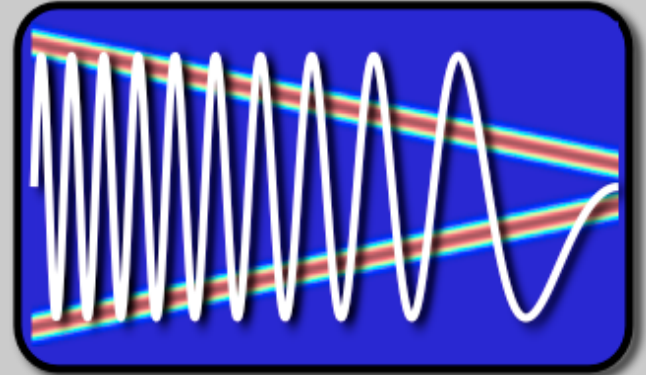


EE123



Digital Signal Processing

Lecture 28

Phase Response

All-Pass and Minimum Phase

GEOMAGNETIC STORM SPARKS AURORAS: Unsettled solar wind conditions + the possible arrival of a weak CME ignited a [G2-class](#) geomagnetic storm during the early hours of April 10th. Northern Lights spilled over the Canadian border into the USA as far south as [Idaho](#), [Montana](#) and [Colorado](#). The storm is subsiding now, but NOAA forecasters estimate a 50% chance that it could flare up again before the end of the day

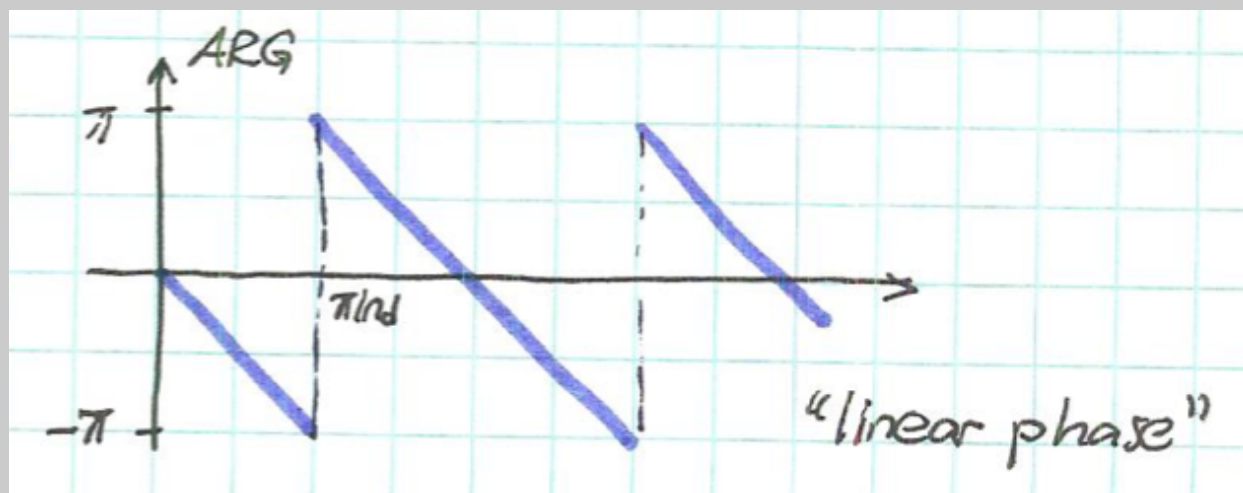
Phase response

Example: $H(e^{j\omega}) = e^{-j\omega n_d} \leftrightarrow h[n] = \delta[n - n_d]$

