Homework 1 Solutions

Email your scores to ee120.gsi@gmail.com

Check the course website for more details on the homework self-grading policy.

Problem 1 (Complex numbers.)

(a)

Cartesian: $e^{j5\pi} = \cos(5\pi) + j\sin(5\pi) = -1$

Polar: $e^{j5\pi} = e^{j\pi}$

Cartesian: $(\frac{1}{3}e^{j\pi/4})^* = \frac{\sqrt{2}}{6} - j\frac{\sqrt{2}}{6}$

Polar: $\left(\frac{1}{3}e^{j\pi/4}\right)^* = \frac{1}{3}e^{j(-\pi/4)}$

Cartesian: $j^{-j} = (e^{j\pi/2})^{-j} = e^{\pi/2}$

Polar: $j^{-j} = e^{\pi/2} = e^{\pi/2} \cdot e^{j0}$

(b)

$$\frac{a+jb}{c+jd} = \frac{(a+jb)(c-jd)}{(c+jd)(c-jd)} = \frac{ac+bd+j(bc-ad)}{c^2+d^2}$$

We must examine the real and imaginary parts to determine the quadrant in the complex plane in which the complex number lies.

$$\frac{a+jb}{c+jd} = \begin{cases} \sqrt{\frac{a^2+b^2}{c^2+d^2}}e^{j\left(\arctan\left(\frac{bc-ad}{ac+bd}\right)+\pi\right)}, & ac+bd<0 \text{ and } bc-ad>0\\ \sqrt{\frac{a^2+b^2}{c^2+d^2}}e^{j\left(\arctan\left(\frac{bc-ad}{ac+bd}\right)-\pi\right)}, & ac+bd<0 \text{ and } bc-ad<0,\\ \sqrt{\frac{a^2+b^2}{c^2+d^2}}e^{j\left(\arctan\left(\frac{bc-ad}{ac+bd}\right)\right)}, & \text{otherwise} \end{cases}$$

(c) See plots in Figure 1.

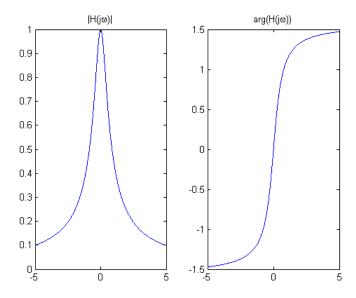


Figure 1: Problem 1(c).

Problem 2 (Implementation of Complex Systems.)

$$\begin{array}{rcl} y[n] & = & (2-j3)\,x[n] + (3+j)\,x[n-1] \\ y_R[n] + jy_I[n] & = & (2-j3)(x_R[n] + jx_I[n]) + (3+j)(x_R[n-1] + jx_I[n-1]) \\ y_R[n] + jy_I[n] & = & 2x_R[n] + j2x_I[n] - j3x_R[n] + 3x_I[n] + 3x_R[n-1] + j3x_I[n-1] + jx_R[n-1] - x_I[n-1] \\ y_R[n] + jy_I[n] & = & (2x_R[n] + 3x_I[n] + 3x_R[n-1] - x_I[n-1]) + j(2x_I[n] - 3x_R[n] + 3x_I[n-1] + x_R[n-1]) \\ y_R[n] & = & 2x_R[n] + 3x_I[n] + 3x_R[n-1] - x_I[n-1] \\ y_I[n] & = & 2x_I[n] - 3x_R[n] + 3x_I[n-1] + x_R[n-1] \end{array}$$

The block diagram is shown in Figure 2

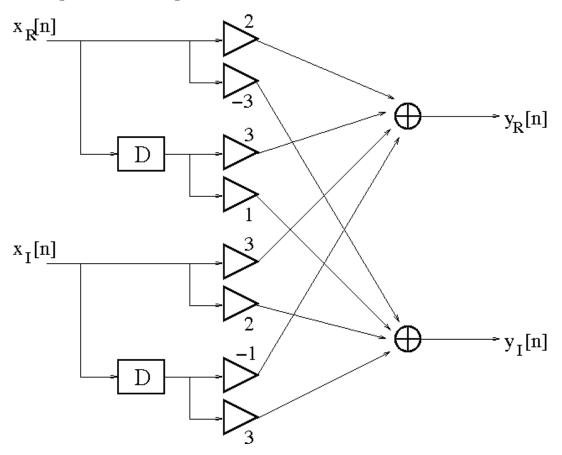


Figure 2: Problem 2.

