## EECS 16B Designing Information Devices and Systems II Spring 2016 Anant Sahai and Michel Maharbiz Discussion 1A

## 1. DFT of pure sinusoids

We can think of a real-world signal that is a function of time x(t). By recording its values at regular intervals, we can represent it as a vector of discrete samples  $\vec{x}$ , of length *n*.

$$\vec{x} = \begin{bmatrix} x[0] \\ x[1] \\ \vdots \\ x[n-1] \end{bmatrix}$$
(1)

Let  $\vec{X} = \begin{bmatrix} X[0] & \dots & X[n-1] \end{bmatrix}^T$  be the signal  $\vec{x}$  represented in the frequency domain, that is

$$\vec{X} = U^{-1}\vec{x} = U^*\vec{x}$$
(2)

where U is a matrix of the DFT basis vectors ( $\omega = e^{i\frac{2\pi}{n}}$ ).

$$U = \begin{bmatrix} | & | \\ \vec{u}_0 & \cdots & \vec{u}_{n-1} \\ | & | \end{bmatrix} = \frac{1}{\sqrt{n}} \begin{bmatrix} 1 & 1 & 1 & \cdots & 1 \\ 1 & \boldsymbol{\omega} & \boldsymbol{\omega}^2 & \cdots & \boldsymbol{\omega}^{n-1} \\ 1 & \boldsymbol{\omega}^2 & \boldsymbol{\omega}^4 & \cdots & \boldsymbol{\omega}^{2(n-1)} \\ \vdots & \vdots & \vdots & \vdots \\ 1 & \boldsymbol{\omega}^{n-1} & \boldsymbol{\omega}^{2(n-1)} & \cdots & \boldsymbol{\omega}^{(n-1)(n-1)} \end{bmatrix}$$
(3)

Alternatively, we have that  $\vec{x} = U\vec{X}$  or more explicitly

$$\vec{x} = X[0]\vec{u}_0 + \dots + X[n-1]\vec{u}_{n-1} \tag{4}$$

In other words,  $\vec{x}$  is a linear combination of the complex exponentials  $\vec{u}_i$  with coefficients X[i].

(a) Consider the continuous-time signal  $x(t) = \cos(\frac{2\pi}{3}t)$ . Suppose that we sampled it every 1 second to get (for n = 3 time steps):

$$\vec{x} = \left[\cos\left(\frac{2\pi}{3}(0)\right) \quad \cos\left(\frac{2\pi}{3}(1)\right) \quad \cos\left(\frac{2\pi}{3}(2)\right)\right]^T.$$

Compute  $\vec{X}$  for this signal.

(b) Now for the same signal as before, suppose that we took n = 6 samples. In this case we would have:

$$\vec{x} = \left[\cos\left(\frac{2\pi}{3}(0)\right) \quad \cos\left(\frac{2\pi}{3}(1)\right) \quad \cos\left(\frac{2\pi}{3}(2)\right) \quad \cos\left(\frac{2\pi}{3}(3)\right) \quad \cos\left(\frac{2\pi}{3}(4)\right) \quad \cos\left(\frac{2\pi}{3}(5)\right)\right]^T$$

Repeat what you did above. What is  $\vec{X}$  for this signal.

(c) Let's do this more generally. For the signal  $x(t) = \cos(\frac{2\pi k}{N}t)$ , compute  $\vec{X}$  of its vector form in discrete time,  $\vec{x}$ , of length n = N:

$$\vec{x} = \left[\cos\left(\frac{2\pi k}{N}(0)\right) \quad \cos\left(\frac{2\pi k}{N}(1)\right) \quad \cdots \quad \cos\left(\frac{2\pi k}{N}(n-1)\right)\right]^T.$$

## 2. The DFT basis and LTI systems

Suppose  $\vec{x}$  is the input signal applied to a linear time-invariant (LTI) system characterized by the impulse response  $\vec{h}$ . The output  $\vec{y}$  is given by  $C_{\vec{h}}\vec{x}$  where  $C_{\vec{h}}$  is a circulant matrix with the first column given by  $\vec{h}$ . Suppose the DFT coefficients of  $\vec{x}$  are given by

$$\vec{X} = \begin{bmatrix} X[0] & X[1] & \dots & X[n-1] \end{bmatrix}^T$$
.

(a) Compute the DFT representation of the impulse response  $\vec{h}$  as well as the eigenvalues of the circulant matrix that defines the system with that impulse response. (n = 3)

$$\vec{h} = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix}^T$$

(b) Compute the DFT representation of the impulse response  $\vec{h}$  as well as the eigenvalues of the circulant matrix that defines the system with that impulse response. (n = 3)

$$ec{h} = \begin{bmatrix} \cos \left( rac{2\pi}{3}(0) 
ight) & \cos \left( rac{2\pi}{3}(1) 
ight) & \cos \left( rac{2\pi}{3}(2) 
ight) \end{bmatrix}^T$$

(c) Compute the DFT representation of the impulse response  $\vec{h}$  as well as the eigenvalues of the circulant matrix that defines the system with that impulse response. (n = N)

$$\vec{h} = \left[\cos\left(\frac{2\pi k}{N}(0)\right) \quad \cos\left(\frac{2\pi k}{N}(1)\right) \quad \cdots \quad \cos\left(\frac{2\pi k}{N}(n-1)\right)\right]^T$$

## 3. Convolution and Duality

There is a term "convolution" that is often used in signal-processing. This is a very close sibling (how are they different?) to the term "correlation" that you have already learned in 16A. We call  $C_{\vec{x}}\vec{y}$  the **circular convolution** of the signals  $\vec{x}$  and  $\vec{y}$ , and this is sometimes denoted by  $\vec{x} \otimes \vec{y}$ .

- (a) Let  $\vec{z} = C_{\vec{x}}\vec{y}$ . Let  $\vec{X}, \vec{Y}, \vec{Z}$  be the DFT-basis representations of the signals  $\vec{x}, \vec{y}, \vec{z}$ . Use the properties of the DFT basis to relate  $\vec{Z}$  to  $\vec{X}$  and  $\vec{Y}$ .
- (b) Now, let's do something seemingly strange. Let's try the same thing in the frequency-domain. Let  $\vec{W} = C_{\vec{x}}\vec{Y}$ . What is the relationship between  $\vec{w}$  and  $\vec{x}$  and  $\vec{y}$ ?