

1. Cosine Transformation Assume that we are dealing with signals of length n .

$$\vec{x} = \begin{bmatrix} x[0] \\ x[1] \\ \vdots \\ x[n-1] \end{bmatrix} = \sum_{p=0}^{n-1} X[p] \vec{u}_p = \begin{cases} \sum_{p=-\lfloor \frac{n}{2} \rfloor}^{\lfloor \frac{n}{2} \rfloor} X[p] \vec{u}_p & \text{if } n \text{ odd} \\ \sum_{p=-\frac{n}{2}+1}^{\frac{n}{2}} X[p] \vec{u}_p & \text{if } n \text{ even} \end{cases} \quad (1)$$

In the lectures, we have seen that we can represent signals with sums of complex exponentials in the DFT basis. One non-intuitive aspect is that, even when the signal is real, the basis is complex. In this problem, we will explore a different representation in which real signals are written as linear combinations of periodic real signals.

Specifically, we will show how to derive

$$x[t] = \alpha_0 + \sum_{m=1}^{\lfloor \frac{n}{2} \rfloor} \alpha_m \cos\left(\frac{2\pi m}{n}t + \phi_m\right). \quad (2)$$

First, we need to understand cosines with phases.

- (a) For a real signal \vec{x} , the DFT coefficients are conjugate symmetric, i.e. $X[m] = X[-m]^*$ (you will show this in the homework). Therefore, suppose $X[m] = re^{i\theta}$, what is $X[-m]$?
- (b) Now assume that, for a real signal \vec{x} , its DFT coefficients are $X[m] = 0$ for $m \neq \pm 5$. Show that we can represent the t -th component of \vec{x} by $x[t] = \alpha \cos(\frac{2\pi}{n}5t + \phi)$. Find α and ϕ . Your answer should be in terms of $|X[5]|$, and $\angle X[5]$.
- (c) Therefore, let \vec{x} be an arbitrary signal of length n , where n is odd. Write it as a sum of cosines where the cosine scaling and phase are written in terms of the DFT coefficients, $X[-\frac{n-1}{2}], \dots, X[\frac{n-1}{2}]$. You can use $\angle z$ and $|z|$ to refer to the angle and magnitude of a complex number, respectively (i.e. $z = |z|e^{i\angle z}$).
- (d) How about when n is even?

2. Phase response Let \vec{x} be a real signal of length n (assume n odd till the end). In the previous part, we showed that we could write

$$x[t] = \alpha_0 + \sum_{p=1}^{\lfloor \frac{n}{2} \rfloor} \alpha_p \cos\left(\frac{2\pi p}{n}t + \phi_p\right).$$

where there are $1 + \frac{n-1}{2}$ different α_p parameters and $\frac{n-1}{2}$ parameters ϕ_p .

Let C be a circulant matrix with eigenvalues $\lambda_0, \dots, \lambda_{n-1}$. In lecture, we have seen that the DFT basis diagonalizes C (which correspond to LTI systems). However, the basis is complex and the eigenvalues are usually complex. However, if we push a real signal through C , we will get a real signal back. Where do all the imaginary parts go, then?

Let \vec{y} be the output of C with the input \vec{x} . The reason why the DFT basis is so useful is that, since C is diagonalized by the basis, we have

$$y[t] = \frac{1}{\sqrt{n}} \sum_{p=0}^{n-1} \lambda_p X[p] e^{j \frac{2\pi p}{n} t} \quad (3)$$

- (a) Use the fact that the complex exponentials are eigenvectors of C to write out what the output of C is when given the input \vec{x} , for the specific case of $x[t] = \cos(\frac{2\pi p}{n}t + \theta)$.
- (b) Using the fact that the eigenvalues λ_p for a real circulant matrix C exhibit conjugate symmetry (from the Homework), what cosine does this output correspond to?
- (c) What does the system C do to cosines in terms of the effect on their magnitude, frequency, and phase?
- (d) Write $y[t]$ entirely as a sum of cosines.
- (e) What changes for n even? (Think about what $\lambda_{\frac{n}{2}}$ must be like for a real C matrix.)

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