Problem 3 (Elementary functions and their graphs.)

(a)
$$Re\{y(t)\} = \cos(2\pi t)$$

$$Im\{y(t)\} = \sin(2\pi t)$$

$$|y(t)| = 1$$

$$\arg(y(t)) = 2\pi(t - \lfloor t \rfloor)$$

See plots in Figure 3.

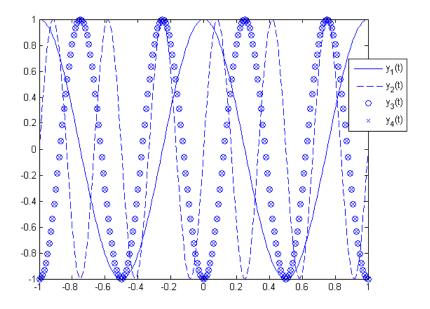


Figure 3: Problem 3(a).

Problem 4 (Periodic continuous-time signals.)

(a)

 $\cos(\frac{\pi}{3}t)$ has a period of $\frac{2\pi}{\pi/3} = 6$.

 $\cos(\frac{2\pi}{5}t)$ has a period of $\frac{2\pi}{2\pi/5} = 5$.

The fundamental period is 30, the least common multiple of 6 and 5.

(b)

 $w(t) = \cos(\frac{\pi}{3}t) + \cos(t)$ is a periodic.

 $\cos(\frac{\pi}{3}t)$ has a period of $\frac{2\pi}{\pi/3}=6$ and $\cos(t)$ has a period of 2π . Since one is rational and one is irrational, they have no common multiples.

Problem 5 (Transformations of functions.)

See Figure 4.

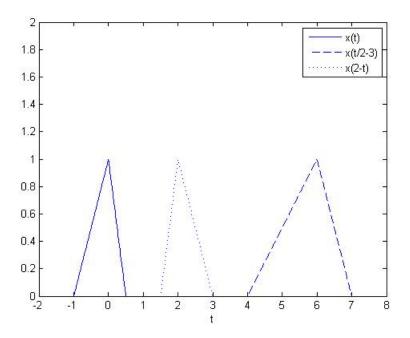


Figure 4: Problem 5.

Problem 6 (Periodic discrete-time signals.)

(a)

 $y_1[n] = \cos(\frac{\pi}{3}n)$ has a period of 6.

 $y_2[n] = \cos(\frac{3\pi}{5}n)$ has a period of 10.

See plots in Figure 5.

(b)

The fundamental period is 30, the least common multiple of 6 and 10.

See plots in Figure 5.

(c)

 $z[n] = \cos(2n)$ is a periodic. There is no integer n such that 2n is divisible by 2π .

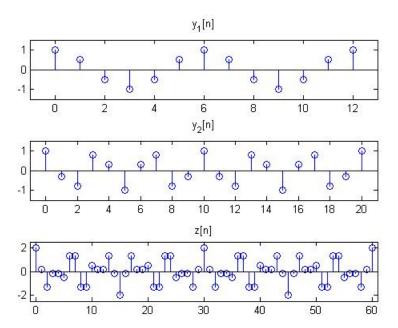


Figure 5: Problem 6.

Problem 7 (Properties of systems.)

Let $f(t) \triangleq k_1 x_1(t) + k_2 x_2(t)$ where k_1, k_2 are constants.

Let $g(t) \triangleq x(t-\tau)$ where τ is a constant.

(a) $\mathcal{H}[x(t)] = x(at+b)$ is linear and not time-invariant.

$$\mathcal{H}[f(t)] = f(at+b) = k_1 x_1 (at+b) + k_2 x_2 (at+b)$$

$$= k_1 \mathcal{H}[x_1(t)] + k_2 \mathcal{H}[x_2(t)]$$

$$\mathcal{H}[g(t)] = g(at+b) = x(at+b-\tau)$$

$$\neq x(a(t-\tau)+b)$$

The last line is the result of shifting the output of $\mathcal{H}[x(t)]$, i.e. x(at+b), by τ .

