega) = \sum_{n=0}^{N-1} h(n) e^{-j\omega n}$$

$$H(\omega) = G(\omega) e^{-j\alpha\omega}$$

## Observations on $G(w)$

- ①  $G(w)$  is a continuous fn of  $w$  and is as many times differentiable as you want.
- ② How many local extrema does  $G(w)$  have?   
  $\rightarrow$  Global maxima



Express  $\cos(n\omega)$  as sum of powers of  $\cos\omega$ .

$$\cos(2\omega) = 2\cos^2\omega - 1$$

$$\cos(3\omega) = \cos(2\omega + \omega) = \cos 2\omega \cos \omega - \sin 2\omega \sin \omega$$

$$= \cos\omega [2\cos^2\omega - 1]$$

$$- 2\sin^2\omega \cos\omega$$

$$= 2\cos^3\omega - \cos\omega$$

$$- 2\cos\omega [1 - \cos^2\omega]$$

$$= 4\cos^3\omega - 3\cos\omega$$

Generally  $\cos(n\omega)$  as sum of power of  $\cos\omega$

$$\cos(n\omega) = \sum_{i=0}^n \eta_i (\cos\omega)^i$$

$\eta_i$  Chebyshev<sup>n</sup>

$$G(\omega) = \sum_{n=0}^M a(n) \left[ \sum_{i=0}^n \eta_i (\cos \omega)^i \right]$$

$$G(\omega) = \sum_{n=0}^M \delta(n) (\cos \omega)^n$$

$\delta$  depends on  $\eta$  and  $a$

To compute local extrema of  $G(\omega)$

take the derivative and set to zero.

$$\frac{dG(\omega)}{d\omega} = 0 \Rightarrow \sum_{n=0}^M \delta(n) n (\cos \omega)^{n-1} (-\sin \omega)$$

$$= 0 \Rightarrow$$



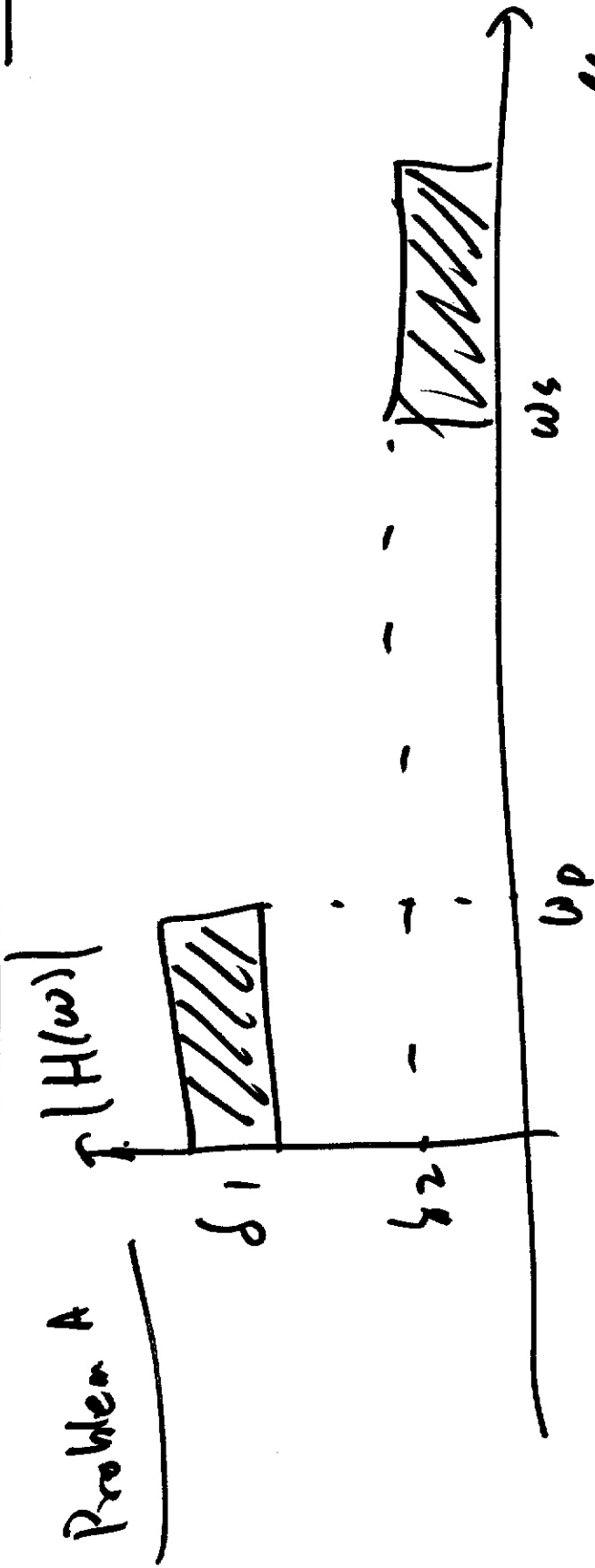
$$\left\{ \begin{array}{l} \sin w = 0 \Rightarrow w = 0, \pi \end{array} \right.$$

$$\sum_{n=0}^M \delta(n) n (\cos w)^{n-1} = 0 \rightarrow \text{Max of } \frac{M-1}{\text{feet.}} \text{ zeros.}$$

