

## Lecture 19

### OUTLINE

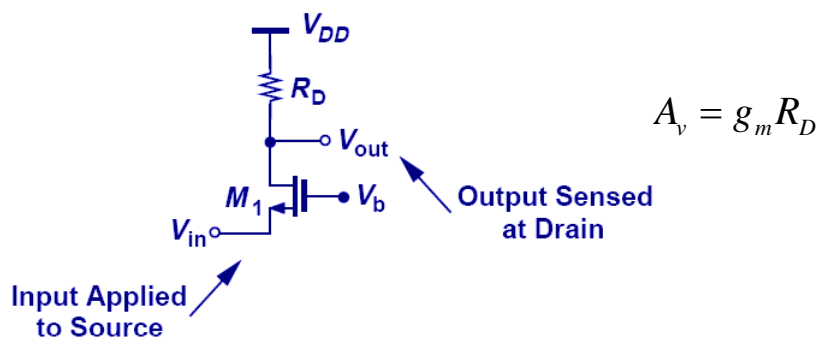
- Common-gate stage
- Source follower
- Reading: Chap. 7.3-7.4

EE105 Spring 2008

Lecture 19, Slide 1

Prof. Wu, UC Berkeley

## Common-Gate Stage



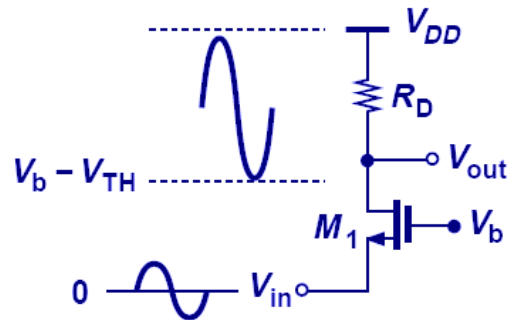
- Common-gate stage is similar to common-base stage: a rise in input causes a rise in output. So the gain is positive.

EE105 Spring 2008

Lecture 19, Slide 2

Prof. Wu, UC Berkeley

## Signal Levels in CG Stage



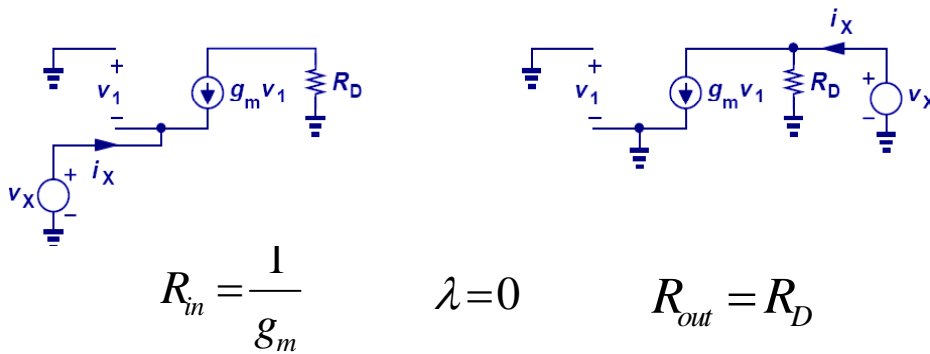
- In order to maintain M1 in saturation, the signal swing at  $V_{out}$  cannot fall below  $V_b - V_{TH}$

EE105 Spring 2008

Lecture 19, Slide 3

Prof. Wu, UC Berkeley

## I/O Impedances of CG Stage



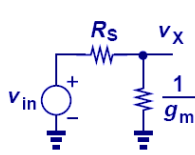
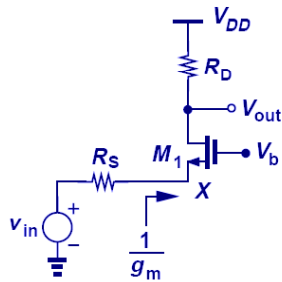
- The input and output impedances of CG stage are similar to those of CB stage.

