

Lecture 23 - Q/A session for MT II.

Administrivia → Grade corrections will be addressed after MT II

→ Justin's review session tonight (5-7) in 101 Morgan

→ Review problem solutions up by tonight

Note: TBP Telebeam advising sessions

→ Tu Apr. 12	} 6 pm - 8 pm Wozniak lounge 4 th floor Soda hall
→ Mon. Apr. 18	
→ Wed. Apr. 20 th	

(FREE FOOD!)

Midterm II → (1) op-amps I (cascade?) (25 points)

(3 points
for filling
out front
page correctly)

(2) Dynamic route I (es: oscillator) (30 pts)

(3) Dynamic route II (es: flip-flop) (30 pts)
(Lecture 20)


(4) op-amp II (difficult?) (12 pts)

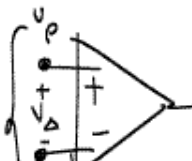
Total: 100 pts

Important Concepts:

Es. points: $f'(t) = 0$

Dynamic route. Illustrate
on $i-v$ graph
 $(i(t), v(t))$

op-amps → Linear $v_o = A(v_p - v_n)$ 
→ Saturation (i.e. rail)

$v_o > 0$  $v_o = +V_{cc}$ rail

$v_o < 0$  $v_o = -V_{cc}$ rail

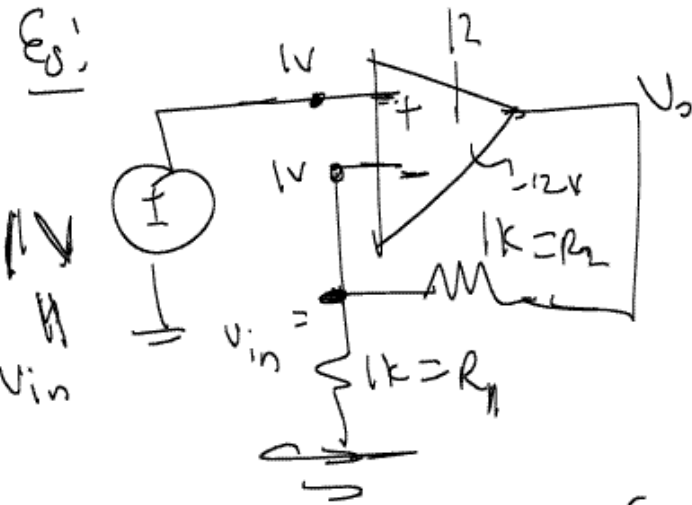
$$V_{in} = 1V, \quad V_o = \left(1 + \frac{1k}{1k}\right) 1V$$

$$= 2 \cdot 1 = \underline{\underline{2V}}$$

Since $2V < 12V$

\Rightarrow op-amp does not rail

$\Rightarrow v_p \approx v_n = 1V$



Non-inverting topology: $V_o = \left(1 + \frac{R_2}{R_1}\right) V_{in}$

Cheat sheet: inverting amplifier

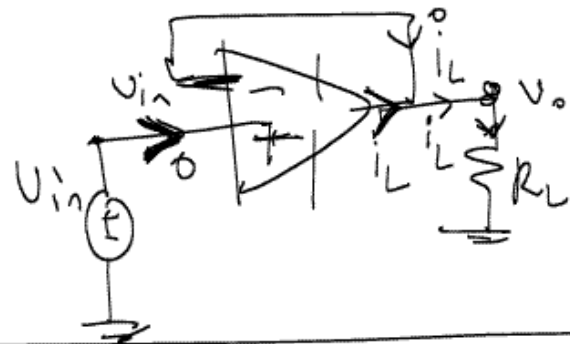
noninverting amplifier

voltage follower

inverting summing amplifiers, differential amplifiers



Note: Purpose of voltage follower:

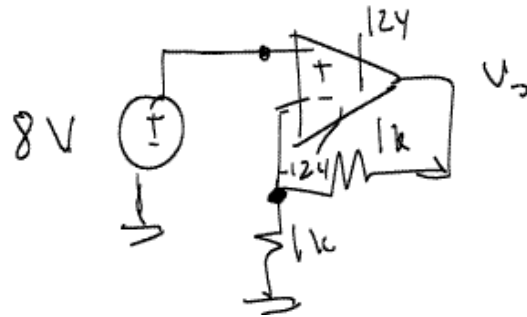


$$V_o = V_{in}, \quad i_L = \frac{V_o}{R_L} = \frac{V_{in}}{R_L}$$

signs
rails

⇒ V_{in} does not put out any current,
op-amp does all the work!!!

Ex 2:

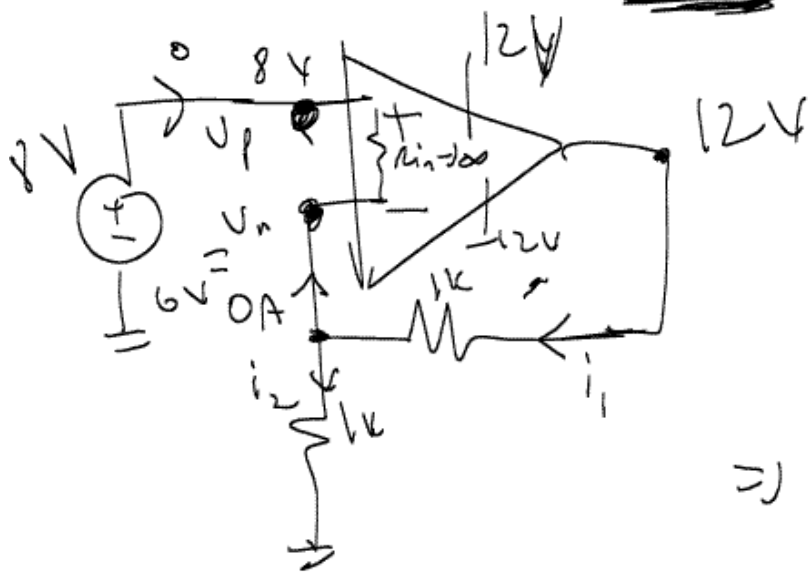


$$V_o = \left(1 + \frac{1k}{1k}\right) 8V$$

$$= 2 \cdot 8 = \underline{\underline{16V}}$$

Oops, $V_o = 12V$ (op-amp rails)