Polynomial in  $\cos w = x$

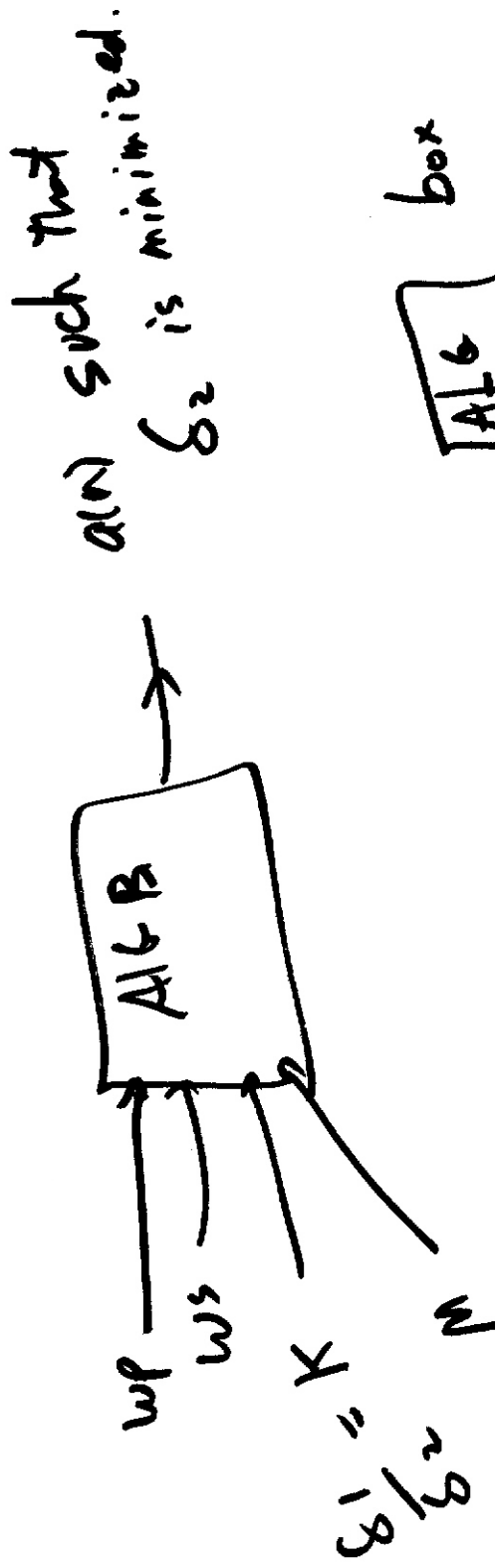
Total # of local extrema for  $\cos w$  is  $(M-1) + 2 = M+1$


# Problem Statement Optimal FIR Filter Design



Given  $\omega_p, \delta_1, \delta_2$ , Determine coeff of  $h(n)$  i.e.  $a(n)$  such that  $M$  is minimized