EE105 Spring 2008

Lecture 19, Slide 4

Prof. Wu, UC Berkeley

## CG Stage with Source Resistance



$$v_x = \frac{\frac{1}{g_m}}{\frac{1}{g_m} + R_S} v_{in}$$

$$A_v = \frac{v_{out}}{v_x} \frac{v_x}{v_{in}} = g_m R_D \frac{\frac{1}{g_m}}{\frac{1}{g_m} + R_S}$$

$$= \frac{R_D}{\frac{1}{g_m} + R_S} = \frac{g_m R_D}{1 + g_m R_S}$$

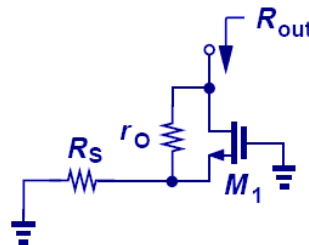
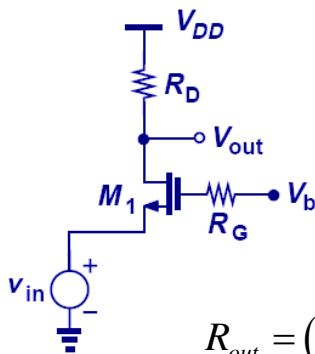
- When a source resistance is present, the voltage gain is equal to that of a CS stage with degeneration, only positive.

EE105 Spring 2008

Lecture 19, Slide 5

Prof. Wu, UC Berkeley

## Generalized CG Behavior



$$R_{out} = (1 + g_m r_o) R_S + r_o$$

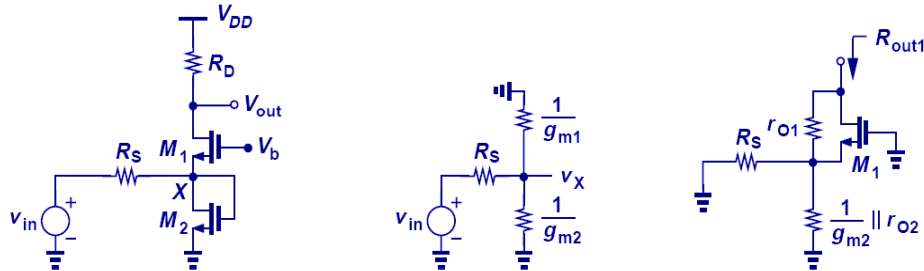
- When a gate resistance is present it does not affect the gain and I/O impedances since there is no potential drop across it (at low frequencies).
- The output impedance of a CG stage with source resistance is identical to that of CS stage with degeneration.

EE105 Spring 2008

Lecture 19, Slide 6

Prof. Wu, UC Berkeley

## Example of CG Stage



$$\frac{v_{out}}{v_{in}} = \frac{g_{m1} R_D}{1 + (g_{m1} + g_{m2}) R_S}$$

$$R_{out} \approx \left[ g_{m1} r_{o1} \left( \frac{1}{g_{m2}} \parallel R_S \right) + r_{o1} \right] \parallel R_D$$

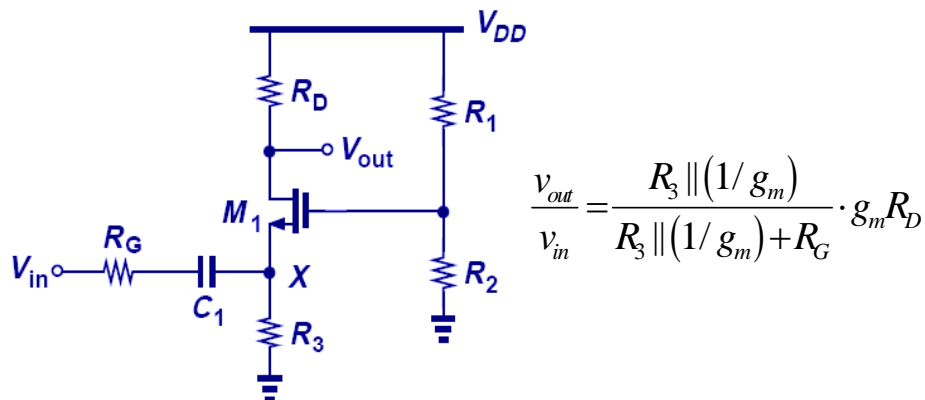
- Diode-connected M2 acts as a resistor to provide the bias current.

