

OCT 24, 03

Optimal FIR Filter Design

For low pass filter, alternations
always occur at ω_p and ω_s .

Fig 7.38

Filter will be equi-ripple except possibly
at $\omega=0$ or $\omega=\pi$.

slope 0

Fig 7.39.

Alg. for optimal Filter Design

Parks/ McClellan.

Problem B

$$G(\omega) = \sum_{n=0}^M a(n) \cos n\omega$$

$$K = \frac{\delta_1}{\delta_2}, \quad M$$

Given: $w_s, w_p,$

find $\underline{a(n)}$ to minimize δ_2

$i = 1, \dots, M+1$

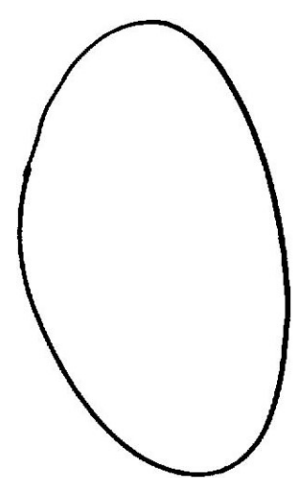
$$E(w_i) = \pm \delta_2$$

$$E(w_i) = -E(w_{i+1})$$

$i = 1, \dots, M.$

$$W(w_i) [G(w_i) - D(w_i)] = (-1)^{i+1} \delta_2$$

$w_i =$ alternation frequencies.



hard on
solve a linear
sys of eqns.

w_i for $i=1, \dots, M+2$
 $M+2$ unknowns

$a(n) \rightarrow M+1$
 $\delta_2 \rightarrow 1$
 $M+2$ unknowns

easy

show "hard" : show, if I know w_i

how to find $a(n)$ and δ_2 .

$$G(w_i) = \sum_{n=0}^M a(n) \cos(w_i n) + \delta_2$$

w_i
 $i=1, \dots, M+2$

$$G(w_i) = \frac{(-1)^{i+1} \delta_2}{W(w_i)} + D(w_i)$$

$$\sum_{n=0}^M a(n) \cos(w_i n) = \frac{(-1)^{i+1} \delta_2 + D(w_i)}{W(w_i)}$$

$i=1$

$$\begin{aligned}
 & a(0) \cos(\omega_0 \cdot 0) + a(1) \cos(\omega_1 \cdot 1) + a(2) \cos(\omega_2 \cdot 2) \\
 & + \dots + \dots + a(M) \cos(\omega_M \cdot M) \\
 & = \frac{(-1)^{i+1} \delta_2}{W(\omega_i)} + D(\omega_i)
 \end{aligned}$$

$$\begin{aligned}
 i=2 & \quad a(0) \cos(\omega_2 \cdot 0) + a(1) \cos(\omega_2 \cdot 1) \dots \dots
 \end{aligned}$$

$$= \left(\begin{array}{c} \text{linear} \end{array} \right)$$

Equations in $a(n)$

$\Rightarrow M+2$

$i=M+2$

$M+1$ unknown $a(n) + \delta_2$
 $M+2$ unknown

Solve a linear syst of Eqns. in $M+2$ unknowns.

$$\begin{bmatrix} D(w_1) \\ D(w_2) \\ \vdots \\ D(w_{m+2}) \end{bmatrix}$$

Known

$$\begin{bmatrix} \frac{(-1)^{i+1}}{W(w_i)} \\ \frac{(-1)^{i+1}}{W(w_2)} \\ \vdots \\ \frac{(-1)^{i+1}}{W(w_{m+2})} \end{bmatrix}$$

Known w_i

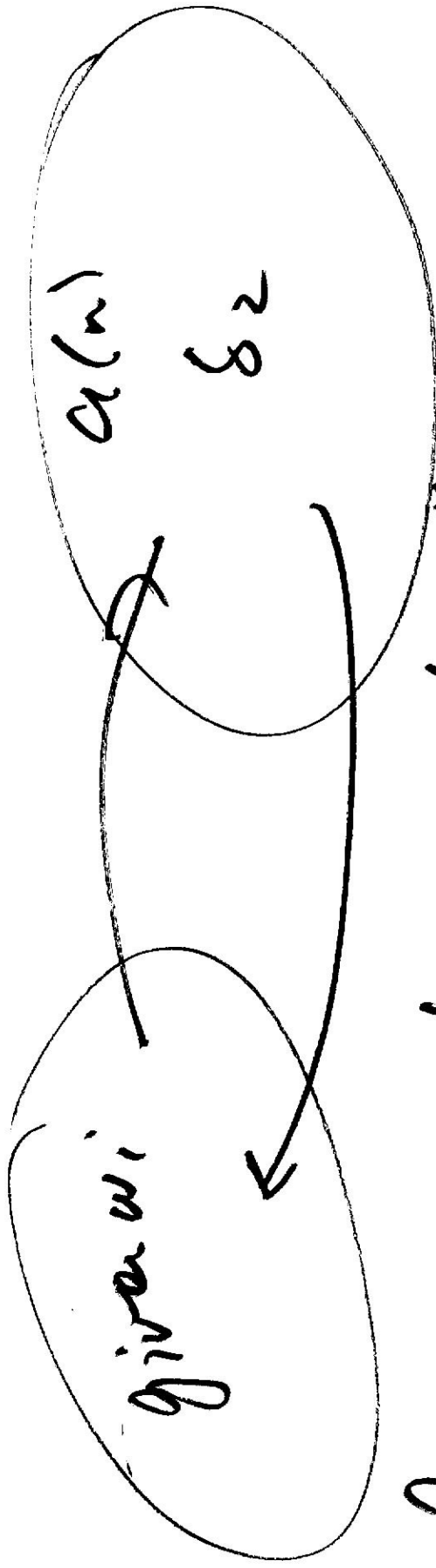
$$= \sum \frac{1}{w_i}$$

$$\begin{bmatrix} a(0) \\ a(1) \\ a(2) \\ \vdots \\ a(M) \end{bmatrix}$$

unknown

$$\begin{bmatrix} \cos w_1 \cdot 0 & \cos w_1 \cdot 1 & \cos w_1 \cdot 2 & \dots & \cos w_1 \cdot M \\ \cos w_2 \cdot 0 & \cos w_2 \cdot 1 & \cos w_2 \cdot 2 & \dots & \cos w_2 \cdot M \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ \cos w_{m+2} \cdot 0 & \cos w_{m+2} \cdot 1 & \cos w_{m+2} \cdot 2 & \dots & \cos w_{m+2} \cdot M \end{bmatrix}$$

Given if w_i are known



Remez exchange algorithm.

P & M: showed. z_2 is given by the following expression if w_i are known:

$$z_2 = \frac{\sum_{k=1}^{M+2} b_k D(w_k)}{\sum_{k=1}^{M+2} b_k (-1)^{k+1}}$$

when $b_k = \prod_{\substack{i=1 \\ i \neq k}}^{M+2} \frac{1}{\cos w_i}$

Empirical Studies

Approx length of filter $\approx 1 - \frac{10 \log_{10}(\delta_1 \delta_2) - 13}{2.3 \Delta \omega}$

$$\Delta \omega = \omega_s - \omega_p$$

Kaiser length $\approx 1 + \frac{A - 8}{2.2 \Delta \omega}$ $A = -20 \log_{10} \delta$

$$\omega_s = 0.6\pi$$

$$\omega_p = 0.4\pi$$

$$\delta_1 = 0.01$$

$$\delta_2 = 0.001$$

optimum filter $\rightarrow N \approx 27 = 2M + 1$

Kaiser filter $\rightarrow N \approx 38$