

EECS C128/ ME C134  
 Final  
 Thu. May 14, 2015  
 1510-1800 pm

Name: \_\_\_\_\_  
 SID: \_\_\_\_\_

- Closed book. One page, 2 sides of formula sheets. No calculators.
  
- There are 8 problems worth 100 points total.

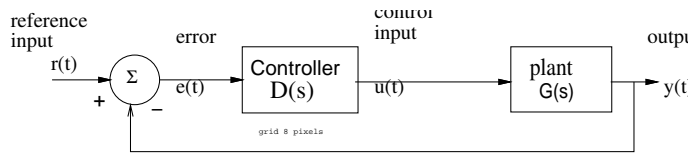
Problem	Points	Score
1	14	
2	14	
3	16	
4	8	
5	13	
6	13	
7	14	
8	8	
Total	100	

In the real world, unethical actions by engineers can cost money, careers, and lives. The penalty for unethical actions on this exam will be a grade of 'F' and a letter will be written for your file and to the Office of Student Conduct.

$\tan^{-1} \frac{1}{2} = 26.6^\circ$	$\tan^{-1} 1 = 45^\circ$
$\tan^{-1} \frac{1}{3} = 18.4^\circ$	$\tan^{-1} \frac{1}{4} = 14^\circ$
$\tan^{-1} \sqrt{3} = 60^\circ$	$\tan^{-1} \frac{1}{\sqrt{3}} = 30^\circ$
$\sin 30^\circ = \frac{1}{2}$	$\cos 60^\circ = \frac{\sqrt{3}}{2}$

$20 \log_{10} 1 = 0dB$	$20 \log_{10} 2 = 6dB$
$20 \log_{10} \sqrt{2} = 3dB$	$20 \log_{10} \frac{1}{2} = -6dB$
$20 \log_{10} 5 = 20dB - 6dB = 14dB$	$20 \log_{10} \sqrt{10} = 10 dB$
$1/e \approx 0.37$	$1/e^2 \approx 0.14$
$1/e^3 \approx 0.05$	$\sqrt{10} \approx 3.16$