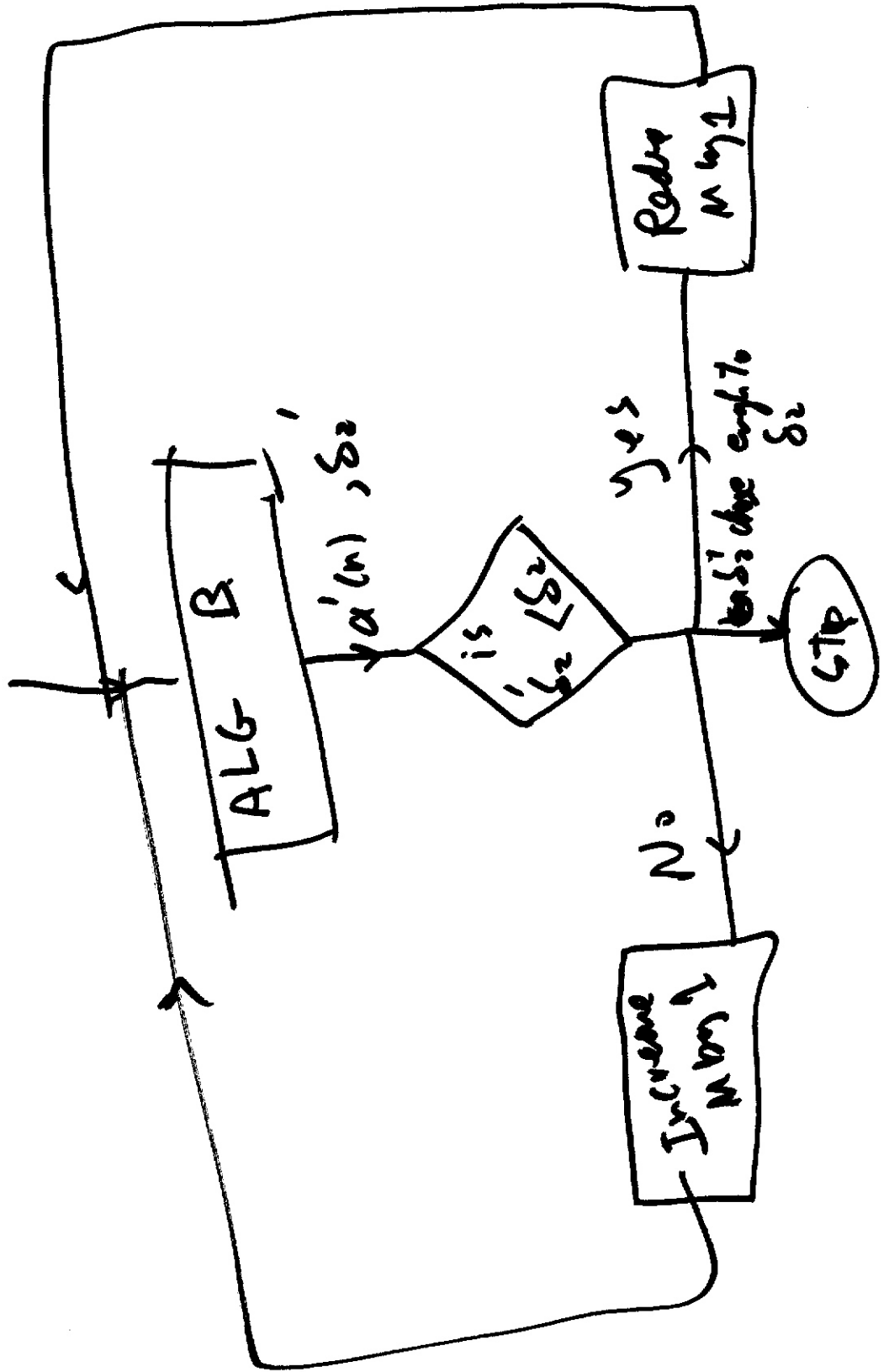
Problem B Given  $w_1, w_2, M, K = \frac{\delta_1}{\delta_2}$   
Determine  $a(n)$  such that  $\delta_2$  is minimized.



Show that if I had a box  then I can use it  
that solves Problem B, then I can use it  
to solve Problem A.

Given  $w, w_s, \delta_1, \delta_2$

compute  $k = \frac{\delta_1}{\delta_2}$ , Guess  $M$



Problem:  $E(\omega) = W(\omega) [G(\omega) - D(\omega)]$

where  $W(\omega) =$  positive weighting function  $\begin{cases} \frac{1}{K} & I_1 \\ 1 & I_2 \end{cases}$

$$G(\omega) = \sum_{n=0}^M a(n) \cos(\omega n)$$

$D(\omega) =$  Desired Freq. Response =  $\begin{cases} 1 & I_1 \\ 0 & I_2 \end{cases}$

Problem: Find  $a(n)$  To minimize  $\text{Max}_{\omega \in F} |E(\omega)|$

where  $F = I_1 \cup I_2$

is a subset of a closed interval  $0 \leq \omega \leq \pi$



$I_1 =$  passband,  $I_2 =$  stopband.

# Optimal FIR Filter Design

Problem C:  $E(\omega) = W(\omega) [G(\omega) - D(\omega)]$

$$G(\omega) = \sum_{n=0}^M \underline{a(n)} \cos(\omega n)$$

$$W(\omega) = \text{positive weighting fn} = \begin{cases} \frac{1}{K} & I_1 \\ 1 & I_2 \end{cases}$$

$$D(\omega) = \text{Desired Freq. Response} = \begin{cases} 1 & I_1 \\ 0 & I_2 \end{cases}$$

$$\text{Max}_{\omega \in F} |E(\omega)|$$

Find  $a(n)$  to minimize.