$$|H(e^{j\omega})| = 1$$

$$\arg[H(e^{j\omega})] = -\omega n_d$$

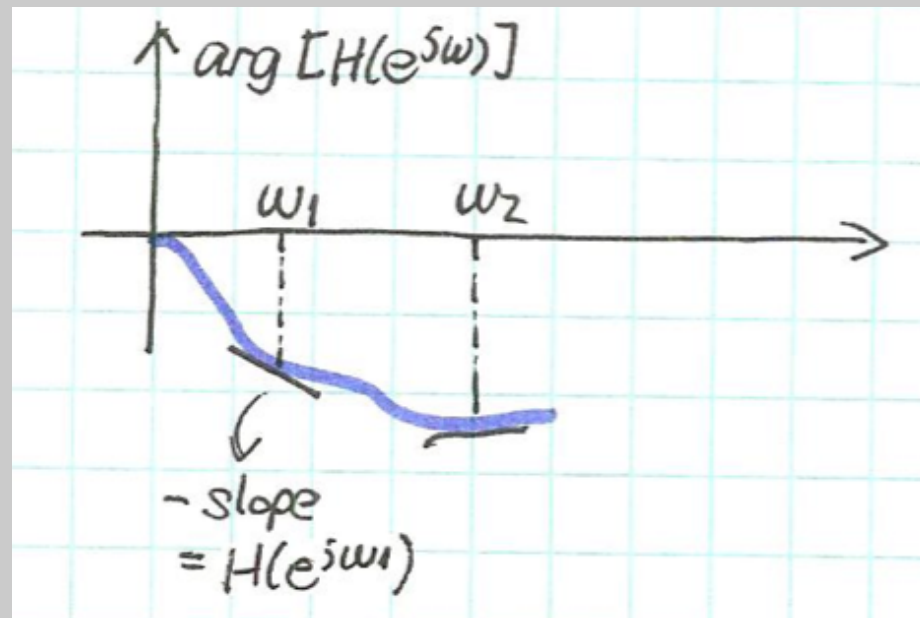
ARG is the wrapped phase
arg is the unwrapped phase



Group delay

To characterize general phase response, look at the group delay:

$$\text{grd}[H(e^{j\omega})] = -\frac{d}{d\omega} \{ \arg[H(e^{j\omega})] \}$$

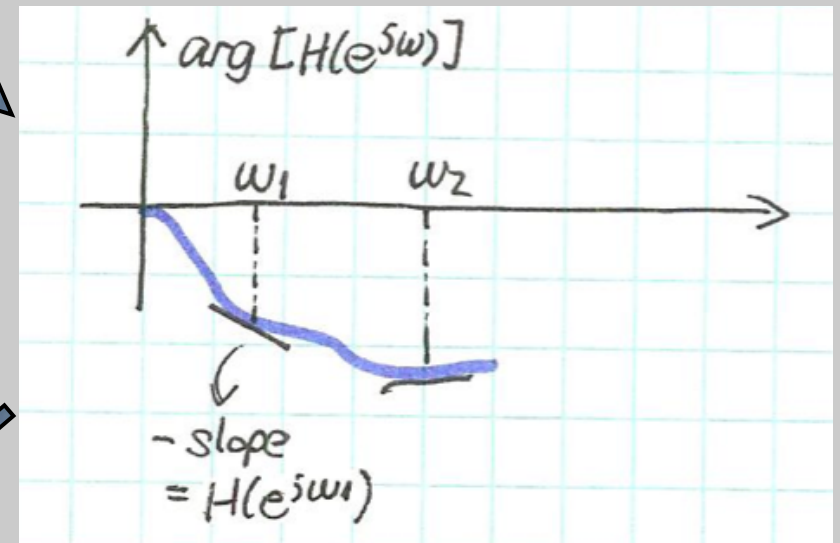
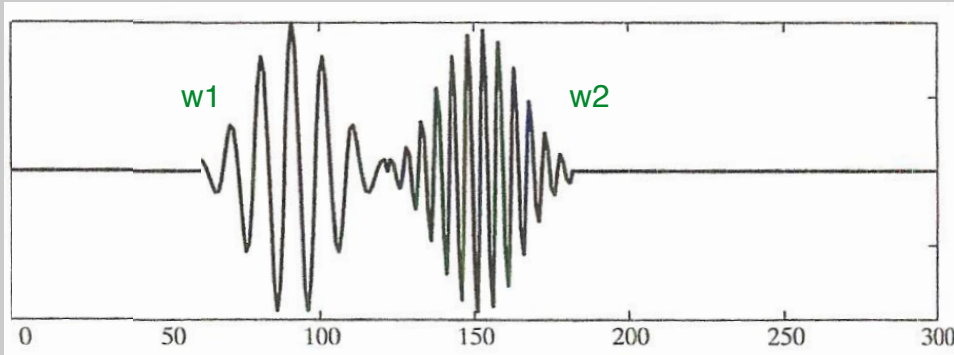