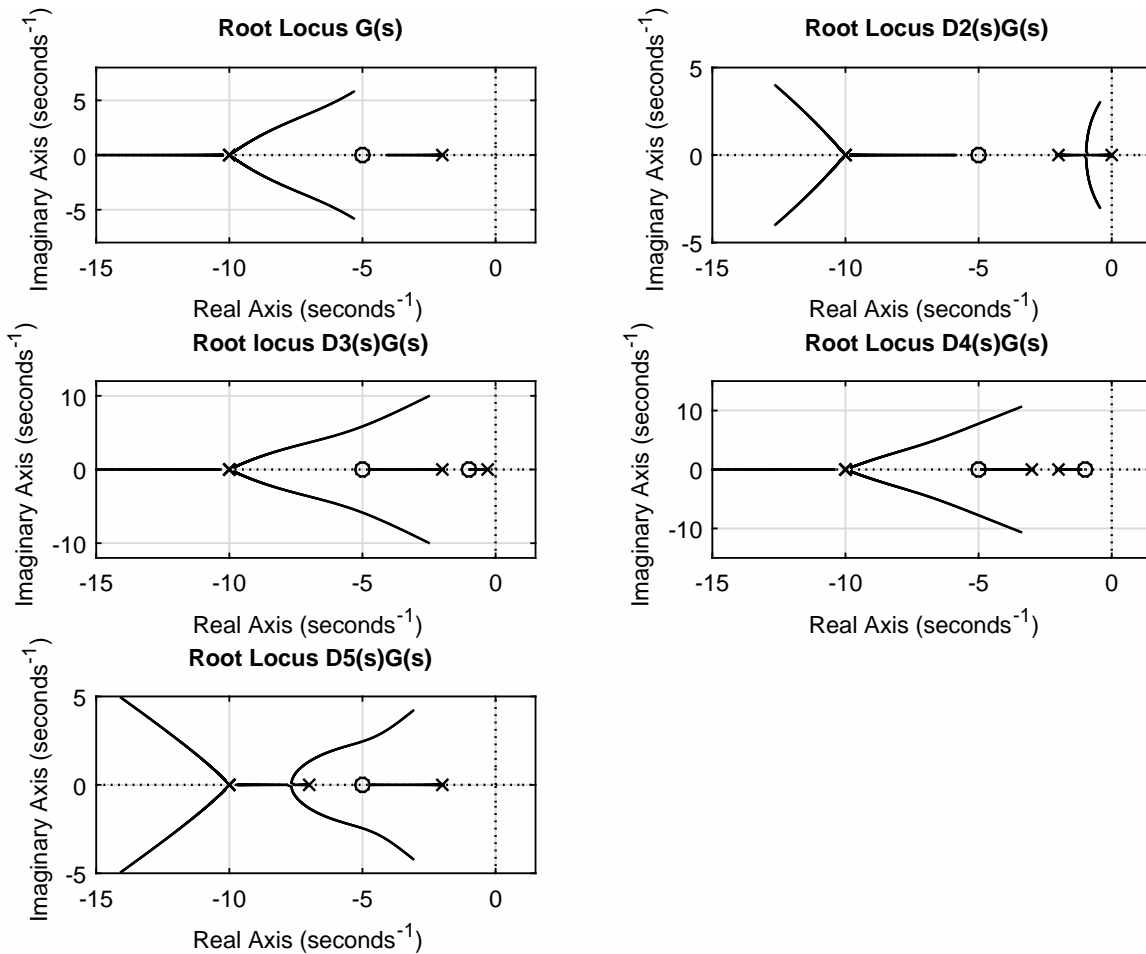
**Problem 1 (14 pts)**



You are given the open-loop plant:

$$G(s) = \frac{s + 5}{(s + 2)(s + 10)^3}$$

For the above system, the partial root locus is shown for 5 different controller/plant combinations,  $G(s), D_2(s)G(s), \dots, D_5(s)G(s)$ . (Note: the root locus shows open-loop pole locations for  $D(s)G(s)$ , and closed-loop poles for  $\frac{DG}{1+DG}$  are at end points of branches).