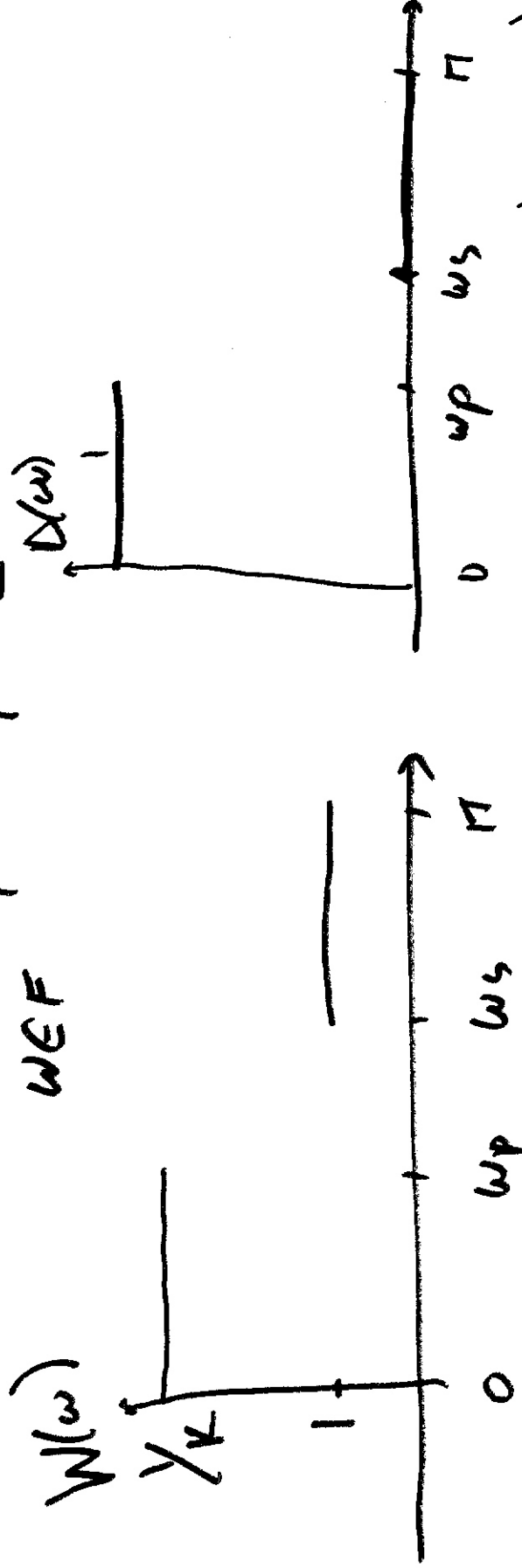
$F \equiv I_1 \cup I_2 = \text{subset of } \omega \text{ interval } [0, \pi] \text{ used in interval optimization}$



Show: Problem C is equivalent to Problem B.