EE105 Spring 2008

Lecture 19, Slide 7

Prof. Wu, UC Berkeley

## CG Stage with Biasing



$$\frac{v_{out}}{v_{in}} = \frac{R_3 \parallel (1/g_m)}{R_3 \parallel (1/g_m) + R_G} \cdot g_m R_D$$

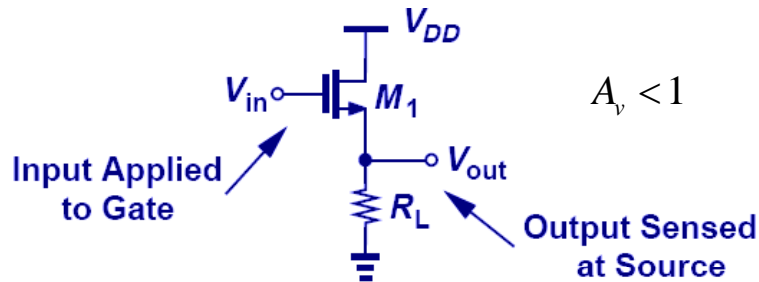
- $R_1$  and  $R_2$  provide gate bias voltage, and  $R_3$  provides a path for DC bias current of  $M_1$  to flow to ground.

EE105 Spring 2008

Lecture 19, Slide 8

Prof. Wu, UC Berkeley

## Source Follower Stage

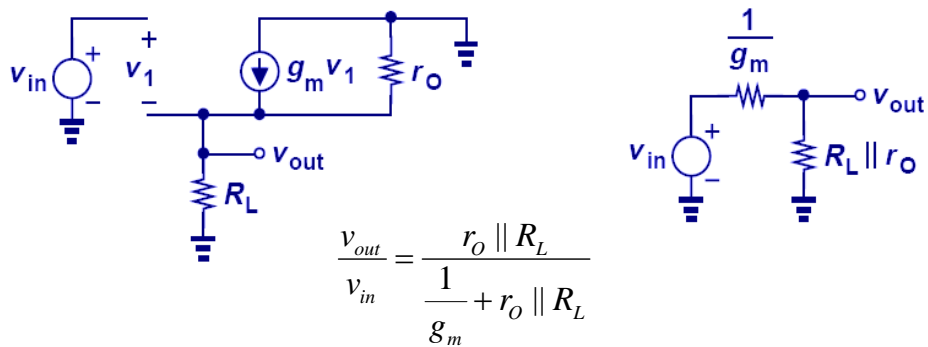


EE105 Spring 2008

Lecture 19, Slide 9

Prof. Wu, UC Berkeley

## Source Follower Core



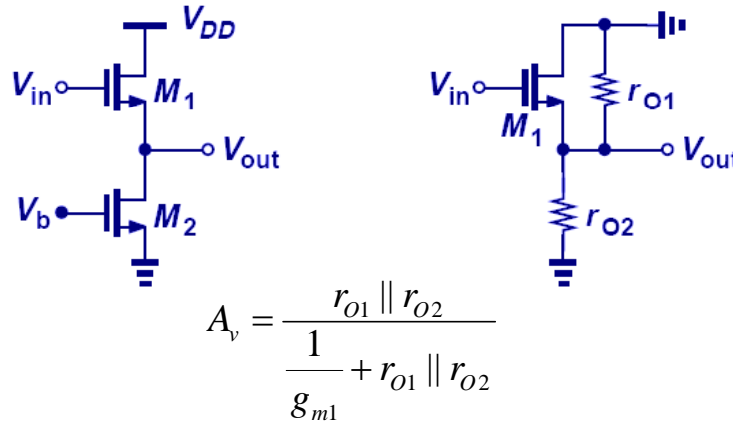
- Similar to the emitter follower, the source follower can be analyzed as a resistor divider.