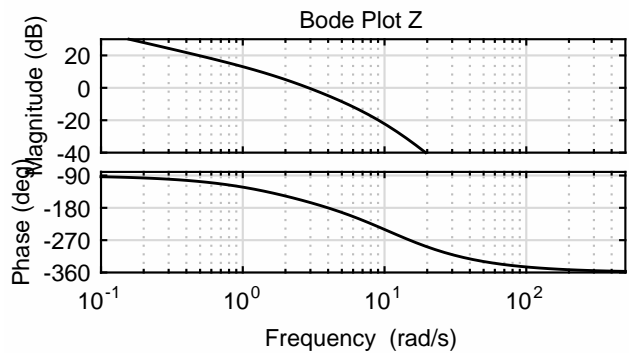
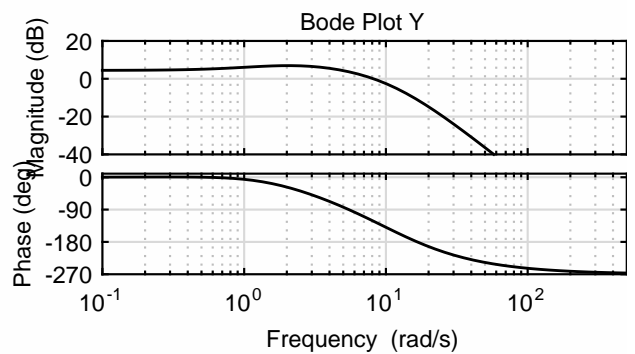
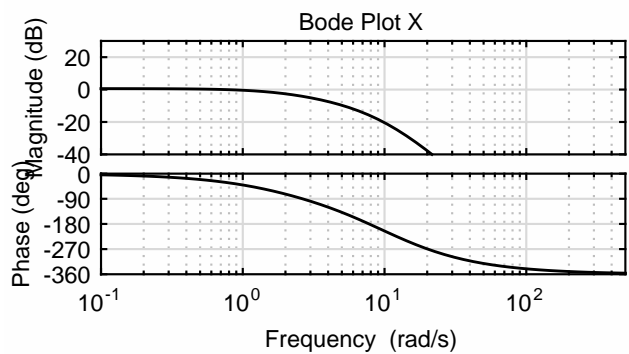
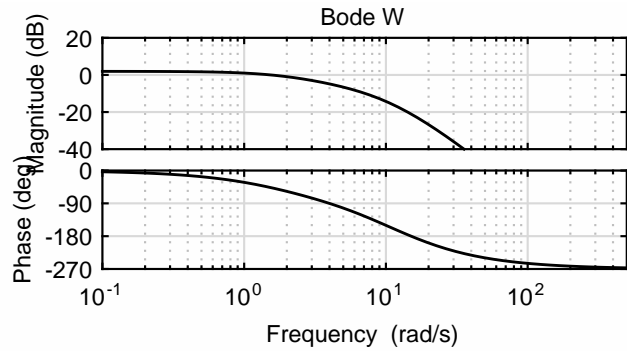
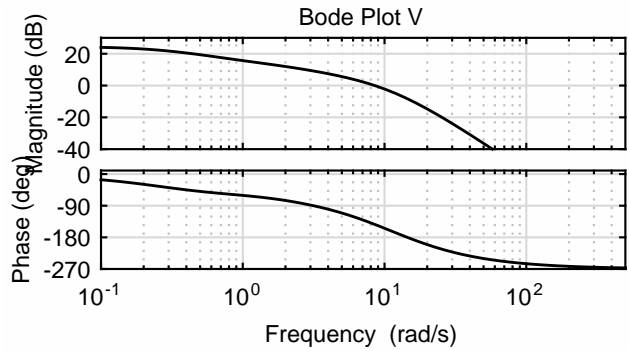


[5 pts] a) For each set of open-loop poles and zeros given above, choose the best corresponding open-loop Bode plot V,W,X,Y, or Z from the next page:

- (i)  $G(s)$ : Bode Plot \_\_\_\_
- (ii)  $D_2(s)G(s)$ : Bode plot \_\_\_\_
- (iii)  $D_3(s)G(s)$ : Bode plot \_\_\_\_
- (iv)  $D_4(s)G(s)$ : Bode Plot \_\_\_\_
- (v)  $D_5(s)G(s)$ : Bode Plot \_\_\_\_

**Problem 1, cont.**

The open-loop Bode plots for 5 different controller/plant combinations,  $D_1(s)G(s), \dots, D_5(s)G(s)$  are shown below.



