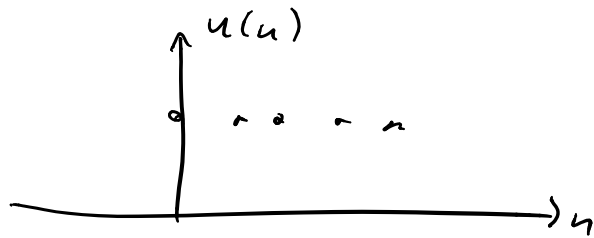


DTFT

$$\begin{array}{ccc}
 x(n) & \xrightarrow{\text{DTFT}} & X(\omega) = \sum_{n=-\infty}^{+\infty} x(n) e^{-j\omega n} \\
 \uparrow n \text{ integ} & & \uparrow \text{Continuous} \\
 & & \text{Real} \\
 & & \text{var}
 \end{array}$$

If $x(n)$ is absolutely summable
 $\sum_n |x(n)| < \infty$ Then F.T. exists
 and converges to a continuous fn.

EX $x(n) = a^n u(n)$

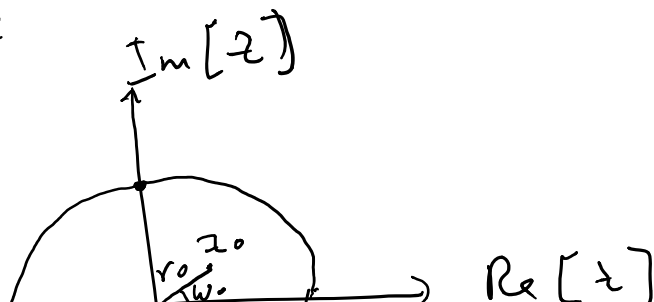


$$x(n) \longleftrightarrow \mathcal{Z}\{x(n)\} = X(z) = \sum_{n=-\infty}^{+\infty} x(n) z^{-n}$$

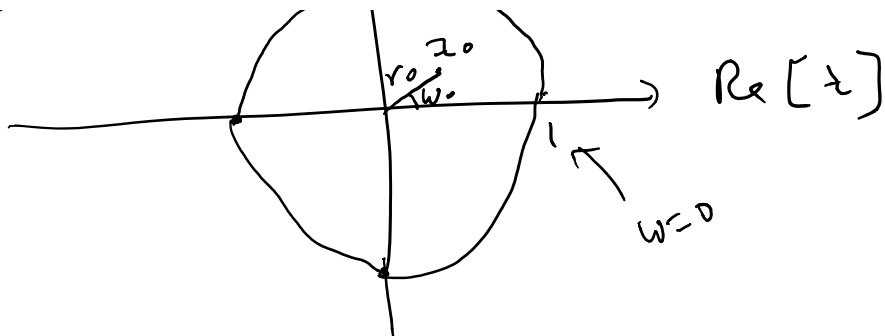
$$X(\omega) = \sum_n x(n) e^{-j\omega n}$$

$$X(\omega) = [X(z)]_{z=e^{j\omega}}$$

$$z_0 = r_0 e^{j\omega_0}$$



$$z_0 = r_0 e^{j\omega_0}$$



$$|z_0| = r_0$$

$$\neq z_0 = \omega_0$$

What values of z does $X(z)$ converge?

$$X(z) = \sum_{n=-\infty}^{+\infty} x(n) (r e^{j\omega})^{-n}$$

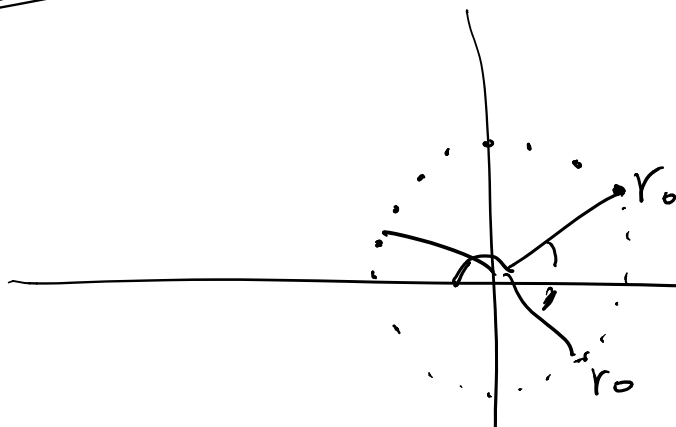
$$X(z) = \sum_{n=-\infty}^{+\infty} \underbrace{(x(n) r^{-n})}_{\text{DTFT of } x(n) r^{-n}} e^{-j\omega n}$$

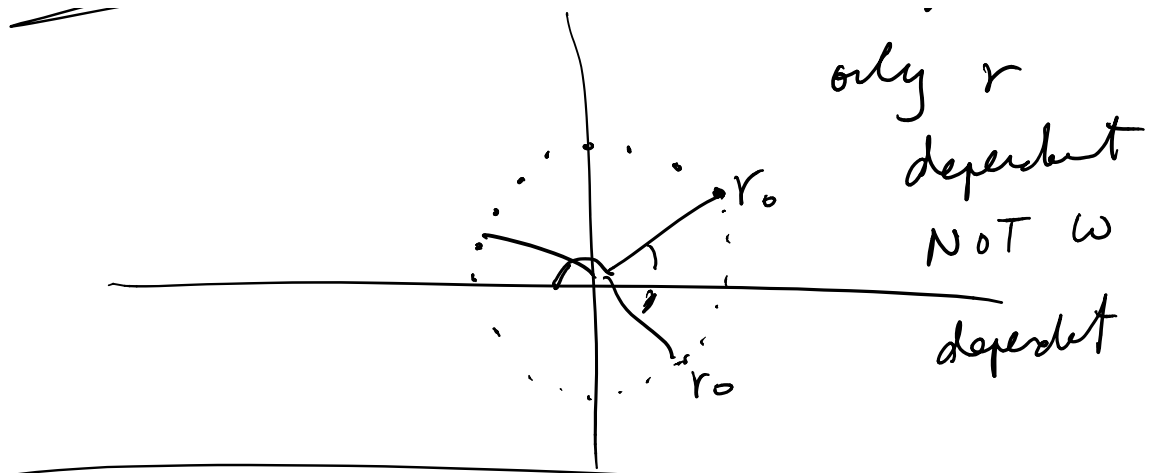
$X(z)$ converges if $x(n) r^{-n}$ is absolutely summable i.e.

$$\sum_n |x(n) r^{-n}| < \infty$$

R.O.C. z -Trajectory is

only r dependent
NOT ω dependent





Possibilities For Region of Convergence ROC

- Inside of some circle
- outside " " "
- Between 2 circles.
- point
- at z

$X(z)$ by itself doesn't uniquely specify a sequence

$$X(z) + \text{ROC} \longleftrightarrow X(n)$$

$$X(z) + \text{ROC 1} \longleftrightarrow X_1(n)$$

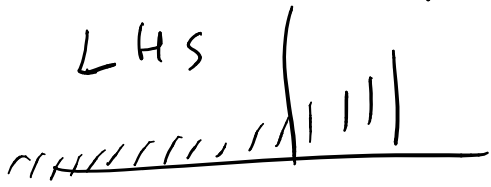
$$X(z) + \text{ROC 2} \longleftrightarrow X_2(n)$$

Example $X(z) = \frac{1}{1 - az^{-1}}$

ROC 1 \rightarrow LHS
Left handed



ROC 1 \rightarrow LHS
Left handed
sequence



ROC 2 \rightarrow RHS



$x(n) = a^n u(n)$

$X(z) = \sum_{n=0}^{\infty} a^n z^{-n}$

RHS

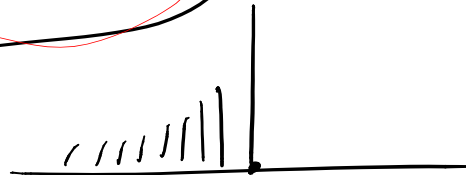
$\sum_{n=0}^{\infty} (a z^{-1})^n$

$\frac{1}{1 - a z^{-1}}$

if $|a z^{-1}| < 1$
 $\Rightarrow |z| > |a| \Rightarrow \text{ROC 2}$

$x(n) = -a^n u(-n-1)$

Left hand seq

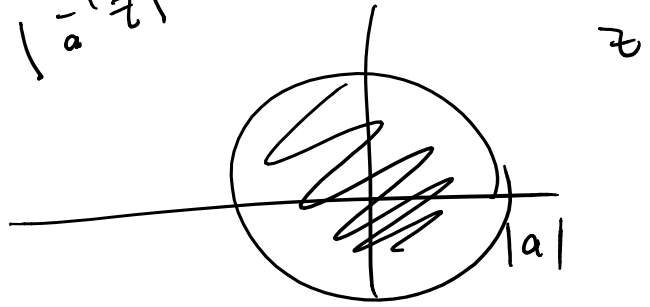


$X(z) = \sum_{n=-\infty}^{-1} a^n z^{-n} = 1 - \sum_{m=0}^{\infty} (a^{-1} z)^m$

check of
via $m = -n$

Converge
 $|a^{-1} z| < 1 \Rightarrow |z| < |a|$

$$|a^{-n} z| < 1$$

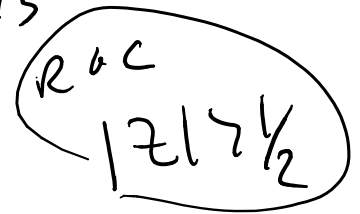


Ex

$$x[n] = \left(\frac{1}{2}\right)^n u[n] + \left(-\frac{1}{3}\right)^n u[n]$$

converge $|z| > \frac{1}{2} \rightarrow \text{ROC1}$

$|z| > \frac{1}{3} \rightarrow \text{ROC2}$



Intersect
ROC 1
and ROC 2

Ex Both Handed Seq.

$$x[n] = \left(-\frac{1}{3}\right)^n u[n] - \left(\frac{1}{2}\right)^n u[-n-1]$$

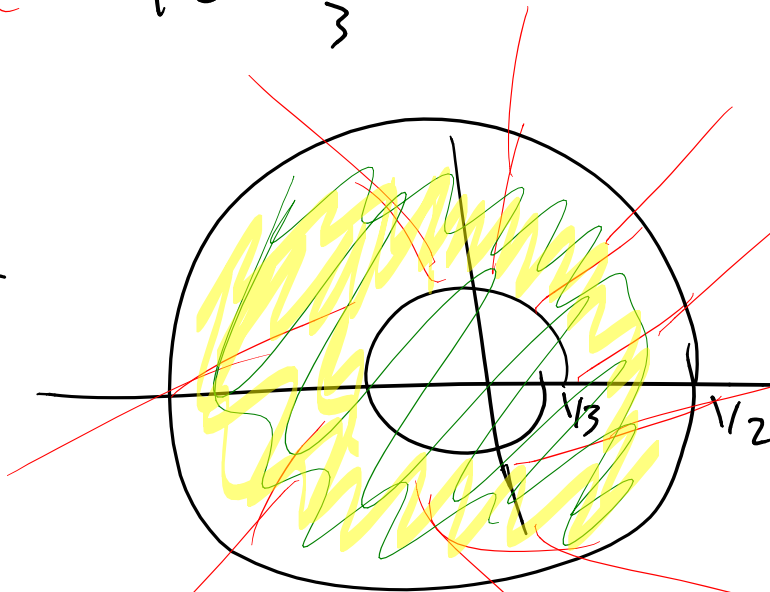
ROC 1 $|z| > \frac{1}{3}$

$|z| < \frac{1}{2}$ ROC 2

$$ROC 1 \quad |z| > \frac{1}{3}$$

$$|z| < \frac{1}{2} \quad ROC 2$$

Ring



ROC is a ring

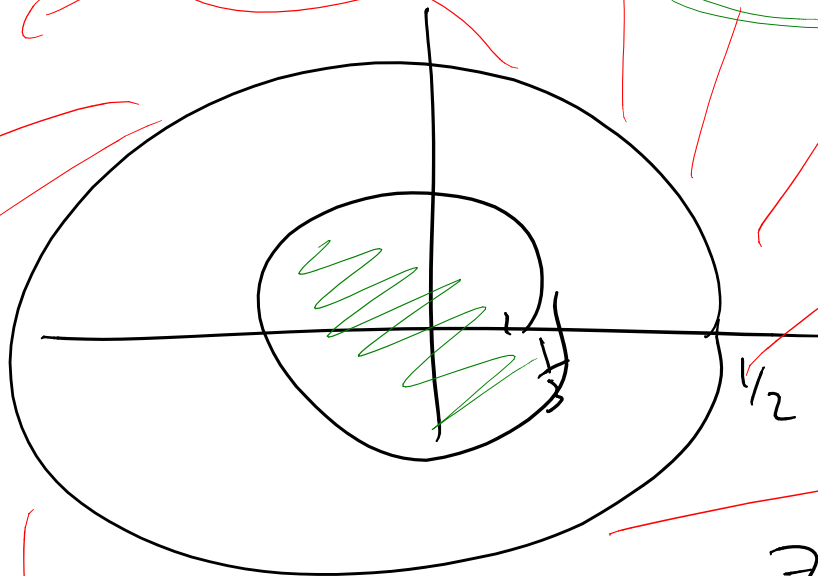
$$ROC 1 \cap ROC 2$$

$$\frac{1}{3} < |z| < \frac{1}{2}$$

$$x(n) = \left(-\frac{1}{2}\right)^n u(n) - \left(\frac{1}{3}\right)^n u(-n-1)$$

ROC 1

ROC 2



⇒ There is no ROC for

Does not exist

→ Z.T.

Ex $x(n)$ Finite length seq.

$$V(z) = \sum_{-N}^N x(n) z^{-n} = x(0)z^0 + x(1)z^{-1} + \dots$$

$$X(z) = \sum_{n=M}^N x(n) z^{-n} = x(0)z^0 + x(1)z^{-1} + x(-1)z^1 + x(-2)z^2 + \dots$$

Converges everywhere
except at zero and ∞

Ex $x(n) = \begin{cases} a^n & 0 \leq n \leq N-1 \\ 0 & \text{otherwise} \end{cases}$

$$X(z) = \sum_{n=0}^{N-1} x(n) z^{-n} = \frac{1 - (a z^{-1})^N}{1 - a z^{-1}}$$

$$= 1 + a z^{-1} + a^2 z^{-2} + \dots + a^{N-1} z^{-(N-1)}$$

$$= 1 + \frac{a}{z} + \frac{a^2}{z^2} + \dots$$

ROC: everywhere except for $z=0$

Ex $x(n) = \begin{cases} a^n & -N \leq n \leq 0 \\ 0 & \text{elsewhere} \end{cases}$

$$X(z) = \sum_{n=-N}^0 a^n z^{-n} = 1 + a^{-1} z + a^{-2} z^2 + \dots$$

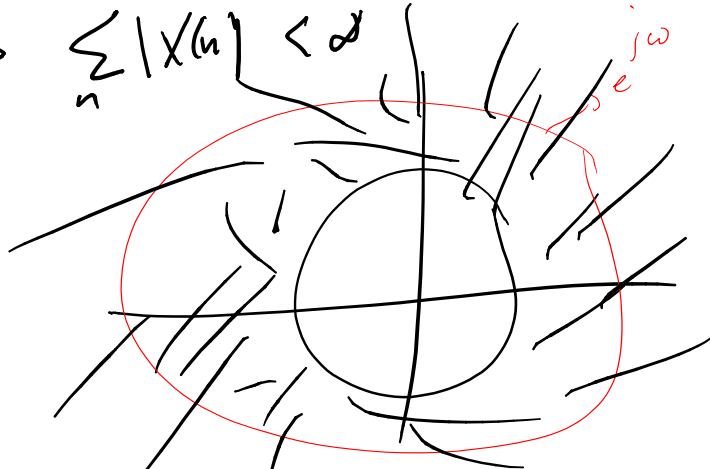
ROC: everywhere except $z=\infty$

Properties of ROC :

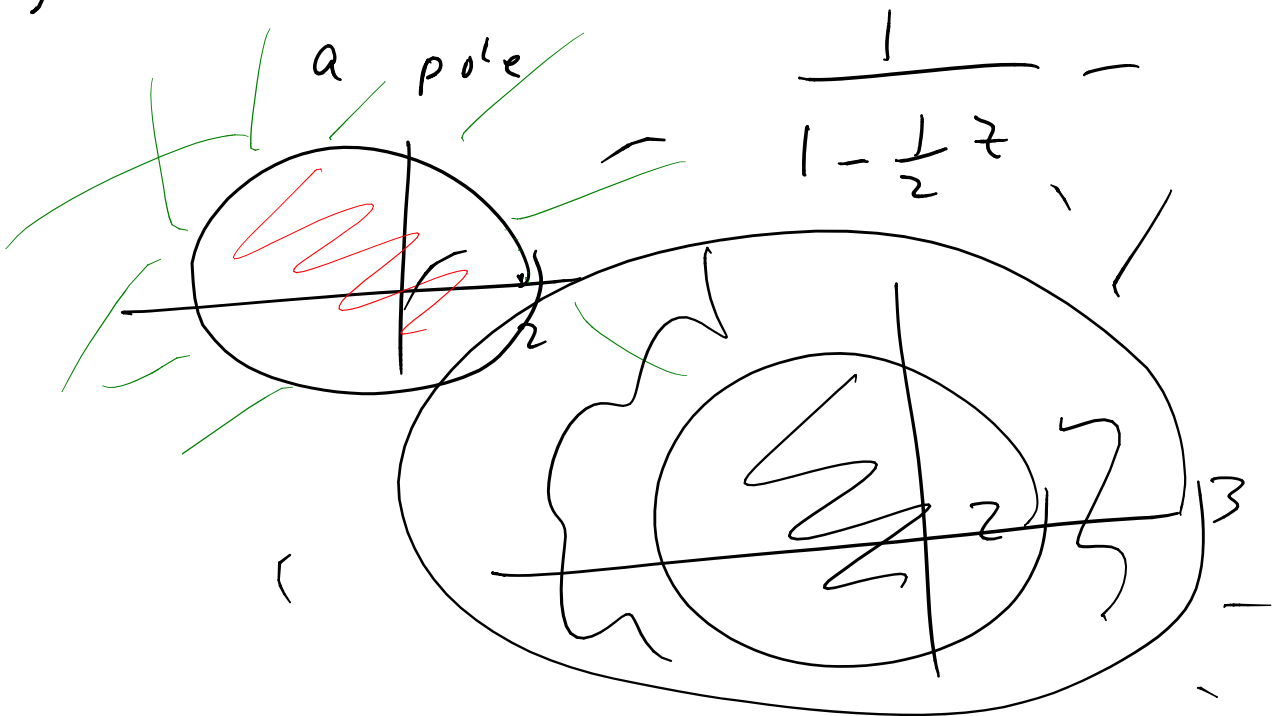
1. ROC is a fn of r and not ω

2. F.T. exists $\sum_n |x(n)| < \infty$

if ROC include unit circle.



3. ROC cannot include a pole

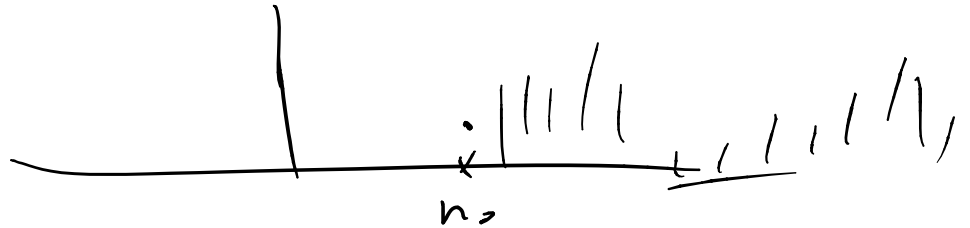


4. Finite length seq.

ROC everywhere except for $z=0$ or $z \rightarrow \infty$

Show: If seq. RHS \rightarrow ROC outside of some circle.

- $X(z)$ RHS \rightarrow non zero for some n_0 and large indices than n_0



- ROC outside of some circle \Leftrightarrow

If $X(z)$ converges for r_0 it also converges for $\forall r > r_0$

Assume $X(z)$ converges for some r_0 .

Show it also converges for $r > r_0$

$$\Rightarrow \sum_{n=-\infty}^{\infty} |X(n) r_0^{-n}| < \infty \quad \checkmark \text{ True}$$

Invoke RHS

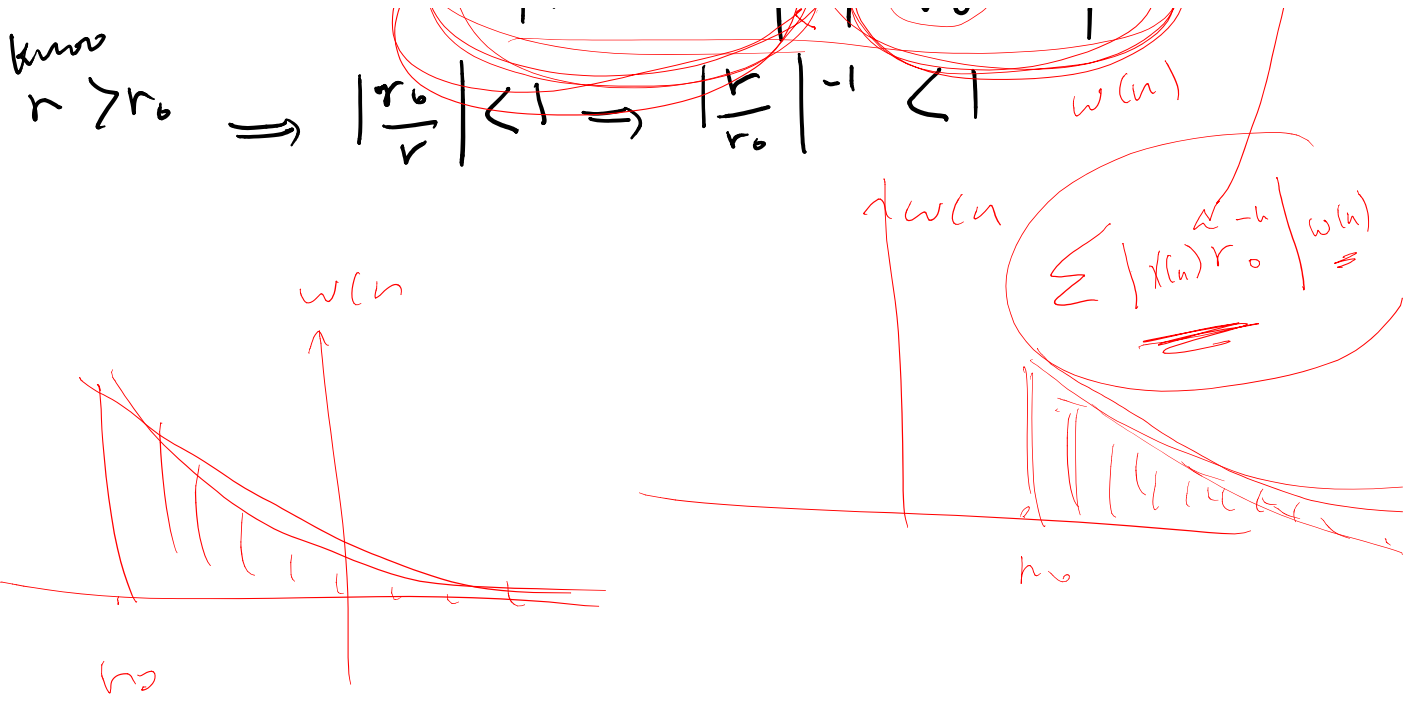
$$= \sum_{n=n_0}^{\infty} |X(n) r_0^{-n}| < \infty \quad \checkmark \text{ True.}$$

Show if $r > r_0$ show: $\sum_{n=n_0}^{\infty} |X(n) r^{-n}| < \infty$
 (Note: "want to show" is written below the equation with an arrow pointing to it)

$$|X(n) r^{-n}| = |X(n) r_0^{-n} \left(\frac{r}{r_0}\right)^{-n}|$$

$$= |X(n) r_0^{-n}| \left|\left(\frac{r}{r_0}\right)^{-n}\right|$$

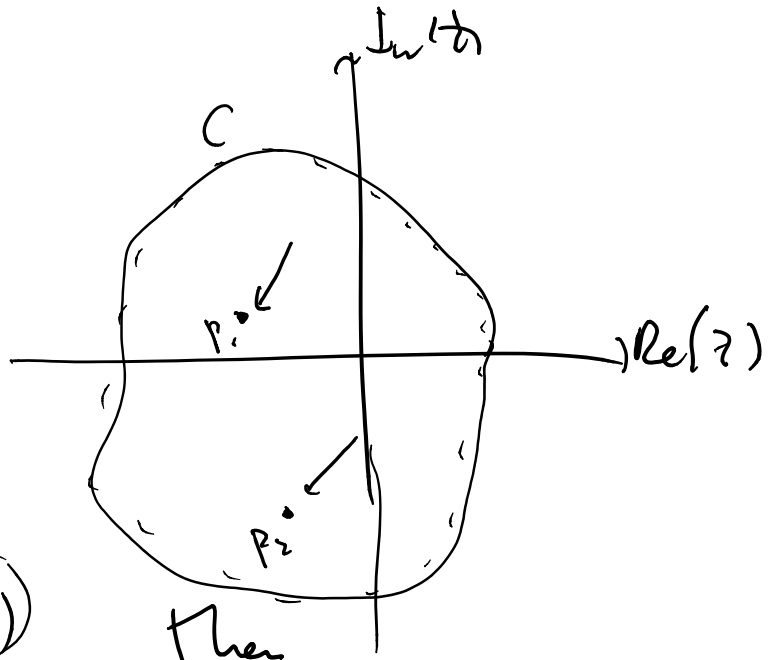
know $n > n_0$



Cauchy's Residue Theorem

$$\int_{\gamma} F(z) dz =$$

\sum Poles Residues of $F(z)$ at each pole inside contour C



If $F(z) = \frac{\phi(z)}{(z-z_0)^s}$ then

Residue $[F(z)]_{z=z_0} = \frac{1}{(s-1)!} \left[\frac{d^{s-1}}{dz^{s-1}} \phi(z) \right]_{z=z_0}$