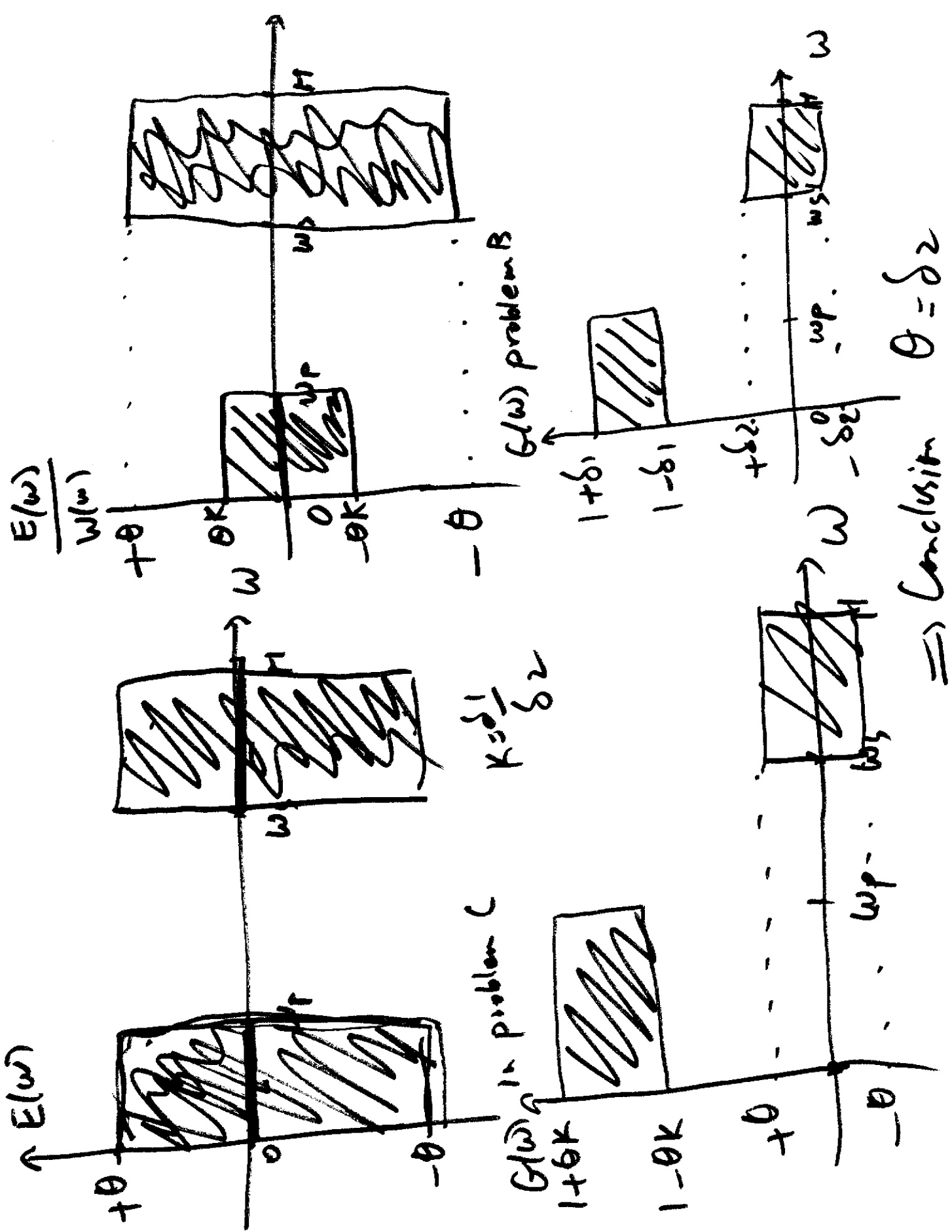
Start to show  $\Theta = \delta_2$

max WEF  $| E(\omega) | \stackrel{??}{=} \delta_2$



$$E(\omega) = W(\omega) [G(\omega) - D(\omega)] \Rightarrow \frac{E(\omega)}{W(\omega)} = G(\omega) - D(\omega)$$

$$\Rightarrow G(\omega) = D(\omega) + \frac{E(\omega)}{W(\omega)}$$





Conclusion :

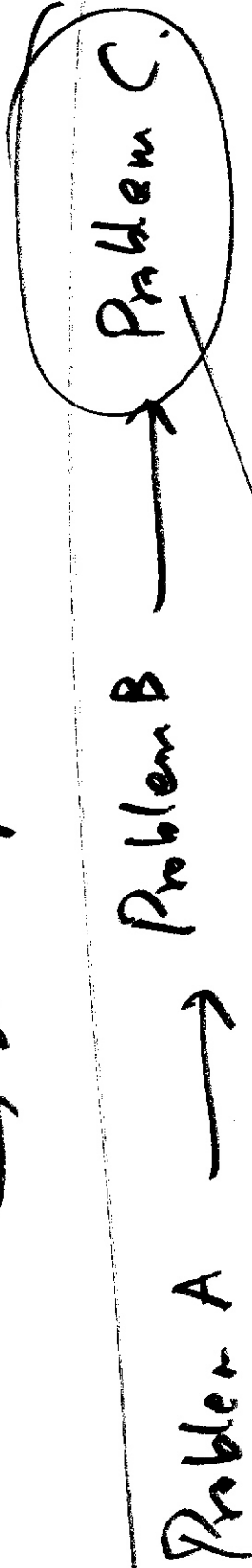
Problem C :

Problem B :

⇒

Same problem  $\theta = \delta_2$

find  $a(a)$  to minimize  $\theta$  }  
find  $a(a)$  to minimize  $\delta_2$  }



Alternation The

$F =$  Union of closed intervals.

Let  $P(x)$  be an  $r$ th order polynomial.