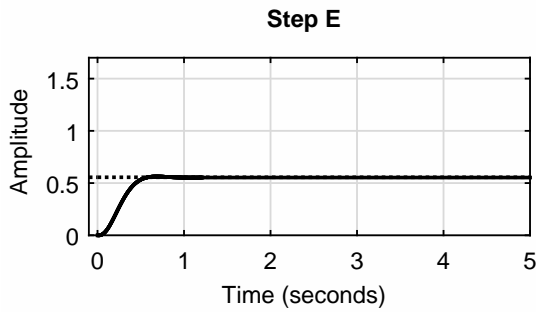
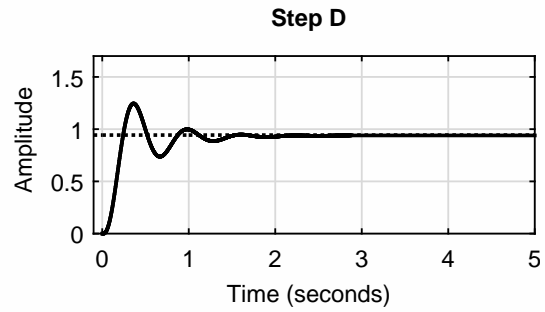
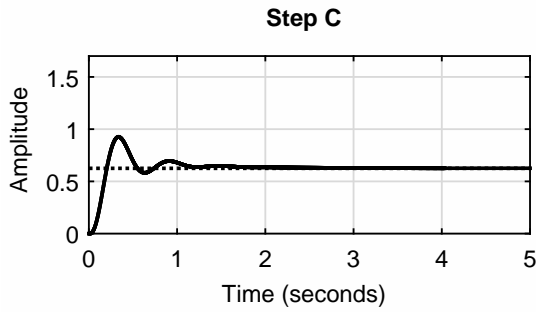
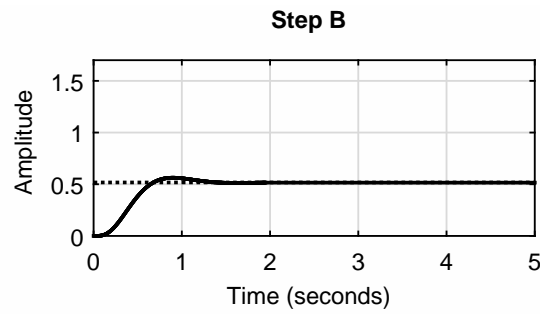
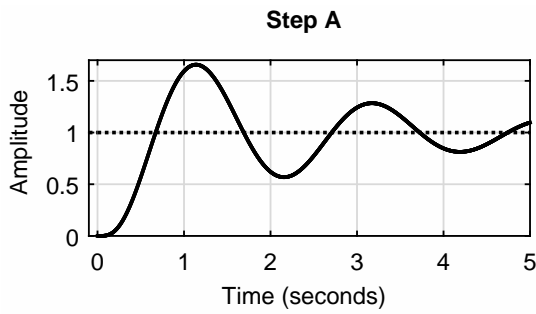
[4 pts] b) For the listed Bode plots, estimate the phase and gain margin:

- (i) Bode plot V: phase margin \_\_\_\_ (degrees) at  $\omega =$  \_\_\_\_  
 Bode plot V: gain margin \_\_\_\_ dB at  $\omega =$  \_\_\_\_
- (ii) Bode plot Z: phase margin \_\_\_\_ (degrees) at  $\omega =$  \_\_\_\_  
 Bode plot Z: gain margin \_\_\_\_ dB at  $\omega =$  \_\_\_\_

**Problem 1, cont.**

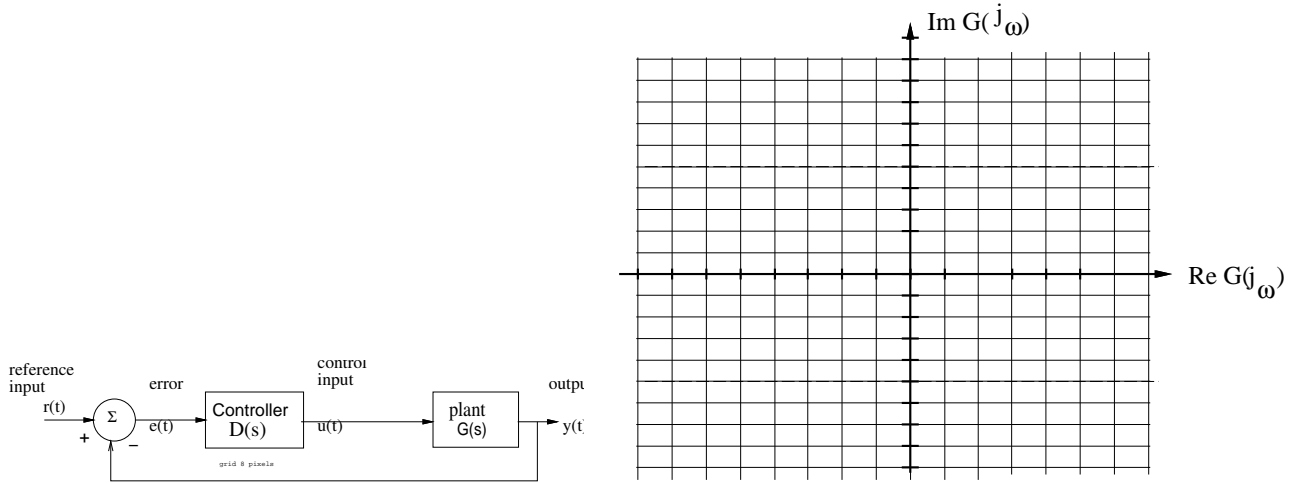
[5 pts] c) For each closed loop controller/plant with root locus as given in part a), choose the best corresponding closed-loop step response (A-E)