EX

$F(z)$

$$\frac{1}{(z-2)^2}$$

$\phi(z) = 1$

$s = 2$

$z_0 = 2$

Residue $[F(z)]$

$$z=2 = \frac{1}{(2-1)!} \frac{d}{dz} [1] = 0$$

EX

$F(z)$

$$\frac{z^3}{z-2}$$

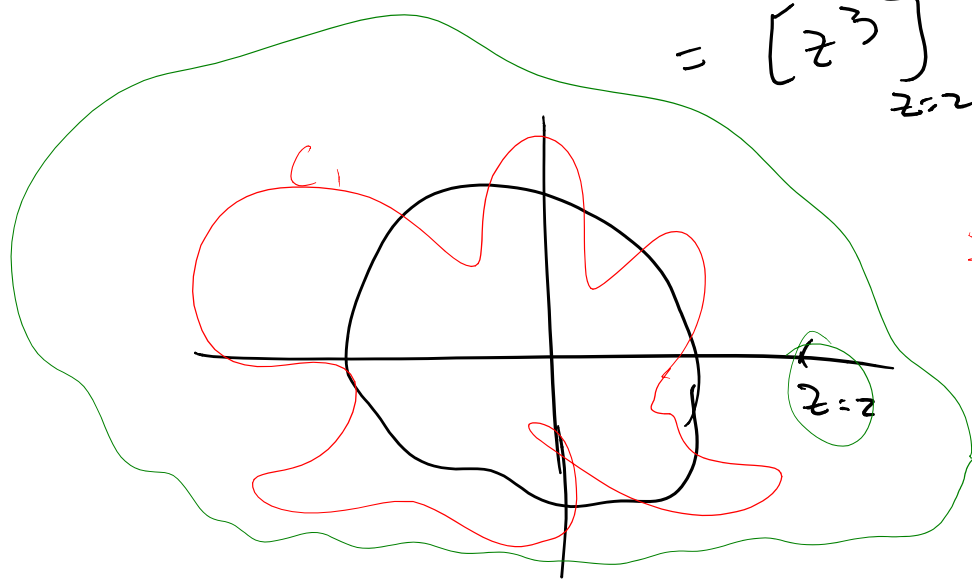
$\phi(z) = z^3$

$z_0 = 2$

$s = 1$

Residue $[F(z)]$

$$z=2 = \frac{1}{1} [\phi(z)]_{z=2} = [z^3]_{z=2} = 8$$



$\frac{1}{2\pi j} \oint_{C1} F(z) dz = 0$

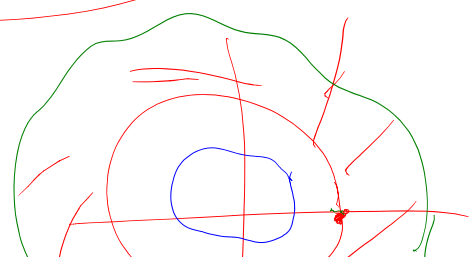
$\frac{1}{2\pi j} \oint_{C2} F(z) dz = 8$

Inverse Z-Transform:

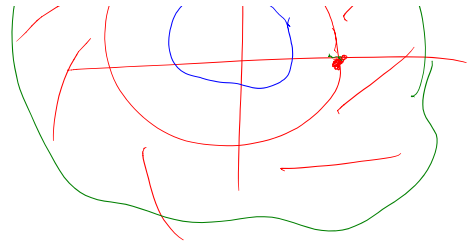
$$X(n) = \frac{1}{2\pi j} \oint_C$$

$$X(z) z^{n-1} dz$$

C: closed contour in z-plane



Contour in
RGC of
 $X(z) z^{n-1}$



Observation

$$\rightarrow \frac{1}{2\pi j}$$

$$\oint_C z^n dz = \begin{cases} 1 & n = -1 \\ 0 & \text{otherwise} \end{cases}$$

any closed contour encircling origin

(1) $n \neq -1$ $\rightarrow z^n$ has no poles inside C

$$\Rightarrow \frac{1}{2\pi j} \oint_C z^n dz = 0$$

(2) $n = -2$ $\frac{1}{2\pi j} \oint_C \frac{1}{z^2} dz$

pole $z=0$, order $2 = 5$ $\phi(z) = 1$

$$\text{Res}[z] = \frac{1}{(5-1)!} \left[\frac{d}{dz} \phi(z) \right] = 0$$

same arg $n = -3, -4, \dots$

(3) $n = -1$ simple pole, order 1 @ $z=0$

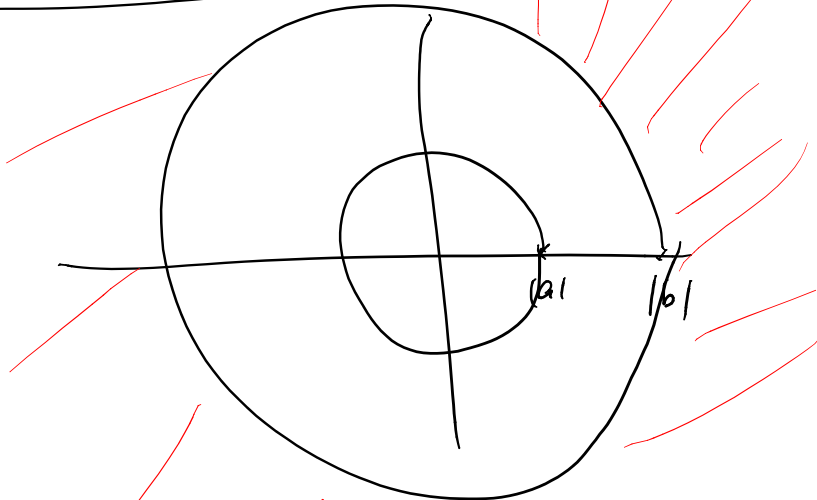
$$\frac{1}{2\pi j} \oint_C \frac{1}{z} dz =$$

$$\text{Res}[z=0] = \frac{1}{(1-1)!} \left[\phi(z) \right]_{z=0} = 1$$

$$\text{Res}[z=0] = \frac{1}{(1-1)^1} \quad \left\{ \begin{array}{l} z=0 \end{array} \right.$$

$$\Rightarrow \frac{1}{2\pi j} \oint_C z^n dz = \delta(n+1)$$

Ex $X(z) = \frac{1}{(1-a\bar{z}')(1-b\bar{z}')} \quad |b| > |a|$
 ROC: $|z| > |b|$



$$X(n) = \frac{1}{2\pi j} \oint_C \frac{z^{n+1}}{(z-a)(z-b)} dz$$

Consider,

- ① $n \geq -1 \Rightarrow$ 2 poles $z=a, z=b$
- ② $n < -1 \Rightarrow$ Repeated poles at the origin and $z=a, z=b$

$$\textcircled{1} \quad n \geq -1 \quad X(n) = \text{Res} \left[\frac{F(z)}{z-a} \right] + \text{Res} \left[\frac{F(z)}{z-b} \right]$$

$$= \left[\frac{z^{n+1}}{(z-b)} \right]_{z=a} + \left[\frac{z^{n+1}}{(z-a)} \right]_{z=b}$$

↑ $t=a$
↑ s

② $n < -1$ $n = -2, -3, \dots$

$n = -2 \rightarrow z=0, z=a, z=b$

$n = -3 \rightarrow$ double pole at $z=0 \rightarrow s=2$
 $z=a, z=b$

$n = -4 \rightarrow$ triple pole $z=0 \quad s=3$
 $z=a, z=b$

$n = -2$

$$X(z) = \left[\frac{1}{(z-a)z(z-b)} \right]_{z=0} + \left[\frac{1}{z(z-b)} \right]_{z=a} + \left[\frac{1}{(z)(z-a)} \right]_{z=b}$$

$$= \dots = 0$$

$n = -3, -4, -5, \dots \Rightarrow X(z) = 0$

$$X(z) = u(z) \left(\frac{a^{n+1}}{a-b} + \frac{b^{n+1}}{b-a} \right)$$

R.H.S.

$$X(z) = \frac{1}{(1-a\bar{z}')(1-b\bar{z}')}$$

A + B = P(z)

$$= \frac{\textcircled{A}}{(1-az^{-1})} + \frac{\textcircled{B}}{(1-bz^{-1})} = \frac{P(z)}{Q(z)}$$

Partial Fraction Expansion

Initial Value Theorem

If $x(n)=0$ for $n < 0$ Then

$$x(0) = \lim_{z \rightarrow \infty} X(z)$$

$$\begin{aligned} \lim_{z \rightarrow \infty} X(z) &= \lim_{z \rightarrow \infty} \left[\sum_{n=0}^{+\infty} x(n) z^{-n} \right] \\ &= \lim_{z \rightarrow \infty} [x(0) + x(1)z^{-1} + x(2)z^{-2} + \dots] \\ &= x(0) \end{aligned}$$

Methods for Inverse Z.T.

- ① Longly Divide the
- ② Partial Fraction Exp
- ③ Direct way: inspect.

$$X(z) = \frac{5 + 20z^{-1}}{z^2}$$

$$X(z) = \sum x(n) z^{-n} = \frac{x(0) + x(1)z^{-1}}{z^2} +$$

$$X(z) = \sum_{n=-\infty}^{\infty} x(n) z^{-n} = \frac{x(0) + x(1)z^{-1} + \dots}{x(2)z^{-2} + \dots}$$

$$x(n) = \begin{cases} 5 & n=0 \\ 20 & n=6 \end{cases}$$

(4) Table look up

(5) Use properties

(6) Series Expansion

$$X(z) = \log(1 + a z^{-1}) \quad \text{Roc } |a z^{-1}| < 1$$

$$\log(1+x) = x - \frac{x^2}{2} + \frac{x^3}{3} - \frac{x^4}{4} + \dots$$

$$|x| < 1$$

$$\log(1 + a z^{-1}) = a z^{-1} - \frac{(a z^{-1})^2}{2} + \frac{(a z^{-1})^3}{3} - \frac{(a z^{-1})^4}{4} + \dots$$

$$X(z) = \sum_n x(n) z^{-n} = \dots + x(-2) z^2 + x(-1) z + x(0) + x(1) z^{-1} + x(2) z^{-2} + x(3) z^{-3} + \dots$$

$$x(2) = -\frac{a^2}{2}$$

$$x(3) = \frac{a^3}{3}$$

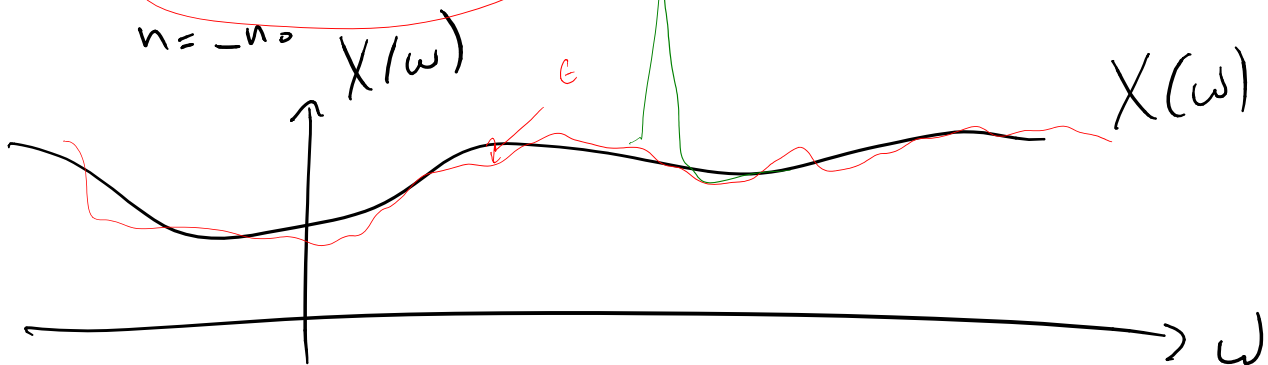
$$x(1) = a$$

$$\sum_n x(n) e^{-j\omega n}$$

\exists a continuous function of ω called $X(\omega)$ s.t.

$\forall \epsilon > 0, \omega \exists n_0$ s.t.

$$\left| \sum_{n=-n_0}^{+n_0} x(n) e^{-j\omega n} - X(\omega) \right| < \epsilon$$

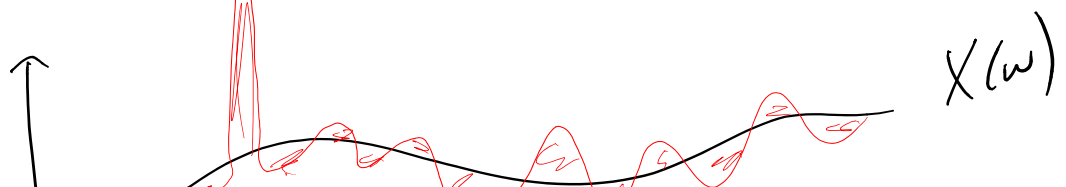


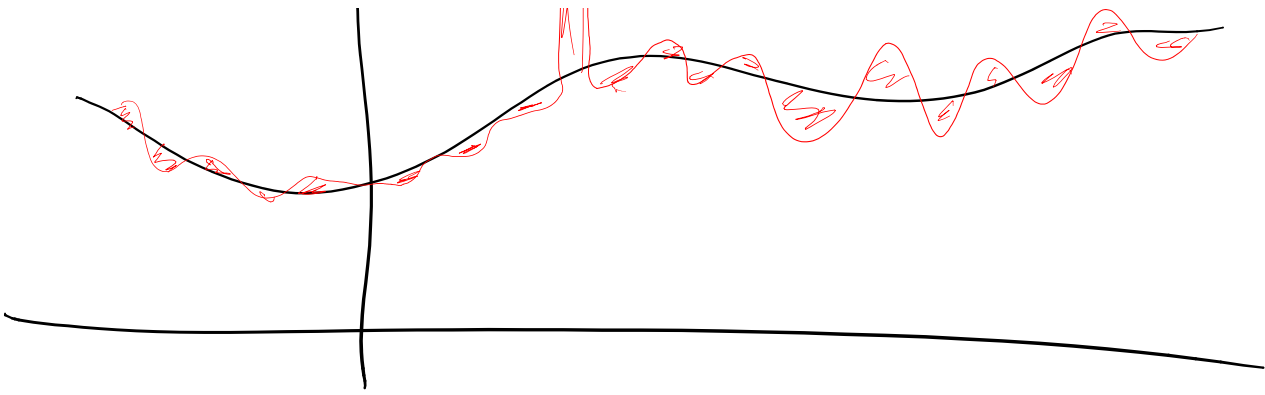
MSE Convergence. (MSE)

$\sum_n x(n) e^{-j\omega n}$ converges in MSE sense to a known given function $X(\omega)$ if

$\forall \epsilon, \exists n_0$ s.t.

$$\int_{-\pi}^{+\pi} \left| \sum_{n=-n_0}^{+n_0} x(n) e^{-j\omega n} - X(\omega) \right|^2 d\omega < \epsilon$$





① If $\sum_n |x(n)| < \infty$ is absolutely summable \Rightarrow Then $\sum_n x(n)e^{-j\omega n}$ converges uniformly \Rightarrow D.T.F.T exist and D.T.F.T is a continuous function

AND whatever it converges to $\sum_n x(n)e^{-j\omega n}$ if I plug into $\int X(\omega)e^{j\omega n} d\omega$ get $x(n)$ back

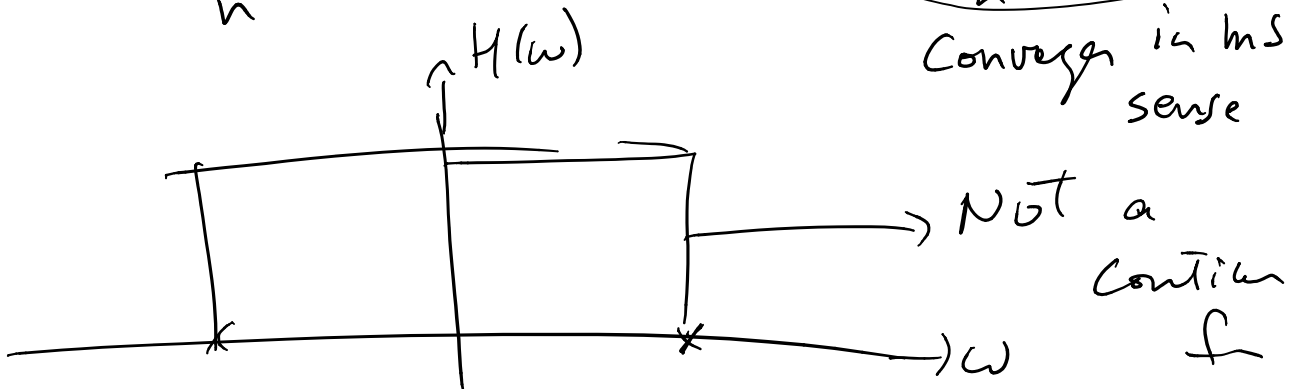
Ex $(\frac{1}{2})^n u(n) = x(n) \rightarrow \frac{1}{1 - \frac{1}{2}e^{-j\omega}}$
 $z^n u(n) = x(n) \rightarrow$ F.T doesn't exist

② If $x(n)$ is square summable $\leftarrow \sum_n |x(n)|^2 < \infty$ $\leftarrow \sum_n |x(n)e^{-j\omega n}|^2 < \infty$

$$\sum_n x^2(n) < \infty$$

$$\rightarrow \sum_n x(n) e^{-j\omega n}$$

Converges in MSE sense



$$\int_{-\pi}^{+\pi} H(\omega) e^{j\omega n} d(\omega) = \frac{\sin \omega_c n}{\pi n} = h(n)$$

$$\sum_n \frac{1}{n} \text{ diverges so does } \sum \left| \frac{\sin \omega_c n}{\pi n} \right| \text{ diverges}$$

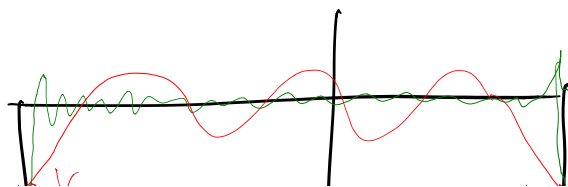
$h(n)$ is not absolutely summable

However $h(n)$ square sum

$$\sum_n \frac{\sin^2 \omega_c n}{\pi^2 n^2} \text{ Converges because } \sum \frac{1}{n^2} \text{ Converges}$$

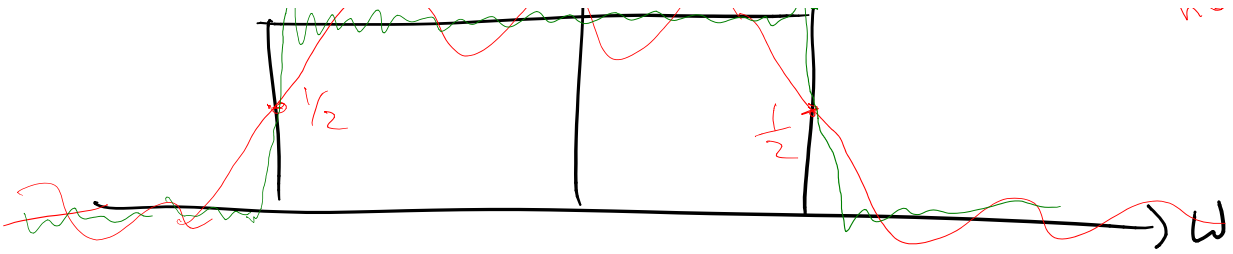
$$\sum_{-n_0}^{+n_0} h(n) e^{-j\omega n} \text{ Converges in a MSE sense.}$$

At $\omega = \omega_c$ $\sum h(n) e^{-j\omega n}$ converges $\rightarrow \frac{1}{2}$ tho.

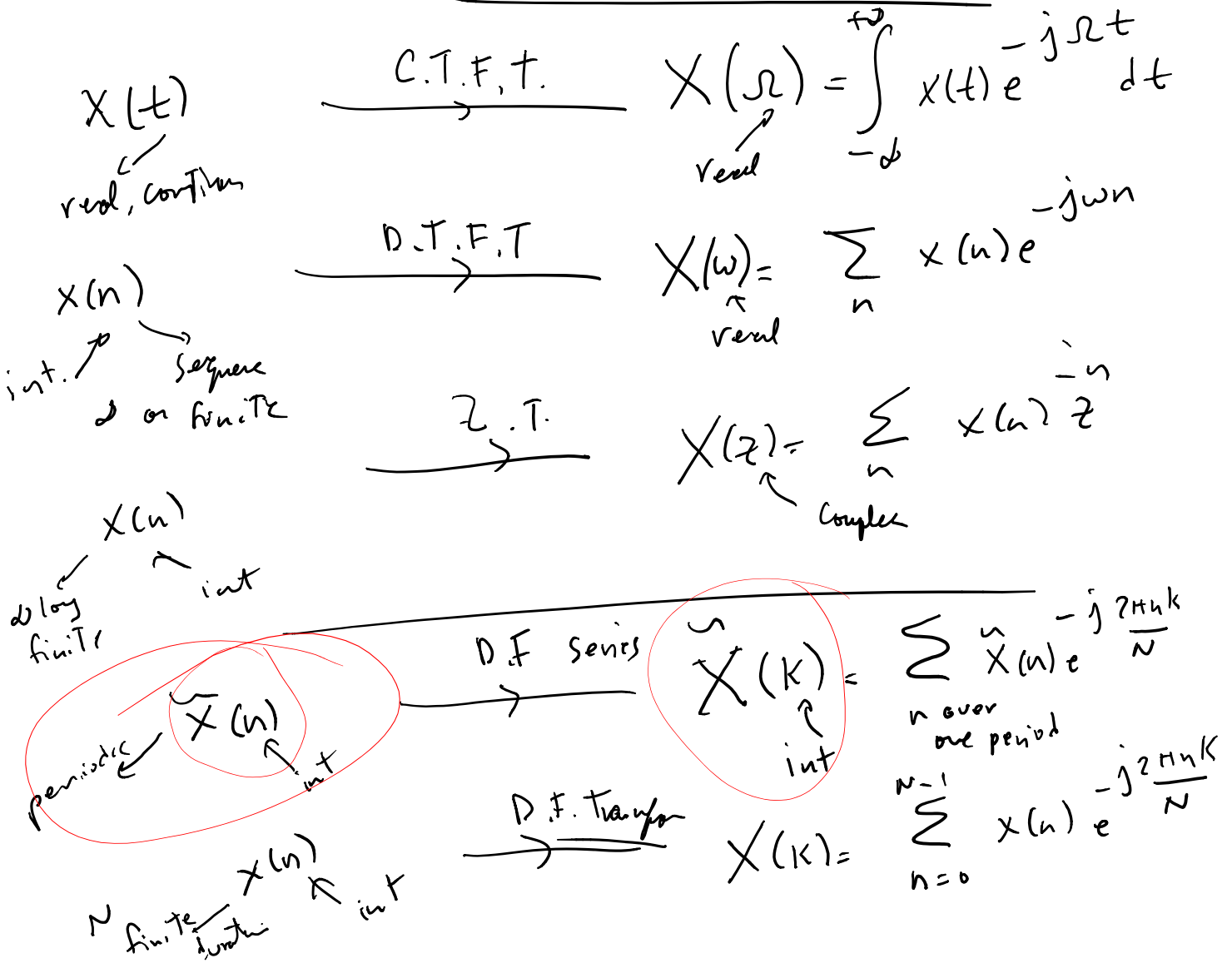


$$\omega_0 = 10$$

$$\omega_c = 60$$



Discrete Fourier Series



DFS = Discrete Fourier Series.