$$P(x) = \sum_{k=0}^r a_k x^k \quad \text{function continuous in } F$$

$D =$  desired

positive fn

$$W(x) [D(x) - P(x)]$$

$$E(x) = W(x) [D(x) - P(x)]$$

Define

$$\|E\| = \max_{x \in F} |E(x)|$$

Necessary + sufficient condition for  $P(x)$

To be a unique  $r$ th order polynomial

is that  $E(x)$

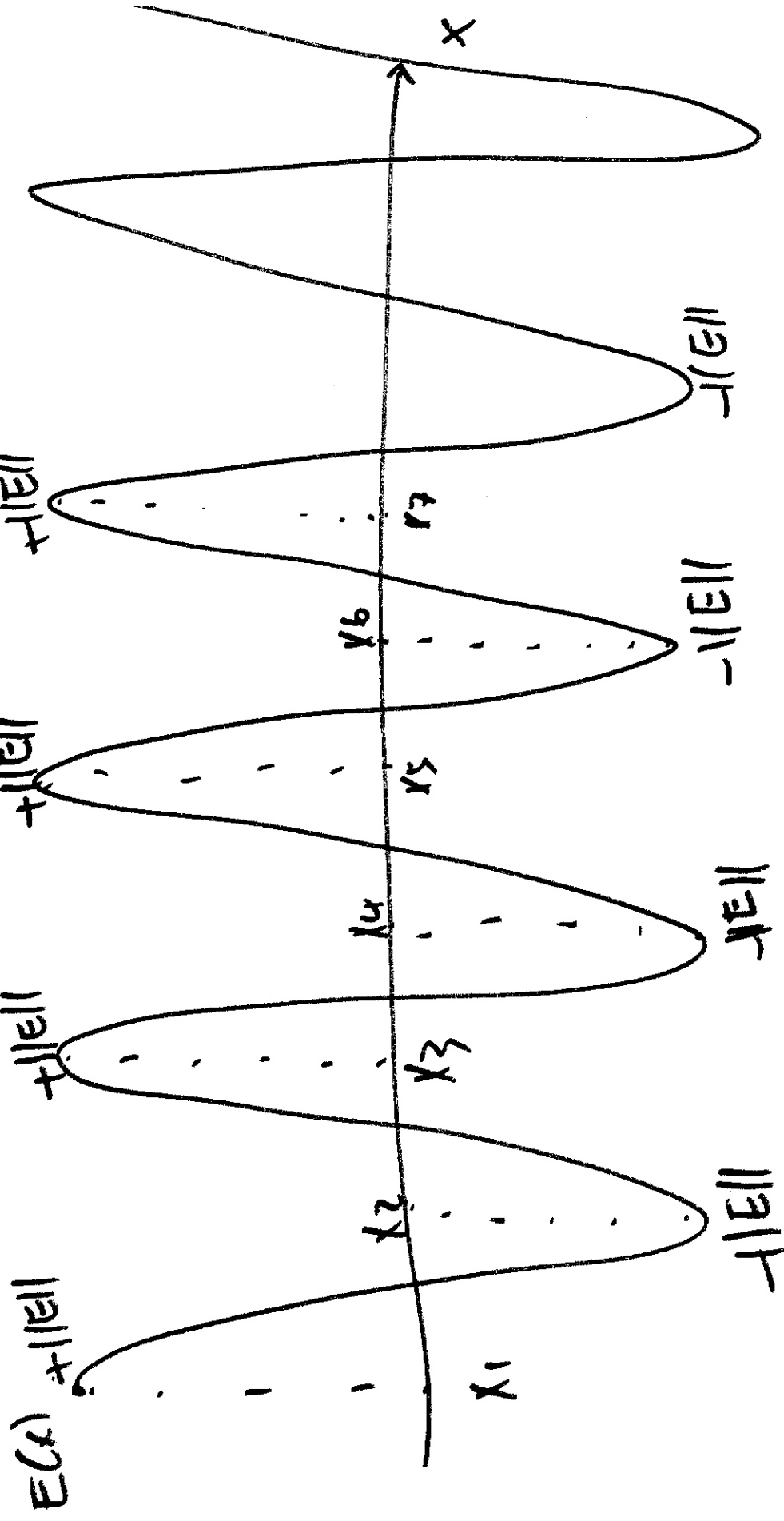
that minimizes  $\|E\|$  is  $r+2$  alternations.

Exhibits at least

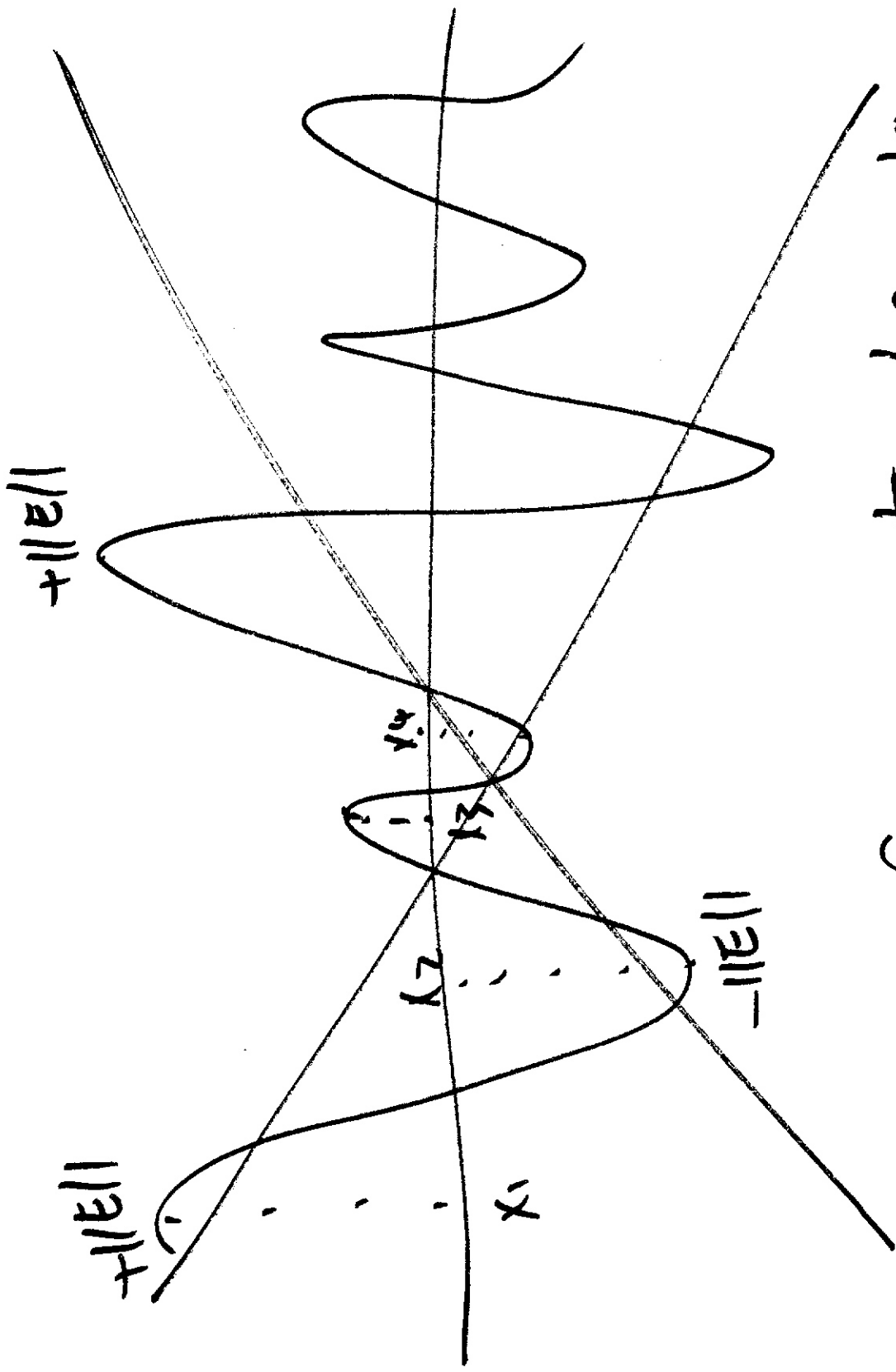
i.e. There is at least  $r+2$  values of  $x$