- (i)  $G(s)$ : step response \_\_\_\_
- (ii)  $D_2(s)G(s)$ : step response \_\_\_\_
- (iii)  $D_3(s)G(s)$ : step response \_\_\_\_
- (iv)  $D_4(s)G(s)$ : step response \_\_\_\_
- (v)  $D_5(s)G(s)$ : step response \_\_\_\_



**Problem 2 (14 pts)**

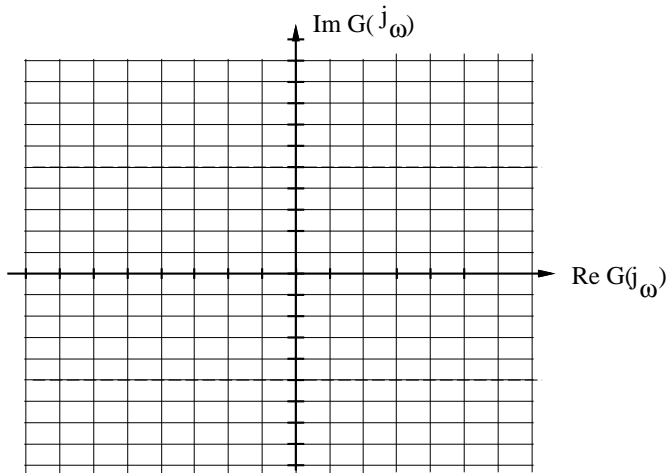
[4 pts] a. You are given the open loop plant:  $G(s) = k \frac{s+6}{s-6}$ , with  $D(s) = 1$ . Sketch Nyquist plot for  $G(s)$  with  $k = 1$ , showing clearly any encirclements.



[2 pts] b. Find the bounds on  $k$  for the system with unity feedback to be stable.

[4 pts] c. You are given the open loop plant  $G(s) = k \frac{(s-6)(s-4)}{(s+6)(s-1)}$ .

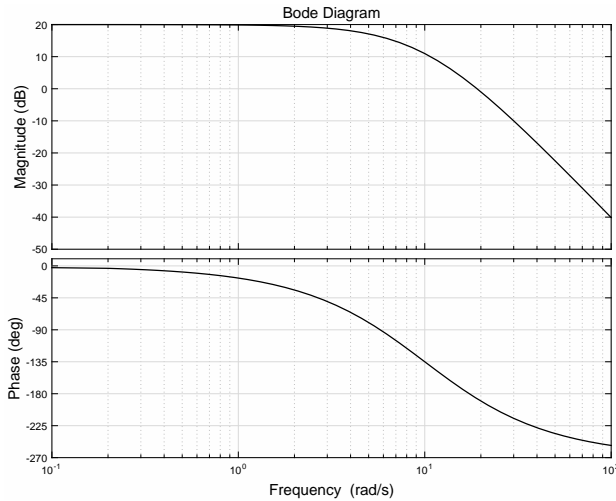
Sketch Nyquist plot for  $G(s)$  with  $k = 1$ , showing clearly any encirclements. Hint: phase of  $G(j\omega = 2)$  is  $+180^\circ$ .