For linear phase system, the group delay is n_d

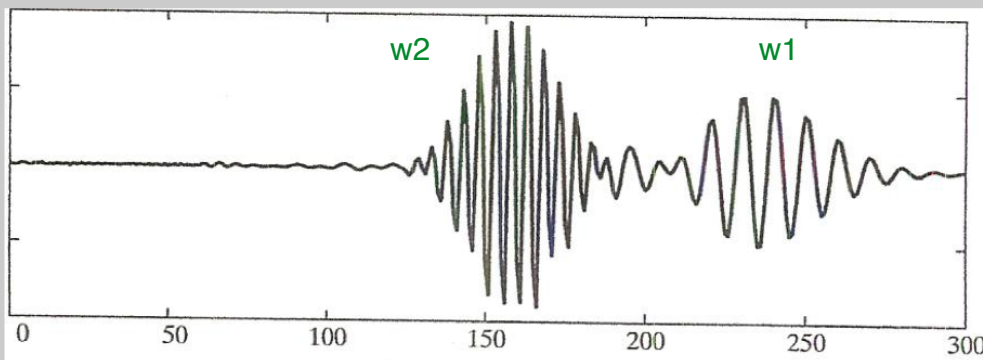
Group delay

$$\text{grd}[H(e^{j\omega})] = -\frac{d}{d\omega} \{ \arg[H(e^{j\omega})] \}$$

Input



Output



For narrowband signals, phase response looks like a linear phase

Group delay math

$$H(z) = \frac{b_0}{a_0} \frac{\prod_{k=1}^M (1 - c_k z^{-1})}{\prod_{k=1}^N (1 - d_k z^{-1})}$$

arg of products is sum of args

$$\arg [H(e^{j\omega})] = -\sum_{k=1}^N \arg [1 - d_k e^{-j\omega}] + \sum_{k=1}^M \arg [1 - c_k e^{-j\omega}]$$

$$\text{grd} [H(e^{j\omega})] = -\sum_{k=1}^N \text{grd} [1 - d_k e^{-j\omega}] + \sum_{k=1}^M \text{grd} [1 - c_k e^{-j\omega}]$$

Group delay math

$$\text{grd}[H(e^{j\omega})] = -\sum_{k=1}^N \text{grd}[1 - d_k e^{-j\omega}] + \sum_{k=1}^M \text{grd}[1 - c_k e^{-j\omega}]$$

