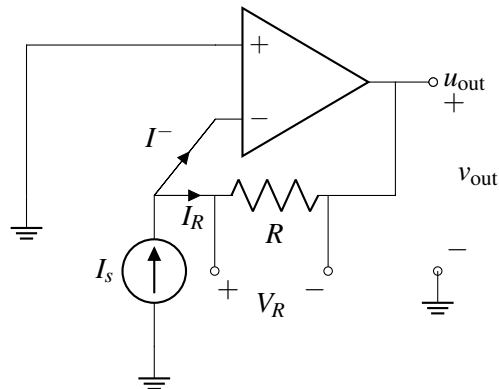


EECS 16A Designing Information Devices and Systems I

Fall 2019 Discussion 11B

1. A Trans-Resistance Amplifier



- (a) Calculate v_{out} as a function of I_s and R .

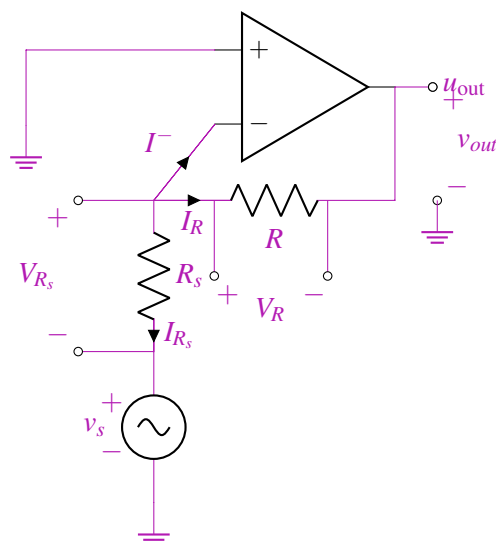
Answer:

$I^- = 0$, so $I_s = I_R$. Thus, $V_R = I_s R$.

Because the op-amp is in negative feedback, we also know that $u^+ = u^- = 0V$. Therefore, $v_{out} = u^- - V_R = -I_s R$.

- (b) Implement the same behavior as the above circuit, but replace the current source with a voltage source and a resistor.

Answer:

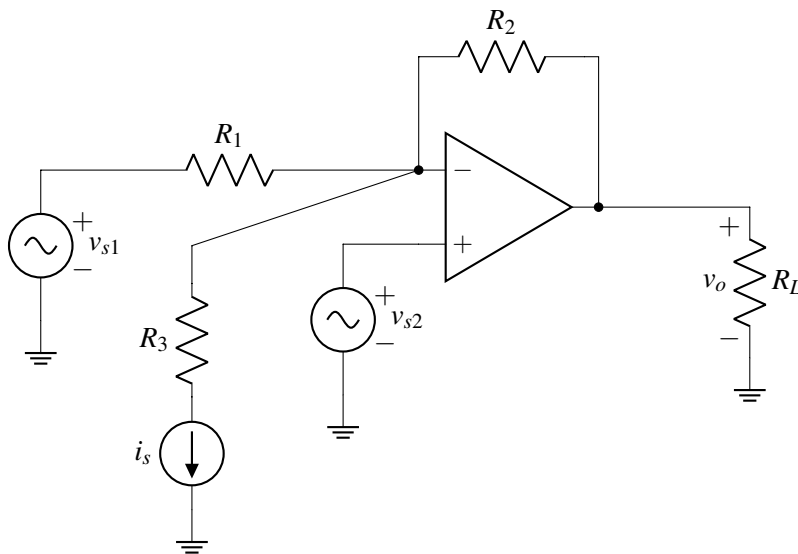


We know that $u^- = 0V$ because of the rules of op-amps in negative feedback. Applying KCL, we see that $I_{R_s} + I_R = 0 \Rightarrow I_{R_s} = -I_R$. In addition, applying Ohm's Law, $I_{R_s} = -\frac{v_s}{R_s}$, so $I_R = \frac{v_s}{R_s}$.

Hence, $v_{out} = -V_R = -\frac{v_s}{R_s} R = -\frac{R}{R_s} v_s$. This is the **inverting amplifier**.

2. Node Voltage Analysis with Op-Amps

Use node voltage analysis to find the output voltage v_o for the circuit shown below.



Answer:

We check if the op-amp is in negative feedback, and if it is ideal. We can see that the op-amp is in negative feedback because there is a path from the output back to the minus terminal, and we assume that the op-amp is ideal. Because both of these conditions are met, we can use the node voltage analysis for op-amp rules.