[4 pts] d. Find the bounds on  $k$  for the system with unity feedback to be stable.

**Problem 3 (16 pts)**

The open-loop system is given by  $G(s) = \frac{10^4}{(s+10)^3}$ , and Bode plot for  $G(s)$  is here (Fig. 3.1):

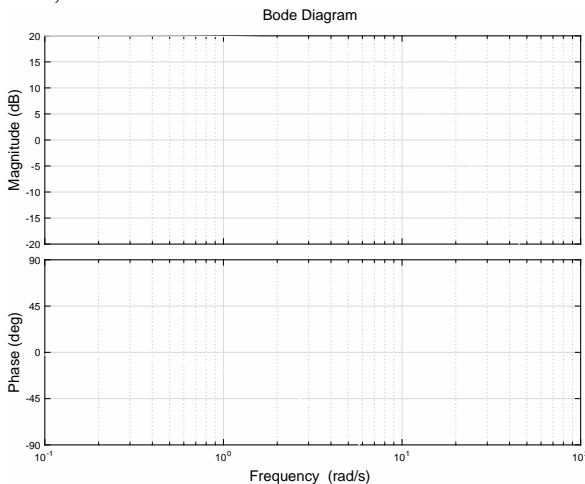


A lag controller  $D(s) = k \frac{s+\alpha}{s+\beta}$  is to be designed such that the unity gain feedback system with openloop transfer function  $D(s)G(s)$  has the same steady state error as with OLTF  $G(s)$  and has a nominal (asymptotic approximation) phase margin  $\phi_m = 55^\circ$  at  $\omega = 8 \text{ rad s}^{-1}$ . Note  $20 \log |G(j\omega = j8)| = 13 \text{ dB}$ .

[6 pts] a. Determine gain, zero, and pole location for the lag network  $D(s)$ :

gain  $k = \underline{\hspace{2cm}}$       zero  $\alpha = \underline{\hspace{2cm}}$       pole  $\beta = \underline{\hspace{2cm}}$

[4 pts] b. Sketch the asymptotic Bode plot for the lag network  $D(s)$  alone on the plot below (Fig. 3.2):



[4 pts] c. Sketch the asymptotic Bode plot for the combined lag network and plant  $D(s)G(s)$  on the plot (3.1) above.

[2 pts] d. Mark the phase and phase margin frequency on the plot of  $D(s)G(s)$  (Fig. 3.1). Explain briefly (1 sentence) how does the actual phase margin compare to the asymptotic prediction?

**Problem 4 (8 pts)**

You are given the following plant

$$\dot{\mathbf{x}} = A\mathbf{x} + Bu, \quad y = C\mathbf{x}$$

where  $A$  is  $N \times N$ ,  $u$  is scalar,  $B$  is  $N \times 1$ ,  $C$  is  $1 \times N$ , and  $\mathbf{x}$  is  $N \times 1$ . The system is observable and controllable.

[2 pts] a. Consider a controller  $u = r - K\mathbf{x}$  where  $r$  is a reference input, and  $K$  is  $1 \times N$ .

Determine the transfer function  $\frac{Y(s)}{R(s)} = \underline{\hspace{2cm}}$

[2 pts] b. Consider a controller  $u = K(r - \mathbf{x})$  where  $r$  is a reference input, and  $K$  is  $1 \times N$ .

