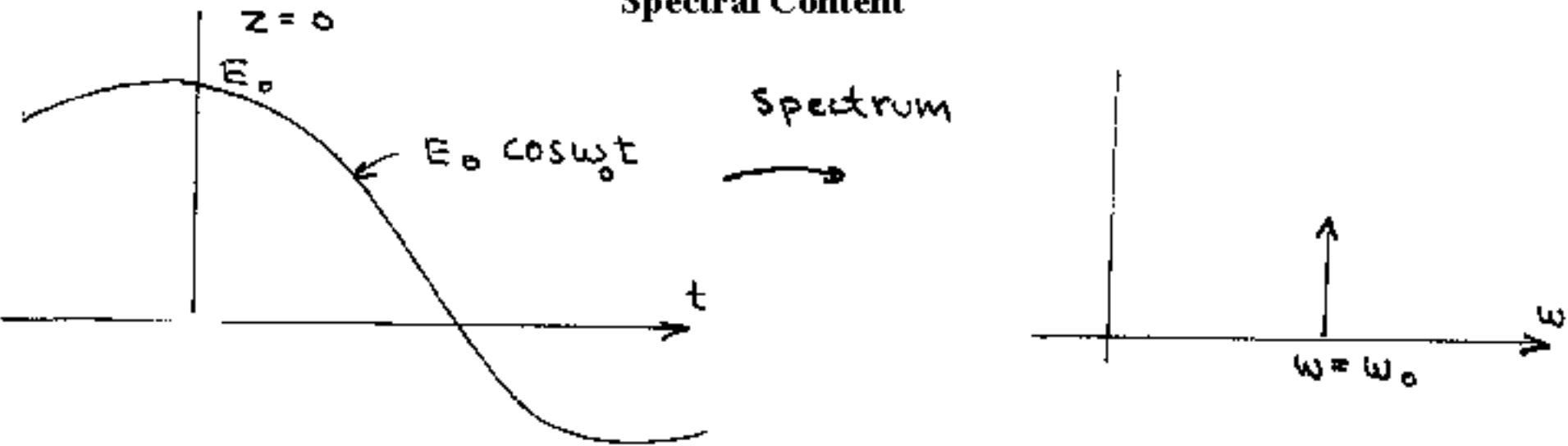


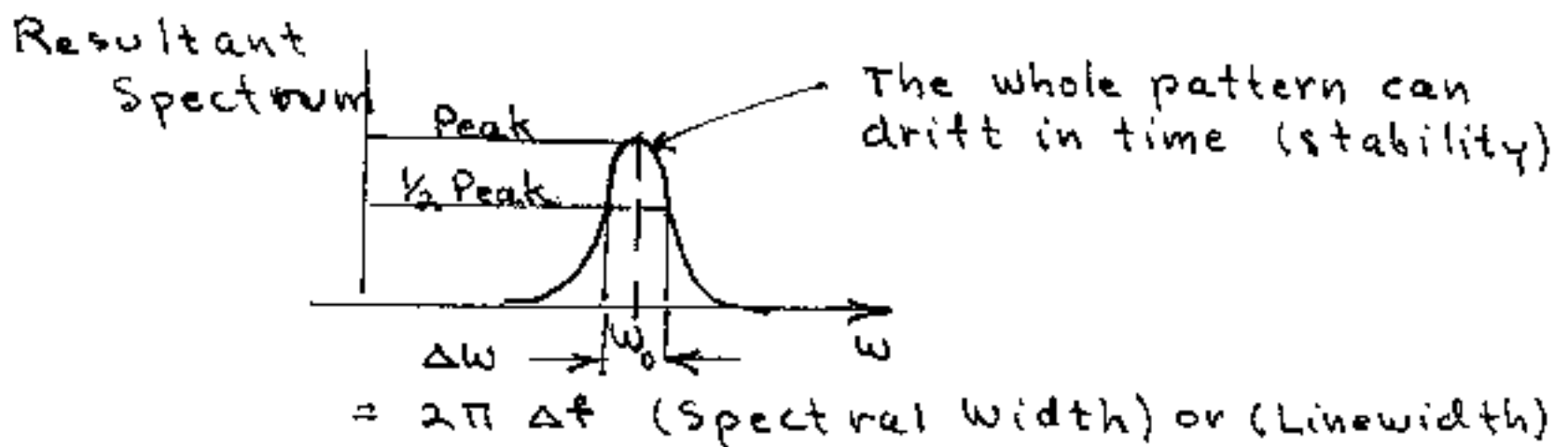
Spectral Content



a) Ideal Continuous Cosine at $z=0$

b) Laser - Non-ideal and Thus Has

- 1) ω_0 varying in time (stability) (drift).
- 2) rapid variation due to "phase fluctuations" and amplitude fluctuations.
- 3) When modulated with a signal the electric field spectrum is further broadened.



Example question!

If $\Delta f = 100 \text{ MHz}$ at $\lambda_0 = 1 \mu\text{m}$, what is $\Delta \lambda$ in nanometers.

Solⁿ $\lambda f = c \Rightarrow \Delta \lambda f + \lambda \Delta f = 0 = \Delta \lambda = -\frac{\lambda}{f} \Delta f$
 Thus $\Delta \lambda = -(\lambda^2/c) \Delta f = -(1 \times 10^{-9})^2 / (3 \times 10^{10} \times (10^8)) \times \frac{1}{100} \times 10^9 \text{ nm}$