i.e., $V_o = \left(1 + \frac{R_2}{R_1}\right) V_{in}$ is invalid

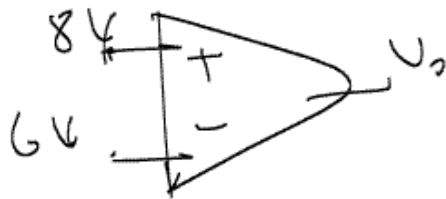


$$i_1 = i_2 \text{ (KCL @ } V_n)$$

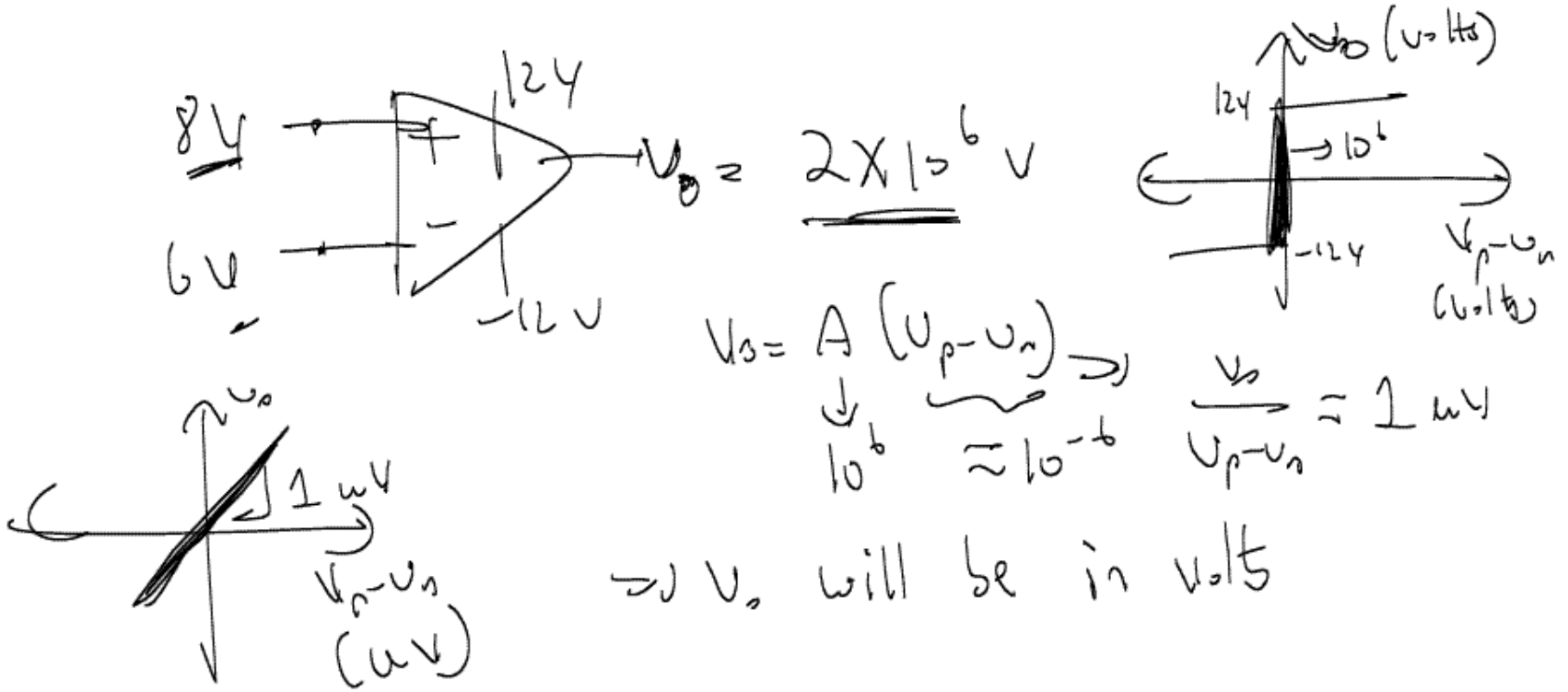
$$\Rightarrow \frac{12 - V_n}{1k} = \frac{V_n}{1k}$$

$$\Rightarrow 12 = 2V_n \Rightarrow \boxed{V_n = 6V}$$

Notice:

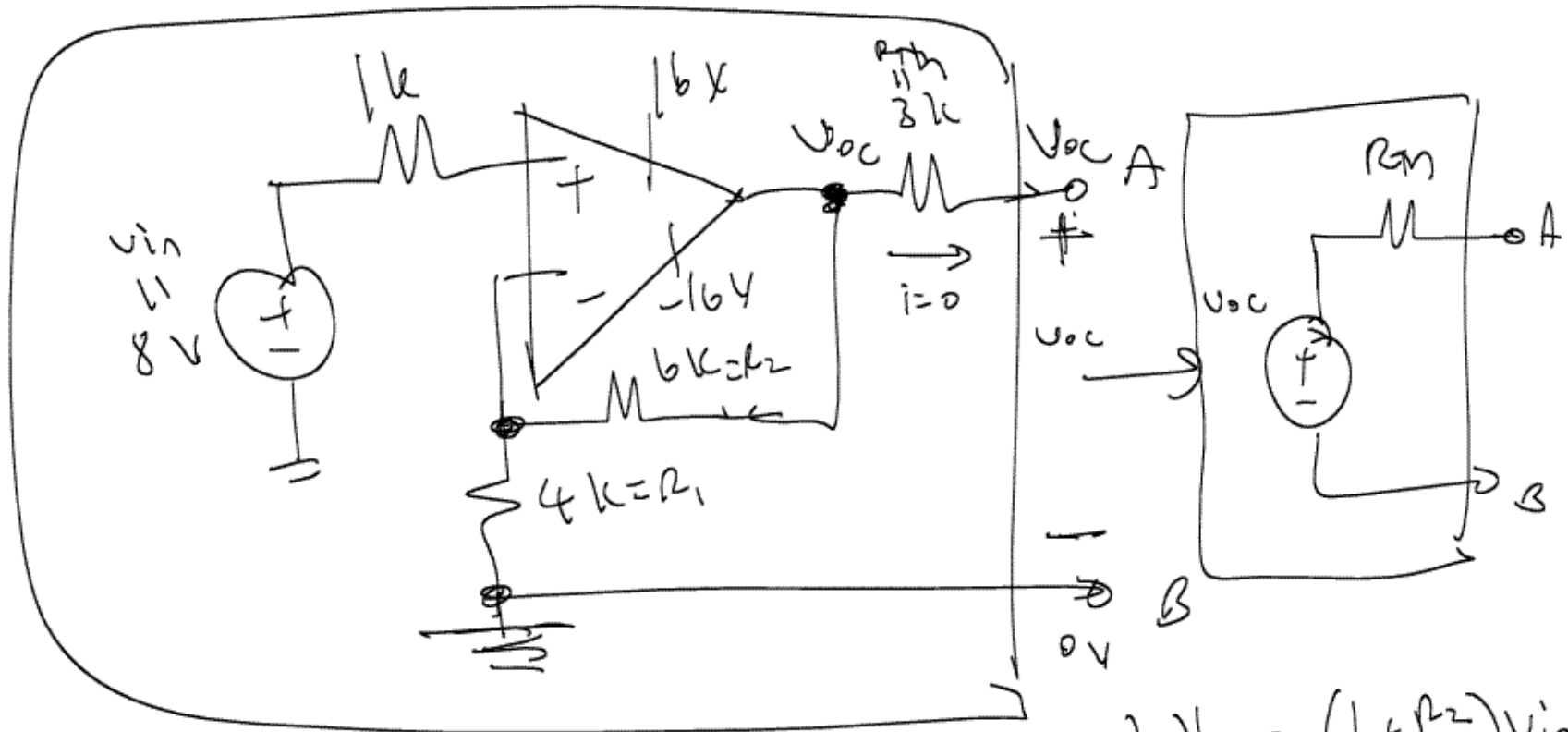


$$V_o = A (V_p - V_n) = 2 \times 10^6 V, \quad A \approx 10^6$$



Questions on Review Problems

2(b): Thev. equivalent at AB:



V_{oc} = Output Voltage of the op-amp

$$V_{oc} = \left(1 + \frac{R_2}{R_1}\right) V_{in}$$

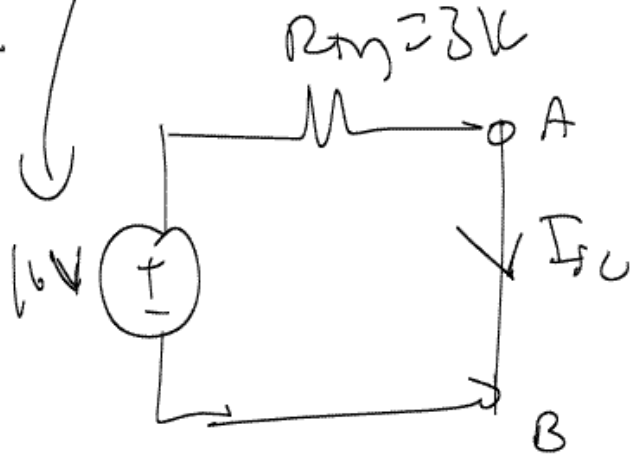
Op-amp verify!