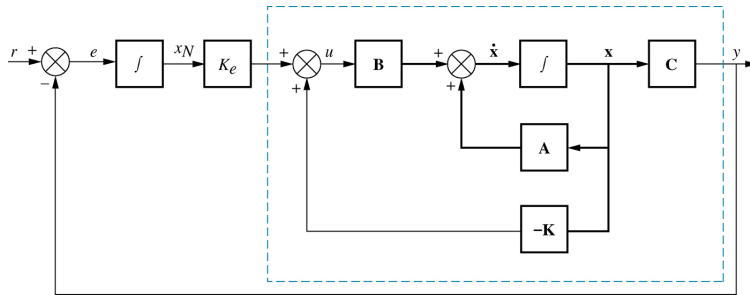
Determine the transfer function  $\frac{Y(s)}{R(s)} = \underline{\hspace{2cm}}$

[2 pts] c. Draw a block diagram of the controlled system using integrators, summing junctions, and scaling functions.

[3 pts] d. If the same  $K$  is used in part a. and b. above, briefly explain any difference in transfer function or behavior:

**Problem 5. (13 pts)**

Consider the following control system:



[3 pts] a. Write the state and output equations for the system, in terms of  $A, B, C, K, K_e$ .

$$\begin{bmatrix} \dot{\mathbf{x}} \\ \dot{x}_N \end{bmatrix} = \begin{bmatrix} & \\ & \end{bmatrix} \begin{bmatrix} \mathbf{x} \\ x_N \end{bmatrix} + \begin{bmatrix} \\ \end{bmatrix} r(t), \quad y = \begin{bmatrix} \\ \end{bmatrix} \begin{bmatrix} \mathbf{x} \\ x_N \end{bmatrix} \quad (1)$$

[6 pts] b. Given  $C = [1 \ 0]$ ,  $B = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$ ,  $A = \begin{bmatrix} 0 & 1 \\ -1 & -2 \end{bmatrix}$

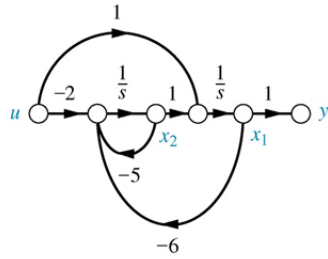
find  $K$  and  $K_e$  such that the closed loop poles are at  $s = -1, -2, -4$ .

$$K = [ \quad ] \quad K_e = \underline{\hspace{2cm}}$$

[4 pts] c. Show, with  $r(t)$  a unit step input, that  $e = 0$  in steady state (with the  $K, K_e$  found above). (Hint: do not use matrix inverse.)



**Problem 6. 13 pts**



Given the following system model:

[3 pts] a. Write the state and output equations for the system above.

$$\dot{\mathbf{x}} = \mathbf{A}\mathbf{x} + \mathbf{B}u = \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} \quad \\ \quad \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} \quad \\ \quad \end{bmatrix} u(t), \quad y = \mathbf{C}\mathbf{x} = \begin{bmatrix} \quad \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

[2 pts] b. Determine if the system  $A, B, C$  is controllable and observable.

[2 pts] c. Provide state equations for an observer which takes as inputs  $u(t), y(t)$ , and provides an estimate of the state  $\hat{\mathbf{x}}(t)$ .

[6 pts] d. Find observer gain  $L$  such that the observer has closed loop poles at  $s_1 = -10, s_2 = -10$ .

**Problem 7 (14 pts)**

[3 pts] a. Given  $G(s) = \frac{1}{s+2}$ . Let  $m(t)$  be the step response of  $g(t)$ , i.e.  $M(s) = \frac{1}{s(s+2)}$ . Let  $x_1(t) = m(t) - m(t - T)$  where  $T$  is the sampling period. Find  $X_1(z)$  the Z transform of  $x_1(t)$ .

$$X_1(z) = \underline{\hspace{2cm}}$$

[3 pts] b. Given  $\dot{x}_2(t) = -2x_2(t) + u(t)$ . Find the discrete time equivalent system using zero-order hold for input  $u(t)$  and sampling period  $T$ :  $x_2((k+1)T) = Gx_2(kT) + Hu(kT)$ .