... Time signal with period N

DTFT of periodic discrete time signal with period N
 $\tilde{X}(n) = \tilde{X}(n + kN)$ any integer

$$e_k(n) = e^{j\frac{2\pi nk}{N}} \quad k=0, \dots, N-1$$

$$\tilde{X}(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) e^{j\frac{2\pi nk}{N}}$$

Discrete $\tilde{X}(n)$ along exponential $e_k(n)$

Show: $e_k(n)$ is periodic:

$$e_k(n) = e_{k+rN}(n) \quad \text{arbitrary int.}$$

$$e_0(n) = e_N(n) = e_{2N}(n) = \dots$$

$$e_1(n) = e_{N+1}(n) = e_{2N+1}(n) = \dots$$

Proposed: $X(k) = \sum_{n=0}^{N-1} X(n) e^{-j\frac{2\pi nk}{N}}$ RMS

Proof:

$$RHS = \sum_{n=0}^{N-1} \left(\frac{1}{N} \sum_{l=0}^{N-1} X(l) e^{j\frac{2\pi nl}{N}} \right) e^{-j\frac{2\pi nk}{N}}$$

$$= \sum_{l=0}^{N-1} X(l) \left[\frac{1}{N} \sum_{n=0}^{N-1} e^{j\frac{2\pi n(l-k)}{N}} \right]$$

$A = \delta(l-k-rN)$ any r

$$X(k+rN) = X(k) = X(k+N)$$

any.

$$= X(k+rN) = X(k) = X(k+N) \\ = \overline{X(k+2N)} = X(k+3N) \\ = \dots$$

Show A: Case 1: $l-k = rN$ is an integer multiple of N

$$A = \frac{1}{N} \sum_{n=0}^{N-1} e^{j2\pi rN \frac{n}{N}} = 1$$

Case 2: $l-k \neq rN$

$$A = \frac{1}{N} \sum_{n=0}^{N-1} e^{j2\pi (l-k) \frac{n}{N}}$$

Recall: $\sum_{n=0}^{N-1} \alpha^n = \frac{1-\alpha^N}{1-\alpha}$ ← math

$$A = \frac{1}{N} \frac{1 - e^{j2\pi (l-k) \frac{N}{N}}}{1 - e^{j2\pi (l-k) \frac{1}{N}}} = 0$$

$$A = \delta(l-k-rN)$$

$$\tilde{X}(n) = \sum_{k=0}^{N-1} \tilde{X}(k) e^{j2\pi nk/N}$$

$$\tilde{X}(k) = \frac{1}{N} \sum_{n=0}^{N-1} \tilde{X}(n) e^{-j2\pi nk/N}$$

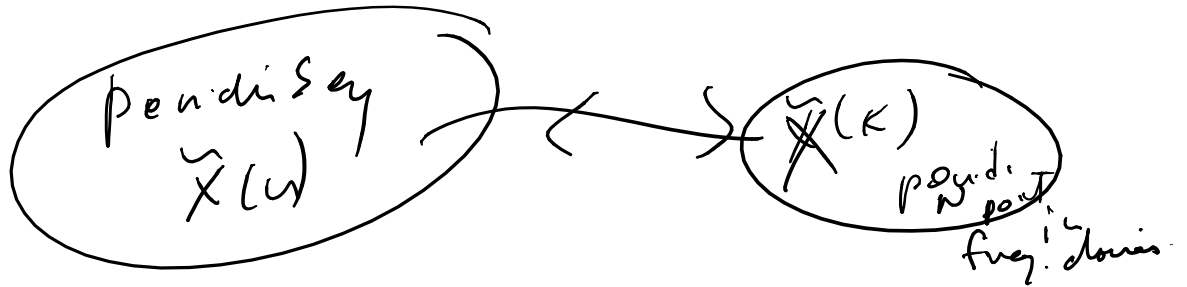
DFT pair

Synthesis

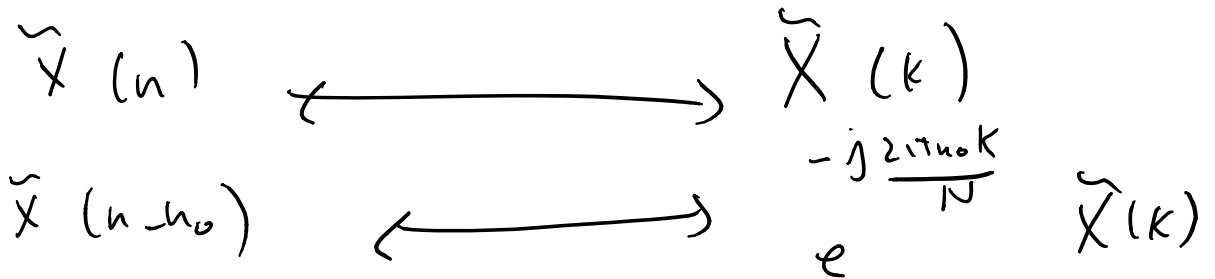
Analysis

DFT Pair

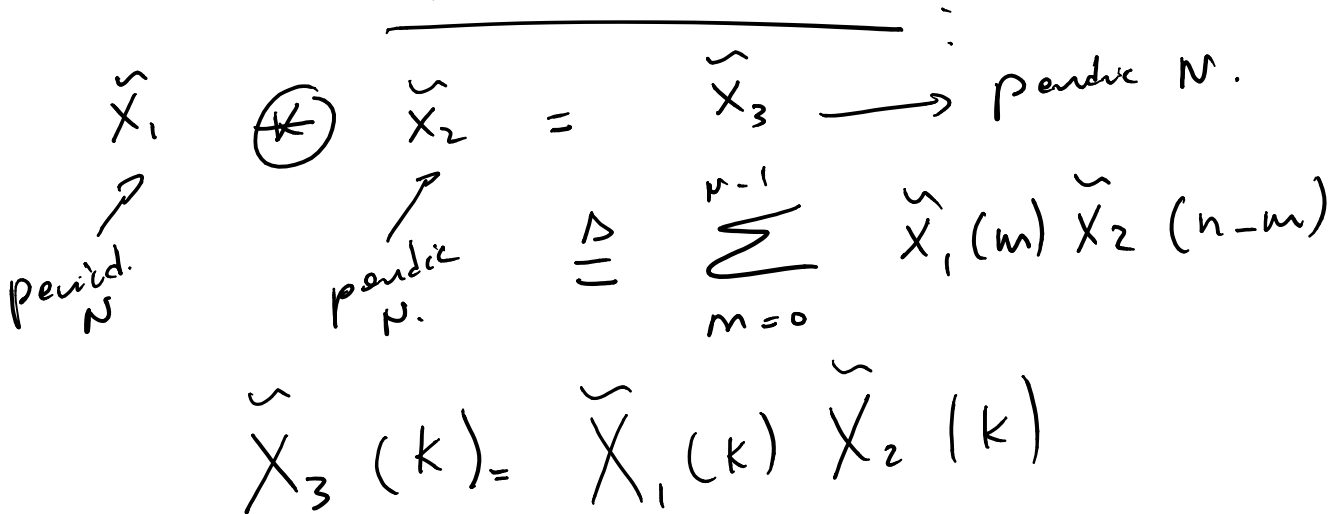
analysis



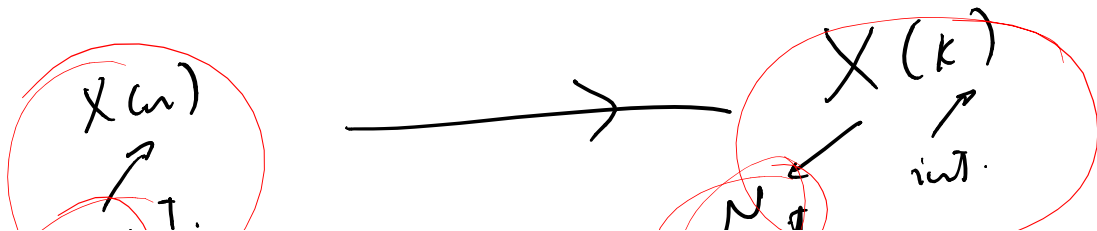
Shift Property.



Period Convolution

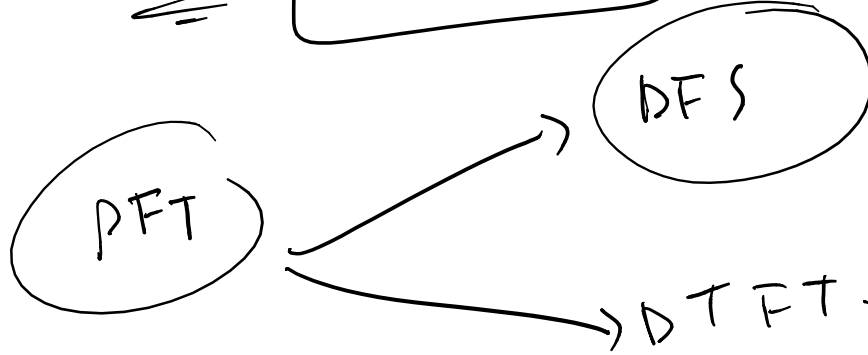


DFT = Discrete Fourier Transform.



N point.

N point int.



First Approx: DFT \leftrightarrow DFS

1. Start with $x(n)$
2. "Periodicize" $x(n)$ to get $\tilde{x}(n)$

$$\tilde{x}(n) = \sum_{r=-\infty}^{+\infty} x(n+rN)$$

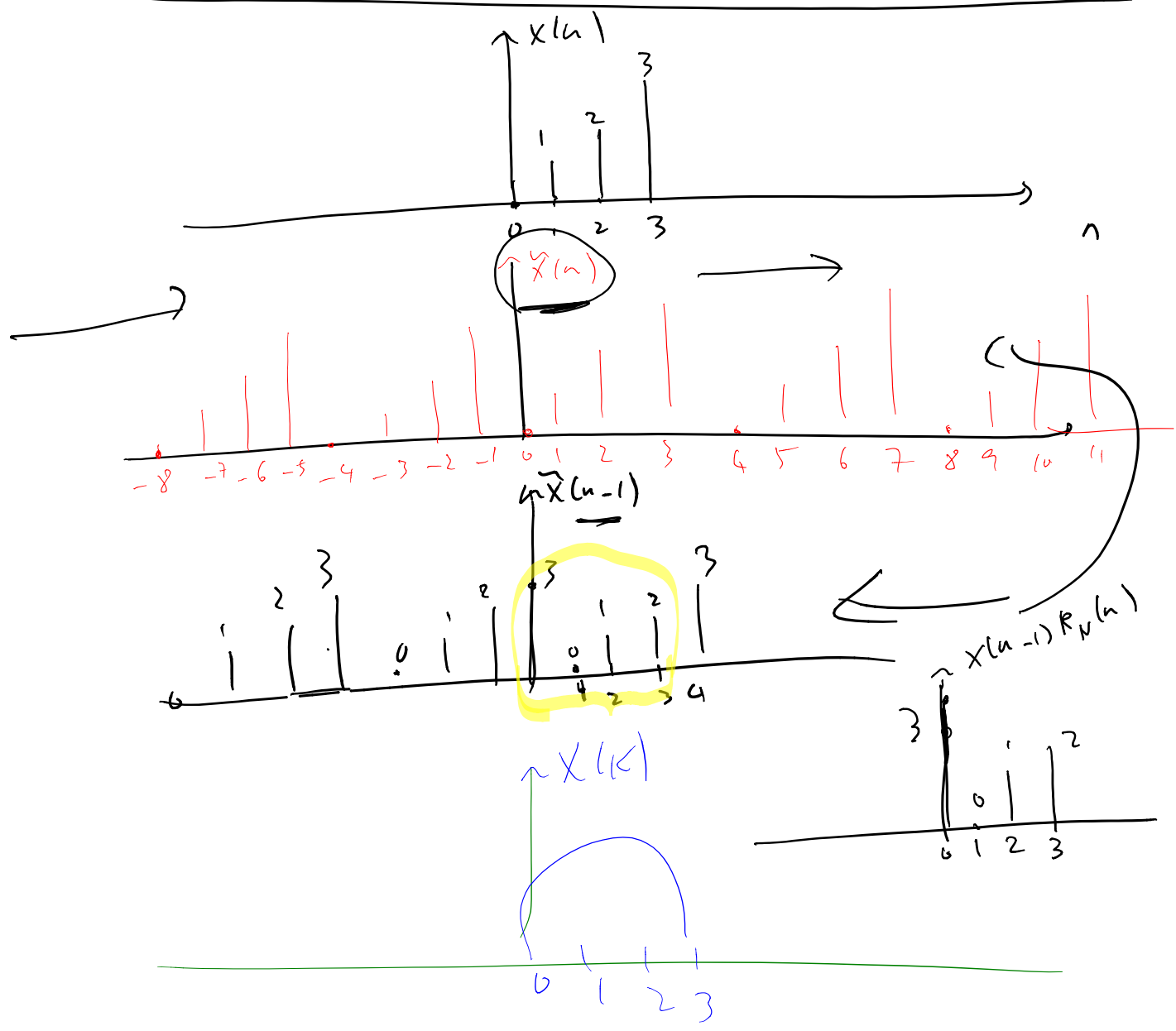
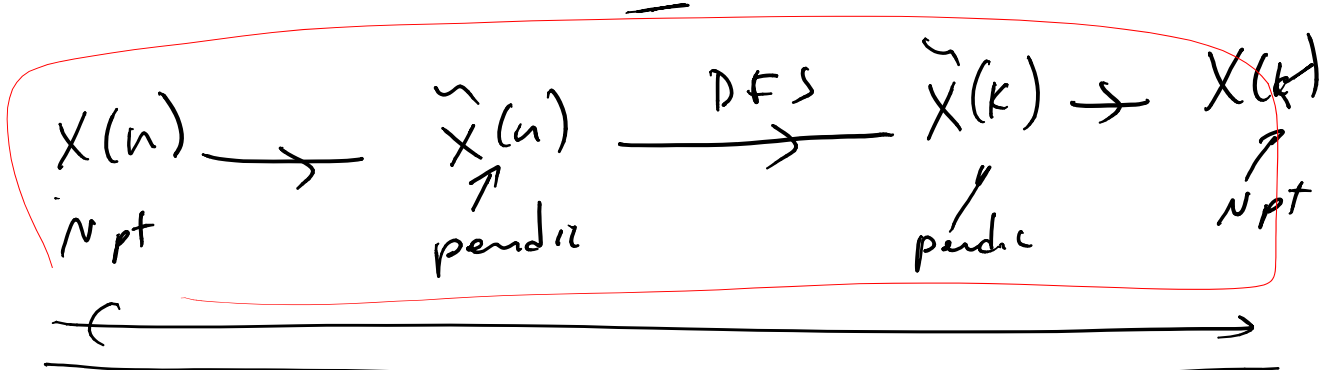
$$x(n) = \tilde{x}(n) \underbrace{R_N(n)}_{\text{Extract one period}}$$

$$R_N(n) = \begin{cases} 1 & n=0, \dots, N-1 \\ 0 & \text{otherwise} \end{cases}$$

3. Compute DFS of $\tilde{x}(n) \rightarrow \tilde{X}(k)$
4. Take one period of $\tilde{X}(k)$ to get $V(k)$ DFT of $x(n)$

$X(k)$ - DFT of $x(n)$

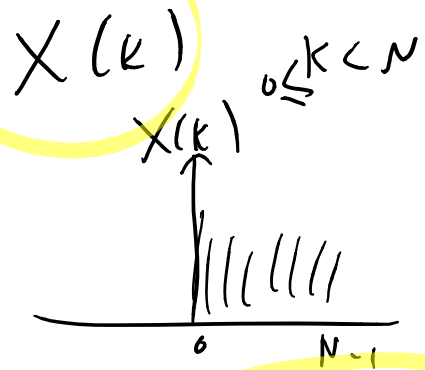
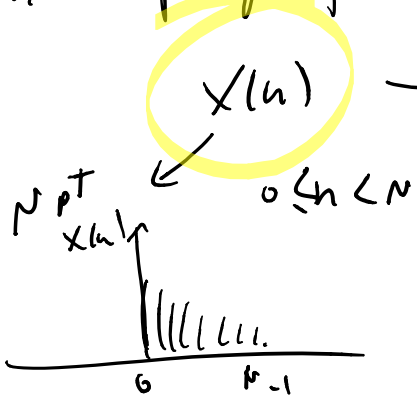
$$X(k) = \sum_{n=0}^{N-1} x(n) R_N(k, n)$$



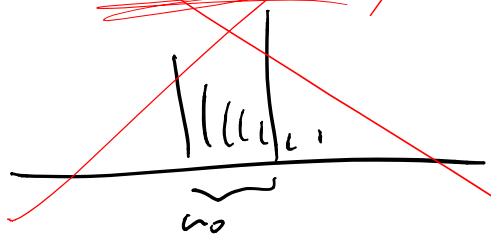
$N_{PT} \longrightarrow N_{PT}$

Properties of DFT

① Shift property



~~$x(n-n_0)$??~~



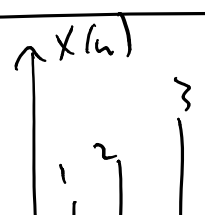
$\xleftarrow{\text{IDFT}}$

$X(k) e^{-j2\pi n_0 k / N}$

$X(n-n_0) R_N(n)$

$\xrightarrow{\text{DFT}}$

$X(k) e^{j2\pi n_0 k / N}$



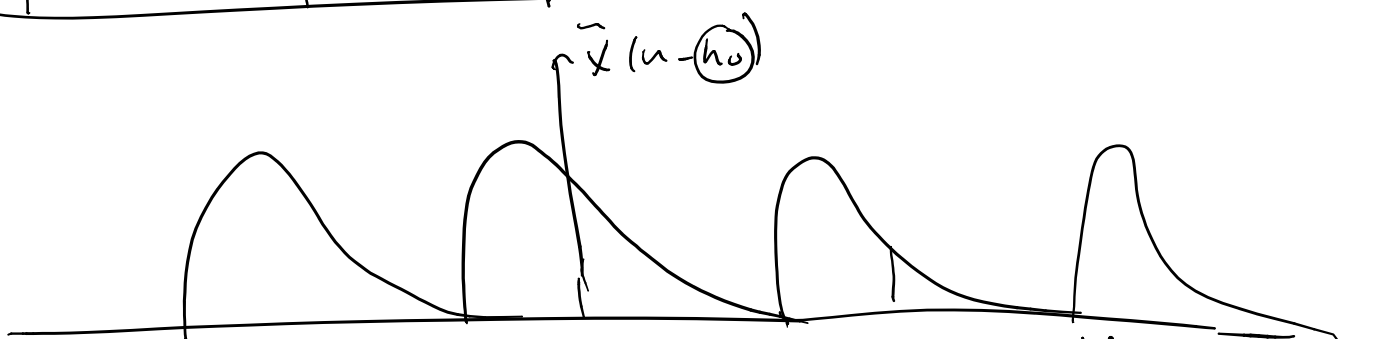
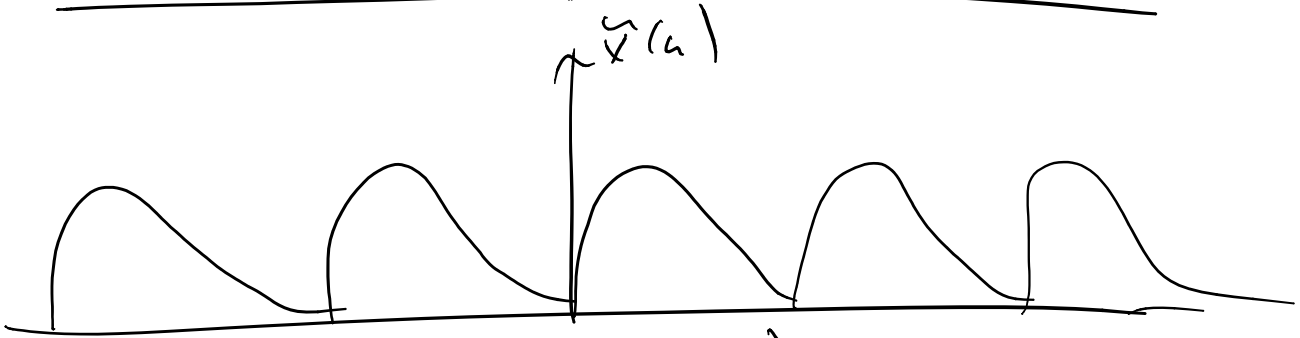
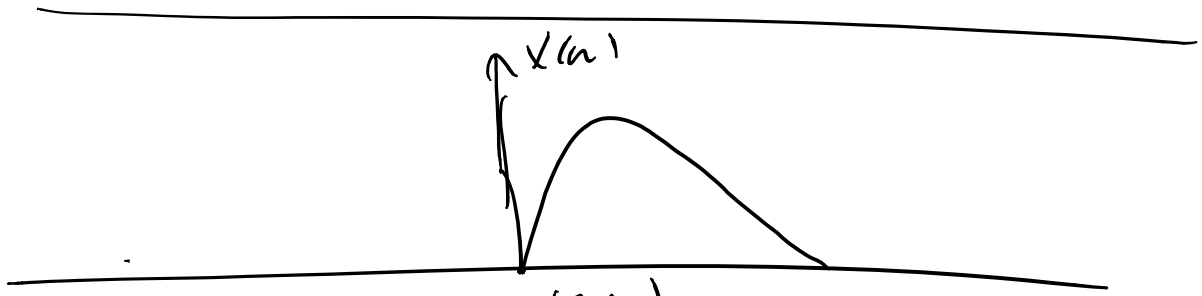
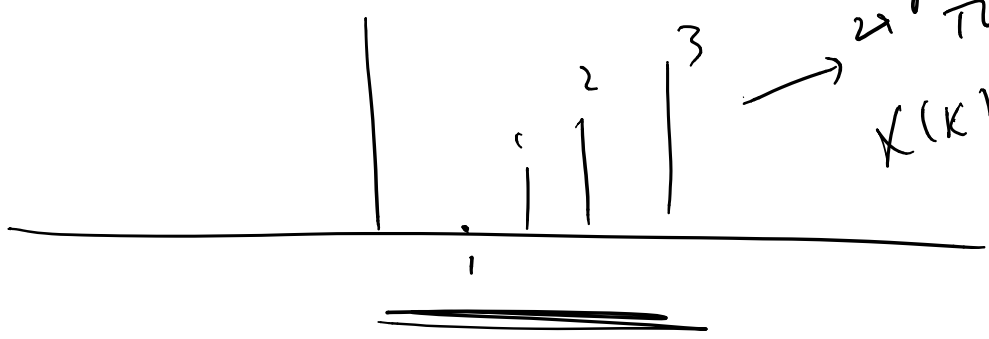
\longrightarrow 4PT & 9.
4PT DFT.

$x(n-1)$

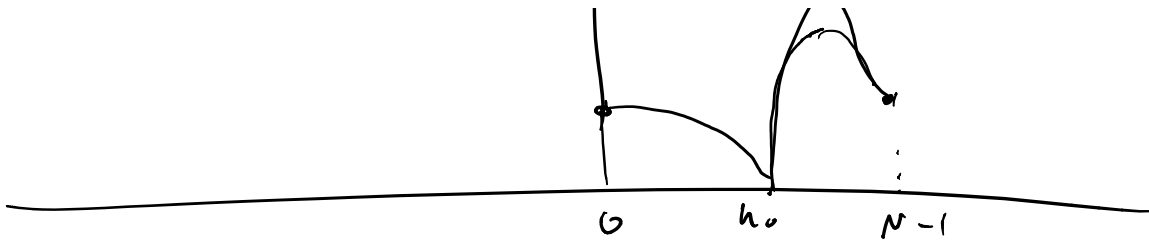


4-PT DFT

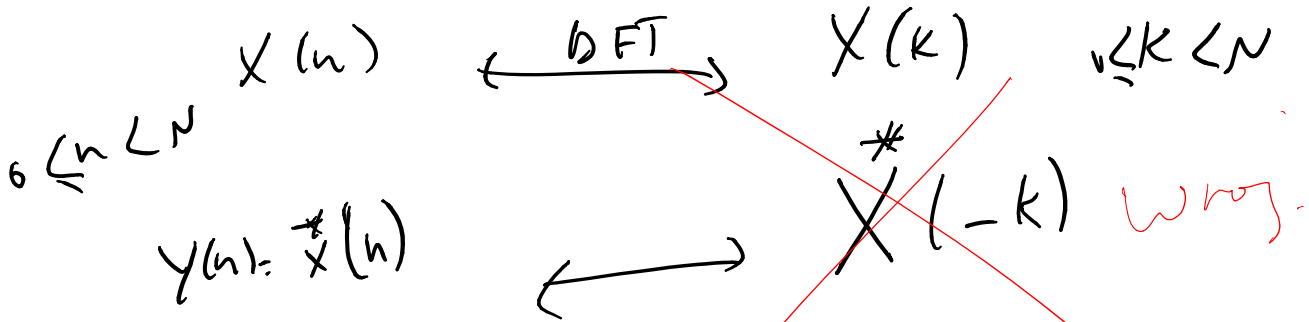
→ 2-PT DFT of this signal
 $X(k)e^{j\omega_0 k}$



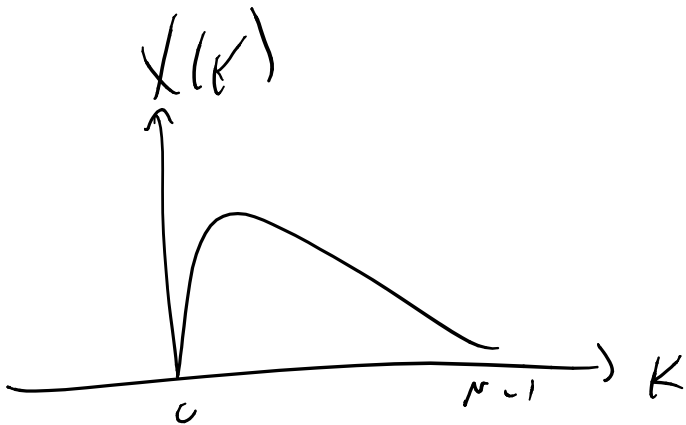
$$\text{IDFT} \left\{ X(k) e^{j2\pi h_0 k} \right\}$$



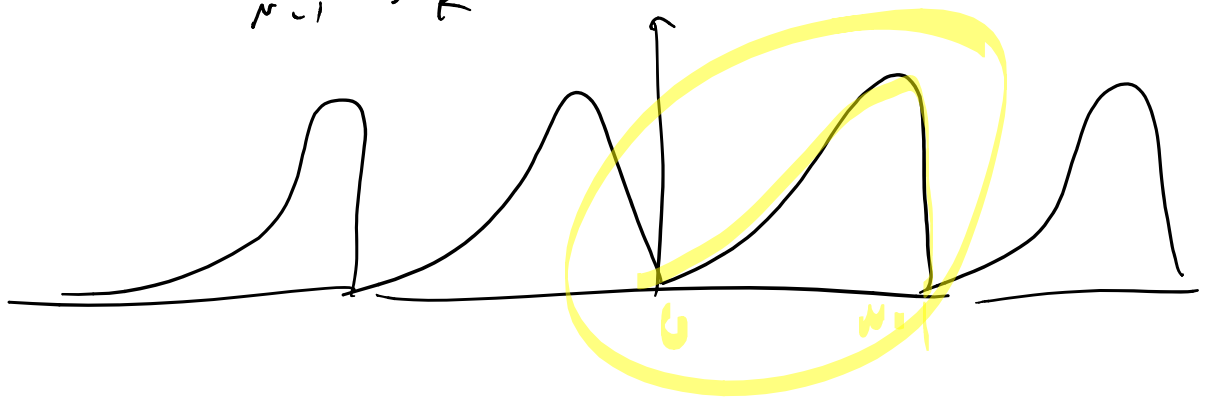
Property of DFT.