(b) $\mathcal{H}[x(t)] = x(at^2 + b)$ is linear and not time-invariant.

$$\mathcal{H}[f(t)] = f(at^{2} + b) = k_{1}x_{1}(at^{2} + b) + k_{2}x_{2}(at^{2} + b)$$

$$= k_{1}\mathcal{H}[x_{1}(t)] + k_{2}\mathcal{H}[x_{2}(t)]$$

$$\mathcal{H}[g(t)] = g(at^{2} + b) = x(at^{2} + b - \tau)$$

$$\neq x(a(t - \tau)^{2} + b)$$

The last line is the result of shifting the output of $\mathcal{H}[x(t)]$, i.e. $x(at^2 + b)$, by τ .

Problem 8 (Properties of systems.)

(a) $\mathcal{H}[x(t)] = x(at) + b$ is not linear and not time-invariant.

$$\mathcal{H}[f(t)] = f(at) + b = k_1 x_1(at) + k_2 x_2(at) + b$$

$$\neq k_1 \mathcal{H}[x_1(t)] + k_2 \mathcal{H}[x_2(t)] = k_1 x_1(at) + k_1 b + k_2 x_2(at) + k_2 b$$

$$\mathcal{H}[g(t)] = g(at) + b = x(at - \tau) + b$$

$$\neq x(a(t - \tau)) + b$$

The last line is the result of shifting the output of $\mathcal{H}[x(t)]$, i.e. x(at) + b, by τ .

(b) $\mathcal{H}[x(t)] = \frac{d}{dt}x(t)$ is linear and time-invariant.

$$\mathcal{H}[f(t)] = \frac{d}{dt}f(a) = k_1 \frac{d}{dt}x_1(t) + k_2 \frac{d}{dt}x_2(t)$$

$$= k_1 \mathcal{H}[x_1(t)] + k_2 \mathcal{H}[x_2(t)]$$

$$\mathcal{H}[g(t)] = \frac{d}{dt}g(t)$$

$$= \frac{d}{dt}x(t-\tau)$$

The above line is the result of shifting the output of $\mathcal{H}[x(t)]$, i.e. $\frac{d}{dt}x(t)$, by τ .

Problem 9 (Properties of continuous-time systems.)

(a)
$$y(t) = x(t) \frac{e^t + e^{-t}}{2}$$

- (i) memoryless (OWN definition): The current output value y(t) is determined by the value of the current input x(t).
 - not memoryless (LV/Handout 1 definition): The output at time t requires knowledge of the value of t.
- (ii) unstable: For the trivially bounded input $x(t) = 1 \ \forall t$, the output y(t) is unbounded as $t \to +\infty$.
- (iii) causal: The current output value y(t) does not depend on future inputs $x(\tau), \tau > t$.
- (iv) linear:

Let
$$y_1(t) = x_1(t) \frac{e^t + e^{-t}}{2}$$
 and $y_2(t) = x_2(t) \frac{e^t + e^{-t}}{2}$.
If the input to the system is $f(t) = k_1 x_1(t) + k_2 x_2(t)$, where k_1, k_2 are constants, then the output is $g(t) = f(t) \frac{e^t + e^{-t}}{2} = (k_1 x_1(t) + k_2 x_2(t)) \frac{e^t + e^{-t}}{2} = k_1 y_1(t) + k_2 y_2(t)$.

(v) not time-invariant:

If the input is $f(t)=x(t-\tau)$, where τ is a constant, then the output is $g(t)=f(t)\frac{e^t+e^{-t}}{2}=x(t-\tau)\frac{e^t+e^{-t}}{2}$. Since $y(t-\tau)=x(t-\tau)\frac{e^{t-\tau}+e^{-(t-\tau)}}{2}$, $g(t)\neq y(t-\tau)$.

