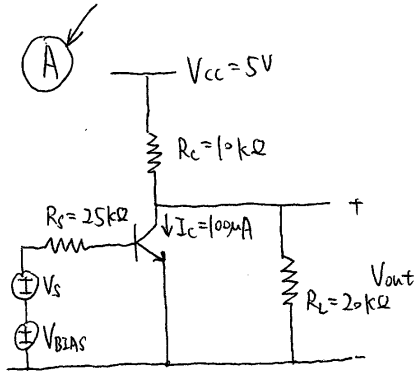
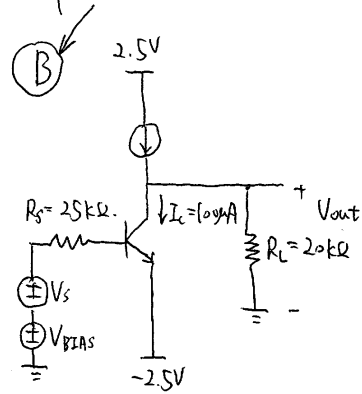


Homework #10 Solution EE105 (Spring 2001) by Xuesong Jiang

8.1. Circuit in the 2nd midterm



Improved circuit



(a) In both circuits

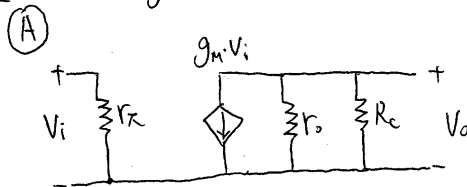
$$V_{BE} = \frac{kT}{q} \ln \frac{I_C}{I_S} = 26 \text{ mV} \cdot \ln \frac{10^{-4} \text{ A}}{10^{-16} \text{ A}} = 718.4 \text{ mV}$$

$$I_B = \frac{I_C}{\beta} = \frac{100 \mu\text{A}}{100} = 1 \mu\text{A}$$

$$\text{In (A), } V_{BIAS} = I_B \cdot R_S + V_{BE} = 1 \mu\text{A} \cdot 25 \text{ k}\Omega + 718.4 \text{ mV} = \boxed{743.4 \text{ mV}}$$

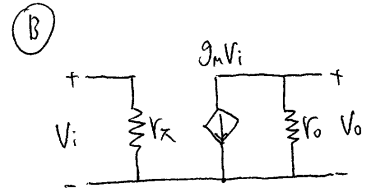
$$\text{In (B), } V_{BIAS} = I_B \cdot R_S + V_{BE} - 2.5 \text{ V} = \boxed{-1.757 \text{ V}}$$

(b) Small signal model



$$r_{\pi} = \frac{\beta}{g_m} = \frac{100}{3.846 \text{ mS}} = 26 \text{ k}\Omega$$

$$g_m = \frac{q \cdot I_C}{kT} = \frac{100 \mu\text{A}}{26 \text{ mV}} = 3.846 \text{ mS}$$



$$r_o = \frac{V_A}{I_C} = \frac{20 \text{ V}}{100 \mu\text{A}} = 200 \text{ k}\Omega$$

For both circuits

$$R_{in} = r_{\pi} = 26 \text{ k}\Omega$$

$$g_m = g_m = 3.846 \text{ mS}$$

In (A),

$$R_{out} = r_{o1} || R_c = \frac{10 \text{ k}\Omega \cdot 200 \text{ k}\Omega}{10 \text{ k}\Omega + 200 \text{ k}\Omega} = 9.524 \text{ k}\Omega$$

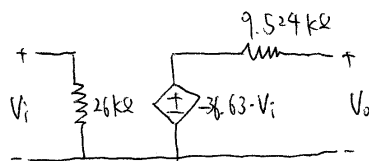
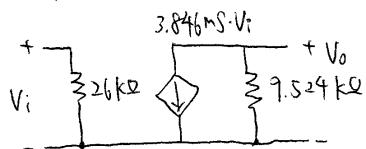
$$A_v = -g_m \cdot R_{out} = -3.846 \text{ mS} \cdot 9.524 \text{ k}\Omega = -36.63$$