$$x_1 < x_2 < \dots < x_{r+2}$$

$$E(x_i) = -E(x_{i+1}) = \pm \|E\|$$

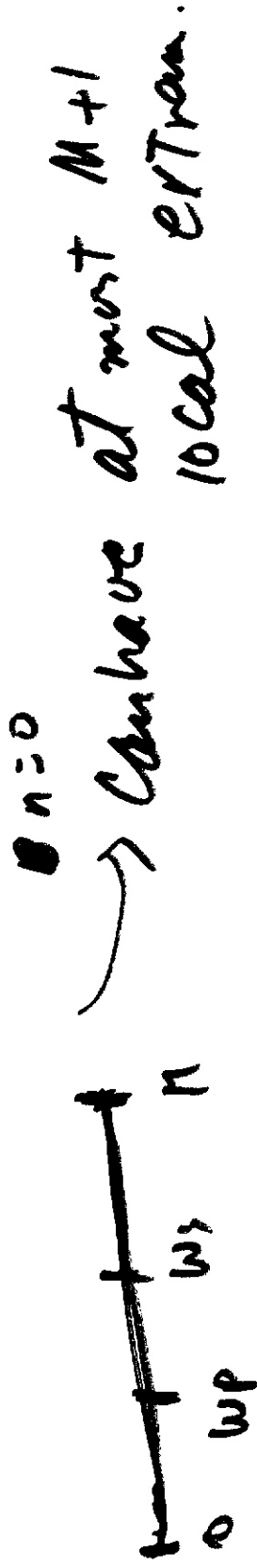


...  $r+2$  of these.