We write three equations for the op-amp: two KCL equations (one at the plus terminal, and one at minus terminal), and one equation setting $v^+ = v^-$. (Note that no current flows into or out of the op-amp's plus or minus terminals). First, we write the KCL equation for the minus terminal of the op-amp. All currents are defined as flowing out of the minus terminal:

$$i_{R_1} + i_{R_2} + i_{R_3} = 0.$$

Second, we note that $I_{v_{s2}} = 0$ because no current can enter the plus terminal. We do not need this equation because v^+ is set by the independent source v_{s2} .

Third, we set $v^+ = v^-$ which means that $v^- = v_{s2}$.

Because of the independent current source, we know:

$$i_{R_3} = i_s$$

By Ohm's law, we know:

$$i_{R_1} = \frac{v^- - v_{s1}}{R_1}$$

and

$$i_{R_2} = \frac{v^- - v_o}{R_2}$$

Then, substituting back into the original KCL equation, we have:

$$\frac{v^- - v_{s1}}{R_1} + \frac{v^- - v_o}{R_2} + i_s = 0$$

and substituting $v^- = v_{s2}$, we have:

$$\frac{v_{s2} - v_{s1}}{R_1} + \frac{v_{s2} - v_o}{R_2} + i_s = 0$$

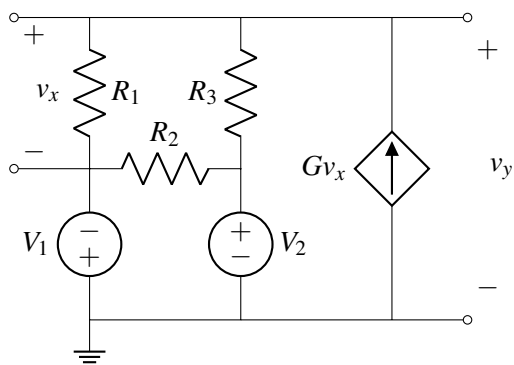
which we rearrange to find v_o , giving:

$$v_o = v_{s2} \left(1 + \frac{R_2}{R_1} \right) + i_s \cdot R_2 - \left(\frac{R_2}{R_1} \right) v_{s1}$$

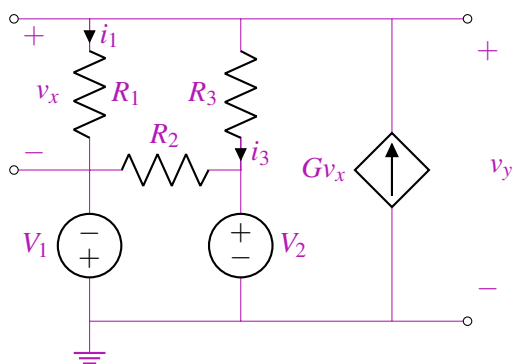
3. Take Node of the Voltage Sources

Use nodal analysis to solve for the voltages v_x and v_y . Use the following values for numerical calculations. **Note the polarity on the voltage sources.**

$$\begin{aligned} V_1 &= 5\text{ V} & R_1 &= 10\ \Omega \\ V_2 &= 5\text{ V} & R_2 &= 50\ \Omega \\ G &= \frac{1}{4}\text{ S} & R_3 &= 40\ \Omega \end{aligned}$$



Answer:



Applying KCL at the node v_y :

$$\begin{aligned} i_1 + i_3 - Gv_x &= 0\text{ A} \\ \frac{v_y + 5\text{ V}}{10\ \Omega} + \frac{v_y - 5\text{ V}}{40\ \Omega} - \left(\frac{1}{4}\text{ S}\right)(v_y + 5\text{ V}) &= 0\text{ A} \\ 4(v_y + 5\text{ V}) + (v_y - 5\text{ V}) - 10(v_y + 5\text{ V}) &= 0\text{ V} \end{aligned}$$

$$4v_y + 20\text{V} + v_y - 5\text{V} - 10v_y - 50\text{V} = 0\text{V}$$

$$-5v_y - 35\text{V} = 0\text{V}$$

$$v_y = -7\text{V}$$

$$v_x = v_y + 5\text{V} = -7\text{V} + 5\text{V} = -2\text{V}$$

Finally, we arrive at:

$$v_x = -2\text{V}$$

$$v_y = -7\text{V}$$