$$V_{oc} = 16V$$

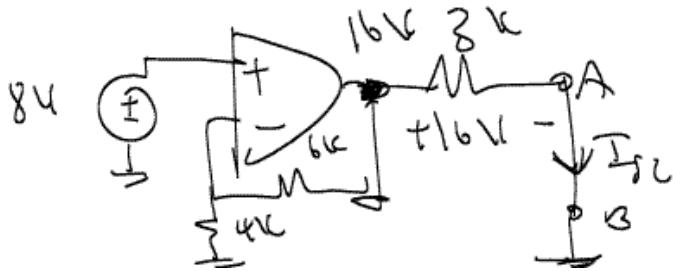
$$= \left(1 + \frac{6k}{4k}\right) 8$$

$$2.5 \cdot 8V = \underline{\underline{20V}}$$

fixed by the op-amp

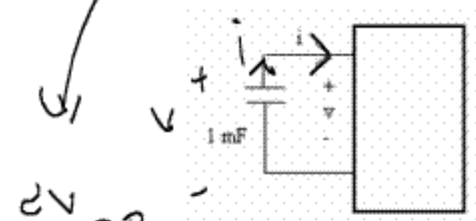


$$I_{sc} = \frac{V_{oc}}{R_m}$$



$$I_{sc} = \frac{16V}{3k}$$

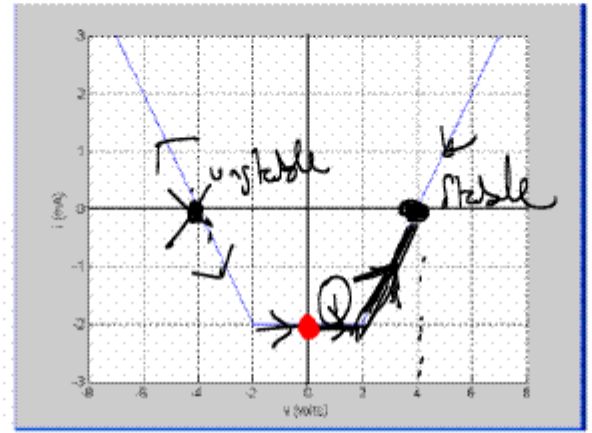
(c) $i = - (1 \text{ mF}) \frac{dv}{dt}$