In (B),

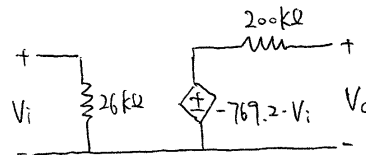
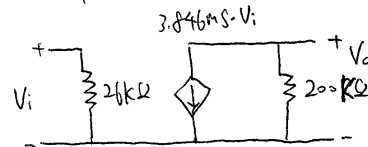
$$R_{out} = r_o = 200 \text{ k}\Omega$$

$$A_v = -g_m \cdot R_{out} = -3.846 \text{ mS} \cdot 200 \text{ k}\Omega = -769.2$$

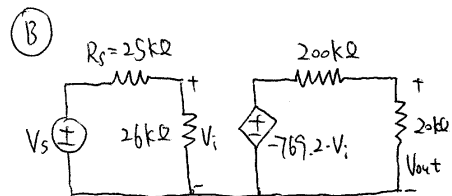
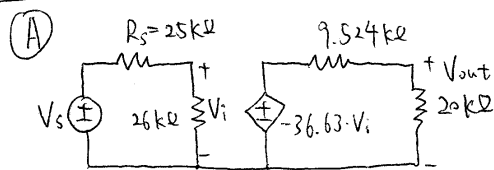
Two port model for (A)



Two port model for (B)



(C). Overall circuit small signal model



For (A),

$$\text{Overall voltage gain} = \frac{V_{out}}{V_s} = \frac{26k\Omega}{25k\Omega + 26k\Omega} \cdot (-36.63) \cdot \frac{20k\Omega}{20k\Omega + 9.524k\Omega}$$
$$= \boxed{-12.65}$$

For (B),

$$\text{Overall voltage gain} = \frac{V_{out}}{V_s} = \frac{26k\Omega}{25k\Omega + 26k\Omega} \cdot (-769.2) \cdot \frac{20k\Omega}{200k\Omega + 20k\Omega}$$
$$= \boxed{-35.65}$$

(d)(i) If $V_{CE} > V_{CE-sat}$, the bipolar transistor is in forward active region, $I_C = \beta \cdot I_B$ and small signal $i_c = \beta \cdot i_b$

$$\text{For (A)} \quad V_o = -i_c \cdot (R_L \parallel R_C \parallel r_o) = -\beta \cdot i_b \cdot (R_L \parallel R_C \parallel r_o) = -\beta \cdot \frac{V_s}{R_s + r_x} \cdot (R_L \parallel R_C \parallel r_o)$$

$$\text{For (B)} \quad V_o = -i_c \cdot (R_L \parallel r_o) = -\beta \cdot i_b \cdot (R_L \parallel r_o) = -\beta \cdot \frac{V_s}{R_s + r_x} \cdot (R_L \parallel r_o)$$

V_o is a linear function of V_i

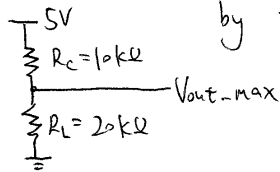
(ii) If $V_{CE} = V_{CE-sat}$, $i_c \neq \beta \cdot i_b$, the overall voltage gain of

both circuit will drop, the input signal is clipped

For (A), the negative output swing will be 0.1V,

For (B), the negative output swing will be -2.4V

(iii) For (A), the maximum positive output swing can be reached when I_c of the bipolar transistor becomes zero, it is determined by the voltage divider.



$$V_{out-max} = 5V \cdot \frac{20k\Omega}{10k\Omega + 20k\Omega} = 3.333V$$

For (B), the maximum positive output swing can be reached when I_c of the bipolar transistor becomes zero, all current from the current source will flow through R_L

$$V_{out-max} = 100\mu A \cdot 20k\Omega = 2V$$

$$\therefore \begin{array}{l} \text{For (A)} \quad V_{out} \in (0.1V, 3.333V) \\ \text{For (B)} \quad V_{out} \in (-2.4V, 2V) \end{array}$$