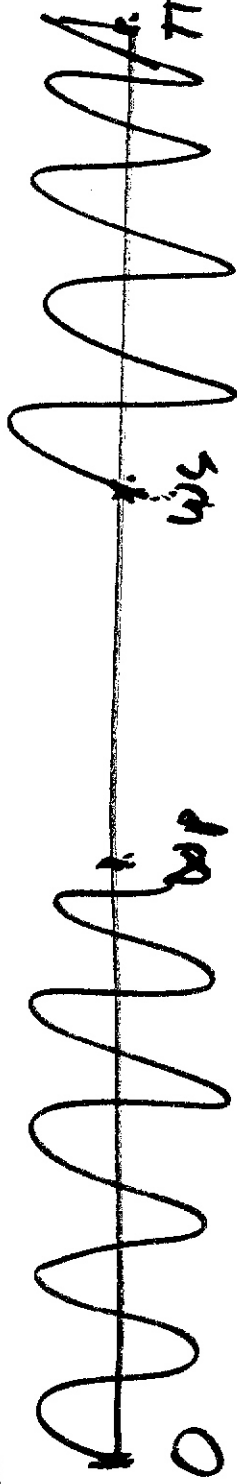


Cannot have this

Overall  $G(\omega) = \sum_{n=0}^M a(n) \cos(\omega n)$

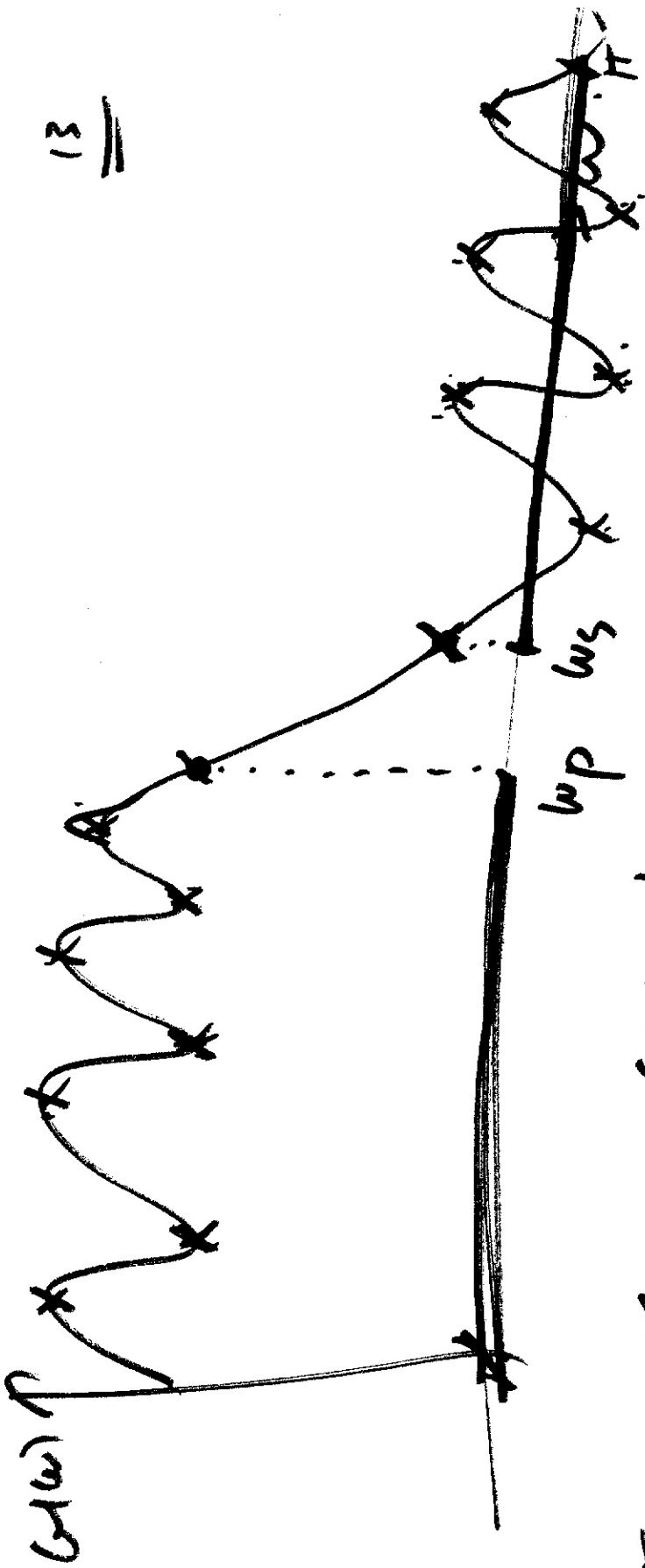


Alt. Thm Caves about  $F = J, U, I, 2$



at least  $M+2$  (r+2) alternations in  $F$ .

even though  $wp$  and  $ws$  are not local extrema of  $G(\omega)$ .



Even though  $G(w)$  between  $0, \pi$  has at most  $M+1$  local extrema, it ~~to~~ can have up to  $M+3$  local extrema in  $I_1 \cup I_2 = F$

Alt Th  $I_1 \cup I_2 \Rightarrow$  at least  $M+2$  alternating local extrema

(b) has either  $M+2$  or  $M+3$   
local extrema in  $F$  that are  
also alternation

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Fig 7.35, 7.37 0 & s

## Properties of optimal Filter

(1) For low pass filter, alternations always occur at  $\omega_p$  and  $\omega_s$ .

Fig 7.38.

$\Rightarrow$  otherwise we "lose" 2 alternations

(2) Filter will be equiripple except possibly at  $\omega = 0, \pi$

slope 0 or local extrem

Fig 7.39



# Alg for optimum Filter Design

Paul Mc Clellan

Problem B: Given  $w_s$ ,  $w_p$ ,  $K = \frac{\delta_1}{\delta_2}$

find  $a(n)$  to minimize  $\delta_2$ .

minimizing  $E(w_i) = \pm \delta_2$   $i=1, \dots, M+2$

alternating fashion  $E(w_i) = -E(w_{i+1})$   $i=1, \dots, M+1$

$$W(w_i) [ \underline{E(w_i)} - D(w_i) ] = (-1)^{i+1} \underline{\delta_2}$$

$$G(w) = \sum_{n=0}^M \underline{a(n)} \cos wn$$

linear sys of eqn.

$w_i$  for  $M+2$   
 $i=1, \dots, M+2$   
 $M+2$  unknown

easy

$a(n)$   $M+1$   
 $\delta_2 \rightarrow 1$   
Total  $M+2$  unknown.