

EE123

# Digital Signal Processing

## Lecture 25 Filter Design

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### Linear Filter Design

- Used to be an art
  - Now, lots of tools to design optimal filters
- For DSP there are two common classes
  - Infinite impulse response IIR
  - Finite impulse response FIR
- Both classes use finite order of parameters for design
- We will cover FIR designs, briefly mention IIR

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### What is a linear filter

- Attenuates certain frequencies
- Passes certain frequencies
- Effects both **phase** and **magnitude**
- IIR
  - Mostly non-linear phase response
  - Could be linear over a range of frequencies
- FIR
  - Much easier to control the phase
  - Both non-linear and linear phase

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### FIR Design by Windowing

- Given desired frequency response,  $H_d(e^{j\omega})$ , find an impulse response

$$h_d[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} H_d(e^{j\omega}) e^{j\omega n} d\omega$$

← ideal

- Obtain the M<sup>th</sup> order causal FIR filter by truncating/windowing it

$$h[n] = \begin{cases} h_d[n]w[n] & 0 \leq n \leq M \\ 0 & \text{otherwise} \end{cases}$$

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### FIR Design by Windowing

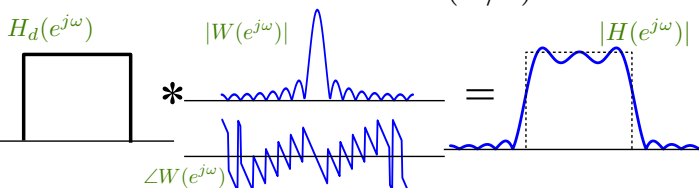
- We already saw that,

$$H(e^{j\omega}) = H_d(e^{j\omega}) * W(e^{j\omega})$$

periodic

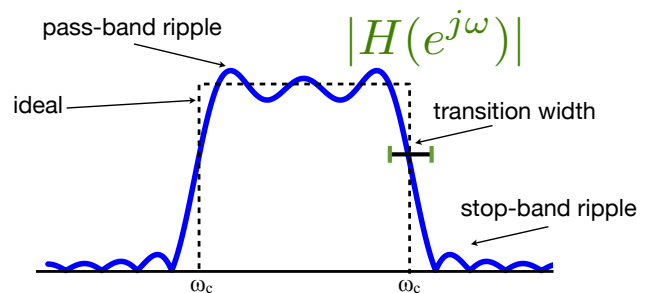
- For Boxcar (rectangular) window

$$W(e^{j\omega}) = e^{-j\omega \frac{M}{2}} \frac{\sin(\omega(M+1)/2)}{\sin(\omega/2)}$$



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### FIR Design by Windowing



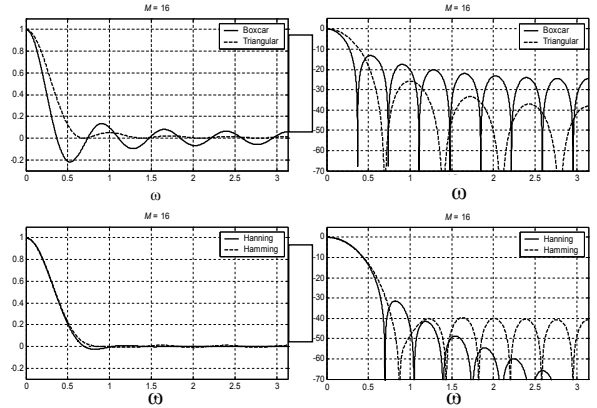
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## Tapered Windows

Name(s)	Definition	MATLAB Command	Graph (M=8)
Hann	$w[n] = \begin{cases} \frac{1}{2} \left[ 1 + \cos\left(\frac{\pi n}{M/2}\right) \right] &  n  \leq M/2 \\ 0 &  n  > M/2 \end{cases}$	hann(M+1)	
Hanning	$w[n] = \begin{cases} \frac{1}{2} \left[ 1 + \cos\left(\frac{\pi n}{M/2+1}\right) \right] &  n  \leq M/2 \\ 0 &  n  > M/2 \end{cases}$	hanning(M+1)	
Hamming	$w[n] = \begin{cases} 0.54 + 0.46 \cos\left(\frac{\pi n}{M/2}\right) &  n  \leq M/2 \\ 0 &  n  > M/2 \end{cases}$	hamming(M+1)	

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## Tradeoff - Ripple vs Transition Width



Python: `scipy.filter.firwin`

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## FIR Filter Design

- Choose a desired frequency response  $H_d(e^{j\omega})$ 
  - non causal (zero-delay), and infinite imp. response
  - If derived from C.T, choose T and use:

$$H_d(e^{j\omega}) = H_c(j\frac{\Omega}{T})$$

- Window:
  - Length M+1  $\Leftrightarrow$  effect transition width
  - Type of window  $\Leftrightarrow$  transition-width/ ripple
  - Modulate to shift impulse response

$$H_d(e^{j\omega})e^{-j\omega\frac{M}{2}}$$

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## FIR Filter Design

- Determine truncated impulse response  $h_1[n]$

$$h_1[n] = \begin{cases} \frac{1}{2\pi} \int_{-\pi}^{\pi} H_d(e^{j\omega}) e^{-j\omega\frac{M}{2}} e^{j\omega n} & 0 \leq n \leq M \\ 0 & \text{otherwise} \end{cases}$$