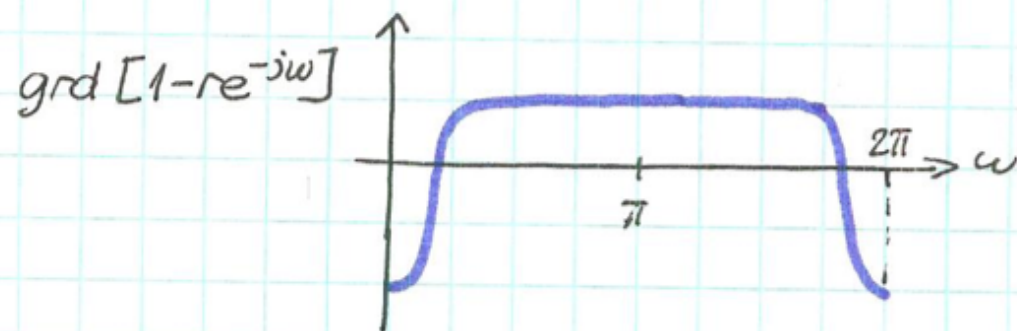
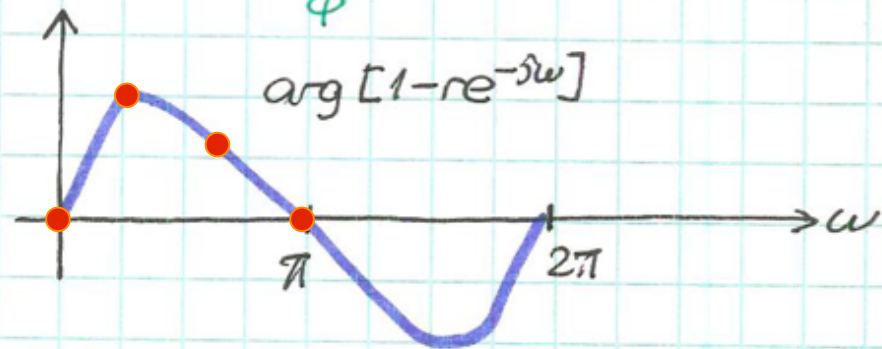
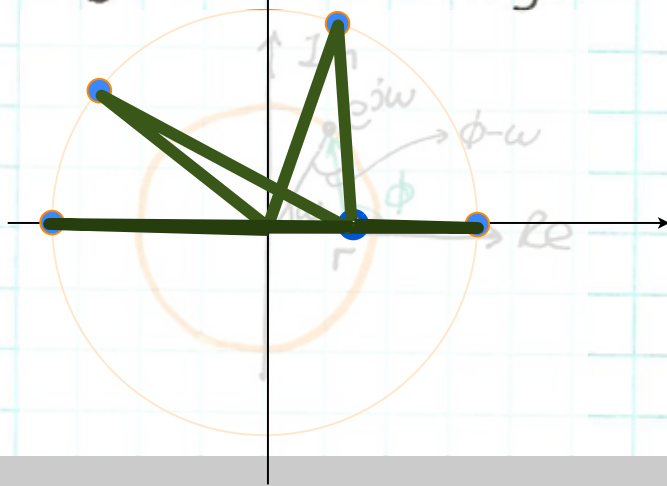
Look at each factor:

$$\text{arg}[1 - \underbrace{r e^{j\theta}}_{c_k \text{ or } d_k} e^{-j\omega}] = \tan^{-1}\left(\frac{r \sin(\omega - \theta)}{1 - r \cos(\omega - \theta)}\right)$$
$$\text{grd}[1 - r e^{j\theta} e^{-j\omega}] = \frac{r^2 - r \cos(\omega - \theta)}{|1 - r e^{j\theta} e^{-j\omega}|^2}$$

Look at a zero lying on the real axis

Geometric Interpretation (for $\theta=0$)

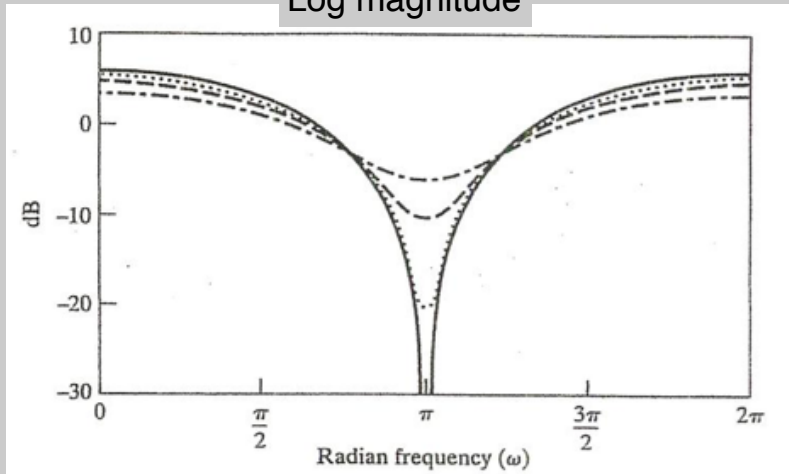
$$\arg [1 - re^{-j\omega}] = \arg [(e^{j\omega} - r)e^{-j\omega}] = \underbrace{\arg [e^{j\omega} - r]}_{\phi} - \underbrace{\arg [e^{-j\omega}]}_{\omega}$$



