Summer 2023

Discussion 6

1 Pre-Check

This section is designed as a conceptual check for you to determine if you conceptually understand and have any misconceptions about this topic. Please answer true/false to the following questions, and include an explanation:

1.1 Simplifying boolean logic expressions will not affect the performance of the hardware implementation.

False. Different gate arrangements that implement the same logic can have different propagation delays, which can affect the allowable clock speed.

1.2 The fewer logic gates, the faster the circuit (assuming each gate has the same propagation delays).

False. Propagation delays add to the allowable clock speed with the depth of the circuit, so a wide circuit with more gates in parallel can have less delay than just a few gates arranged in sequence.

1.3 The time it takes for clock-to-q and register setup can be greater than one clock cycle.

False. This can result in instability if registers are connected to each other, as register outputs may not have propagated properly before the next rising edge.

1.4 Every possible combinational logic circuit can be expressed by some combination of NOR gates.

True. NOR can be used to express AND, OR, and NOT gates. Thus, NOR is 'functionally complete' and can be used to represent any possible Boolean expression, and thus any CL circuit.

1.5 The shortest combinational logic path between two state elements is useful in determining circuit frequency and minimum clock cycle.

False. The minimum clock cycle has to allow enough time for every CL delay to settle on an output, so the frequency is based off of the **longest** CL delay possible in any area between state elements.

2 Boolean Logic

In digital electronics, it is often important to get certain outputs based on your inputs, as laid out by a truth table. Truth tables map directly to Boolean expressions, and Boolean expressions map directly to logic gates. However, in order to minimize the number of logic gates needed to implement a circuit, it is often useful to simplify long Boolean expressions.

We can simplify expressions using the nine key laws of Boolean algebra:

Name	AND Form	OR form
Commutative	AB = BA	A + B = B + A
Associative	AB(C) = A(BC)	A + (B + C) = (A + B) + C
Identity	1A = A	0 + A = A
Null	0A = 0	1 + A = 1
Absorption	A(A + B) = A	A + AB = A
Distributive	(A + B)(A + C) = A + BC	A(B + C) = AB + AC
Idempotent	A(A) = A	A + A = A
Inverse	$A(\overline{A}) = 0$	$A + \overline{A} = 1$
De Morgan's	$\overline{AB} = \overline{A} + \overline{B}$	$\overline{A + B} = \overline{A}(\overline{B})$

2.1 Simplify the following Boolean expressions:

(a)
$$(A+B)(A+\overline{B})C$$

$$(A+B)(A+\overline{B})C = (A+B\overline{B})C$$

= AC

(b)
$$\overline{A}\overline{B}\overline{C} + \overline{A}B\overline{C} + AB\overline{C} + A\overline{B}\overline{C} + ABC + A\overline{B}C$$

$$\begin{split} \overline{A}\overline{C}(\overline{B}+B) + A\overline{C}(B+\overline{B}) + AC(B+\overline{B}) &= \overline{A}\overline{C} + A\overline{C} + AC \\ &= \overline{A}\overline{C} + A\overline{C} + A\overline{C} + AC \\ &= (\overline{A}+A)\overline{C} + A(\overline{C}+C) \\ &= \overline{C} + A \end{split}$$

(c)
$$\overline{A(\overline{B}\overline{C} + BC)}$$

$$\overline{A(\overline{B}\overline{C} + BC)} = \overline{A} + \overline{B}\overline{C} + BC$$

$$= \overline{A} + (\overline{B}\overline{C})\overline{B}\overline{C}$$

$$= \overline{A} + (B + C)(\overline{B} + \overline{C})$$

$$= \overline{A} + B\overline{C} + \overline{B}C$$

(d)
$$\overline{A}(A+B) + (B+AA)(A+\overline{B})$$

$$\overline{A}(A+B) + (B+AA)(A+\overline{B}) = (\overline{A}A + \overline{A}B) + (B+AA)(A+\overline{B})$$

$$= \overline{A}B + (B+A)(A+\overline{B})$$

$$= \overline{A}B + (BA + AA + B\overline{B} + A\overline{B})$$

$$= \overline{A}B + (BA + A + A\overline{B})$$

$$= \overline{A}B + A$$

$$= A + B$$

Alternatively,

$$\overline{A}(A+B) + (B+AA)(A+\overline{B}) = \overline{A}(A+B) + (A+B)(A+\overline{B})$$