Ans.
 $\frac{1}{3000} \text{ nm}$

Representation of a time varying signal by a sum of sinusoidal signals, leading to a "broadened spectrum" ①

simplest case: pure tone at $z=0$

$$f(t) = A \cos(\omega t + \phi) = \frac{A e^{i\phi}}{2} e^{i\omega t} + \frac{A e^{-i\phi}}{2} e^{-i\omega t}$$

Amplitude
phase
complex amplitude
All real

$z \neq 0$ (single dimension)

$$f(t, z) = A \cos(\omega t - kz + \phi) = \frac{A e^{i\phi}}{2} e^{i\omega t - kz} + \text{c.c.}$$

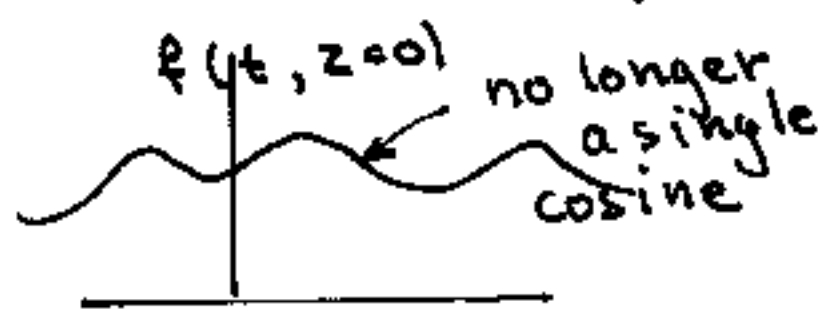
complex conjugate

$$\frac{\omega}{k} = v = c$$

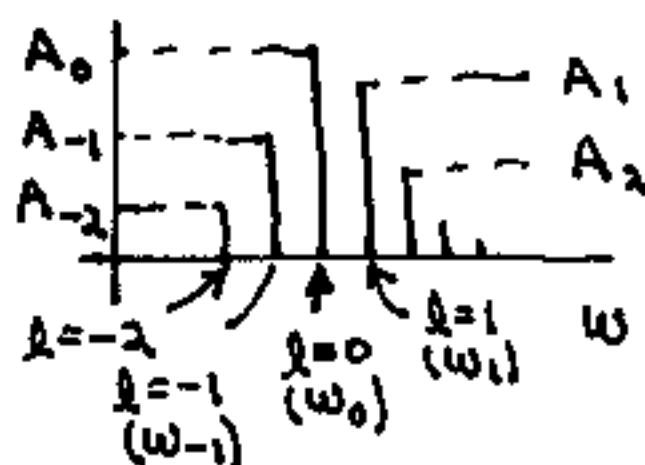
This is a single frequency wave. The exponential form is used because of its convenience and ease in analysis