$\theta \neq 0 \Rightarrow$ shift to the right by θ

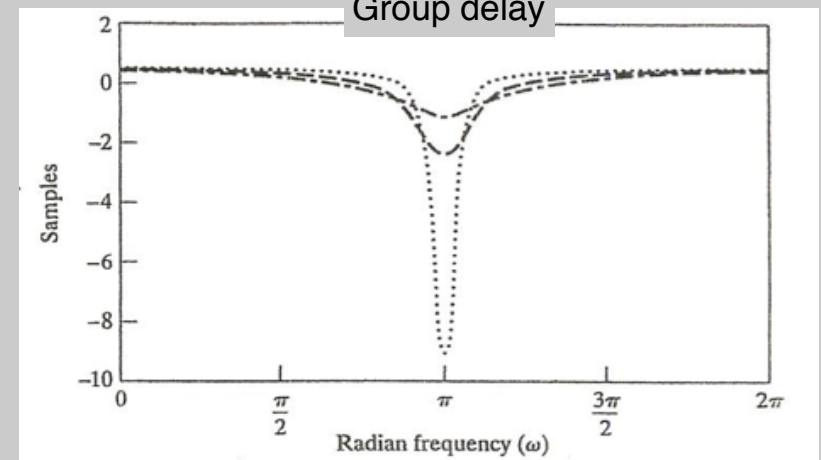
- * Poles increase magnitude, but introduce phase lag and group delay.
- * Zeros do the opposite.
- * These effects are more marked when $r \rightarrow 1$.