= $(A+B)(\overline{A}+A+\overline{B})$
= $A+B$

3 Combinational Logic Design

Logic gates can be connected together to create a variety of useful functions. In this question, we will implement a simplified version of the memory write mask for the RISC-V CPU. The memory write mask looks at the store instruction given and decides which of the four bytes (in one word of memory) to write to. It is four bits long - each bit is one if we should write to the corresponding byte, but zero if we shouldn't. For simplicity, assume that all memory addresses used in store instruction are word-aligned. Here's a truth table for the simplified memory mask:

Instruction	funct3	Out
sb	000	0001
sh	001	0011
sw	010	1111
(undefined)	011-111	XXXX

The x's for the final entry of the table indicates that any output is fine in that case.

3.1 Write out and simplify boolean expressions for each of the output bits in terms of the funct3 (input) bits f_2 , f_1 , f_0 .

We'll work on the bits from right to left. For Out[0], its value is one in all input cases that are defined, so we can just set Out[0] = 1.

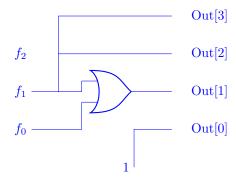
For the other bits, we can write out the base canonical form first, but then we can bring in any amount of terms from the undefined cases (equivalent to setting this bit to one in that undefined case) to simplify the expression.

Out[1] = $\overline{f_2}\overline{f_1}f_0 + \overline{f_2}f_1\overline{f_0}$ from the cases given. Since all of the cases where f_2 =1 are undefined, we can bring in some of those terms to cancel out the f_2 's: Out[1] = $\overline{f_2}\overline{f_1}f_0 + f_2\overline{f_1}f_0 + \overline{f_2}f_1\overline{f_0} + f_2f_1\overline{f_0} = \overline{f_1}f_0 + f_1\overline{f_0}$. We can go one step further: the input 011 is also undefined, so if we bring in that term (and the corresponding f_2 =1 term 111), we end up with Out[1] = $\overline{f_1}f_0 + f_1\overline{f_0} + f_1f_0 = f_1 + f_0$.

Following a similar process, the final two bits simplify to $Out[3] = Out[2] = f_1$.

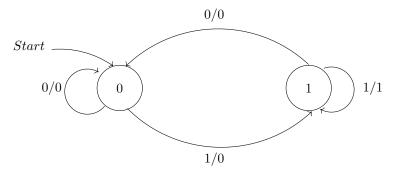
4 FSM, SDS

3.2 Draw out the boolean circuit for this memory write mask based on your simplified expressions above. You may use constants 0 and 1, and the logic gates AND, OR, NOT.



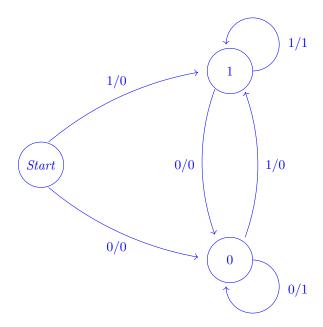
4 FSM

4.1 What pattern in a bitstring does the FSM below detect? What would it output for the input bitstring "0110010011110"?

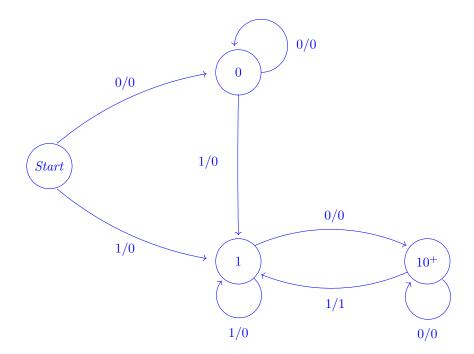


The FSM outputs a 1 if it detects the pattern "11". The FSM would output "001000000110"

4.2 Fill in the following FSM for outputting a 1 whenever we have two repeating bits as the most recent bits, and a 0 otherwise. You may not need all states.

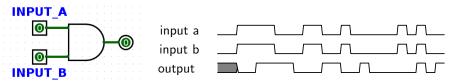


4.3 Draw an FSM that will output a 1 if it recognizes the regex pattern {10+1}. (That is, if the input forms a pattern of a 1, followed by one or more 0s, followed by a 1.)