$X^*(-k) R_N(k)$

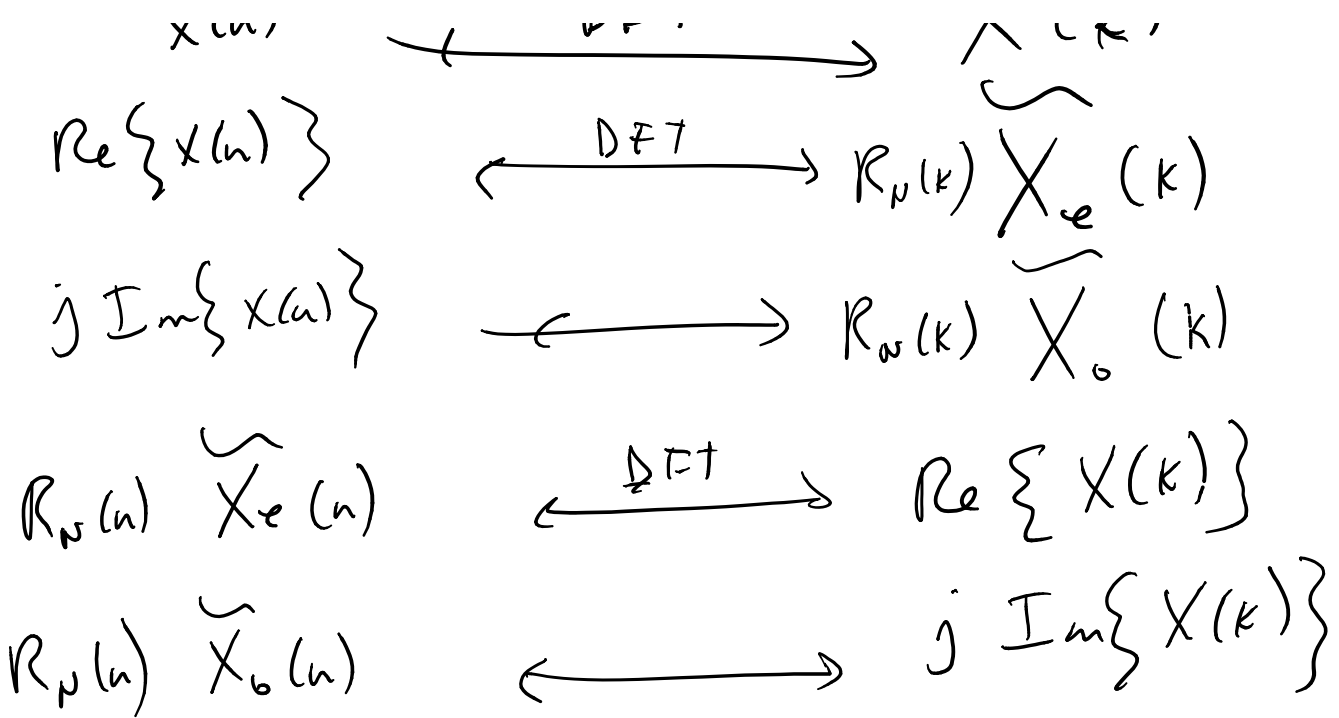


$X^*(-k) R_N(k)$



Symmetry Properties





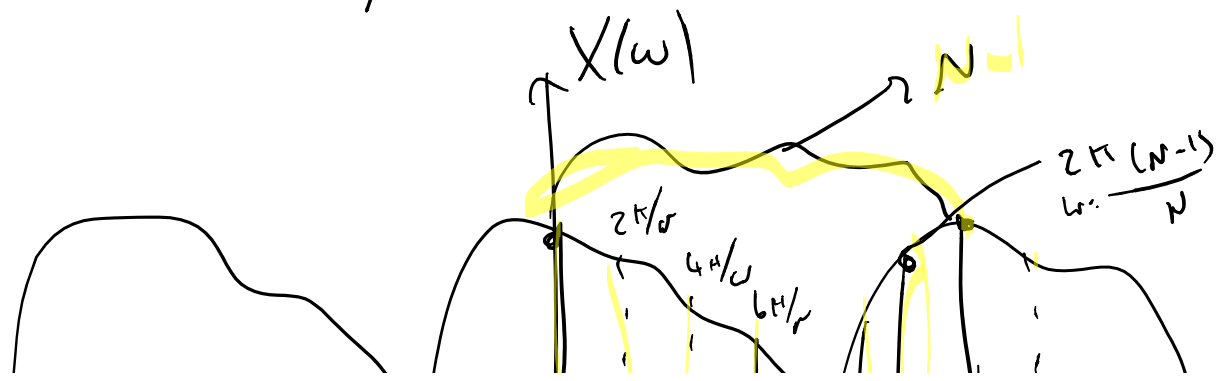
DFT To DTFT.



$x(n) \rightarrow X(\omega) = \sum x(n) e^{-j\omega n}$

$X(k) = \sum_{n=0}^{N-1} x(n) e^{-j \frac{2\pi n k}{N}}$ ← DFT pair.

$X(k) = \text{DFT} = [X(\omega)]_{\omega = \frac{2\pi k}{N}}$





$$X(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) e^{+j \frac{2\pi n k}{N}}$$

$$X(k) = \sum_{n=0}^{N-1} X(n) e^{-j \frac{2\pi n k}{N}}$$

sy
n/ks

analysis

Defn of DFT

$$X(k) = N \text{pt DFT of } x = \begin{cases} \sum_{n=0}^{N-1} x(n) e^{-j \frac{2\pi n k}{N}} & 0 \leq k < N \\ 0 & \text{otherwise} \end{cases}$$

$$x(n) = \begin{cases} \frac{1}{N} \sum_{k=0}^{N-1} X(k) e^{j \frac{2\pi n k}{N}} & 0 \leq n < N \\ 0 & \text{otherwise} \end{cases}$$

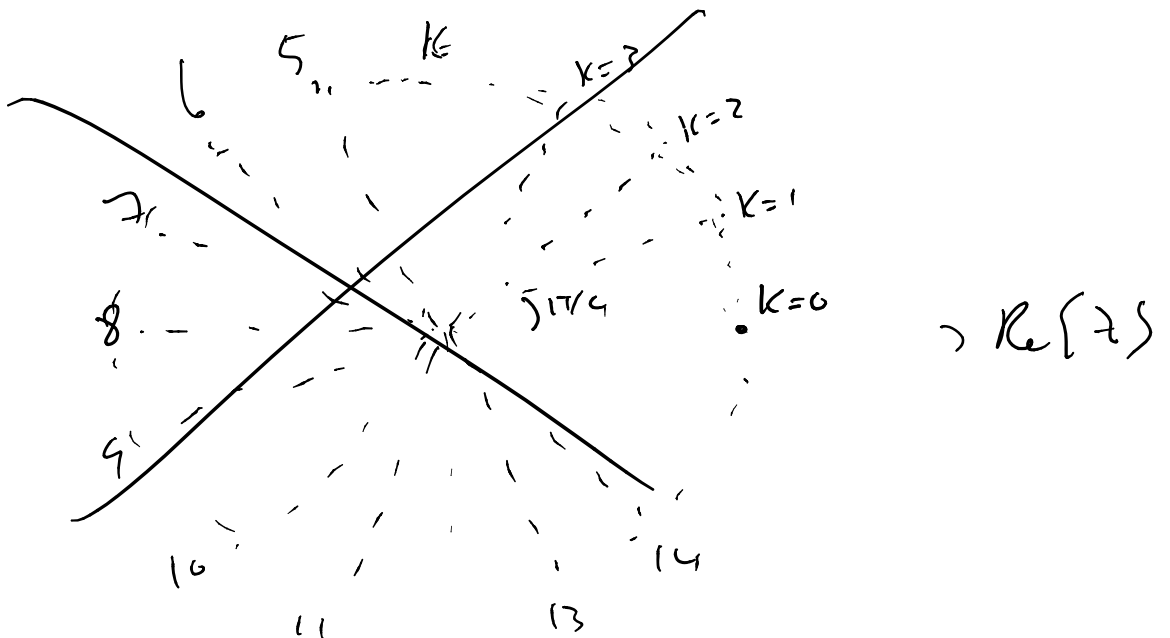
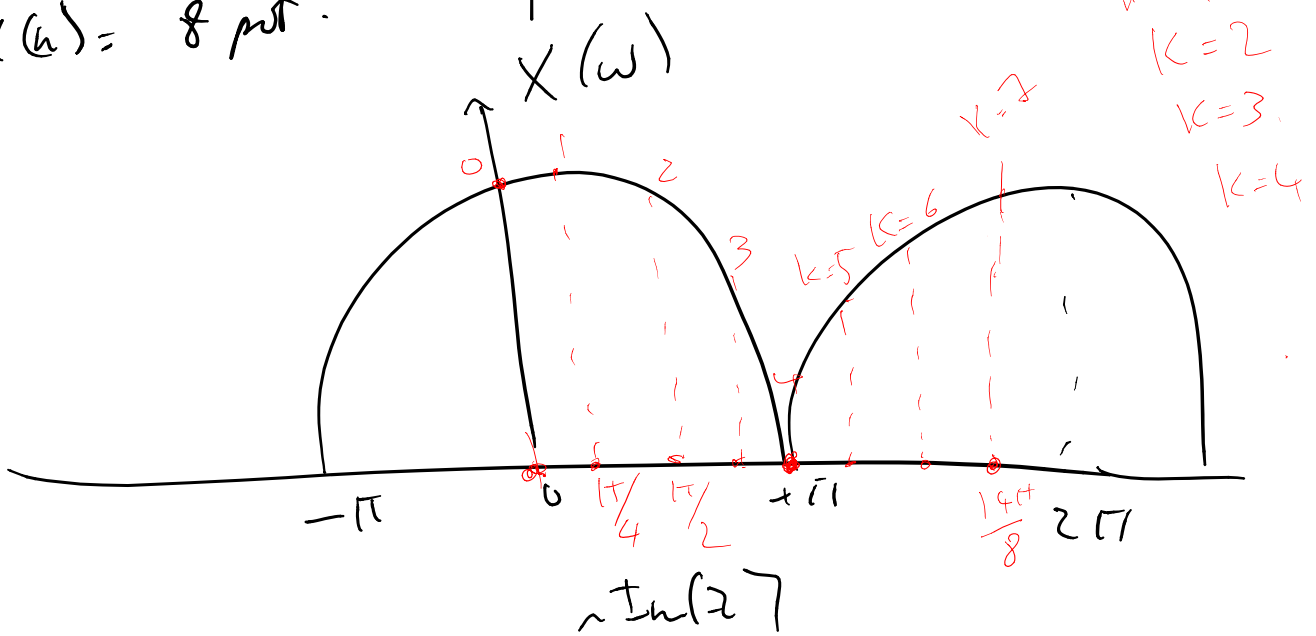
DFT ↔ DTFT

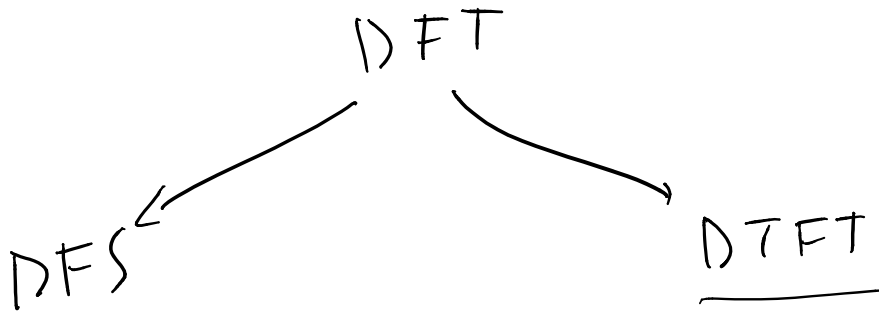
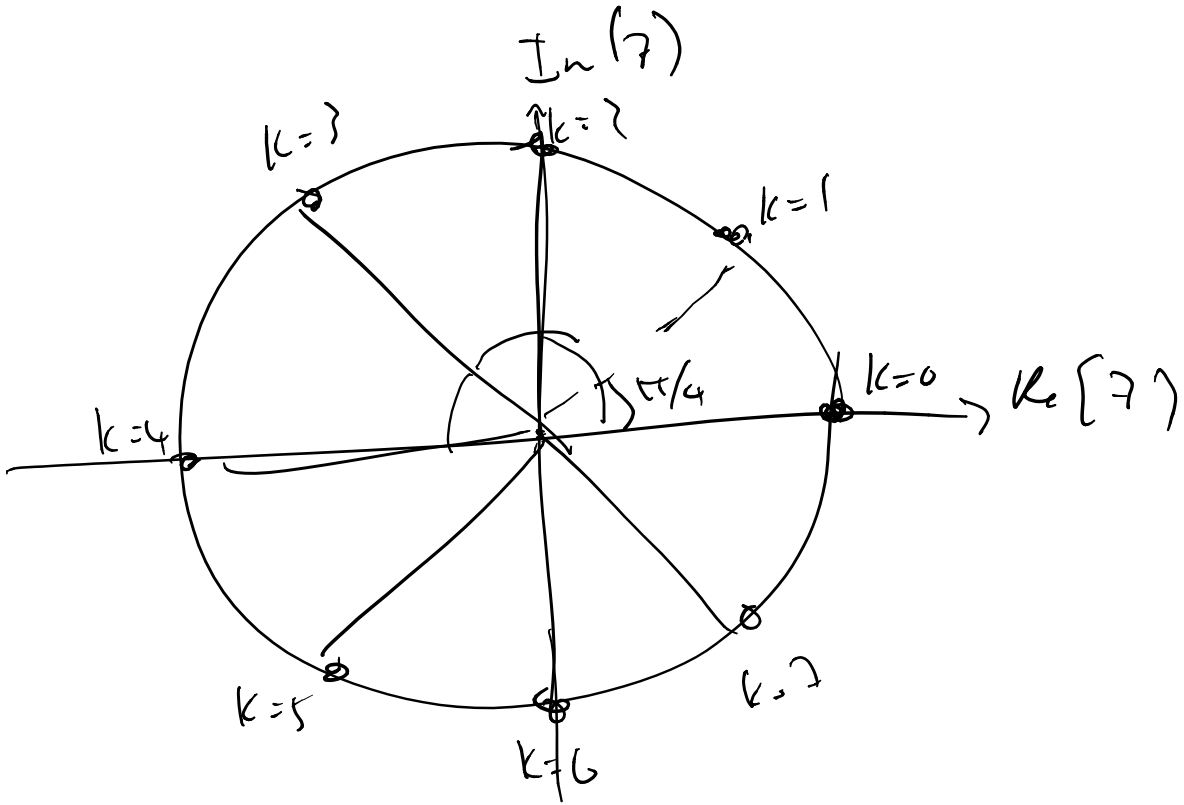
$$X(\omega) = \sum_{n=-\infty}^{\infty} x(n) e^{-j\omega n}$$

$$\left. \begin{aligned}
 X(\omega) &= \sum_n x(n) e^{-j\omega n} \\
 X(k) &= \sum_n x(n) e^{-j \frac{2\pi n k}{N}}
 \end{aligned} \right\} \propto k \text{ or } \omega$$

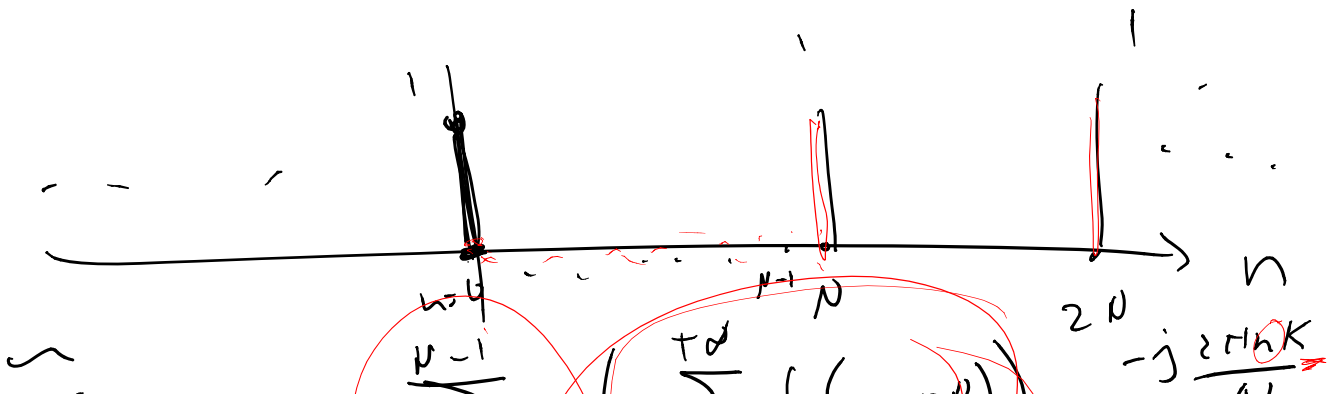
$$X(k) = \begin{cases} [X(\omega)]_{\omega = \frac{2\pi k}{N}} & 0 \leq k < N \\ 0 & \text{otherwise} \end{cases}$$

$x(n) = 8 \text{ pts.}$





$$\tilde{X}(n) = \sum_{r=-\infty}^{\infty} \delta(h + rN)$$



$$\tilde{X}(k) = \sum_{n=0}^{N-1} \left(\sum_{r=-\infty}^{+\infty} \delta(n+rN) \right) e^{-j \frac{2\pi n k}{N}}$$

$k=0$

$$\tilde{X}(k) = \{ \dots \}$$

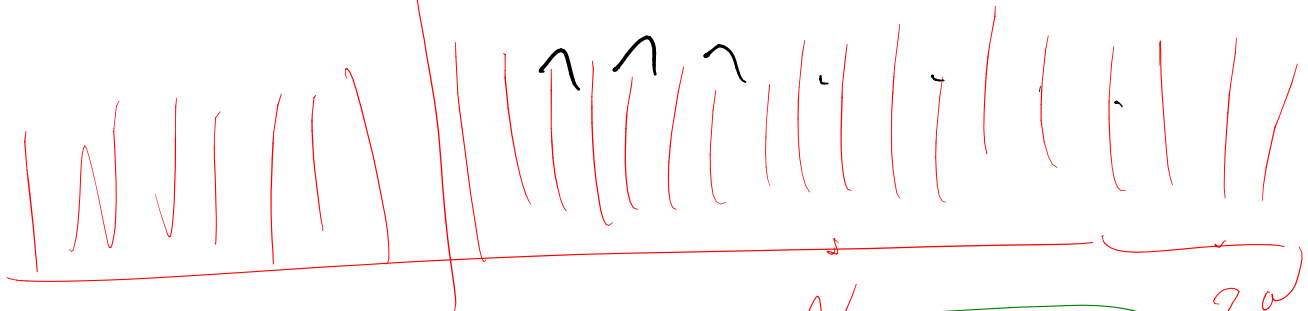
$k=0$

$k=1$

$k=N-1$

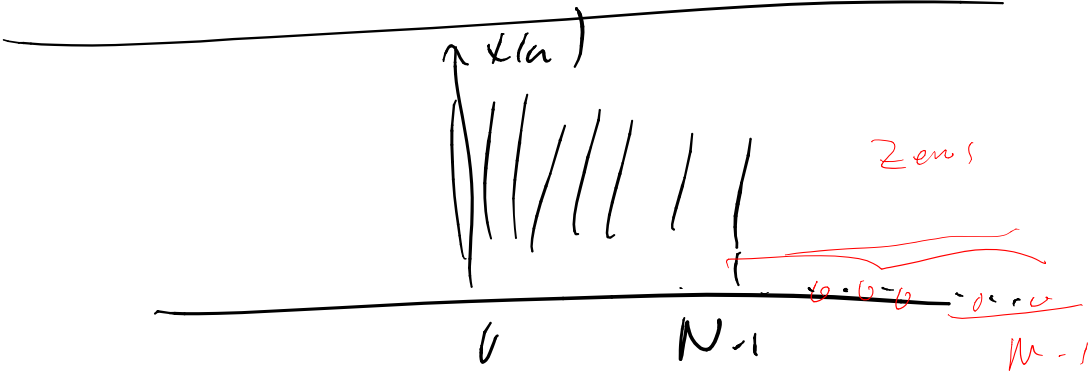
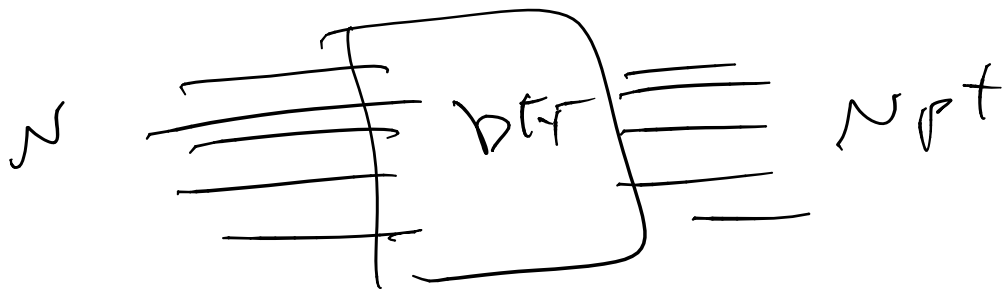
$$\tilde{X}(k) = 1$$

$$\tilde{X}(k)$$



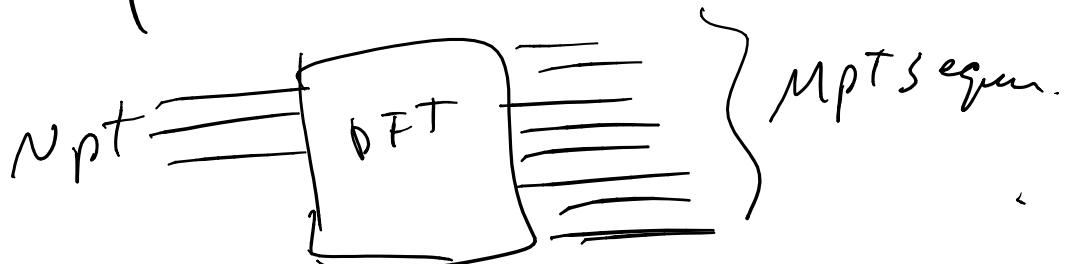
$$\tilde{X}(n) = \frac{1}{N} \sum_{k=0}^{N-1} e^{j \frac{2\pi n k}{N}} = \sum_{r=-\infty}^{+\infty} \delta(n+rN)$$

identity of δ
Fourier
fun.

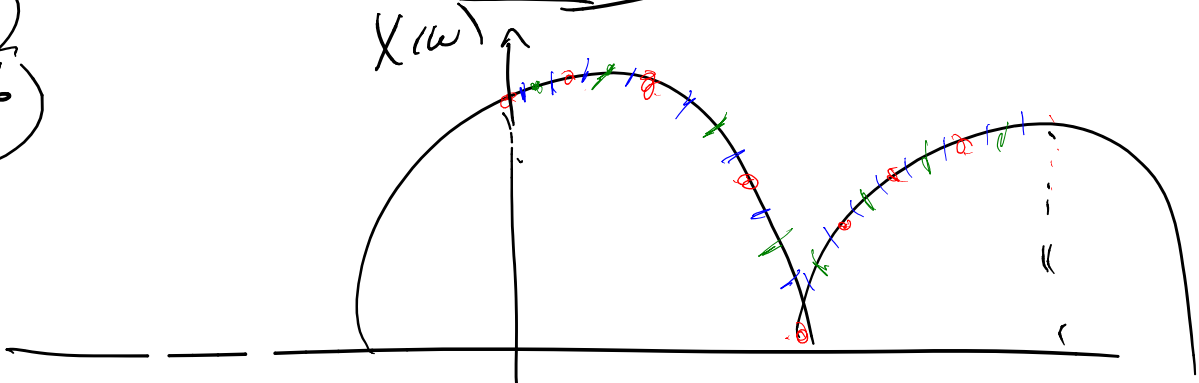


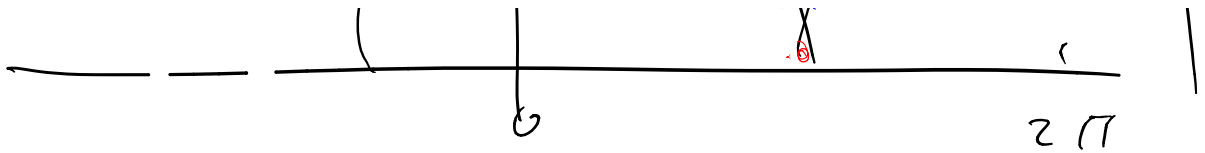
M pt DFT of $x(n)$

$$X(k) = \begin{cases} \sum_{n=0}^{N-1} x(n) e^{-j \frac{2\pi n k}{M}} & 0 \leq k < M \\ 0 & \text{otherwise} \end{cases}$$



$N=8$
 $M=16$



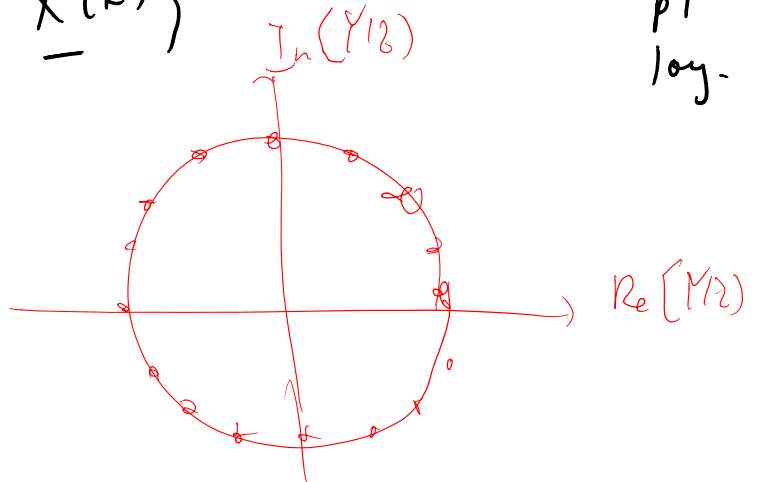


$$X(\omega) = \sum_n x(n) e^{-j\omega n}$$

$$\Rightarrow = \sum_{n=0}^{N-1} x(n) e^{-j\omega n} = \sum_{n=0}^{N-1} x(n) e^{-j\omega n}$$

Thought Exp

1. $y(n)$ either ∞ or finite \leftarrow length unknown
2. $Y(\omega) = \text{DTFT} \{ y(n) \}$
3. Sample $Y(\omega)$ at N equally spaced points
 $\left[Y(\omega) \right]_{\omega = \frac{2\pi k}{N}} = X(k) \leftarrow$ N pt seq
4. $\text{IDFT} \left\{ \underset{N \text{ pt}}{X(k)} \right\} = x(n) \leftarrow$ N pt seq.



