

Digital Signal Processing

Lecture 14 Sampling

Announcements

- Ham exam Th 3/12 7-10+, The Woz Soda hall
- Lab:

-Who is having trouble?

What is this Phenomena?



Sampling of Continuous Time Signals (Ch.4)

- Sampling:
 - -Conversion from C.T (not quantized) into D.T (usually quantized)
- Reconstruction
 - -D.T (quantized) to C.T
- Why?
 - -Digital storage (audio, images, videos)
 - -Digital communications (fiber optics, cellular...)
 - -DSP (compression, correction, restoration)
 - -Digital synthesis (speech, graphics)
 - Learning

Sampling of C.T. Signals

• Typical System:





Ideal Sampling Model

$$x_s(t) = x_c \sum_{n=-\infty}^{\infty} \delta(t - nT)$$

Not physical: used for modeling & derivations

$$x[n] \leftrightarrow x_s(t) \leftrightarrow x_c(t)$$

• How is x[n] related to x_s(t) in freq. domain?

• How is x[n] related to $x_s(t)$ in the Freq. Domain?

$$\begin{split} x_s(t) &: \text{C.T} & X_s(j\Omega) = \sum_n x_c(nT) e^{-j\Omega nT} \\ x[n] &: \text{D.T} & X(e^{j\omega}) = \sum_n x[n] e^{-j\omega n} & \omega = \Omega T \end{split}$$

$$X(e^{j\omega}) = X_s(j\Omega)|_{\Omega = \omega/T}$$

$$X_s(j\Omega) = X(e^{j\omega})\big|_{\omega = \Omega T}$$

• How is $x_s(t)$ related to $x_c(t)$?

$$x_s(t) = x_c(t) \underbrace{\sum_{n} \delta(t - nT)}_{\triangleq s(t)}$$

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$$X_{s}(j\Omega) = \frac{1}{2\pi}X_{c}(j\Omega) * S(j\Omega)$$
$$= \frac{1}{T}\sum_{k=-\infty}^{\infty}X_{c}(j(\Omega - \Omega_{s})) | \Omega_{s} = \frac{2\pi}{T}$$

• X_s is replication of X_c !





Aliasing

Q: What is the difference in acquisition between the two images ?







Reconstruction of Bandlimited Signals

• Nyquist Sampling Thm: suppose $x_c(t)$ is bandlimited

$$X_c(j\Omega) = 0 \quad \forall \quad |\Omega| \ge \Omega_N$$

if $\Omega_s \ge 2\Omega_N$, then $x_c(t)$ can be uniquely determined from its samples $x[n] = x_c(nT)$

 $x_{c}(t)$

Bandlimitedness is the key to uniqueness

 \mathcal{N}

multiple signals go through the samples, but only one is bandlimited!



Reconstruction in Time Domain

$$h_{r}(t) = \frac{1}{2\pi} \int_{-\Omega_{s}/2}^{\Omega_{s}/2} Te^{j\Omega t} d\Omega$$

$$= \frac{T}{2\pi} \frac{1}{jt} s^{j\Omega t} \Big|_{-\Omega_{s}/2}^{\Omega_{s}/2}$$

$$= \frac{T}{2\pi} \frac{e^{j\frac{\Omega_{s}}{2}t} - e^{-j\frac{\Omega_{s}}{2}t}}{2j}$$

$$= \frac{T}{\pi t} \frac{e^{j\frac{\Omega_{s}}{2}t} - e^{-j\frac{\Omega_{s}}{2}t}}{2j}$$

$$= \frac{T}{\pi t} \sin(\frac{\Omega_{s}}{2}t) = \frac{T}{\pi t} \sin(\frac{\pi}{T}t)$$

$$= \sin(\frac{t}{T})$$
Music ECS US Berkely

Reconstruction in Time Domain

$$x_r(t) = x_s(t) * h_r(t) = \left(\sum_n x[n]\delta(t - nT)\right) * h_r(t)$$
$$= \sum_n x[n]h_r(t - nT)$$

The sum of "since gives $x_r(t) \Rightarrow$ Unique signal bandlimited by Ω_s



Aliasing

• If $\Omega_N > \Omega_s/2$, $x_r(t)$ an aliased version of $x_c(t)$



$X_r(j\Omega) = \begin{cases} TX_s(j\Omega) & \text{if } |\Omega| \le \Omega_s/2\\ 0 & \text{otherwise} \end{cases}$

Anti-Aliasing





SDR non-perfect anti-Aliasing Demo