5 State Intro

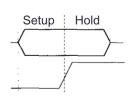
There are two basic types of circuits: combinational logic circuits and state elements. **Combinational logic** circuits simply change based on their inputs after whatever propagation delay is associated with them. For example, if an AND gate (pictured below) has an associated propagation delay of 2ps, its output will change based on its input as follows:



You should notice that the output of this AND gate *always* changes 2ps after its inputs change.

State elements, on the other hand, can *remember* their inputs even after the inputs change. State elements change value based on a clock signal. A rising edge-triggered register, for example, samples its input at the rising edge of the clock (when the clock signal goes from 0 to 1).

Like logic gates, registers also have a delay associated with them before their output will reflect the input that was sampled. This is called the **clk-to-q** delay. ("Q" often indicates output). This is the time between the rising edge of the clock signal and the time the register's output reflects the input change.



The input to the register samples has to be stable for a certain amount of time around the rising edge of the clock for the input to be sampled accurately. The amount of time before the rising edge the input must be stable is called the **setup** time, and the time after the rising edge the input must be stable is called the **hold** time. Hold time is generally

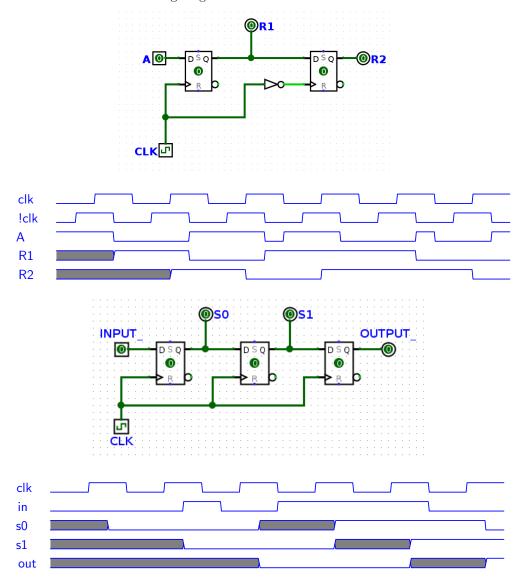
included in clk-to-q delay, so clk-to-q time will usually be greater than or equal to hold time. Logically, the fact that clk-to-q \geq hold time makes sense since it only takes clk-to-q seconds to copy the value over, so there's no need to have the value fed into the register for any longer.

For the following register circuit, assume **setup** of 2.5ps, **hold** time of 1.5ps, and a **clk-to-q** time of 1.5ps. The clock signal has a period of 13ps.

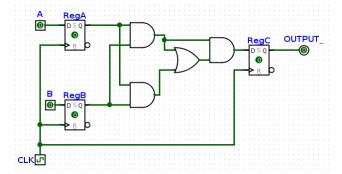


You'll notice that the value of the output in the diagram above doesn't change immediately after the rising edge of the clock. Until enough time has passed for the output to reflect the input, the value held by the output is garbage; this is represented by the shaded gray part of the output graph. Clock cycle time must be small enough that inputs to registers don't change within the hold time and large enough to account for clk-to-q times, setup times, and combinational logic delays.

5.1 For the following 2 circuits, fill out the timing diagram. The clock period (rising edge to rising edge) is 8ps. For every register, clk-to-q delay is 2ps, setup time is 4ps, and hold time is 2ps. NOT gates have a 2ps propagation delay, which is already accounted for in the !clk signal given.



[5.2] In the circuit below, RegA and RegB have setup, hold, and clk-to-q times of 4ns, all logic gates have a delay of 5ns, and RegC has a setup time of 6ns. What is the maximum allowable hold time for RegC? What is the minimum acceptable clock cycle time for this circuit, and clock frequency does it correspond to?



The maximum allowable hold time for RegC is how long it takes for RegC's input to change, so (clk-to-q of A or B) + shortest CL time = 4 + (5 + 5) = 14 ns.

The minimum acceptable clock cycle time is clk-to-q + longest CL time + setup time = 4 + (5 + 5 + 5) + 6 = 25 ns.

25 ns corresponds to a clock frequency of $(1/(25*10^{-9}))s^{-1} = 40MHz$