For more complex signals, simply add other frequency components to the wave. Then $f(t, z)$ becomes

$$f(t, z) = \sum_{\text{over index } l} A_l \cos(\omega_l t - k_l z + \phi_l) = \sum_l A_l e^{i(\omega_l t - k_l z + \phi_l)} + \text{c.c.}$$



"Spectrum"
 simple the A_l plotted versus ω



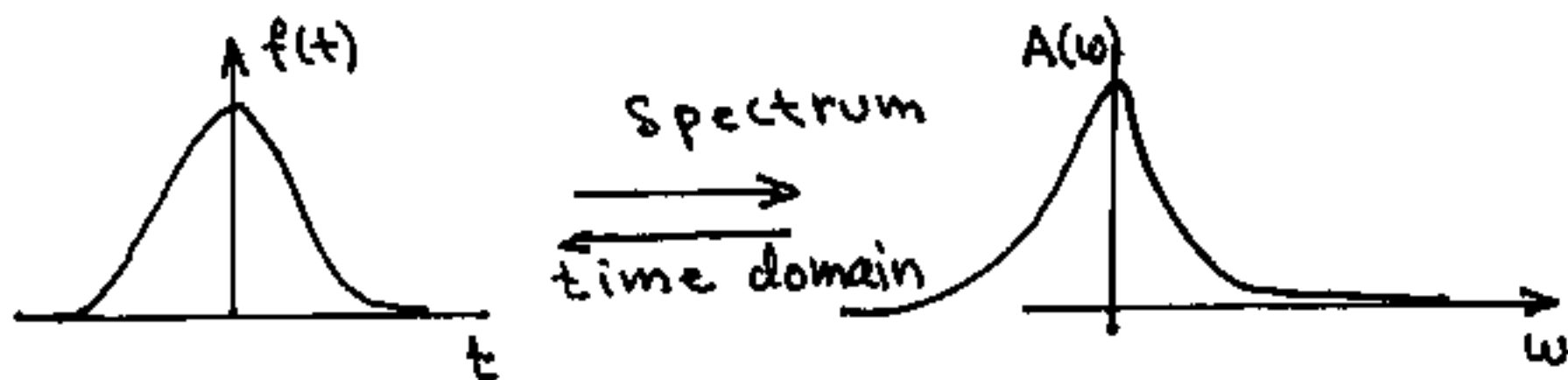
One can also write this as

$$f(t, z) = \int A(\omega) e^{i(\omega t - kz + \phi_l)} \frac{d\omega}{2\pi} + \text{c.c.}$$

where $\frac{A(\omega)}{2\pi} = \sum_l A_l \delta(\omega - \omega_l)$ since the integral just eliminates the delta "distributions" leaving the sum. We often plot $A(\omega)$ indicating the delta "distributions" with arrows



The form $f(t) = \int A(\omega) e^{i(\omega t - k z + \phi)} \frac{d\omega}{2\pi} + c.c.$ ②
 is then true also when $A(\omega)$ is not a sequence (sum) of δ -distributions but is a "continuous" amplitude
 (Note: k and ϕ are functions of ω)



To calculate $A(\omega)$ simply multiply $f(t)$ by $e^{-i\omega_0 t}$ (where ω_0 is the frequency one wishes to find $A(\omega)$) and integrate over all time. Thus

$$\int_{-\infty}^{+\infty} f(t) e^{-i\omega_0 t} dt = \int_{-\infty}^{+\infty} e^{-i\omega_0 t} \int_{-\infty}^{+\infty} d\omega A(\omega) e^{i(\omega t - k z + \phi)} \frac{d\omega}{2\pi}$$