- $(b) y(t) = Im\{x(t)\}$
 - (i) memoryless: The output at time t_0 depends only on the input at the same time t_0 .
 - (ii) stable: If $|x(t)| < \infty$, then $|Im\{x(t)\}| < \infty$.
 - (iii) causal: At each time, the output depends only on the current input.
 - (iv) non-linear: Consider an input $x_1(t) = jt$ which produces the output $y_1(t) = t$ The input $x_2(t) = jx_1(t) = -t$ produces the output $y_2(t) = 0$, which is not equal to $jy_1(t)$
 - (v) time-invariant: If the input is $f(t) = x(t \tau)$, where τ is a constant, then the output is $g(t) = Im\{f(t)\} = Im\{x(t \tau)\}$. Since $y(t \tau) = Im\{x(t \tau)\}$, the system is time-invariant.

Problem 10 (Properties of continuous-time systems.)

(a)
$$y(t) = \int_{-\infty}^{t/2} x(3\alpha) d\alpha = \frac{1}{3} \int_{-\infty}^{3t/2} x(\beta) d\beta$$

- (i) not memoryless: The current output y(t) uses many past values of the input $x(\tau), \tau < t$.
- (ii) unstable: $y(t) = \infty \ \forall t \text{ if } x(t) = u(-t)$.
- (iii) not causal: For t>0, the current output value y(t) depends on future inputs $x(\tau), \tau>t$.
- (iv) linear: Integration is linear.
- (v) not time-invariant: If the input is $f(t) = x(t-\tau)$, where τ is a constant, then the output is $g(t) = \frac{1}{3} \int_{-\infty}^{3t/2} f(\beta) d\beta = \frac{1}{3} \int_{-\infty}^{3t/2} x(\beta-\tau) d\beta = \frac{1}{3} \int_{-\infty}^{3t/2-\tau} x(\sigma) d\sigma$. Since $y(t-\tau) = \frac{1}{3} \int_{-\infty}^{3(t-\tau)/2} x(\beta) d\beta$, $g(t) \neq y(t-\tau)$.
- (b) y(t) = x(-t/2)
 - (i) not memoryless: The output at time t cannot be determined from the input at time t.

- (ii) stable: If x(t) is bounded, then y(t) is bounded.
- (iii) noncausal: For t < 0, the current output y(t) depends on a future value of x(t).
- (iv) linear:

Let
$$y_1(t) = x_1(-t/2)$$
 and $y_2(t) = x_2(-t/2)$.

If the input is $f(t) = k_1 x_1(t) + k_2 x_2(t)$, where k_1, k_2 are constants,

then the output is $g(t) = f(-t/2) = k_1 x_1(-t/2) + k_2 x_2(-t/2) = k_1 y_1(t) + k_2 y_2(t)$.

(v) not time-invariant:

If the input is $f(t) = x(t - \tau)$, where τ is a constant,

then the output is $g(t) = f(-t/2) = x(-t/2 - \tau)$.

Since $y(t - \tau) = x(-(t - \tau)/2), \ g(t) \neq y(t - \tau).$

Problem 11 (Properties of discrete-time systems.)

(a)
$$y[n] = \sum_{k=-\infty}^{n} x[k+1] = \sum_{l=-\infty}^{n+1} x[l]$$

- (i) not memoryless: The current output y[n] depends on many past input values $x[\tau], \tau < n$.
- (ii) not stable: $y[n] = \infty \ \forall n \text{ if } x[n] = u[-n]$.
- (iii) not causal: The current output y[n] depends on future input x[n+1].
- (iv) linear:

Let
$$y_1[n] = \sum_{l=-\infty}^{n+1} x_1[l]$$
 and $y_2[n] = \sum_{l=-\infty}^{n+1} x_2[l]$.

Let $y_1[n] = \sum_{l=-\infty}^{n+1} x_1[l]$ and $y_2[n] = \sum_{l=-\infty}^{n+1} x_2[l]$. If the input is $f[n] = k_1 x_1[n] + k_2 x_2[n]$, where k_1, k_2 are constants, then the output is $g[n] = \sum_{l=-\infty}^{n+1} f[l] = \sum_{l=-\infty}^{n+1} (k_1 x_1[l] + k_2 x_2[l]) = k_1 y_1[n] + k_2 y_2[n]$.