$\frac{dv}{dt} = 0$

$\Rightarrow i = 0$

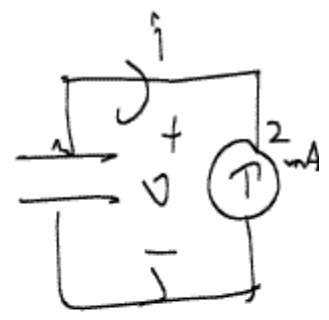
$i > 0, v' < 0$
 $i < 0, v' > 0$

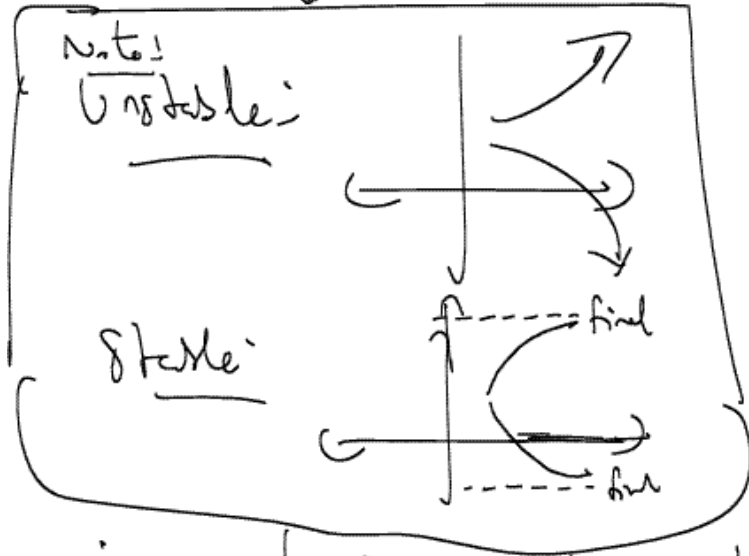
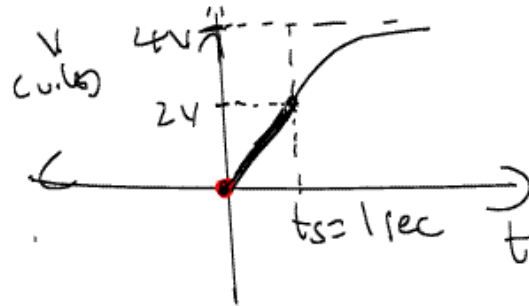
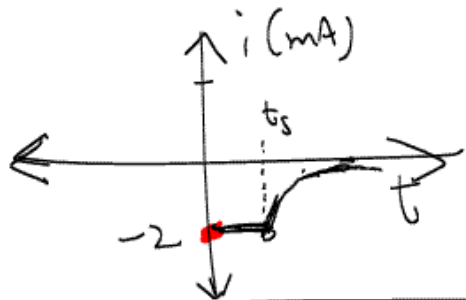


The i-v graph of the nonlinear circuit element is shown in the figure above. Assuming $v(0) = 0 \text{ V}$:

- (a) Find the equilibrium points and sketch the dynamic route.
- (b) Sketch $i(t)$ and $v(t)$, make sure to mark the switching times on the graph(s).

① \Rightarrow





$$i(t) = -2 \text{ mA for } t < t_s$$

$$\Rightarrow -(1 \text{ mF}) \frac{dv}{dt} = -2 \text{ mA}$$

$$\Rightarrow \frac{dv}{dt} = 2 \Rightarrow v(t) = 2t + v_{\text{initial}}$$

$$v_{\text{initial}} = v(0) = 0V$$

$$v(t) = 2t \text{ volts}$$

$$v(t_s) = 2V \Rightarrow [t_s = 1 \text{ second}]$$

$$i(t) = \begin{cases} -2 \text{ mA} & \text{if } t < t_s \\ -2e^{-\frac{(t-t_s)}{\tau}} \text{ mA} & \text{if } t > t_s \end{cases}$$

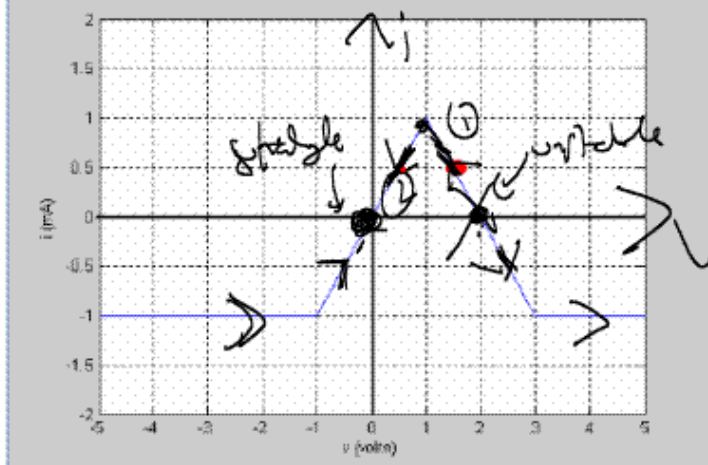
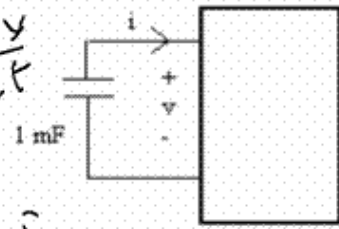
Pages Attachments Comments