Note: t only occurs in the exponent so it can be integrated.
 Thus one only needs $\int_{-\infty}^{+\infty} e^{-i\omega_0 t} e^{i\omega t} dt = \int_{-\infty}^{+\infty} e^{i(\omega - \omega_0)t} dt$

To "get" this to converge write it as $\int_{-\infty}^{+\infty} e^{i(\omega - \omega_0)t} dt = \int_{-\infty}^0 e^{i(\omega - \omega_0)t} dt + \int_0^{+\infty} e^{i(\omega - \omega_0)t} dt$

Add small $\pm \epsilon t$ terms so that the $-\infty$ and $+\infty$ limits give 0. For example

$$\int_{-\infty}^0 e^{i(\omega - \omega_0)t + \epsilon t} dt = \frac{1}{i(\omega - \omega_0) + \epsilon}; \text{ the } -\infty$$

limit = 0 since it has the factor $e^{\epsilon t} \rightarrow 0$ as $t \rightarrow -\infty$

For the other term use $-\epsilon t$ in the exponent.

Now sum the two terms: and take the limit

$$\lim_{\epsilon \rightarrow 0} \left(\frac{1}{i(\omega - \omega_0) + \epsilon} - \frac{1}{i(\omega - \omega_0) - \epsilon} \right) = \lim_{\epsilon \rightarrow 0} \frac{-2\epsilon}{-\epsilon^2 + -(\omega - \omega_0)^2}$$

Note: $\frac{2\epsilon}{\epsilon^2 + (\omega - \omega_0)^2}$ ① - if $(\omega - \omega_0) \neq 0$ this $\rightarrow 0$ as $\epsilon \rightarrow 0$ ③

② - if $(\omega - \omega_0) = 0$ this $\rightarrow \frac{2}{\epsilon} \rightarrow \infty$

as $\epsilon \rightarrow 0$

$$\begin{aligned} \textcircled{3} \int_{-\infty}^{+\infty} \frac{2\epsilon}{\epsilon^2 + (\omega - \omega_0)^2} d\omega &= \int_{-\infty}^{+\infty} \frac{2\epsilon^2 d(\omega - \omega_0)}{\epsilon^2 + (\omega - \omega_0)^2} \quad \leftarrow \text{constant} \\ &= \int_{-\infty}^{+\infty} \frac{2}{1 + \left(\frac{\omega - \omega_0}{\epsilon}\right)^2} d\left(\frac{\omega - \omega_0}{\epsilon}\right) = \int_{-\infty}^{+\infty} \frac{2}{1 + x^2} dx \quad ; \quad x = \frac{\omega - \omega_0}{\epsilon} \\ &= 2 \tan^{-1} \Big|_{-\infty}^{+\infty} = 2 \left(\frac{\pi}{2} - \left(-\frac{\pi}{2}\right) \right) = 2\pi \end{aligned}$$

Thus we have shown that $\int_{-\infty}^{+\infty} dt e^{i(\omega - \omega_0)t} = 2\pi \delta(\omega - \omega_0)$

Thus from the previous page (the 2π cancels)

$$\begin{aligned} \int_{-\infty}^{+\infty} f(t) e^{-i\omega_0 t} dt &= \int d\omega A(\omega) e^{-ikz + i\phi} \delta(\omega - \omega_0) \\ &= A(\omega) e^{-ikz + i\phi} \end{aligned}$$

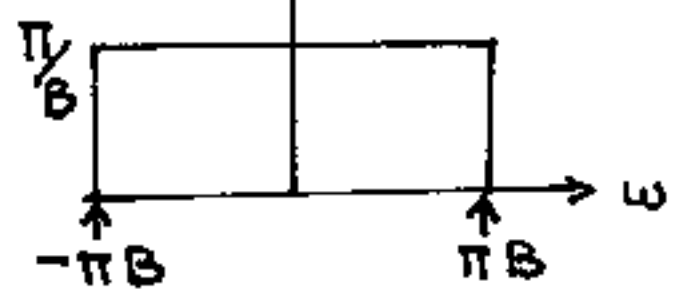
Thus one can obtain $A(\omega) e^{i\phi}$ by knowing $f(t)$ (and e^{-ikz} of course).