Log magnitude

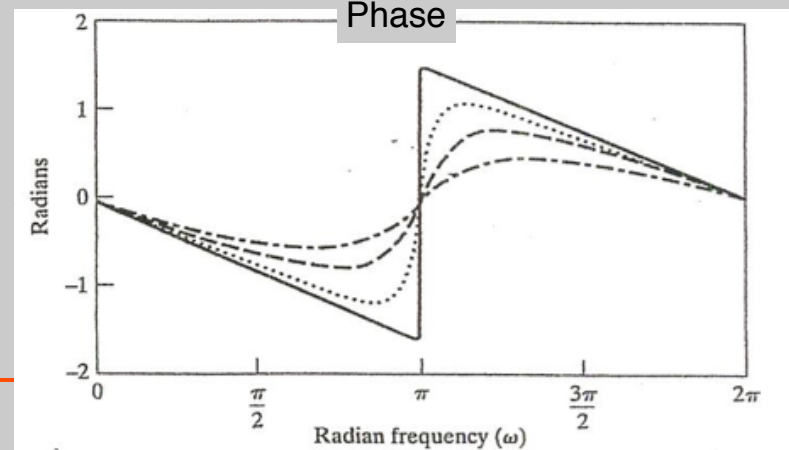


- $r = 0.5$
- .- $r = 0.7$
- ... $r = 0.9$
- $r = 1$

Group delay



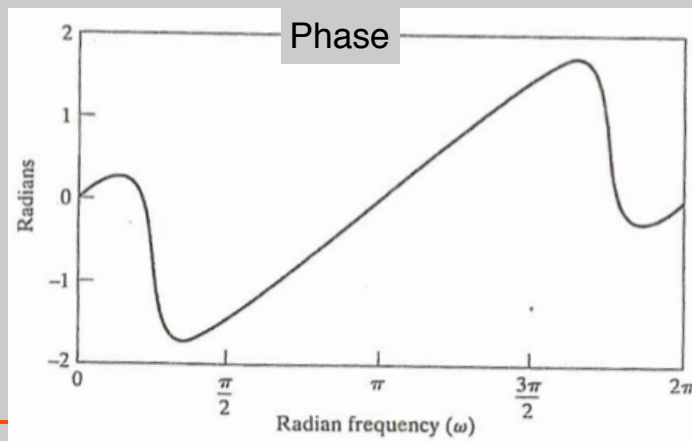
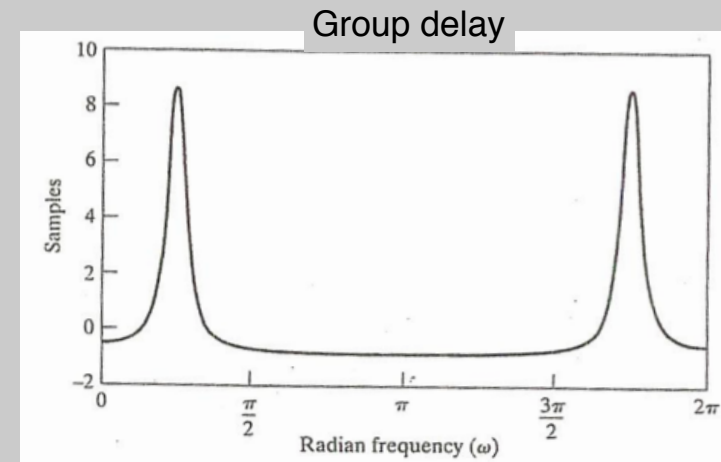
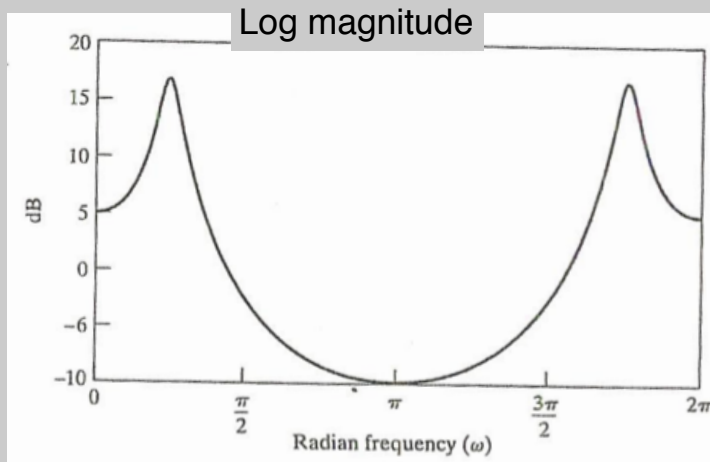
Phase



2nd order IIR example

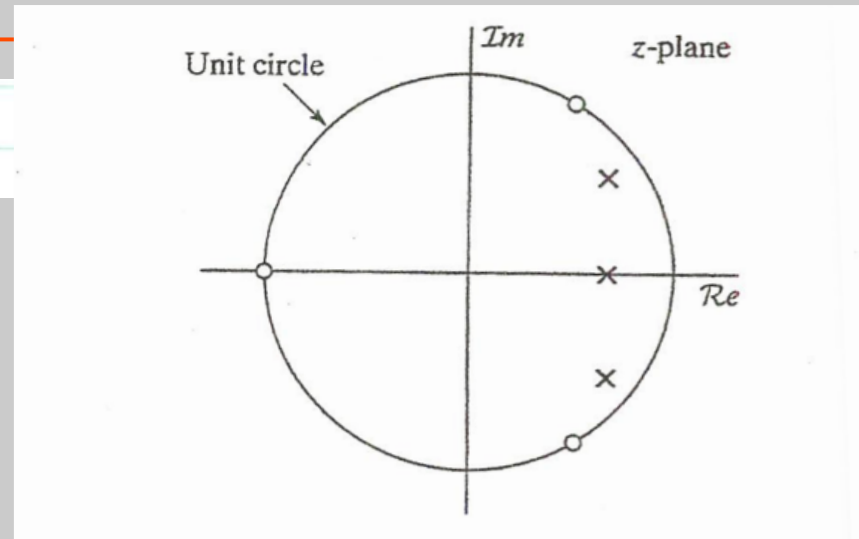
Example: 2nd order IIR with complex poles