EE105 Spring 2008

Lecture 19, Slide 10

Prof. Wu, UC Berkeley

## Source Follower Example



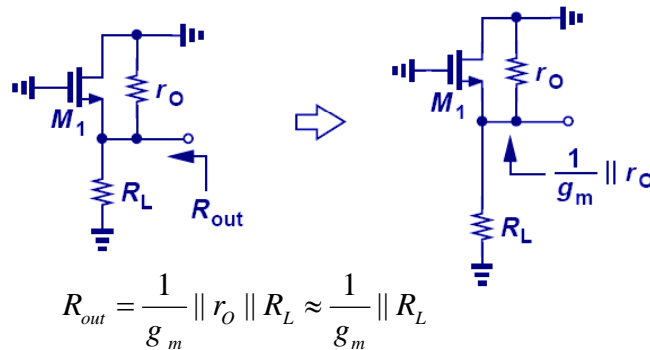
- In this example,  $M_2$  acts as a current source.

EE105 Spring 2008

Lecture 19, Slide 11

Prof. Wu, UC Berkeley

## Output Resistance of Source Follower



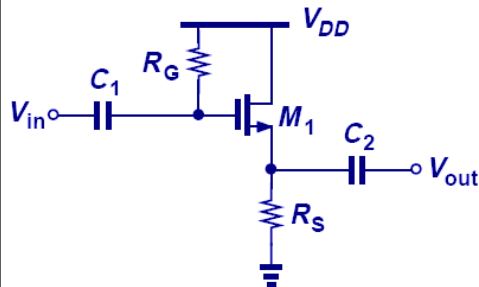
- The output impedance of a source follower is relatively low, whereas the input impedance is infinite (at low frequencies); thus, a good candidate as a buffer.

EE105 Spring 2008

Lecture 19, Slide 12

Prof. Wu, UC Berkeley

## Source Follower with Biasing



$$I_D = \frac{1}{2} \mu_n C_{ox} \frac{W}{L} (V_{DD} - I_D R_S - V_{TH})^2$$

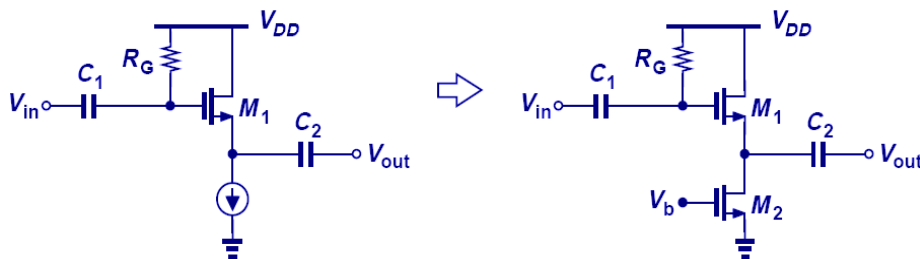
- $R_G$  sets the gate voltage to  $V_{DD}$ , whereas  $R_S$  sets the drain current
- The quadratic equation above can be solved for  $I_D$

EE105 Spring 2008

Lecture 19, Slide 13

Prof. Wu, UC Berkeley

## Supply-Independent Biasing



- If  $R_S$  is replaced by a current source, drain current  $I_D$  becomes independent of supply voltage.

EE105 Spring 2008

Lecture 19, Slide 14

Prof. Wu, UC Berkeley