- Apply window

$$h_w[n] = w[n]h_1[n]$$

- Check:
  - Compute  $H_w(e^{j\omega})$ , if does not meet specs increase M or change window

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## Example: FIR Low-Pass Filter Design

$$H_d(e^{j\omega}) = \begin{cases} 1 & |\omega| \leq \omega_c \\ 0 & \text{otherwise} \end{cases}$$

Choose M  $\Rightarrow$  Window length and set

$$H_1(e^{j\omega}) = H_d(e^{j\omega})e^{-j\omega\frac{M}{2}}$$

$$h_1[n] = \begin{cases} \frac{\sin(\omega_c(n-M/2))}{\pi(n-M/2)} & 0 \leq n \leq M \\ 0 & \text{otherwise} \end{cases}$$

$$\frac{\omega_c}{\pi} \text{sinc}\left(\frac{\omega_c}{\pi}(n-M/2)\right)$$

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## Example: FIR Low-Pass Filter Design

- The result is a windowed sinc function

$$h_w[n] = w[n]h_1[n]$$

- High Pass Design:
  - Design low pass  $h_w[n]$
  - Transform to  $h_w[n](-1)^n$

$$\frac{\omega_c}{\pi} \text{sinc}\left(\frac{\omega_c}{\pi}(n-M/2)\right)$$

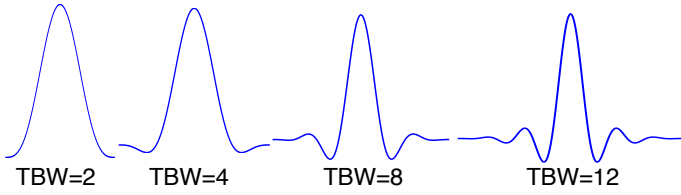
- General bandpass
  - Transform to  $2h_w[n]\cos(\omega_0 n)$

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## Characterization of Filter Shape

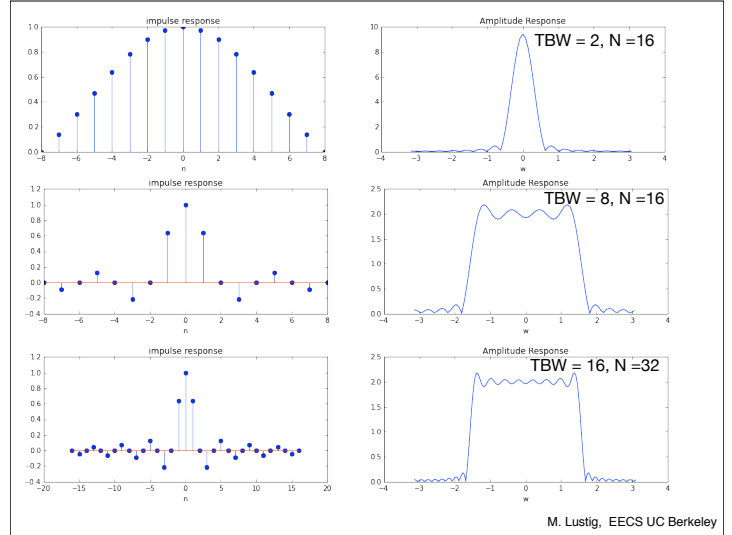
Time-Bandwidth Product, a unitless measure

$$T(BW) = (M+1)\omega/2\pi \Rightarrow \text{also, total \# of zero crossings}$$



Larger TBW  $\Rightarrow$  More of the “sinc” function  
hence, frequency response looks more like a rect function

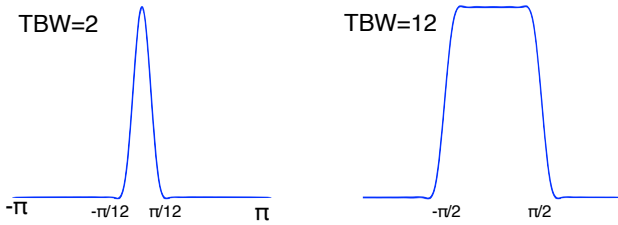
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## Frequency Response Profile

Q: What are the lengths of these filters in samples?



$$2 = (M+1) \cdot (\pi/6) / (2\pi) \Rightarrow M=23 \quad 12 = (M+1) \cdot (\pi) / (2\pi) \Rightarrow M=23$$

Note that transition is the same!

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## Optimal Filter Design

- Window method
  - Design Filters heuristically using windowed sinc functions
- Optimal design
  - Design a filter  $h[n]$  with  $H(e^{j\omega})$
  - Approximate  $H_d(e^{j\omega})$  with some optimality criteria - or satisfies specs.

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