$$H(z) = \frac{1}{(1 - re^{j\theta}z^{-1})(1 - re^{-j\theta}z^{-1})}$$

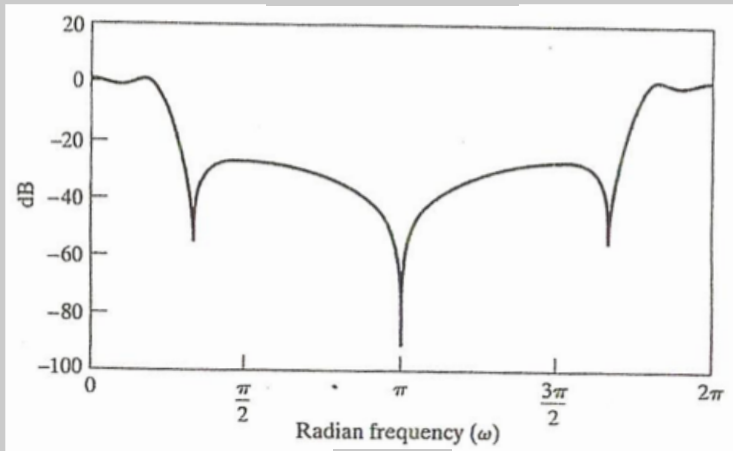


3rd order IIR example

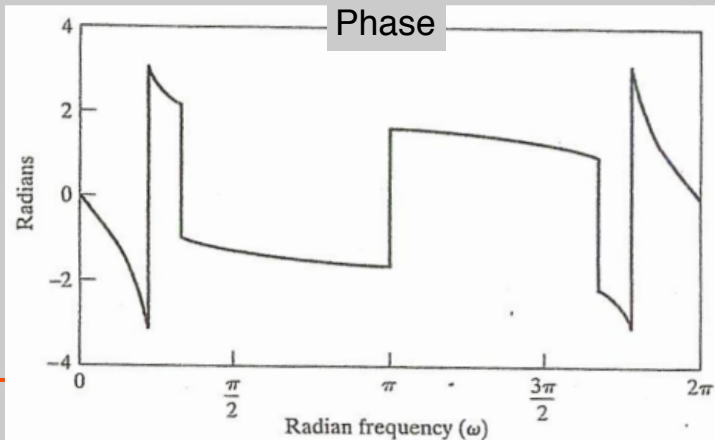
Example: 3rd order IIR



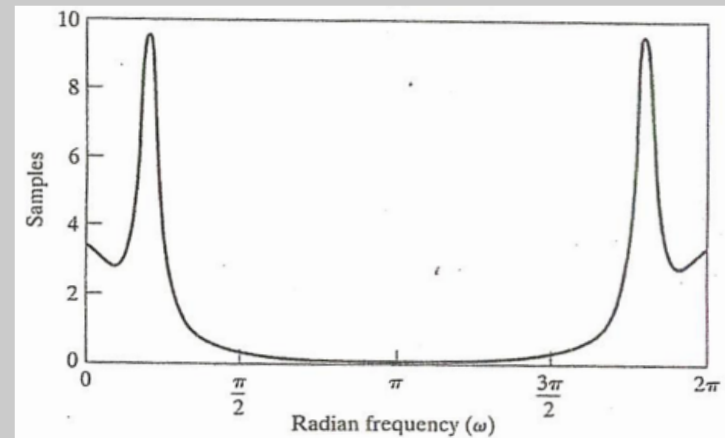
Log magnitude



Phase



Group delay

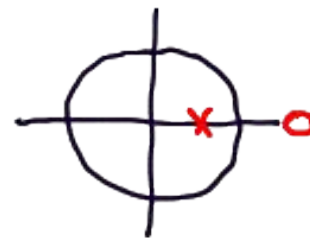


All-Pass Systems

Q

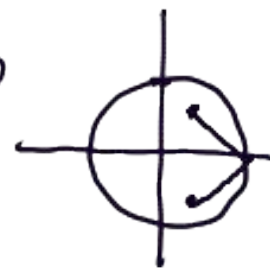
what is the magnitude response of

$$H(z) = \frac{z^{-1} - a^*}{1 - az^{-1}}$$



$$|H(e^{j\omega})| = \frac{|e^{-j\omega} - a^*|}{|1 - ae^{-j\omega}|} = \frac{|e^{-j\omega}(1 - a^*e^{j\omega})|}{|1 - ae^{-j\omega}|} =$$

$$= \frac{|1 - a^*e^{j\omega}|}{|1 - (a^*e^{j\omega})^*|} = 1 \quad \forall \omega$$