(v) time-invariant:

If the input is $f[n] = x[n-\tau]$, where τ is a constant,

then the output is $g[n] = \sum_{l=-\infty}^{n+1} f[l] = \sum_{l=-\infty}^{n+1} x[l-\tau] = \sum_{m=-\infty}^{n+1-\tau} x[m]$.

Therefore $y[n-\tau] = g[n]$.

(b)
$$y[n] = x[n] \cdot \sum_{k=-\infty}^{\infty} \delta[n-3k]$$

Observe that y[n] = x[n] if n is a multiple of 3, and y[n] = 0 otherwise.

(i) memoryless (OWN definition): The current output y[n] does not depend on past or future inputs x[n].

not memoryless (LV/Handout 1 definition): The output at time n requires knowledge of the value of n.

- (ii) stable: y[n] is bounded if x[n] is bounded.
- (iii) causal: The current output does not depend on future inputs.
- (iv) linear:

Let $y_1[n] = x_1[n] \sum_{k=-\infty}^{\infty} \delta[n-3k]$ and $y_2[n] = x_2[n] \sum_{k=-\infty}^{\infty} \delta[n-3k]$.

If the input is $f[n] = k_1x_1[n] + k_2x_2[n]$, where k_1, k_2 are constants, then the output is $g[n] = f[n] \sum_{k=-\infty}^{\infty} \delta[n-3k] = k_1y_1[n] + k_2y_2[n]$.

(v) not time-invariant:

If the input is $f[n] = x[n-\tau]$, where τ is a constant,

then the output is $g[n] = f[n] \sum_{k=-\infty}^{\infty} \delta[n-3k] = x[n-\tau] \sum_{k=-\infty}^{\infty} \delta[n-3k]$. Since $y[n-\tau] = x[n-\tau] \sum_{k=-\infty}^{\infty} \delta[n-\tau-3k]$, $g[n] \neq y[n-\tau]$.

Problem 12 (Convolution of step functions.)

$$\begin{array}{rcl} (a) & h(t) = u(t) - u(t-1) \,, \; x(t) = u(t) \\ \\ & h(t) * x(t) & = & \int_{-\infty}^{\infty} \left(u(\tau) - u(\tau-1) \right) \cdot u(t-\tau) d\tau \\ \\ & = & \int_{-\infty}^{t} u(\tau) - u(\tau-1) d\tau \\ \\ & = & u(t) \cdot \int_{0}^{\min(t,1)} d\tau \\ \\ & = & \begin{cases} 0, & t < 0 \\ t, & 0 \leq t \leq 1 \\ 1, & t > 1 \end{cases} \end{array}$$

See Figure 6.

(b)
$$h(t) = u(t) - u(t-1)$$
, $x(t) = u(t) - u(t-1)$

We will use the linearity and time invariance properties of the system h(t) to make use of part (a) of this problem. Suppose u(t) is the input to the system as in part (a). Call the output z(t). Now, the input is x(t) = u(t) - u(t-1), the difference of u(t) and a shifted copy of u(t). Using the fact that the system is linear and time invariant, the output should now be z(t) - z(t-1).

$$\begin{array}{lcl} h(t)*x(t) & = & [u(t)*(u(t)-u(t-1))] - [u(t-1)*(u(t)-u(t-1))] \\ & = & z(t)-z(t-1) \\ & = & \left\{ \begin{array}{ll} t, & 0 \leq t \leq 1 \\ 2-t, & 1 \leq t \leq 2 \\ 0, & t < 0 \text{ or } t > 2 \end{array} \right. \end{array}$$

See Figure 7.

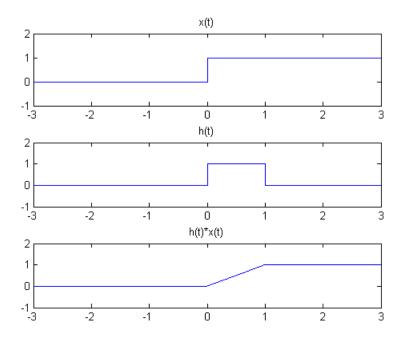


Figure 6: Problem 12(a).