Answer:

$$x(n) = \tilde{w}(n) R_N(n)$$

Answer:

$$x(n) = \tilde{w}(n) R_N(n)$$

$$\tilde{w}(n) = \sum_{k=-\infty}^{+\infty} y(n+kN)$$

$$x(n) = \begin{cases} \frac{1}{N} \sum_{k=0}^{N-1} X(k) e^{j \frac{2\pi nk}{N}} & 0 \leq n < N \\ 0 & \text{otherwise} \end{cases}$$

$$x(n) = \frac{1}{N} \sum_{k=0}^{N-1} \left(\sum_{m=-\infty}^{+\infty} y(m) e^{-j \frac{2\pi mk}{N}} \right) e^{j \frac{2\pi nk}{N}} \quad 0 \leq n < N$$

otherwise

$$x(n) = \sum_{m=-\infty}^{+\infty} y(m) \frac{1}{N} \sum_{k=0}^{N-1} e^{-j \frac{2\pi k(m-n)}{N}}$$

otherwise

$$\sum_{r=-\infty}^{+\infty} \delta(n-m+rN)$$

works
identifies

name: $f(n) = \sum_{r=-\infty}^{+\infty} \delta(n+rN)$

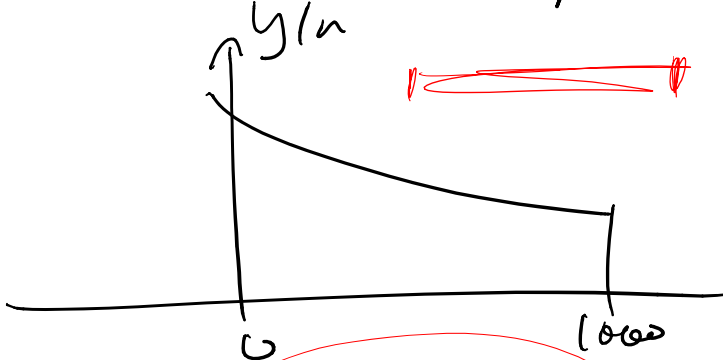
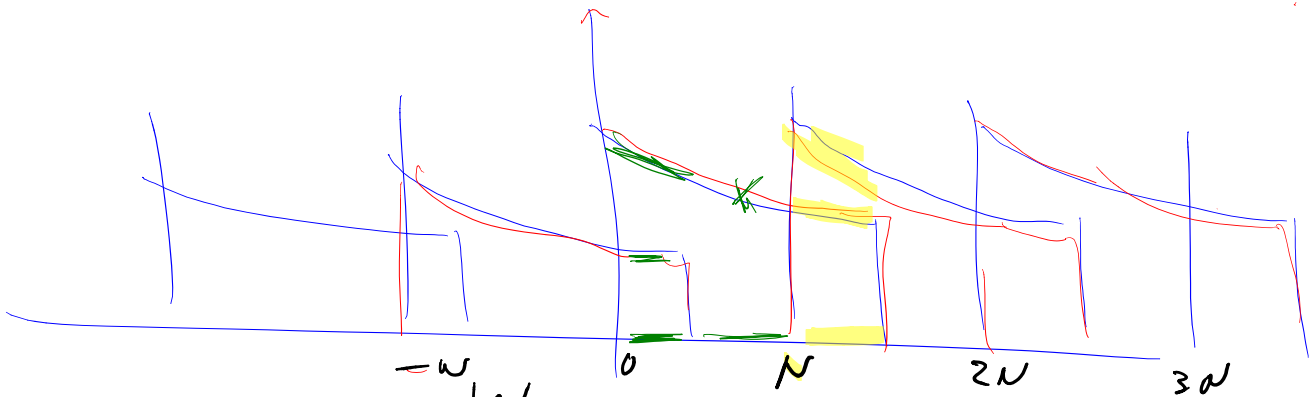
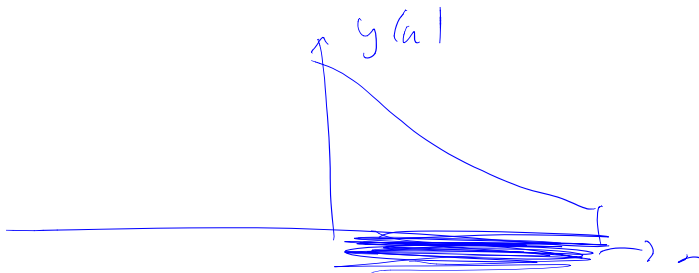
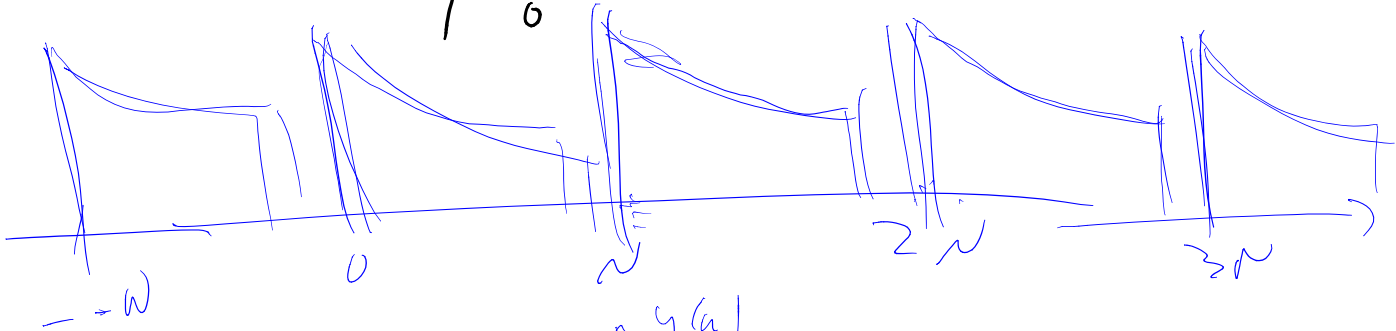
$$f(n-m)$$

$$x(n) = \sum_{m=-\infty}^{+\infty} y(m) f(n-m) \quad 0 \leq n < N$$

otherwise

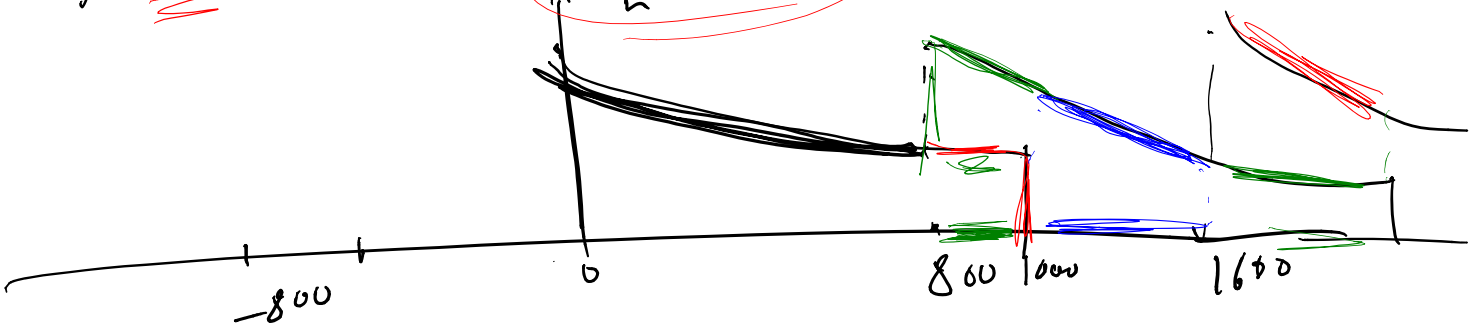
$$\sum_{r=-\infty}^{+\infty} y(n+rN) \quad 0 \leq n < N$$

$$1 = \begin{cases} \sum_{r=-p}^{+p} y(n+rN) & 0 \leq n < N \\ 0 & \text{otherwise} \end{cases}$$



N = 800

$$\sum_r y(n+rN)$$



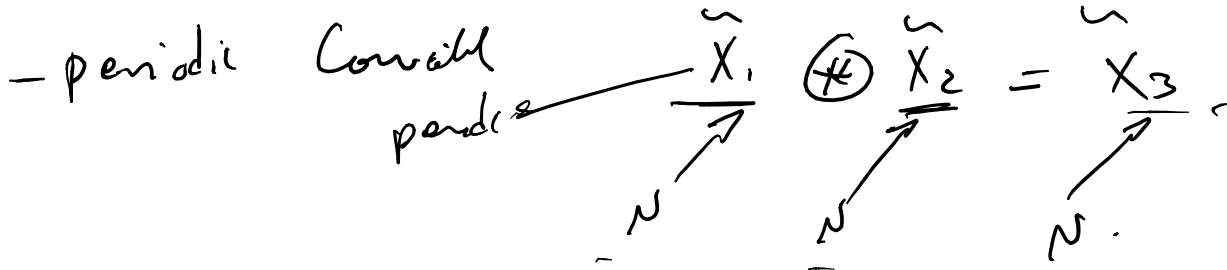
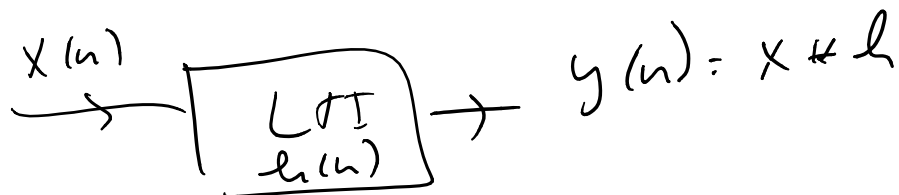
If $N \gg$ duration of $y(n)$
get back $y(n)$ exactly.

If $N <$ duration of $y(n)$
one period of $x(n)$ will be aliased version of $y(n)$

How To use DFT To do Convolution

$$x_1 * x_2 = x_3$$

Linear Conv $x_3(n) = \sum_k x_1(k) x_2(n-k)$



- Circular Convolution

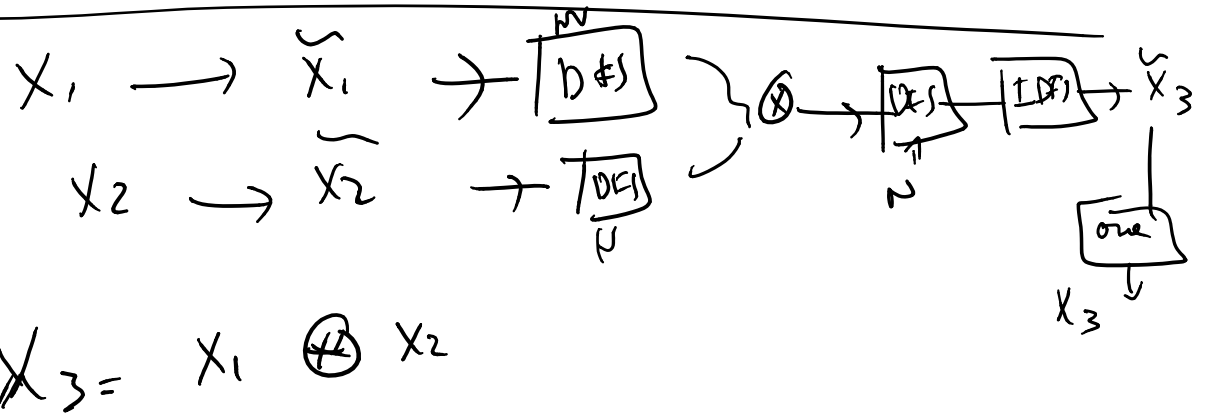
Diagram illustrating circular convolution. It shows two discrete-time Fourier transform (DTFT) spectra, \tilde{x}_1 and \tilde{x}_2 , each plotted over a period of length N . The spectra \tilde{x}_1 and \tilde{x}_2 are circled with an asterisk ($*$) to indicate multiplication. The equation $\tilde{x}_1 * \tilde{x}_2 = \tilde{x}_3$ is shown, with arrows pointing from each term to its corresponding spectrum. The entire diagram is enclosed in a red oval.

If multiply N pt DFT of 2
finite length N pt signals \rightarrow

finite length N PT sequence \rightarrow

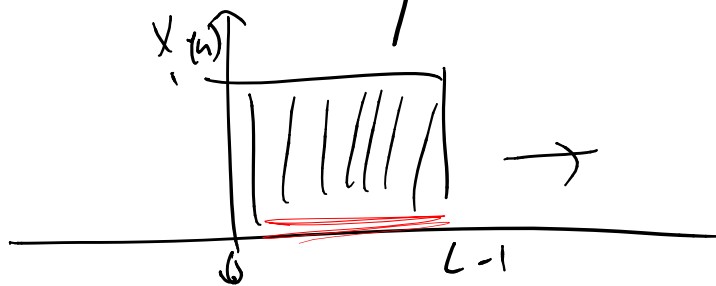
DFT of them circular convolution

NOT DFT of them linear convolution



Ex. Not. Circular Convolution of

$$x_1(n) = x_2(n) = \begin{cases} 1 & 0 \leq n \leq L-1 \\ 0 & \text{other} \end{cases}$$



2 cases:

① $N = L$

② $N = 2L$

Case ① : $N = L$ Do DFT for circular convolution.

... DFT of x_1

conv.

N pt = L pt DFT of x_1

$$X_L(k) = \sum_{n=0}^{L-1} x_1(n) e^{-j \frac{2\pi n k}{N}}$$

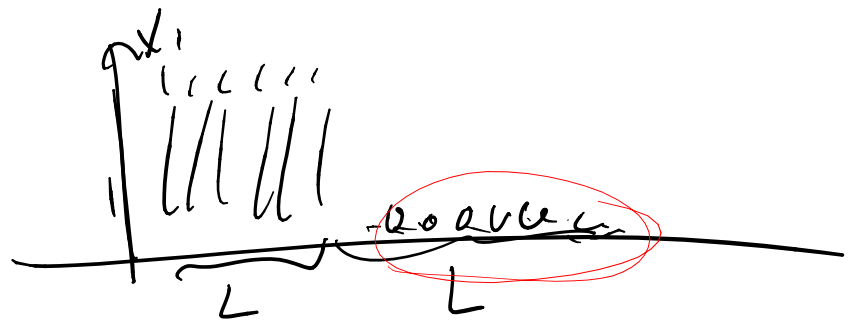
$$= \begin{cases} L & k=0 \\ 0 & \text{other} \end{cases}$$

$X_3(k)$: L pt circular conv of x_1 and x_2

$$= X_L(k) X_L(k) = \begin{cases} L^2 & k=0 \\ 0 & \text{other} \end{cases}$$

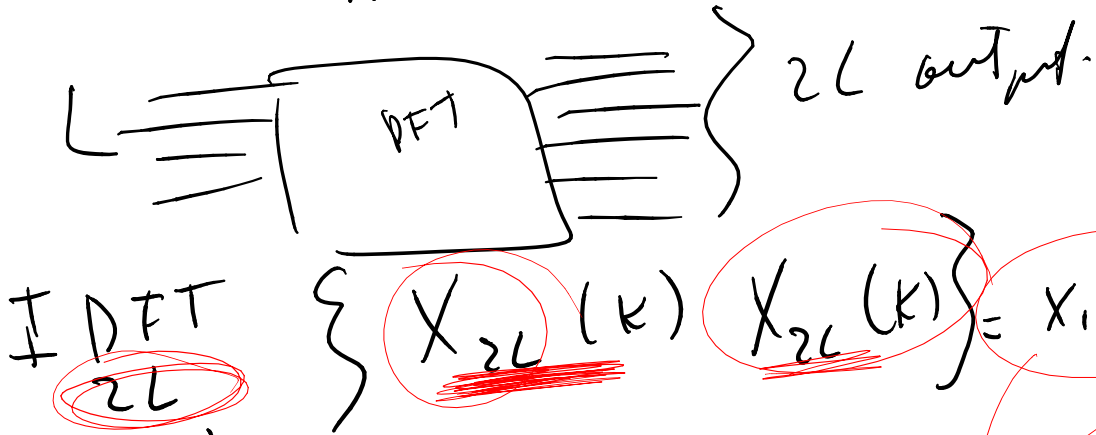
L pt IDFT of $\{X_3(k)\}$, $\begin{cases} L & 0 \leq n \leq L-1 \\ 0 & \text{other} \end{cases}$

Case (2) $N = 2L$



$2L$ pt DFT of x_1 and x_2
 $j \frac{2\pi n k}{N}$

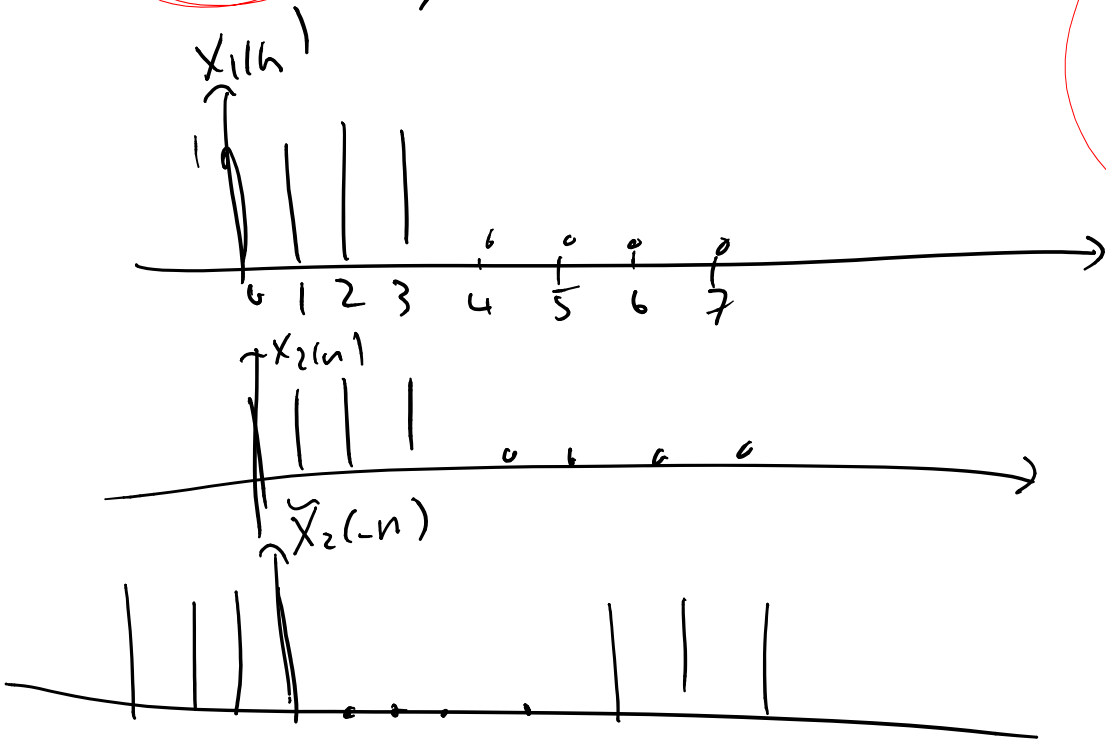
$$X_{2L}(k) = \sum_{n=0}^{L-1} x_1(n) e^{-j \frac{2\pi n k}{2L}}$$



↑ DFT
2L

$$\left\{ \begin{array}{l} X_{2L}(k) \\ X_{2L}(k) \end{array} \right\} = X_1 \oplus X_2$$

↑
2L PTS.



→ Discrete
Linear
Convolution

$$x_f(n) \xrightarrow{\quad} L$$

$$x_2(n) \xrightarrow{\quad} P$$

$$N > L$$

$$N > P$$

Garl

$$x_3 \triangleq x_1 * x_2$$

$$\dots \rightarrow \dots \rightarrow \dots \rightarrow \sum x_2(n) e^{-j\omega n}$$

$$\text{DTFT } \{x_3\} = X_3(\omega) = \sum_n x_3(n) e^{-j\omega n}$$

$$X_3(\omega) = X_1(\omega) X_2(\omega)$$

Sample $X_3(\omega)$ at $\frac{2\pi k}{N}$ equally spaced cycles = $Y(k)$

$$Y(k) = \left[X_3(\omega) \right]_{\omega = \frac{2\pi k}{N}}$$

$$\text{IDFT } \{Y(k)\} = \begin{cases} \sum_{r=-\infty}^{+\infty} x_3(n+rN) & 0 \leq n < N \\ 0 & \text{otherwise} \end{cases}$$

↑
N pt

$$Y(k) = \left[X_1(\omega) \right]_{\omega = \frac{2\pi k}{N}} \left[X_2(\omega) \right]_{\omega = \frac{2\pi k}{N}}$$

↑
N pt

$$\text{IDFT } \{Y(k)\} = x_1 \otimes x_2$$

↑
N pt

$$x_1 \otimes x_2 = \begin{cases} \sum_{r=-\infty}^{+\infty} x_3(n+rN) & 0 \leq n < N \\ 0 & \text{otherwise} \end{cases}$$

↑
N pt

↑ 1 ... (cont. 1)

to get Linear Convolution

of x_1 and x_2
 $L \rightarrow$ $P \rightarrow$

(1) pad x_1 lots of zeros $\rightarrow N$
pad x_2 $\rightarrow N$

$$N \geq P + L - 1$$

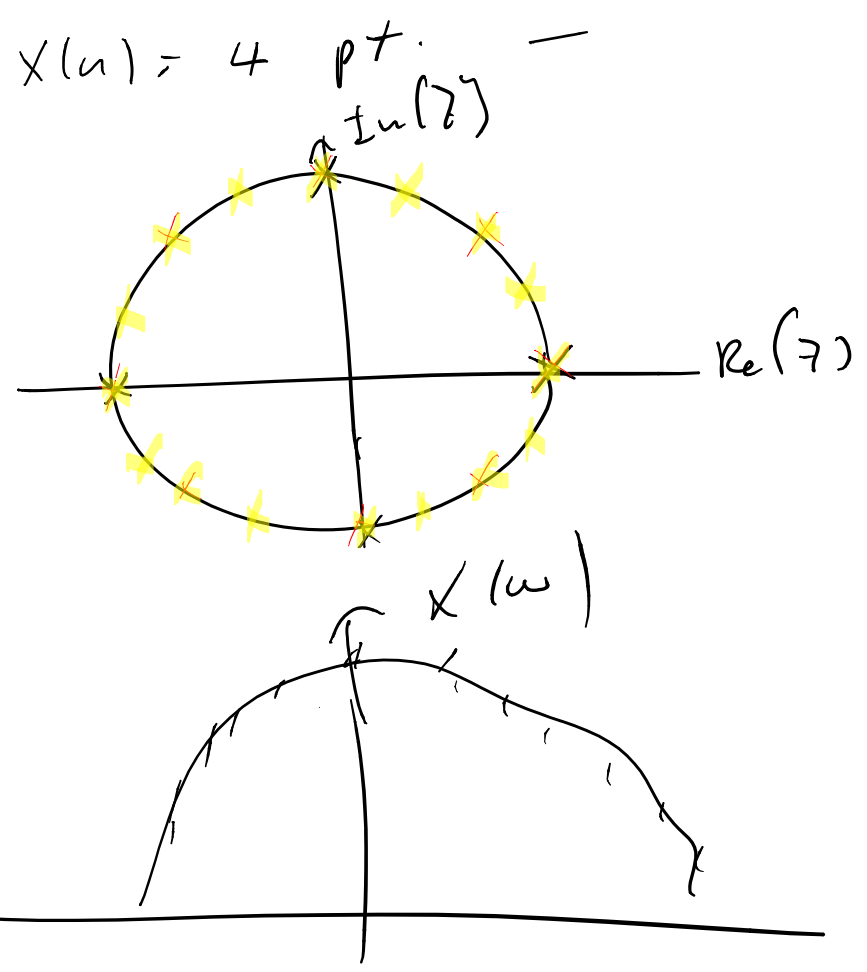
(3) take DFT of both
 $N \rightarrow F$ multiply

IDFT $\rightarrow \underline{N}$
 $\rightarrow N$

x_1 L pt
 x_2 P pt

- (1) Pad x_1 with $N-L$ zeros to get N pt.
- (2) Pad x_2 with $N-P$ zeros to get N pt.
- (3) Take N pt DFT of $x_1 \rightarrow X_1(k)$
" " " " " $x_2 \rightarrow X_2(k)$
" multiply $X_1(k) X_2(k) \rightarrow N$ pt

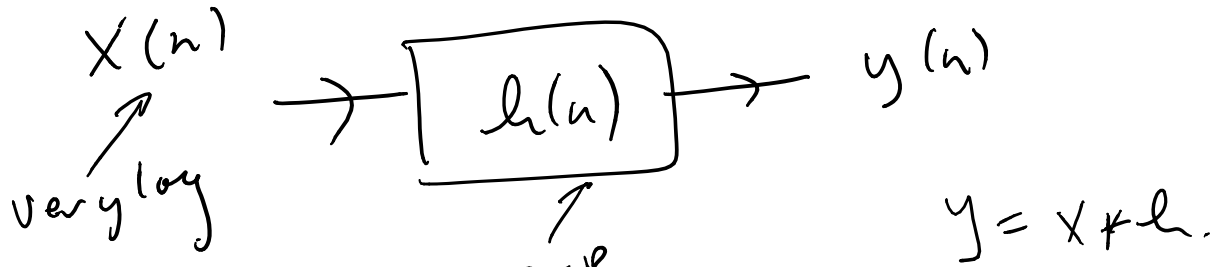
(4) Multiply $X_1(k) X_2(k) \rightarrow N$ pt
 (5) IDFT $\{ X_1(k) X_2(k) \} \rightarrow$ answer.
 If $N \gg L + P - 1$
 $X_3 = X_1 * X_2 =$ The answer