Problem 1

(a)

$$i = (-1 \text{ mF}) \frac{dv}{dt}$$

$$\frac{dv}{dt} = 0 \Rightarrow \dot{i} = 0$$



The i-v graph of the nonlinear circuit element is shown in the figure above. Assuming $i(0) = 0.5 \text{ mA}$:

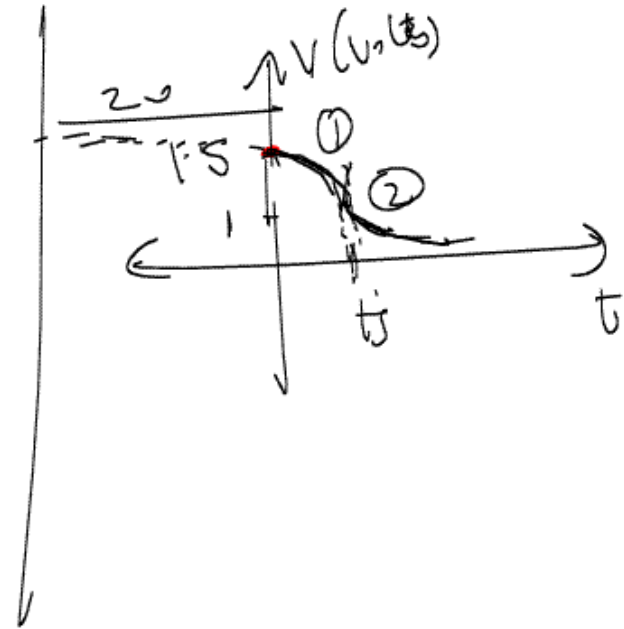
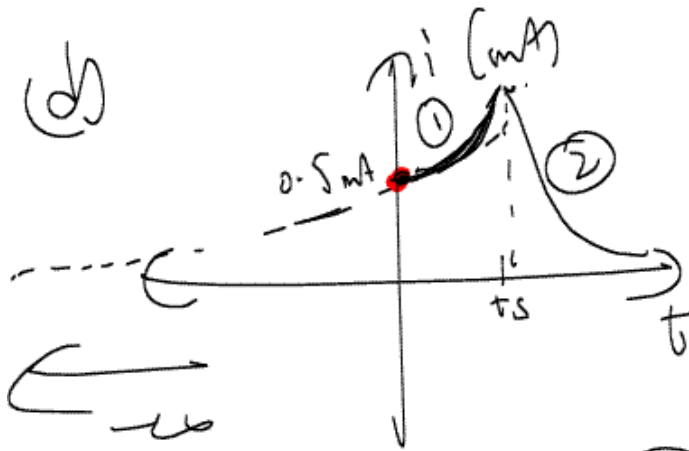
$i > 0, v < 0$
 $i < 0, v > 0$

$v(0) = 1.5 \text{ V}$

- (c) Find the equilibrium points and sketch the dynamic route.
 - (d) Sketch $i(t)$ and $v(t)$, make sure to mark the switching times on the graph(s).
- (Hint: for the unstable part, think carefully about the final values of i and v).

You should look over the flip-flop example from Lecture 20 on the EE100 homepage. **MAKE SURE YOU UNDERSTAND THIS PROBLEM FOR THE MIDTERM!** Of course, you could surf <http://hkn.eecs.berkeley.edu/student/onlineexams.shtml> for online exams from EE100 and EE100, but even if these problems are way too difficult and

(d)



$$i_i(t) = i_f + (i_i - i_f) e^{-t/\tau}$$

$$i_{cb}(t) = i_f + (i_i - i_f) e^{-t/\tau}$$

Annotations for the equation above: i_f is labeled 0 V ; i_i is labeled 0.5 mA ; τ is labeled $t/2$.