$$G = \underline{\hspace{2cm}} \qquad H = \underline{\hspace{2cm}}$$

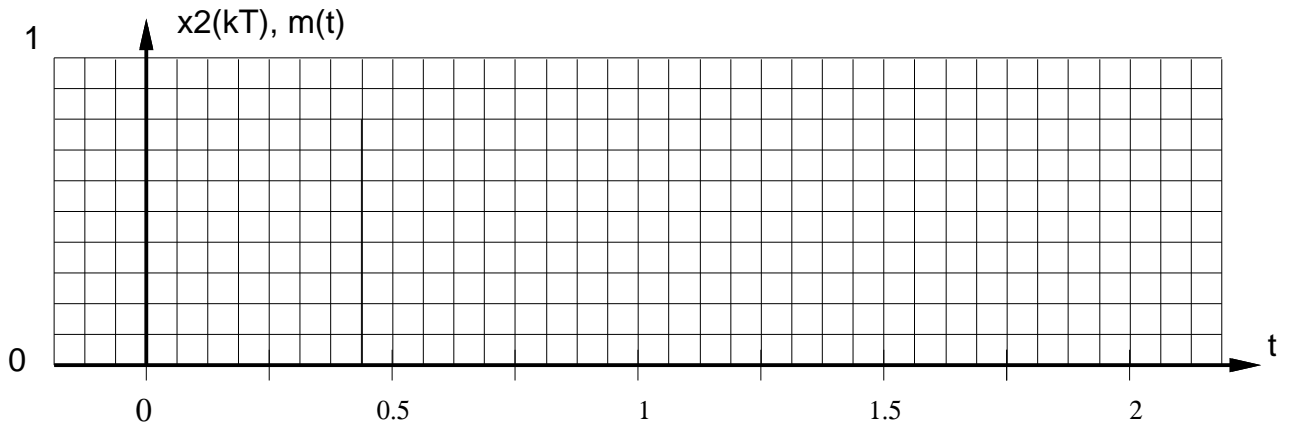
[2 pts] c. Find the  $\frac{X_2(z)}{U(z)}$  the discrete time transfer function from input  $u$  to state  $x_2$  using the state-space form.

$$\frac{X_2(z)}{U(z)} \underline{\hspace{2cm}}$$

**Problem 7, cont.**

[2 pts] d. Does  $\frac{X_2(z)}{U(z)} = X_1(z)$ ? Why or why not?

[4 pts] e. With zero initial conditions (ZSR),  $T = 0.25$ , and a unit step input for  $x_2(kT)$ , sketch  $m(t)$  and  $x_2(kT)$  on the plot below in the interval shown:



**Problem 8 Short Answers (8 pts)**

[4 pts] a. Given the discrete time system below, find  $\lim_{k \rightarrow \infty} x(k)$  for a unit step input  $u(k) = 1$ .

$$x(k+1) = \begin{bmatrix} 1/6 & 1/2 \\ 2/3 & 0 \end{bmatrix} x(k) + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(k)$$

$$\lim_{k \rightarrow \infty} x(k) = \begin{bmatrix} \quad \\ \quad \end{bmatrix}.$$

[4 pts] b. Given  $\dot{x}(t) = -10x(t) + u(t)$ . The discrete time equivalent system using zero-order hold for input  $u(t)$  and sampling period  $T$  is of the form  $x((k+1)T) = Gx(kT) + Hu(kT)$ . The discrete time system has a state feedback controller  $u(kT) = r(kT) - 20x(kT)$  applied.

- i. Find the eigenvalue for the closed loop system: \_\_\_\_
- ii. Find the largest value of  $T$  for which the system will be stable. (May be left in terms of  $\ln$ .)  
:  $T < \underline{\hspace{2cm}}$

Blank page for scratch work.