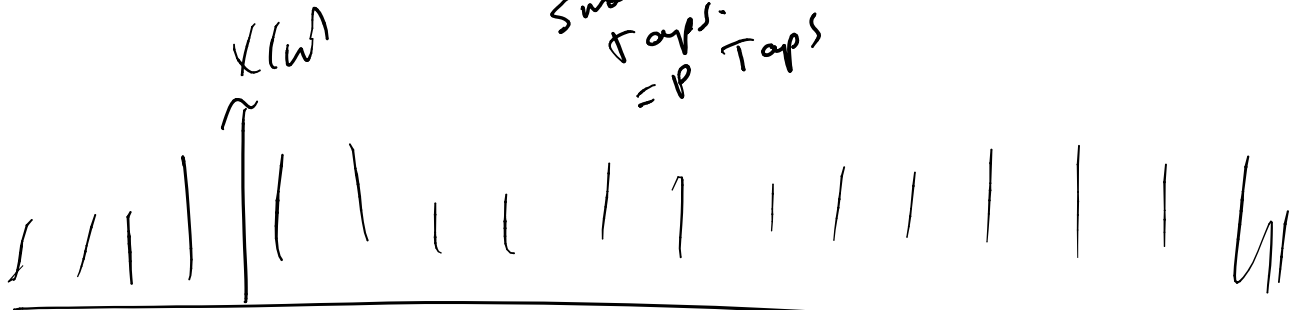
4 pt DFT
 8 pt DFT
 16 pt

Using DFT for filtering sig long sequence

long sequence

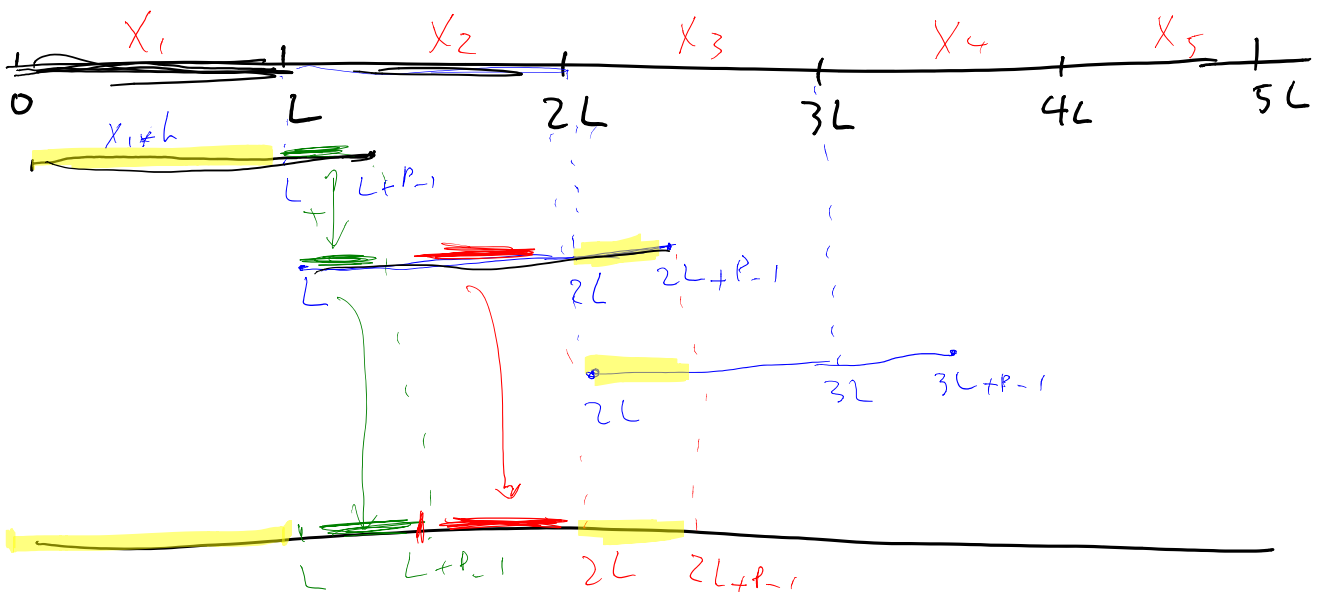


FIR.
Small # of
Taps.
= P Taps



Explicit linearity of convolution

$\rightarrow [X_1(n) + X_2(n)] * h = X_1 * h + X_2 * h$
Overlap Add



Overlap save



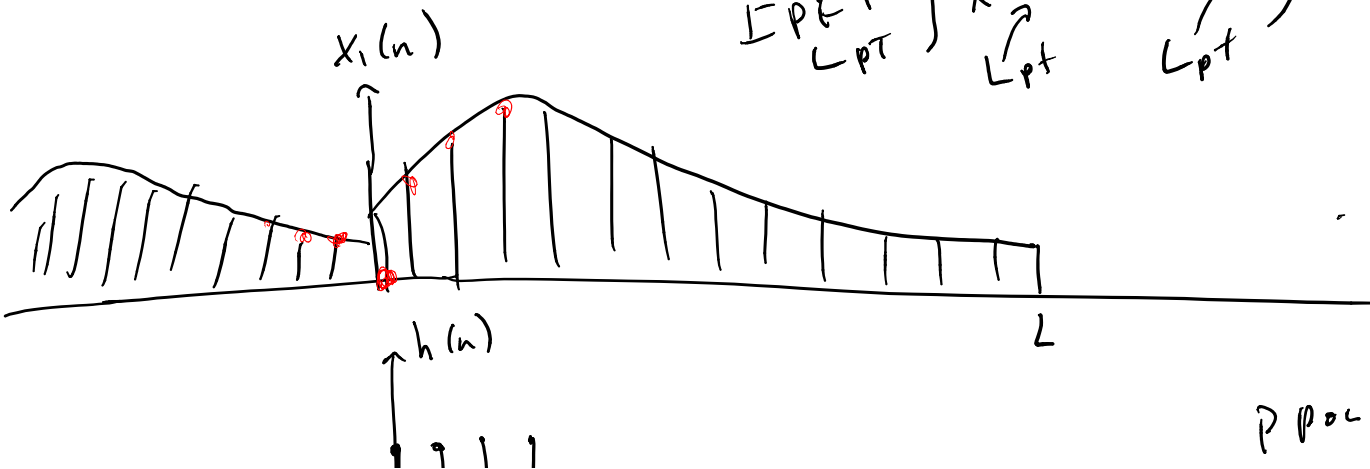
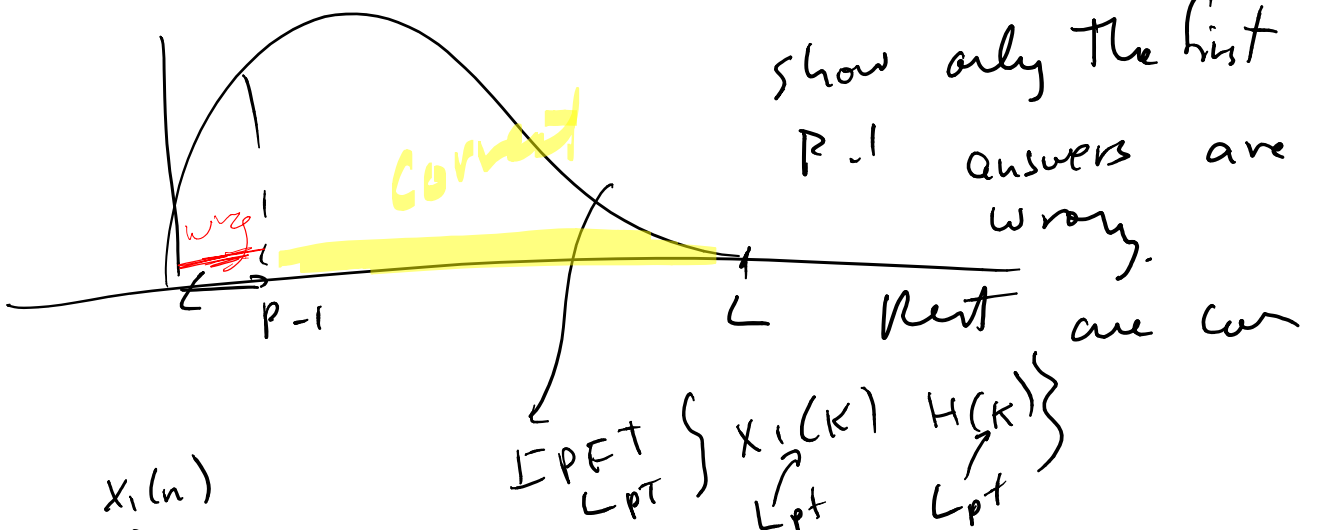
Comment : pad with zeros to get $L+P-1$ seq.

- multiply these.
- Inverse DFT.

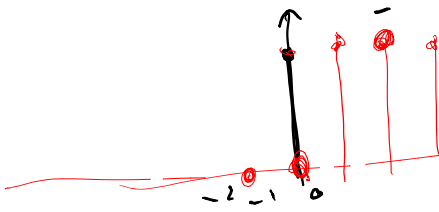
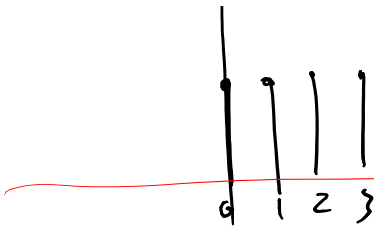
$L \gg P$

Suppose take L pt DFT of X_1 and X_2

- multiply these
- take L pt IDFT.



P POC

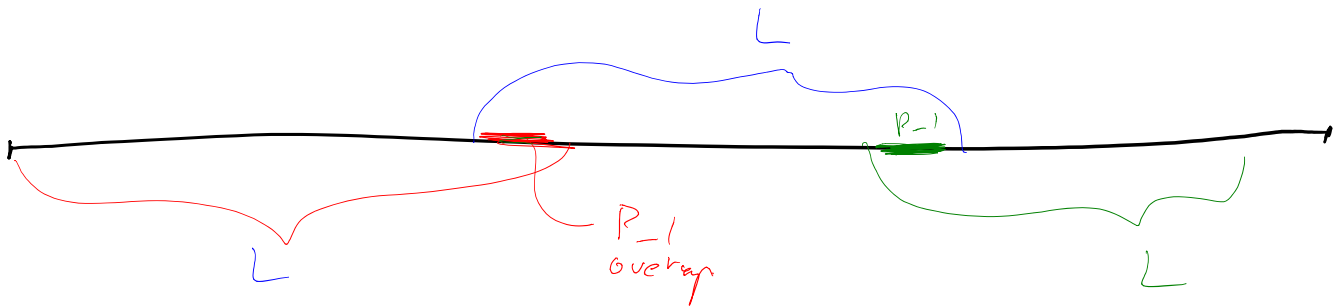


$x[n]$ ← why

If $L \gg P$
 If LPT DFT of Both
 are mutually

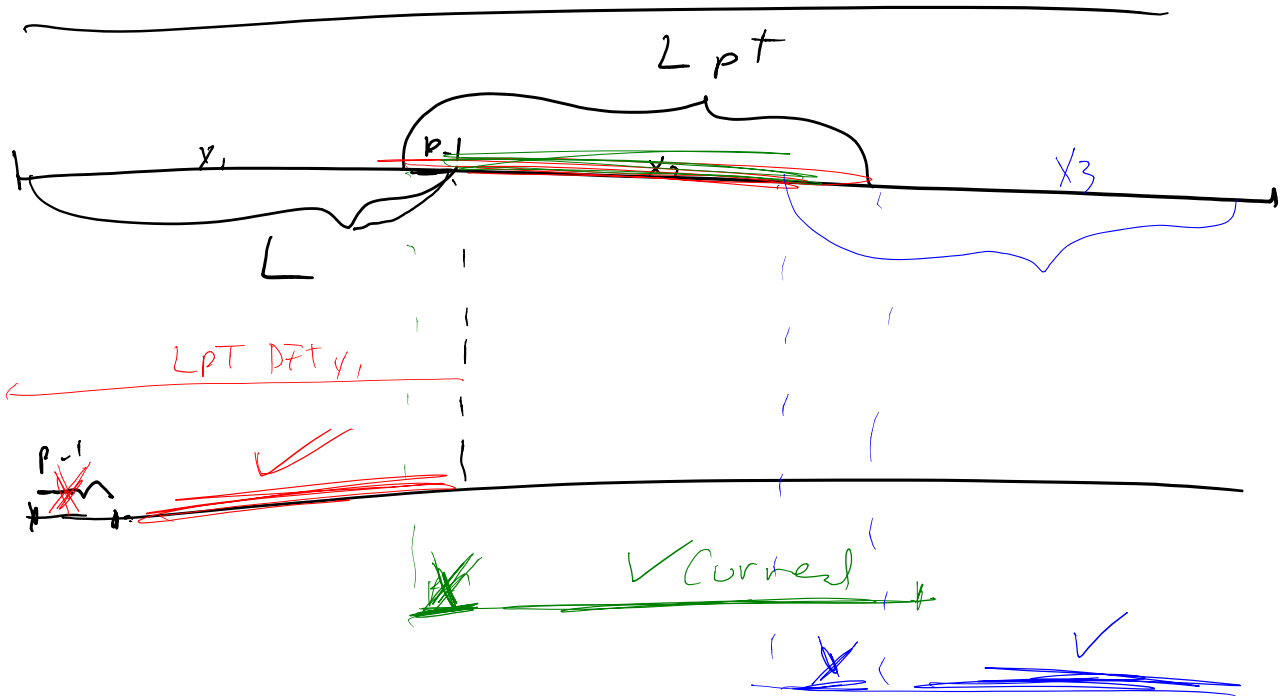
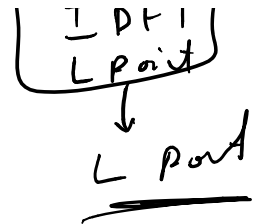
Then only $P-1$ points are wrong
 first $P-1$ points are wrong
 Remaining answer $L - (P-1)$ are correct

Overlap Save



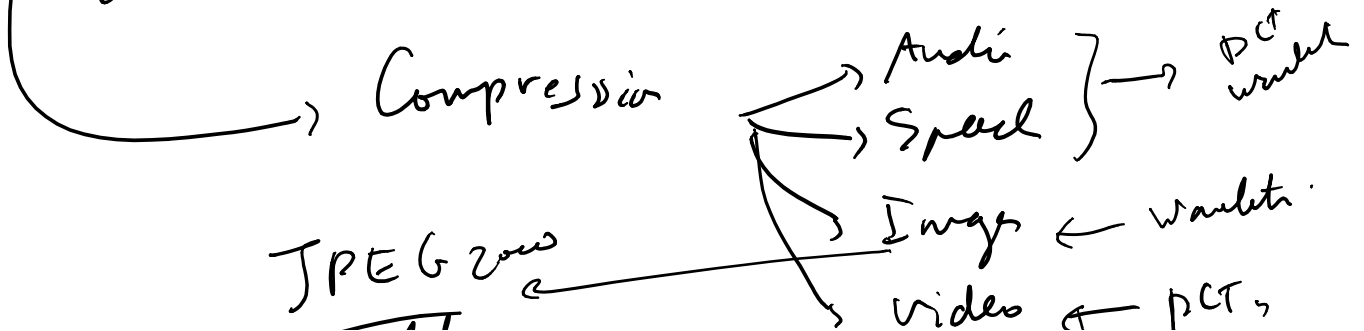
- ① Segment into L point chunks. Overlap by $P-1$
- ② Take L point circular convolution of each chunk
 multiply LPT DFT of chunk } → L pt
 LPT DFT of h }
 IDFT
 L point

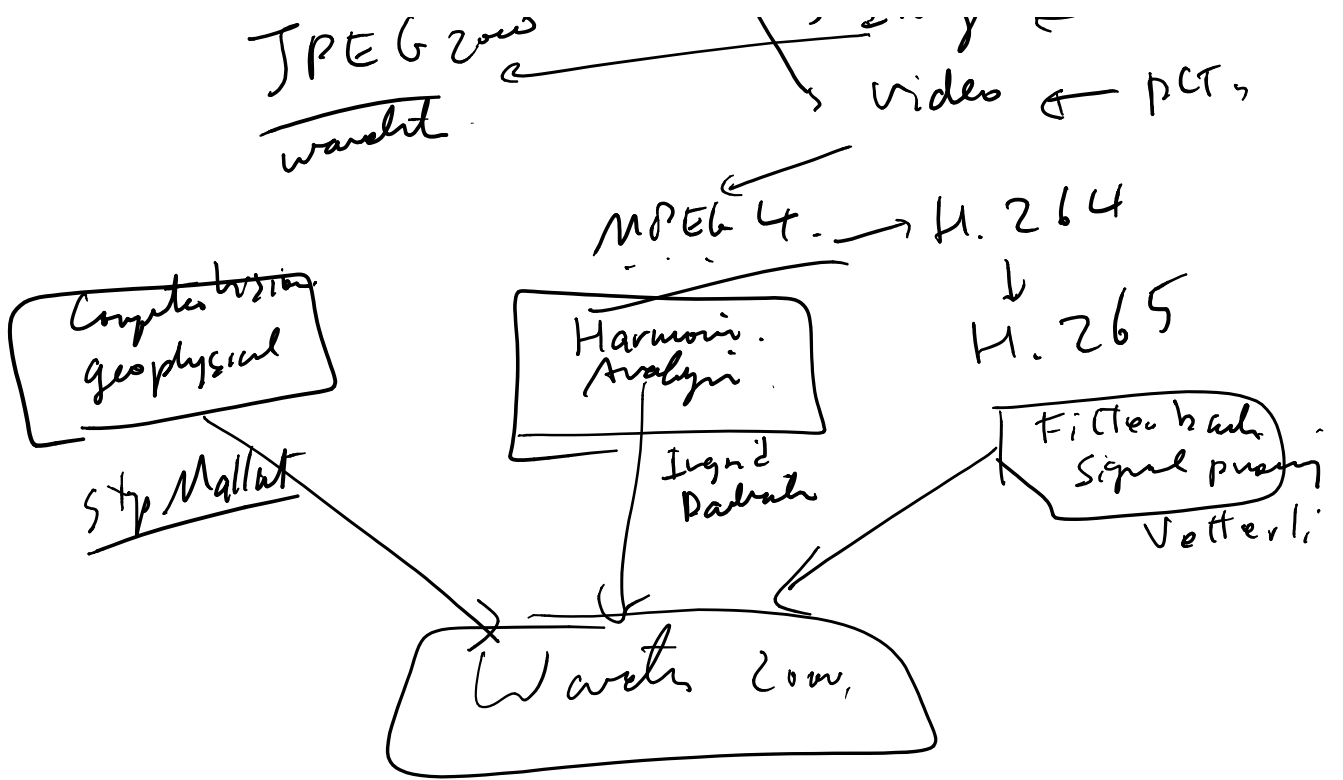
③ Throw away first $P-1$ points of the answer in part 2, Replace with previous segment



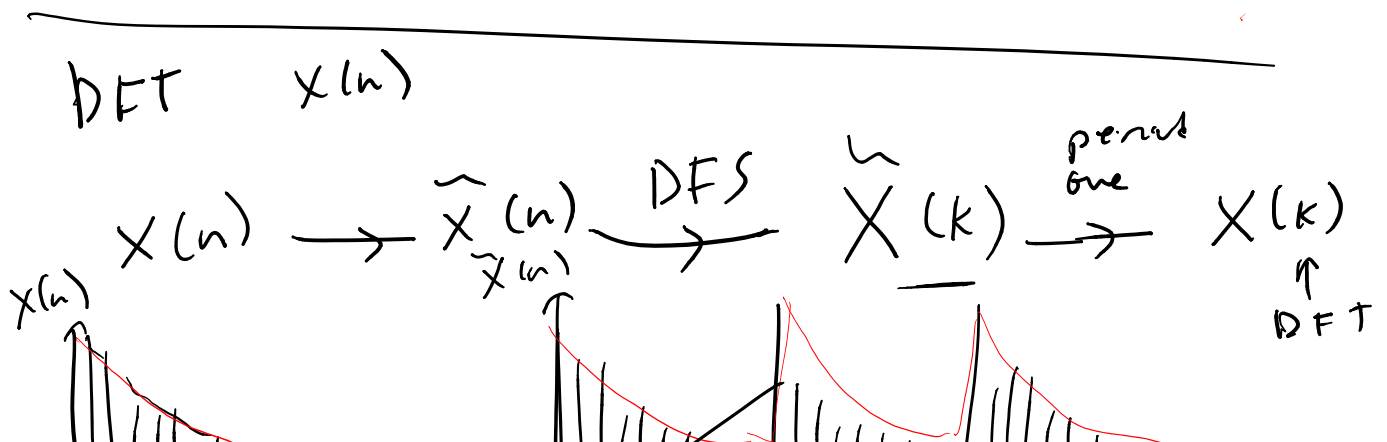
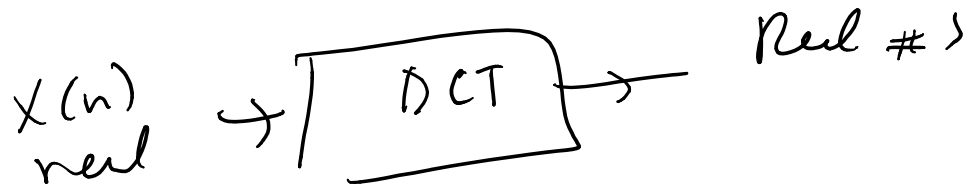
Discrete Cosine Transform

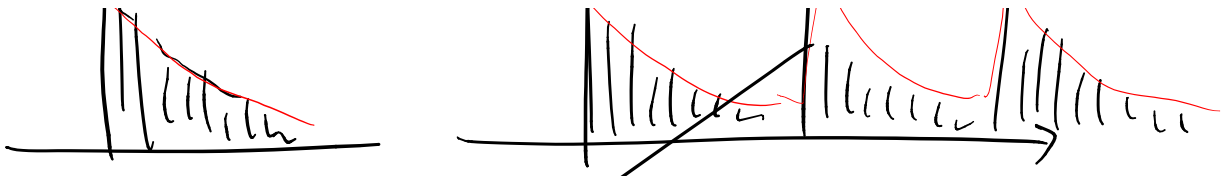
- * Compactness property
- * Coefft in freq. domain are uncorrelated with each other



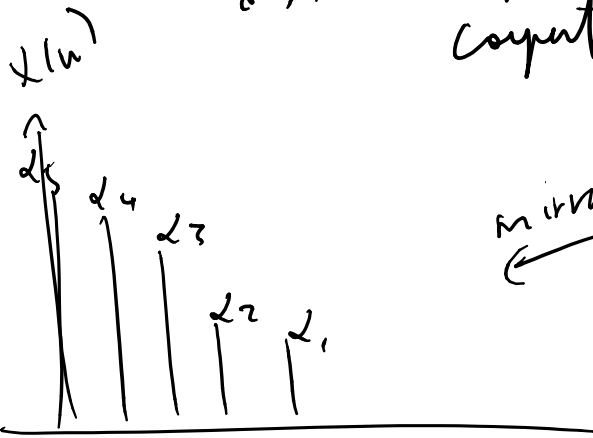


Haar, Hartley, Hadamard, Walsh

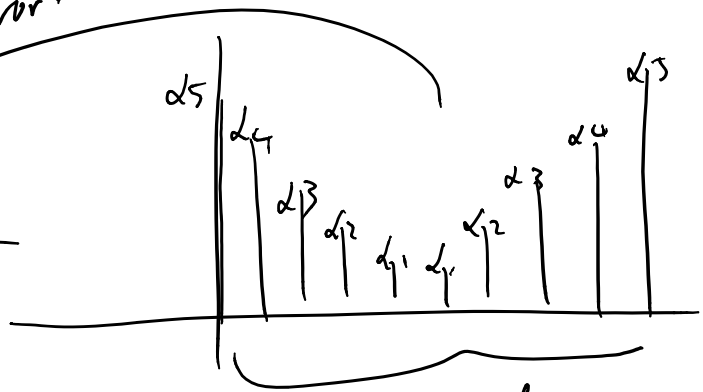




expand $e^{j2\pi k \frac{n}{N}}$
 sharp discontinuity results in high frequency content



mirror



DFS of mirrors signal \rightarrow DCT

4 Kinds of DCT

Analysis \rightarrow

$$X(k) = \sum_{n=0}^{N-1} x(n) \cos\left(\frac{\pi k (2n+1)}{2N}\right)$$

Synthesis \leftarrow

$$x(n) = \sum_{k=0}^{N-1} \beta(k) X(k) \cos\left(\frac{\pi k (2n+1)}{2N}\right)$$

$$x(n) = \frac{1}{N} \sum_{k=0}^{N-1} \beta(k) X(k) \cos\left(\frac{\pi k(2n+1)}{2N}\right)$$

$$\beta(k) = \begin{cases} \frac{1}{2} & k=0 \\ 1 & 1 \leq k < N \end{cases}$$

Relate DFT of $x(n)$ to IDCT 2

- Proposal 1:
- ① $x(n)$ N pt seq. Real
 - ② Take $2N$ pt DFT $\rightarrow X(k)$
 - ③ $2 \operatorname{Re} \left\{ X(k) e^{-j \frac{2\pi nk}{2N}} \right\}$

Proposal 2

- ① start with N pt Real seq $x(n)$
- ② Pad it with N zeros $\rightarrow X_{2N}(n)$
- ③ Form a periodic seq. $\rightarrow X_8(n) = X_{2N}(n) + X_{2N}(-n-1)$
- ④ Take $2N$ pt DFT of one period

④ Take ~~2N pt DFT~~ of ~~one period~~ of $x_2(n) \rightarrow X_2(k)$

⑤ Relate $X_2(k)$ to $X^{e_2}(k)$

Proposal 1

Proof:

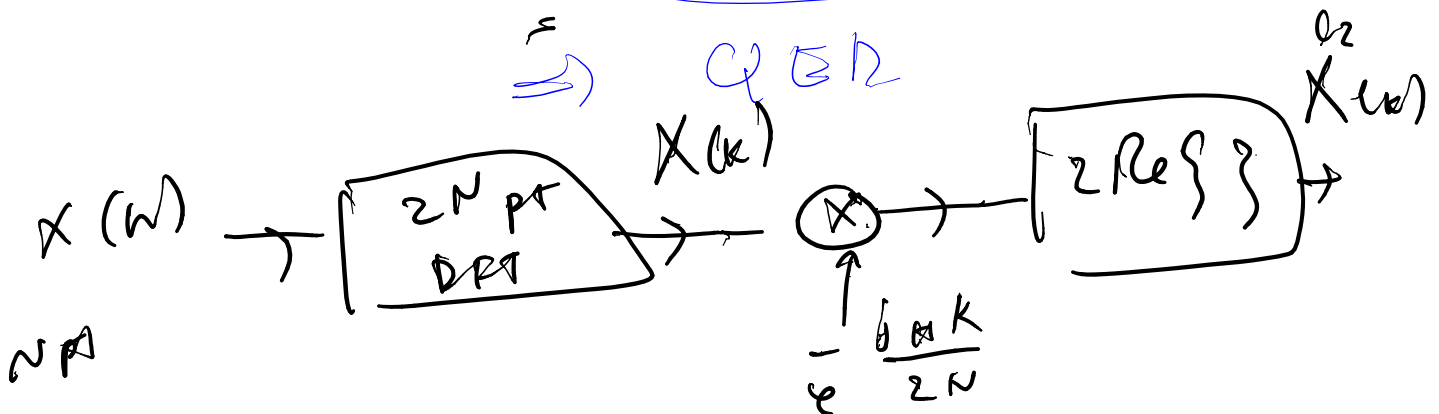
2N point DFT of $x(n)$ is

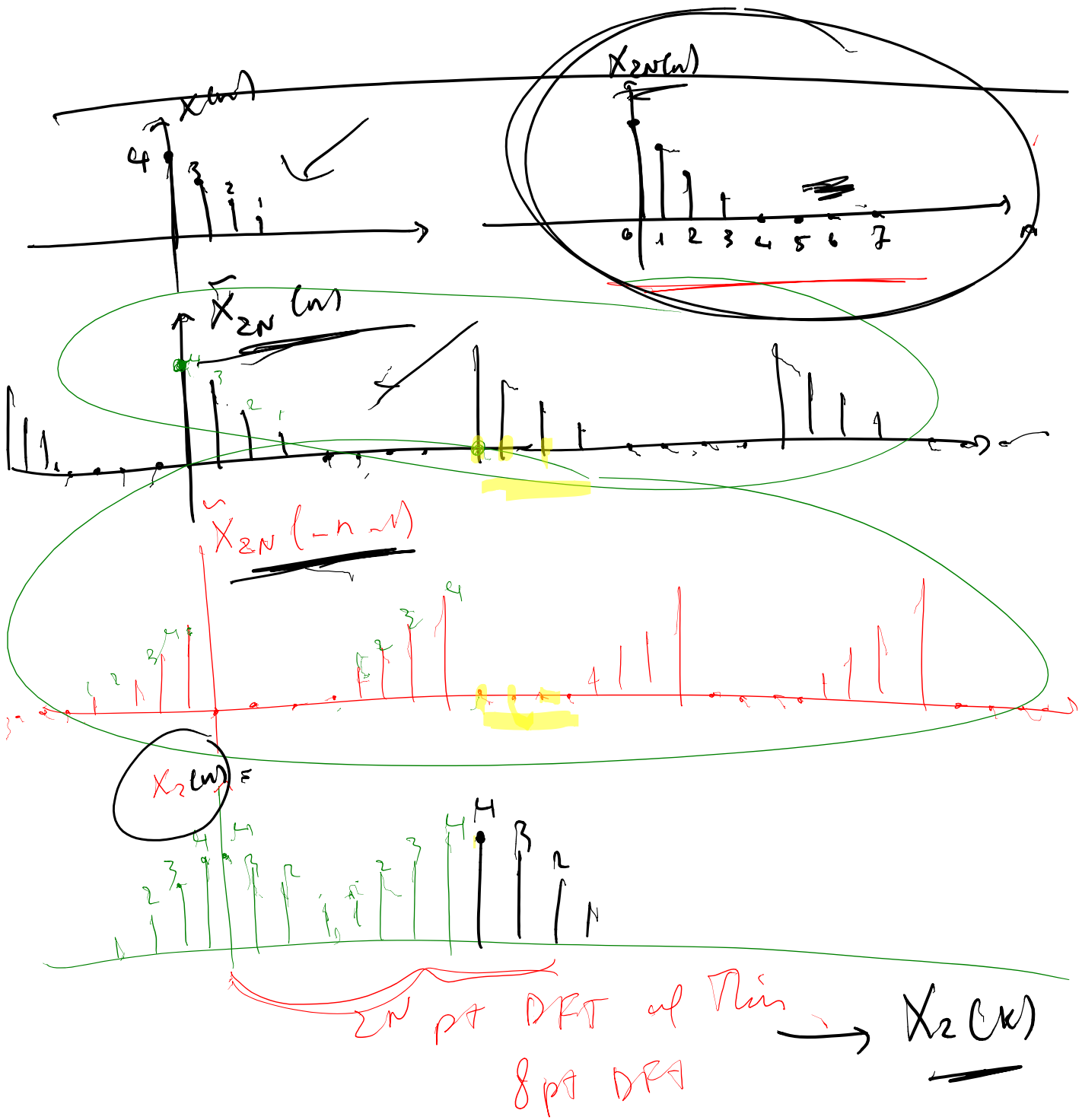
$$X(k) = \sum_{n=0}^{N-1} x(n) e^{-j \frac{2\pi n k}{2N}} \Rightarrow \sum_{n=0}^{N-1} x(n) e^{-j \omega k (2n+1)}$$

$$X(k) e^{-j \frac{\pi k}{2N}} = \sum_{n=0}^{N-1} x(n) e^{-j \frac{2\pi n k}{2N}}$$

$$\underbrace{2 \operatorname{Re} \left\{ \right.}_{\downarrow} \left. \right\}} = 2 \sum_{n=0}^{N-1} x(n) \cos \frac{\omega k (2n+1)}{2N}$$

$\Rightarrow \mathbb{Q} \in \mathbb{R}$





How is $X_2(k)$ related to $X^c(k)$

$X(k)$ is 2N pt DFT of $x(n)$

\dots ω \dots $X(-n-1)$

$$\tilde{x}_2(n) = \tilde{x}_{2N}(n) + \tilde{x}_{2N}(-n-1)$$

$$X_2(k) = X(k) + X(k) e^{j\frac{2\pi k}{2N}}$$

$$= e^{j\frac{\pi k}{2N}} \left[X(k) e^{-j\frac{\pi k}{2N}} + X(k) e^{j\frac{\pi k}{2N}} \right]$$

$$\underline{X_2(k)} = \underline{e^{j\frac{\pi k}{2N}} \left\{ 2 \operatorname{Re} \left\{ X(k) e^{-j\frac{\pi k}{2N}} \right\} \right\}}$$

$$X_2(k) e^{-j\frac{\pi k}{2N}} = X^{c_2}(k)$$

$X^{c_2}(k) \leftarrow \text{part 2}$
 $X^{c_2}(k)$
 \leftarrow

Fast Fourier Transform

FFT

$$\text{DFT: } X(k) = \sum_{n=0}^{N-1} x(n) e^{-j\frac{2\pi nk}{N}}$$

Direct Computation

Direct Computation

For each k : N complex mult
 $N-1$ complex adds $\approx N$

N values of k

$\Rightarrow N(N-1)$ adds $\approx O(N^2)$
 N^2 mult $\Rightarrow O(N^2)$

FFT $\rightarrow N \log N$

$N \leq 10^6$ Direct Computation $O(10^{12})$

FFT $\rightarrow 10^6 \log_2(10^6) \approx 2 \times 10^7$

5 orders of magnitude

Decomposition in Time Frequency

Decomposition in Time

$$X(k) = \sum_{n=0}^{N-1} x(n) e^{-j \frac{2\pi n k}{N}}$$

$$X(n) = \sum_{k=0}^{N-1} x(k) e^{j \frac{2\pi n k}{N}}$$

$0 \leq k \leq N$

$$X(k) = \sum_{n \text{ even}} x(n) e^{-j2\pi kn/N} + \sum_{n \text{ odd}} x(n) e^{-j2\pi kn/N}$$

$$n = 2r$$

$$r: 0 \rightarrow \frac{N}{2} - 1$$

$$n = 2r + 1$$

$$r: 0 \rightarrow \frac{N}{2} - 1$$

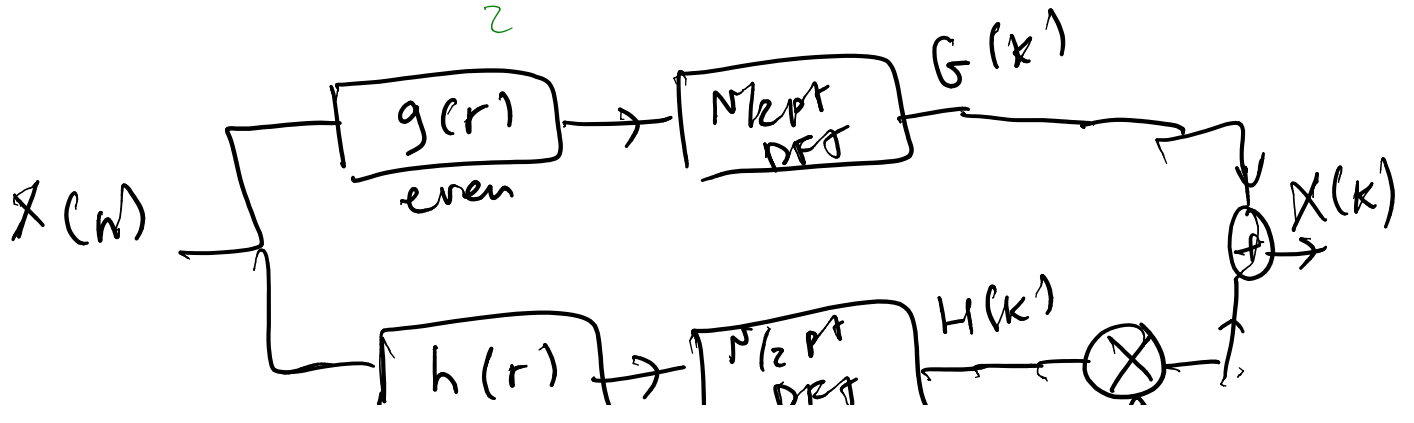
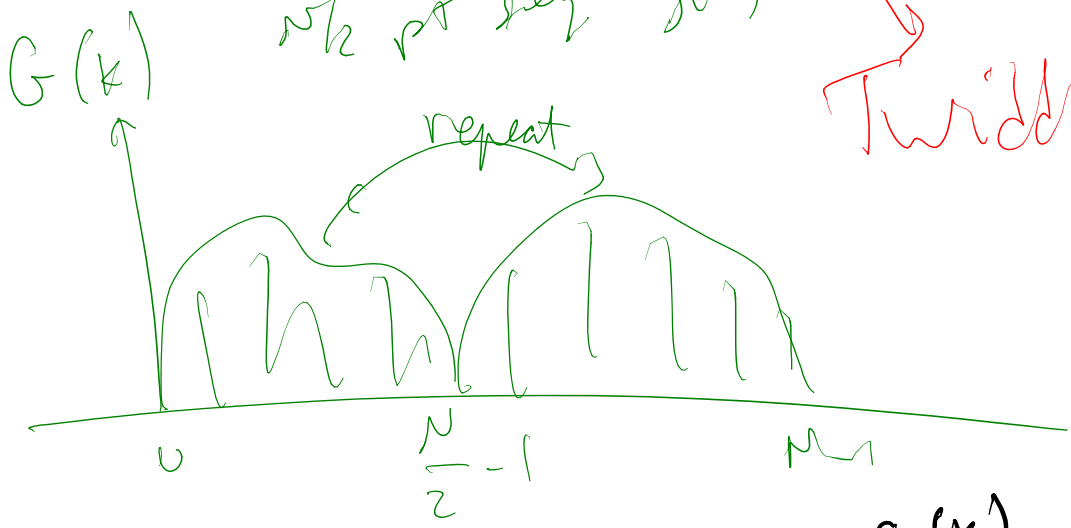
$$= \sum_{r=0}^{\frac{N}{2}-1} x(2r) e^{-j2\pi 2rk/N} + \sum_{r=0}^{\frac{N}{2}-1} x(2r+1) e^{-j2\pi (2r+1)k/N}$$

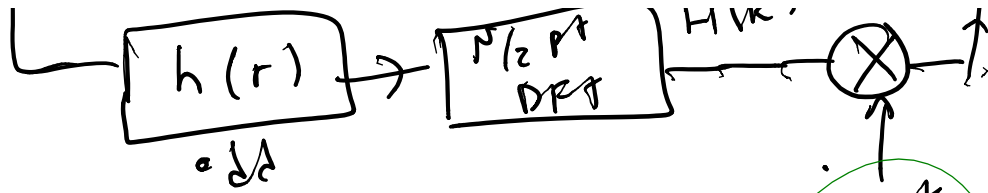
g(r) *h(r)*

$$X(k) = \sum_{r=0}^{\frac{N}{2}-1} g(r) e^{-j2\pi kr/N} + e^{-j2\pi k/N} \sum_{r=0}^{\frac{N}{2}-1} h(r) e^{-j2\pi kr/N}$$

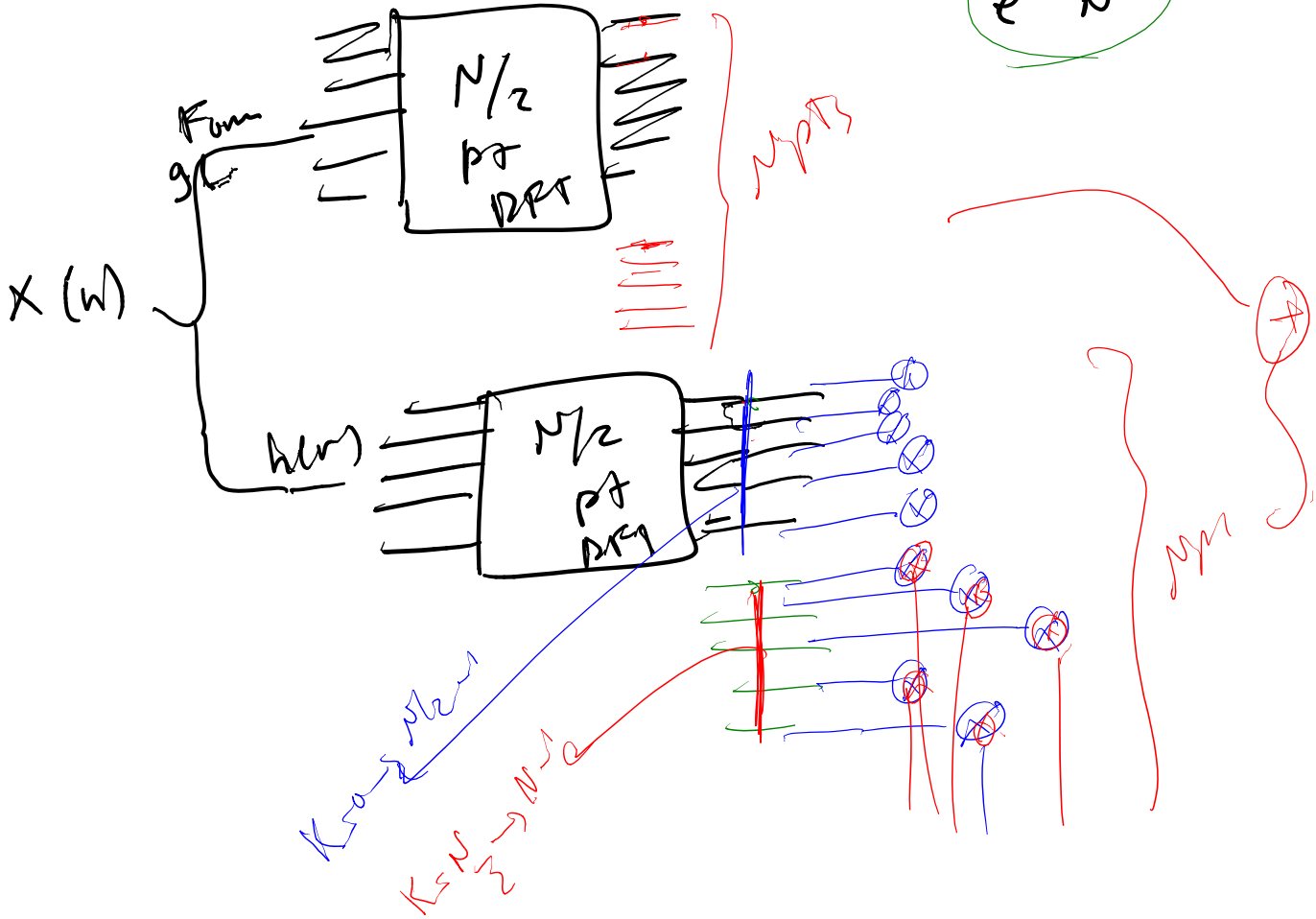
N/2 pt DFT of N/2 pt seq g(r) *N/2 pt DFT of N/2 pt seq h(r)*

Twiddle

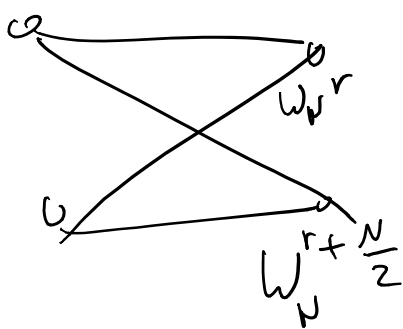




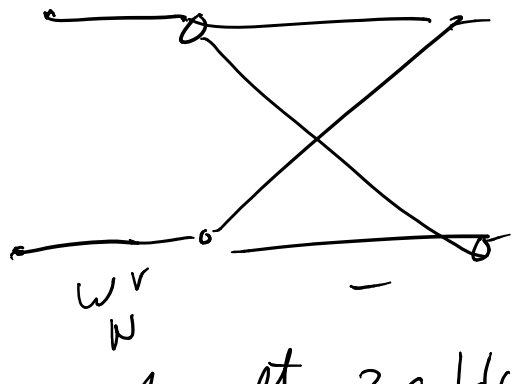
$$e^{-j2\pi rk \frac{N}{2}}$$



W_N^k Twiddle factor



with

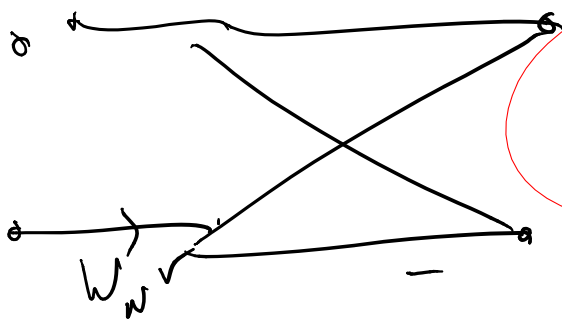


w_N^{-1}
2 mults 2 adds

w_N^i
1 mult 2 adds

Operation Count

- # of stages $\log_2 N$
- Each stage: N inputs, N outputs
 $N/2$ butterflies
 Each butterfly has 2 inputs, 2 outputs



1 mult
2 adds

$\frac{N}{2}$ mults
-
 N adds

Total: $\frac{N}{2} \log_2 N$ mults, $N \log_2 N$ adds

Decimation in Frequency FFT

$$X(k) = \sum_{n=0}^{N-1} X(n) e^{-j \frac{2\pi n k}{N}}$$

$\rightarrow 2r$

① k even $k = 2r$

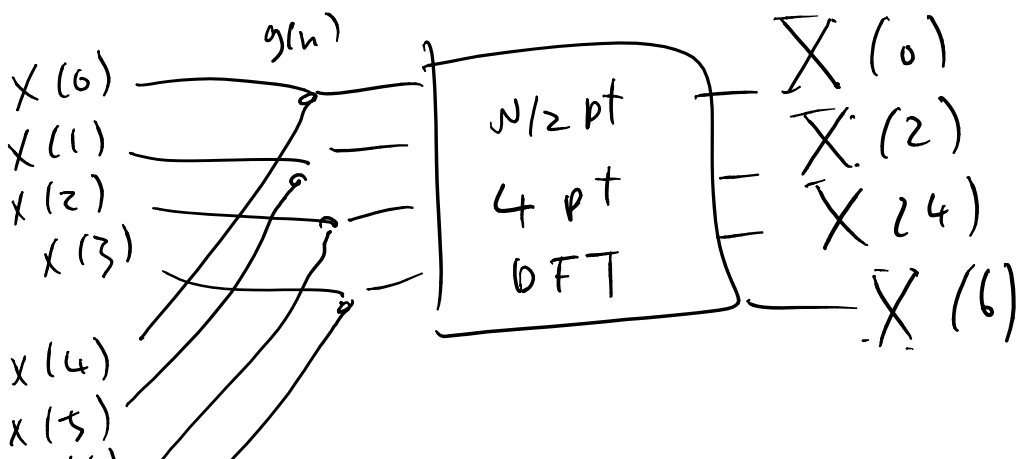
$$e^{-j \frac{2\pi n \cdot 2r}{N}}$$

(1) K even

$$\begin{aligned}
 X(z^r) &= \sum_{n=0}^{N-1} x(n) e^{-j \frac{2\pi r n}{N}} \\
 &= \sum_{n=0}^{N/2-1} x(n) e^{-j \frac{2\pi r n}{N}} + \sum_{n=N/2}^{N-1} x(n) e^{-j \frac{2\pi r n}{N}} \\
 &= \sum_{n=0}^{N/2-1} x(n) e^{-j \frac{2\pi r n}{N}} + \sum_{m=0}^{N/2-1} x(m + \frac{N}{2}) e^{-j \frac{2\pi r (m + \frac{N}{2})}{N}} \\
 &= \sum_{n=0}^{N/2-1} x(n) e^{-j \frac{2\pi r n}{N}} + \sum_{n=0}^{N/2-1} x(n + \frac{N}{2}) e^{-j \frac{2\pi r n}{N}} \\
 X(z^r) &= \sum_{n=0}^{N/2-1} \underbrace{[x(n) + x(n + \frac{N}{2})]}_{g(n)} e^{-j \frac{2\pi r n}{N}}
 \end{aligned}$$

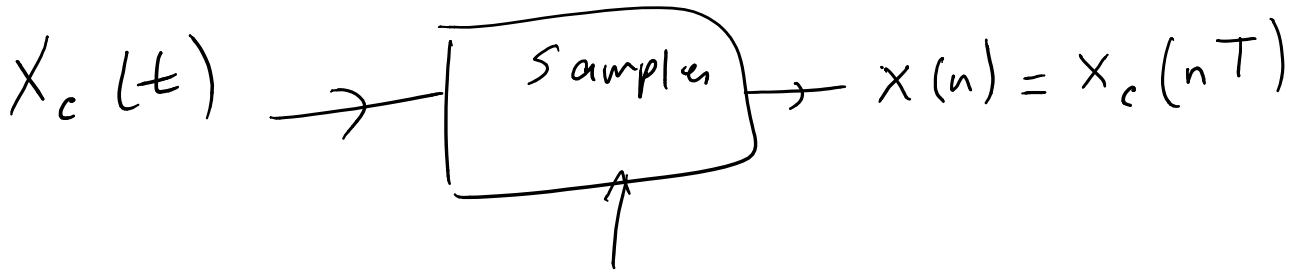
$0 \leq r < N/2$

$\rightarrow N/2$ pt DFT of $g(n)$

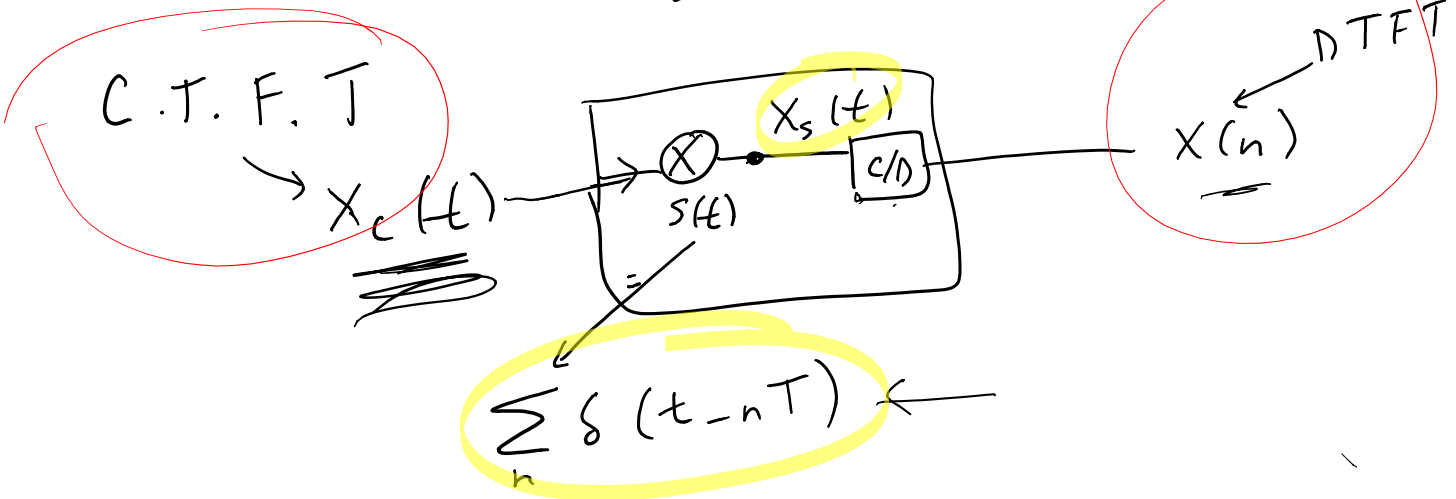


$x(4)$ /
 $x(5)$ /
 $x(6)$ /
 $x(7)$ /

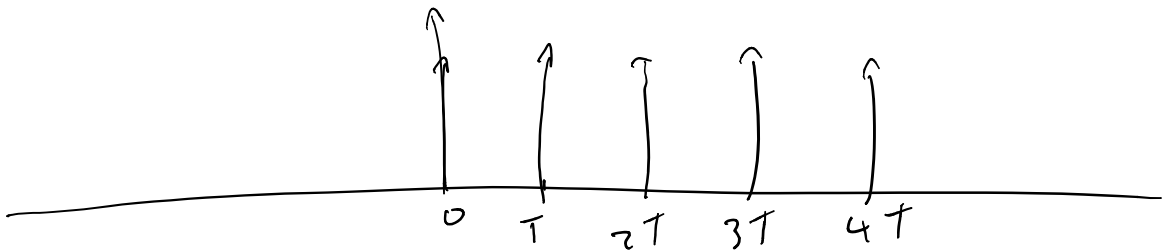
Sampling . Chap 4 O&S



T
 Continue to
 discrete



$$\sum_n \delta(t - nT)$$



C.T.F.T $x_c(t)$

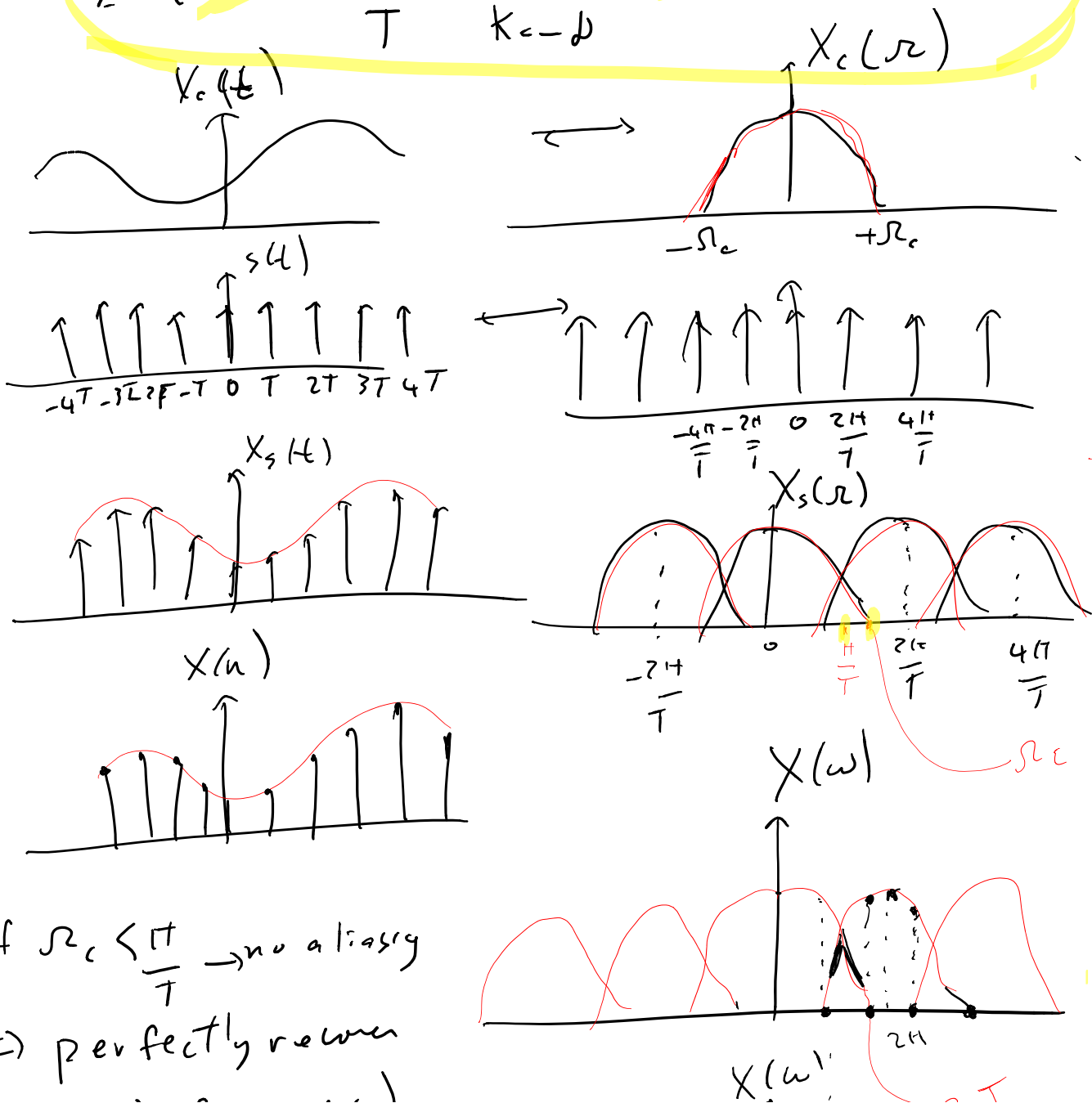
$$X_c(\omega) = \text{C.T.F.T} \{ x_c(t) \} = \int_{-\infty}^{+\infty} x_c(t) e^{-j\omega t} dt$$

$$X_c(\omega) = \dots$$

$$X(\omega) = \text{D.T.F.T.}\{x(n)\} = \sum_{n=-\infty}^{+\infty} x(n) e^{-j\omega n}$$