A general all-pass system:

(3)

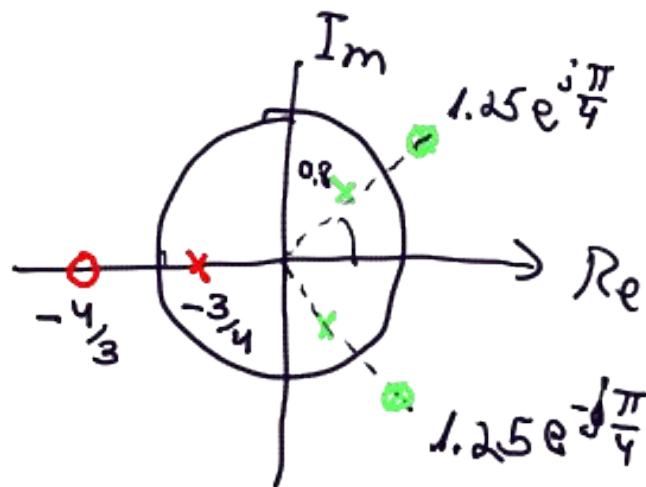
$$H_{ap}(z) = \prod_{k=1}^{M_R} \frac{z^{-1} d_k}{1 - d_k z^{-1}} \cdot \prod_{k=1}^{M_C} \frac{z^{-1} - e_k^*}{1 - e_k z^{-1}} \cdot \frac{z^{-1} - e_k}{1 - e_k^* z^{-1}}$$

d_k : real Poles

e_k : complex poles paired w/ conjugate e_k^*

$$|H_{ap}(e^{j\omega})| \equiv 1$$

Example



phase response of an all-pass:

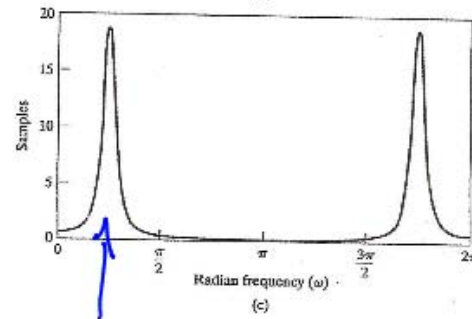
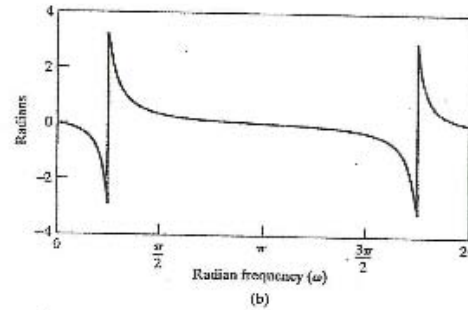
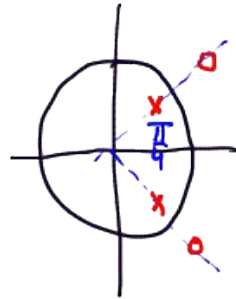
(4)

$$\arg \left[\frac{e^{-j\omega} - re^{-j\theta}}{1 - re^{j\theta} e^{-j\omega}} \right] = \arg \left[\frac{e^{-j\omega} (1 - re^{-j\theta} e^{-j\omega})}{1 - re^{j\theta} e^{-j\omega}} \right]$$
$$= \underbrace{\arg[e^{-j\omega}]}_{-\omega} - 2 \arg[1 - re^{j\theta} e^{-j\omega}]$$

$$\text{grad} \left[\frac{e^{-j\omega} - re^{-j\theta}}{1 - re^{j\theta} e^{-j\omega}} \right] = 1 - 2 \text{grad} [1 - re^{j\theta} e^{-j\omega}]$$

Example: < Figure 5.20 >

5



can be used to compensate phase distortion.

Claim: for a stable op system $H_{op}(z)$: ⑥

$$(i) \text{grad} [H_{op}(e^{j\omega})] > 0$$

$$(ii) \text{arg} [H_{op}(e^{j\omega})] \leq 0$$

Delay positive \rightarrow causal
phase negative \rightarrow phase lag.

proof in book.