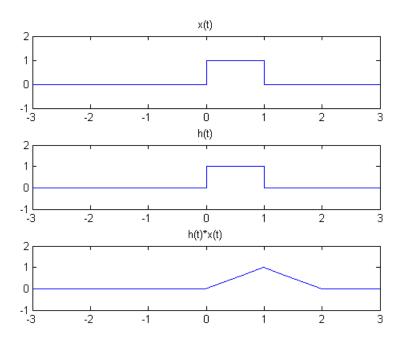


Figure 7: Problem 12(b).

Problem 13 (LTI systems.)

We observe that z(t) = -x(t-1) + 2x(t-2). Because the system is LTI, the new output is equal to -y(t-1) + 2y(t-2). The output is plotted in Figure 8.

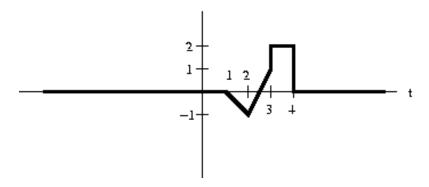


Figure 8: Problem 13.

Problem 14 (Convolution via matlab.)

The following Matlab code generates Figure 9.

```
t = -2:.01:2;
xa = (t>=0);
ha = zeros(size(t));
ha(t>=0) = exp(-(0:.01:2));
za = conv(ha,xa);
xb = ha;
hb = zeros(size(t));
hb(t>=0) = exp(-.5.*(0:.01:2));
zb = conv(hb,xb);
```

The Matlab command $\mathbf{z} = \mathtt{conv}(\mathbf{h}, \mathbf{x})$ convolves vectors \mathbf{h} (of length a) and \mathbf{x} (of length b), and produces a vector \mathbf{z} of length a+b-1. Matlab effectively computes the convolution sum of the discrete-time signals g[n] and y[n], whose nonzero values are contained in \mathbf{h} and \mathbf{x} , and assumes that the signals are zero otherwise. Thus \mathbf{z} contains the nonzero values of g[n]*y[n]. Since our functions h(t) and x(t) are not zero outside of $t \in [-2, 2]$, we need to discard the extraneous (and incorrect) outputs.

Furthermore, because we are approximating $\int h(\tau)x(t-\tau)d\tau$ with $\sum_{\tau}h(\tau)x(t-\tau)\Delta\tau$, we need to multiply the output of conv, $\sum_{\tau}h(\tau)x(t-\tau)$, by $\Delta\tau=0.01$.

```
za = za(201:601)./100;

zb = zb(201:601)./100;
```

Now we are ready to plot the figures.

```
subplot(2,3,1); plot(t,ha); axis([-2 2 -.5 1.5]); title('8a: h(t)');
subplot(2,3,2); stairs(t,xa); axis([-2 2 -.5 1.5]); title('8a: x(t)');
subplot(2,3,3); plot(t,za); axis([-2 2 -.5 1.5]); title('8a: conv(h,x)');
subplot(2,3,4); plot(t,hb); axis([-2 2 -.5 1.5]); title('8b: h(t)');
subplot(2,3,5); plot(t,xb); axis([-2 2 -.5 1.5]); title('8b: x(t)');
subplot(2,3,6); plot(t,zb); axis([-2 2 -.5 1.5]); title('8b: conv(h,x)');
```

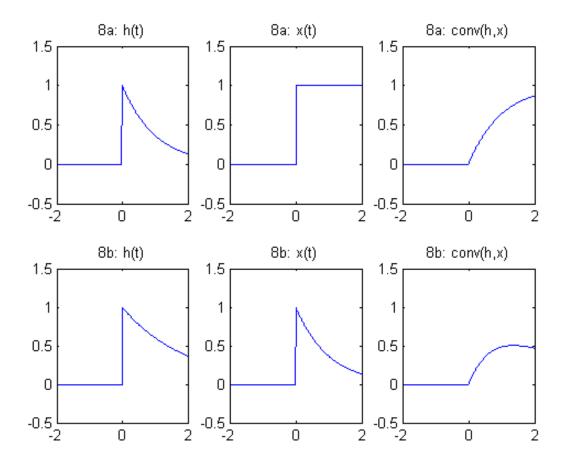


Figure 9: Problem 14.