Examples Filter Response

$$f(t) = \frac{\sin \pi B t}{B t}$$

$$\rightarrow A(\omega) e^{i\phi(\omega)} = \int_{-\infty}^{+\infty} e^{-i\omega t} \frac{\sin \pi B t}{t} dt$$

gives $A(\omega) e^{i\phi(\omega)}$



Check

$$f(t) = \int_{-\pi B}^{+\pi B} \frac{d\omega}{2\pi} \underbrace{e^{i\omega t}}_{A(\omega) e^{i\phi(\omega)}} = \frac{e^{i\pi B t} - e^{-i\pi B t}}{2i B t} = \frac{\sin \pi B t}{B t}$$

The "Uncertainty Relation" (Relationship between a function and its Fourier Spectrum)

relates to "conjugate variables" (ω, t) , (p_x, x) , (k_y, y) , (E_x, H_x) etc.

Consider the function $w - xt$ where w and t can be complex and x is complex

$$(w - xt)(w^* - x^*t^*) \geq 0$$

Multiplying and averaging

$$\overline{|w|^2 - wt^*x^* - w^*tx + |x|^2|t|^2} \geq 0$$

Pick $x = \frac{\overline{wt^*}}{|t|^2}$ (Note: x is a constant and thus not averaged).

Substituting for x gives

$$\overline{|w|^2 - 2\frac{\overline{wt^*}w^*t}{|t|^2} + \frac{\overline{wt^*}w^*t}{|t|^2}} \geq 0 \Rightarrow \boxed{\overline{|w|^2 |t|^2} \geq \overline{wt^*} \overline{w^*t}}$$

Example: w and t . Let the signal be of the form $f(t)$ (real) (normalized to 1). Let $\bar{t} = \bar{\omega} = 0$ (for simplicity) and $F(\omega) =$ Fourier Transform of $f(t)$. Then

$$\begin{aligned} \overline{wt} &= \int_{-\infty}^{+\infty} t \overline{w(t) f(t)} dt = \int_{-\infty}^{+\infty} \overbrace{w(t)}^{(?)^2} \overbrace{f(t)}^{\text{Amplitude}} dt \\ &= \int_{-\infty}^{+\infty} t \overbrace{w(t)}^{(?)^2} \int_{-\infty}^{+\infty} \underbrace{F(\omega)}_{\text{frequency}} e^{i\omega t} d\omega dt \quad \left(\begin{array}{l} \text{to interpret } w(t) \\ \text{properly express} \\ \text{it as a derivative} \\ \text{with respect to } (it) \end{array} \right) \\ &= \int_{-\infty}^{+\infty} t \frac{d}{dt} \int_{-\infty}^{+\infty} F(\omega) e^{i\omega t} d\omega dt \\ &= \int_{-\infty}^{+\infty} \frac{d}{dt} \left(t \int_{-\infty}^{+\infty} F(\omega) e^{i\omega t} d\omega \right) dt - \int_{-\infty}^{+\infty} \left(\int_{-\infty}^{+\infty} F(\omega) e^{i\omega t} d\omega \right) dt \\ &= 0 \text{ at } t = -\infty \text{ and } t = +\infty \quad \underbrace{\int_{-\infty}^{+\infty} f(t) dt = 1} \end{aligned}$$

Thus $\overline{wt} = i$. Thus $\overline{\omega^2} \overline{t^2} \geq 1$.

Example: Gaussian $A e^{-t^2/2T^2}$ has the Fourier transform $e^{-\omega^2 T^2/2}$. These are Gaussian distributions with $\bar{t} = \bar{\omega} = 0$ and $\overline{t^2} = T^2$, $\overline{\omega^2} = \frac{1}{T^2}$ so $\overline{\omega^2} \overline{t^2} = 1$