Can be shown:

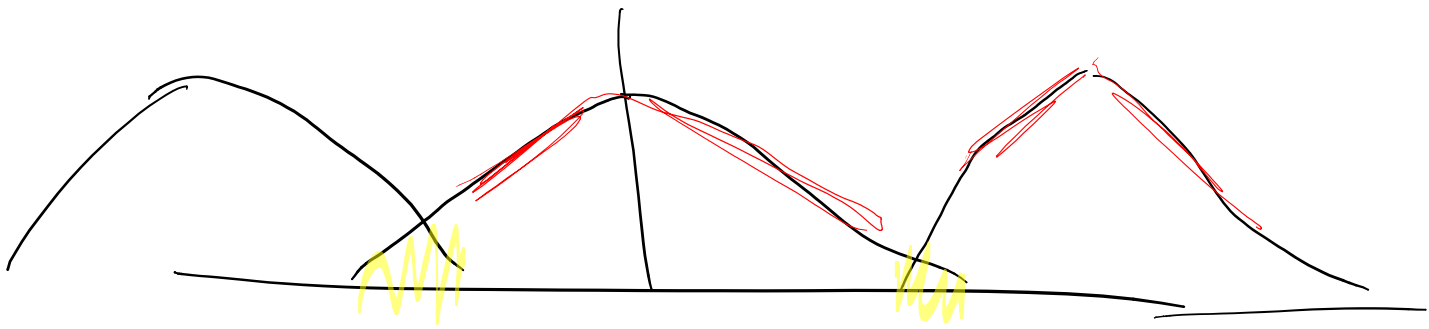
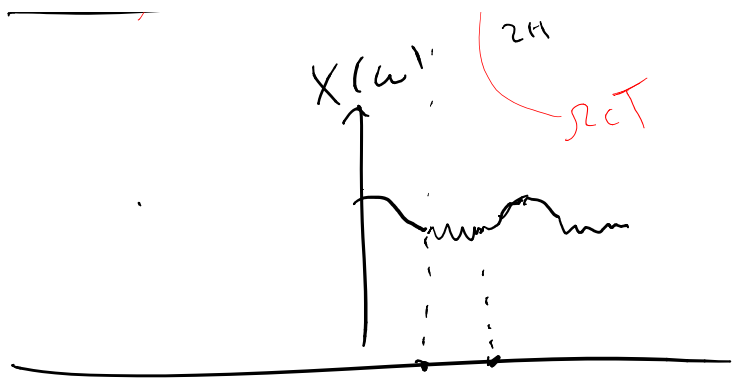
$$X(\omega) = \frac{1}{T} \sum_{k=-\infty}^{+\infty} X_c\left(\frac{\omega - 2\pi k}{T}\right)$$



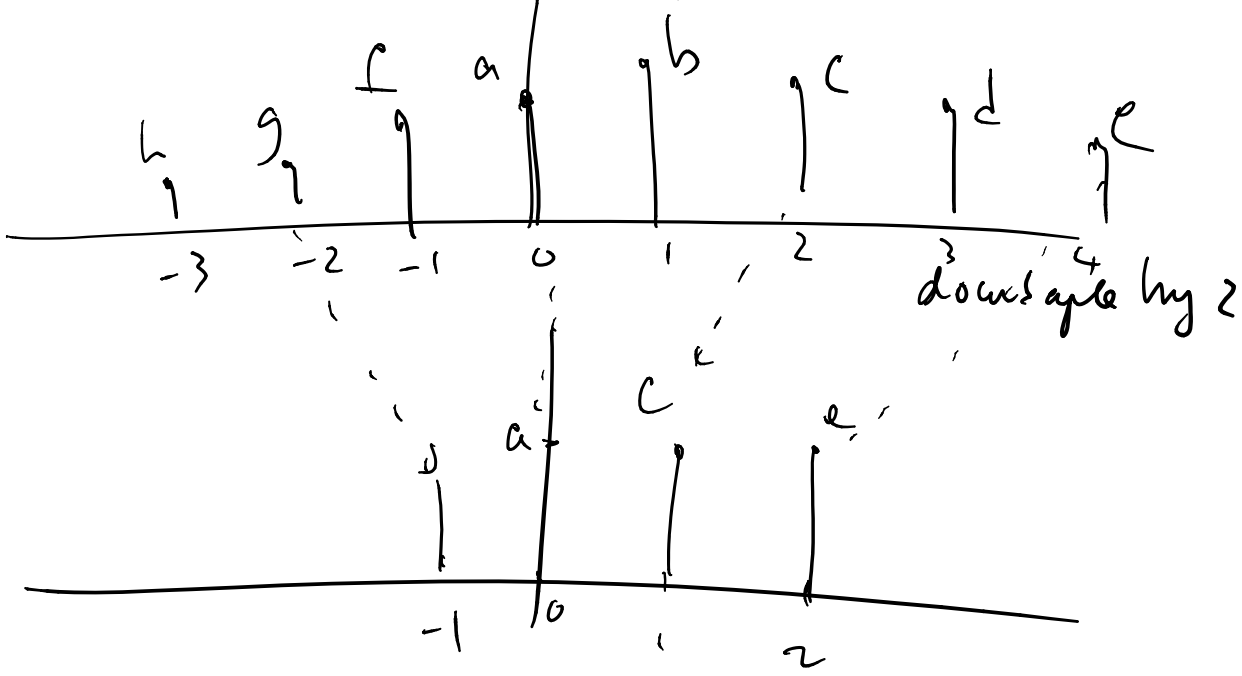
If $\Omega_c < \frac{\pi}{T} \rightarrow$ no aliasing
 \Rightarrow perfectly recover

⇒ perfectly recover $x_c(t)$ from $x(n)$

- If $\Omega_c > \frac{\pi}{T} \rightarrow$
aliasing

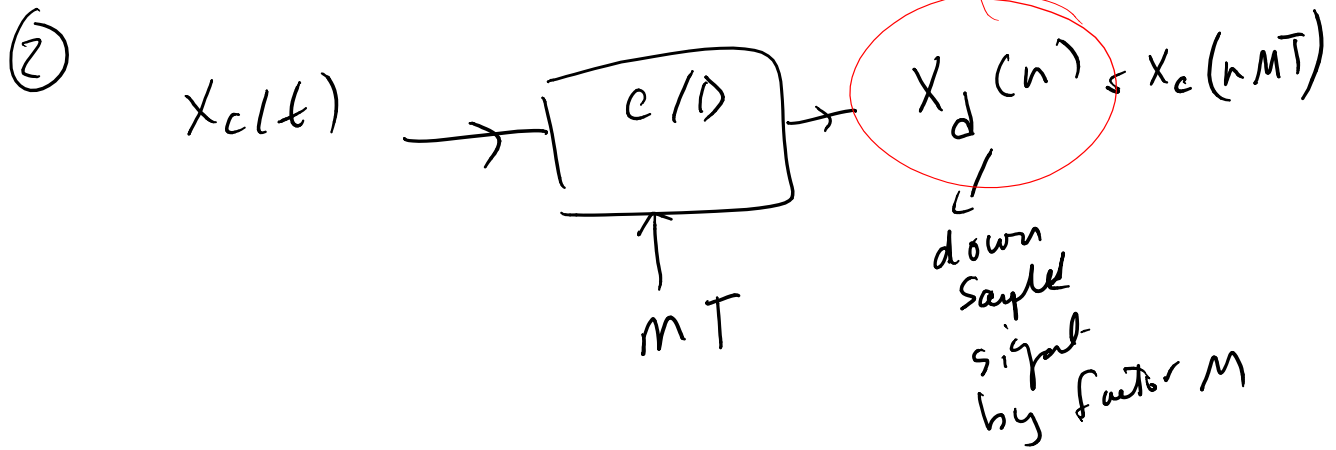
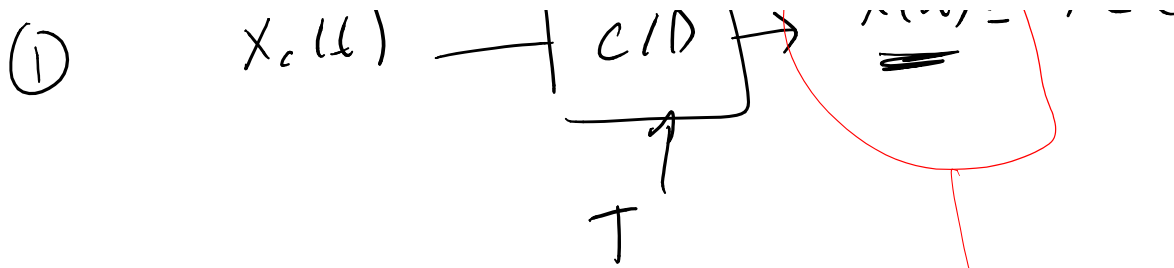


Downsampling



$x_c(t)$

(1) $x_c(t) \xrightarrow{CID} \underline{x(n)} = x_c(nT)$



$$X(\omega) = \text{DTFT}\{x(n)\} = \frac{1}{T} \sum_{k=-\infty}^{+\infty} X_c\left(\frac{\omega}{T} - \frac{2\pi k}{T}\right)$$

$$X_d(\omega) = \text{DTFT}\{X_d(n)\} = \frac{1}{MT} \sum_{k=-\infty}^{+\infty} X_c\left(\frac{\omega}{MT} - \frac{2\pi k}{MT}\right)$$

change of variable $r = i + kM$

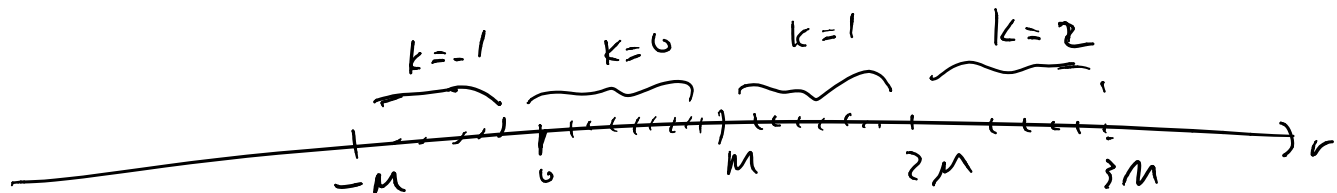
$$-\infty < k < +\infty$$

$$0 \leq i \leq M-1$$

$$\left. \begin{array}{l} k=0 \\ i: 0 \rightarrow M-1 \end{array} \right\} \rightarrow r: 0 \rightarrow M-1$$

$$\left. \begin{array}{l} k=1 \\ i: 0 \rightarrow M-1 \end{array} \right\} \rightarrow r: M \rightarrow 2M-1$$

$$\left. \begin{array}{l} k=2 \\ i: 0 \rightarrow M-1 \end{array} \right\} \rightarrow r: 2M \rightarrow 3M-1$$



$$X_d(\omega) = \frac{1}{M} \sum_{i=0}^{M-1} \frac{1}{T} \sum_{k=-\infty}^{+\infty} X_c \left(\frac{\omega}{MT} - \frac{2\pi k}{T} - \frac{2\pi i}{MT} \right)$$

$$X_d(\omega) = \frac{1}{M} \sum_{i=0}^{M-1} \frac{1}{T} \sum_{k=-\infty}^{+\infty} X_c \left(\frac{\omega - 2\pi i}{MT} - \frac{2\pi k}{T} \right)$$

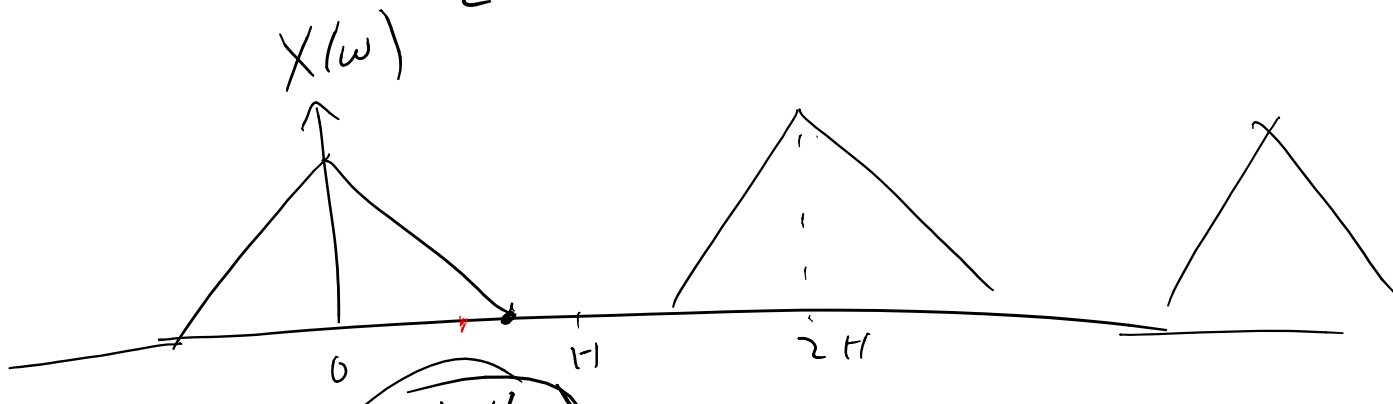
$[X(\omega)]$ $\omega \leftarrow \frac{\omega - 2\pi i}{M}$

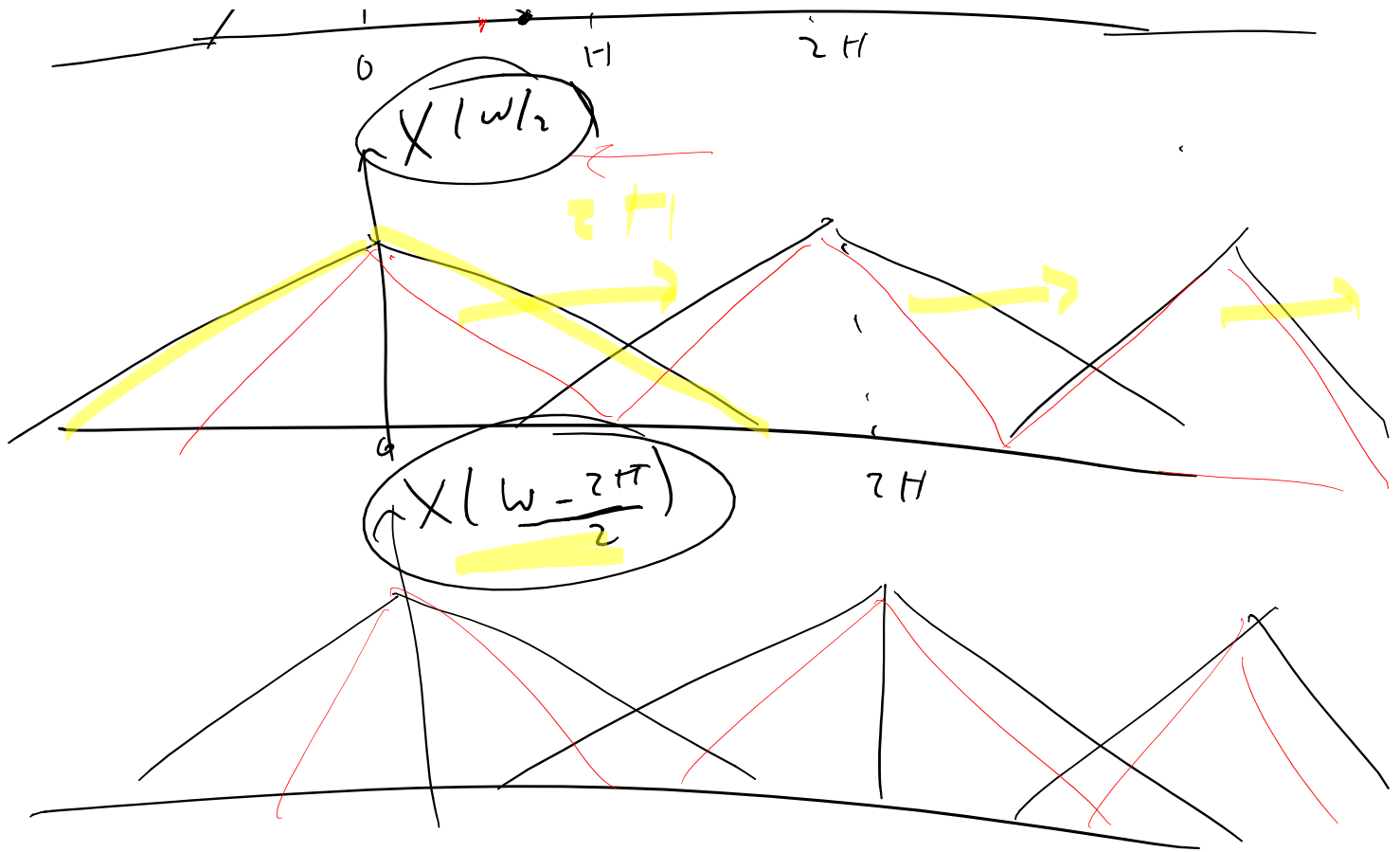
$$X_d(\omega) = \frac{1}{M} \sum_{i=0}^{M-1} X \left(\frac{\omega - 2\pi i}{M} \right)$$

$$M=2$$

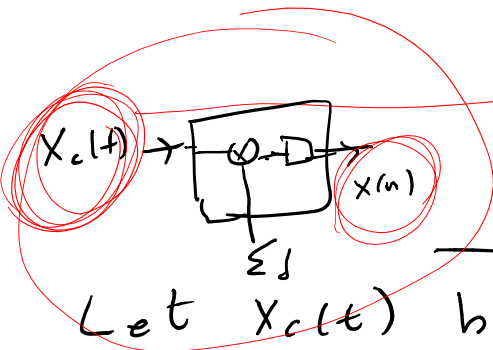
$$X_d(\omega) = \frac{1}{2} \sum_{i=0}^1 X \left(\frac{\omega}{2} - \frac{2\pi i}{2} \right)$$

$$= \frac{1}{2} \left[X \left(\frac{\omega}{2} \right) + X \left(\frac{\omega}{2} - \pi \right) \right]$$





words



Nyquist-Shannon theorem

Let $X_c(t)$ be a bandlimited signal with

$$\text{C.T.F.T } X_c(\Omega) = 0 \text{ for } |\Omega| > \Omega_m$$

Then $X_c(t)$ is uniquely determined by its

$$\text{samples } x(n) = X_c(t) \text{ } n=0, \pm 1, \pm 2, \pm 3, \dots$$

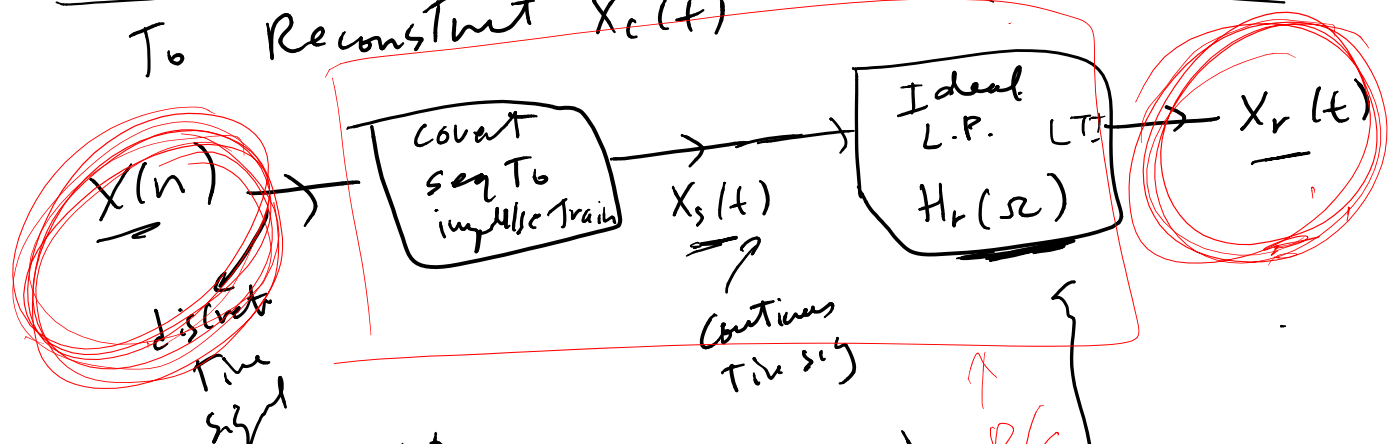
Sampler $X(n) = X_c(t) \quad n=0, \pm 1, \pm 2, \pm 3 \dots$

If $\Omega_s \triangleq \frac{2\pi}{T} \gg 2 \Omega_m$

DTFT of $X(n)$ $X_d(\omega) = \frac{1}{T} \sum_{k=-\infty}^{+\infty} X_c \left(\frac{\omega}{T} - \frac{2\pi k}{T} \right)$

Annotations: $X_d(\omega)$ is circled. $\frac{1}{T}$ is labeled "Scaling". $\frac{2\pi k}{T}$ is labeled "Periods".

To Reconstruct $X_c(t)$



$$X_s(t) = \sum_{n=-\infty}^{+\infty} X(n) \delta(t - nT)$$

$$X_r(t) = \sum_{n=-\infty}^{+\infty} X(n) h_r(t - nT)$$

Annotations: $h_r(t - nT)$ is circled in red and labeled "inputs rep.". A red arrow points from this equation to the LTI block in the diagram above.

$$X_r(\omega) = \sum_n X(n) \text{ e.t.f.t. } \left\{ h_r(t - nT) \right\}$$

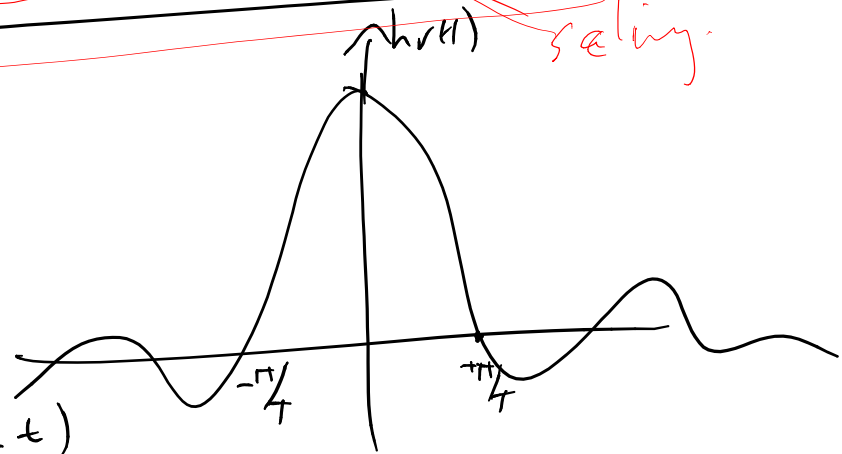
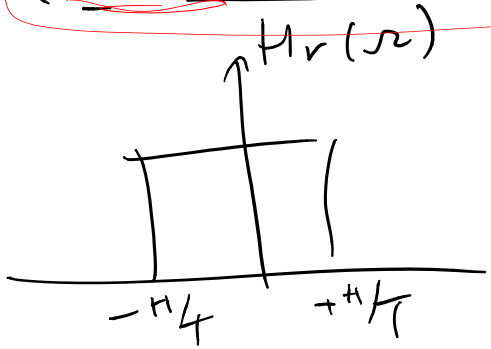
$$X_r(\omega) = \sum_{n=-\infty}^{+\infty} X(n) H_r(\omega) e^{-j\omega nT}$$

$$= H_r(\omega) \sum_{n=-\infty}^{+\infty} X(n) e^{-j\omega nT}$$

DTFT $\left[X_d(\omega) \right]_{\omega = \Omega T}$ $n = -\infty$

$$X_r(\Omega) = H_r(\Omega) X_d(\Omega T)$$

scaling



$$h_r(t) = \frac{\sin\left(\frac{\pi}{T}t\right)}{\frac{\pi}{T}t}$$

$$X_r(t) = \sum_{n=-\infty}^{\infty} x(n) \frac{\sin\left[\frac{\pi}{T}(t-nT)\right]}{\pi(t-nT)}$$

