

**EE16B, Spring 2018
UC Berkeley EECS**

Maharbiz and Roychowdhury

Lectures 6A & 6B: Overview Slides

Controllability and Feedback

Where We Were Before

continuous AND discrete systems

pendulum

Any kind of system
(EE, mech., chem.,
optical, multi-domain,
...)

STATE SPACE
FORMULATION

NONLINEAR

RLC ckt.

LINEAR

LINEARIZATION

Stability

Bounded Input Bounded Output

Where We Were Before

continuous AND discrete systems

pendulum

Any kind of system
(EE, mech., chem.,
optical, multi-domain,
...)

STATE SPACE
FORMULATION

NONLINEAR

RLC ckt.

LINEAR

LINEARIZATION

Stability

Bounded Input Bounded Output
IC: blows up or dies down

Where We Were Before

continuous AND discrete systems

pendulum

Any kind of system
(EE, mech., chem.,
optical, multi-domain,
...)

STATE SPACE
FORMULATION

NONLINEAR

RLC ckt.

LINEAR

LINEARIZATION

Stability

Bounded Input Bounded Output

IC: blows up or dies down

**eigendecomposition turns
the system into n scalar ones**

Where We Were Before

continuous AND discrete systems

pendulum

Any kind of system
(EE, mech., chem.,
optical, multi-domain,
...)

STATE SPACE
FORMULATION

NONLINEAR

RLC ckt.

LINEAR

LINEARIZATION

Stability

Bounded Input Bounded Output

IC: blows up or dies down

**eigendecomposition turns
the system into n scalar ones**

C.T: $\text{Re}(\lambda_i) < 0$

Where We Were Before

continuous AND discrete systems

pendulum

Any kind of system
(EE, mech., chem.,
optical, multi-domain,
...)

STATE SPACE
FORMULATION

NONLINEAR

RLC ckt.

LINEAR

LINEARIZATION

Stability

Bounded Input Bounded Output

IC: blows up or dies down

**eigendecomposition turns
the system into n scalar ones**

C.T: $\text{Re}(\lambda_i) < 0$

D.T.: $|\lambda_i| < 1$

Where We Were Before

continuous AND discrete systems

pendulum

Any kind of system
(EE, mech., chem.,
optical, multi-domain,
...)

STATE SPACE
FORMULATION

NONLINEAR

RLC ckt.

LINEAR

LINEARIZATION

Stability

Bounded Input Bounded Output

IC: blows up or dies down

**eigendecomposition turns
the system into n scalar ones**

C.T: $\text{Re}(\lambda_i) < 0$

D.T.: $|\lambda_i| < 1$

TODAY

Controllability
and Feedback

Controllability

- Given (linearized) S.S.R: $\Delta \vec{x}[t + 1] = \overset{\text{nxn matrix}}{A} \Delta \vec{x}[t] + \overset{\text{nx1 vector}}{\vec{b}} \Delta u[t]$
- can you drive $\Delta \vec{x}[t]$ to any value you want (using $\Delta \vec{u}[t]$)?
 - ie, can you **control** $\Delta \vec{x}[t]$ completely?

Controllability

nxn matrix **nx1 vector**

- Given (linearized) S.S.R: $\Delta \vec{x}[t + 1] = A \Delta \vec{x}[t] + \vec{b} \Delta u[t]$
- can you drive $\Delta \vec{x}[t]$ to any value you want (using $\Delta \vec{u}[t]$)?
 - ie, can you **control** $\Delta \vec{x}[t]$ completely?

$$\Delta \vec{x}[t] = A^t \Delta \vec{x}[0] + \sum_{i=1}^t A^{t-i} \vec{b} \Delta u[i - 1]$$

Controllability

nxn matrix **nx1 vector**

- Given (linearized) S.S.R: $\Delta \vec{x}[t + 1] = A \Delta \vec{x}[t] + \vec{b} \Delta u[t]$

- can you drive $\Delta \vec{x}[t]$ to any value you want (using $\Delta \vec{u}[t]$)?

→ ie, can you **control** $\Delta \vec{x}[t]$ completely?

$$\Delta \vec{x}[t] = A^t \Delta \vec{x}[0] + \sum_{i=1}^t A^{t-i} \vec{b} \Delta u[i - 1]$$

$$\Delta \vec{x}[n] = A^n \Delta x[0] + \begin{bmatrix} A^{n-1} \vec{b}, & A^{n-2} \vec{b}, & \dots, & A \vec{b}, & \vec{b} \end{bmatrix} \begin{bmatrix} \Delta u[0] \\ \Delta u[1] \\ \vdots \\ \Delta u[n - 2] \\ \Delta u[n - 1] \end{bmatrix}$$

Controllability

nxn matrix **nx1 vector**

• Given (linearized) S.S.R: $\Delta \vec{x}[t + 1] = A \Delta \vec{x}[t] + \vec{b} \Delta u[t]$

• can you drive $\Delta \vec{x}[t]$ to any value you want (using $\Delta \vec{u}[t]$)?

→ ie, can you **control** $\Delta \vec{x}[t]$ completely?

• say $\Delta \vec{x}[0] = 0$ (w.l.o.g, see notes)

$$\Delta \vec{x}[t] = A^t \Delta \vec{x}[0] + \sum_{i=1}^t A^{t-i} \vec{b} \Delta u[i - 1]$$

$$\Delta \vec{x}[n] = A^n \Delta x[0] + \begin{bmatrix} A^{n-1} \vec{b}, A^{n-2} \vec{b}, \dots, A \vec{b}, \vec{b} \end{bmatrix} \begin{bmatrix} \Delta u[0] \\ \Delta u[1] \\ \vdots \\ \Delta u[n - 2] \\ \Delta u[n - 1] \end{bmatrix}$$

Controllability

nxn matrix **nx1 vector**

- Given (linearized) S.S.R: $\Delta \vec{x}[t + 1] = A \Delta \vec{x}[t] + \vec{b} \Delta u[t]$

- can you drive $\Delta \vec{x}[t]$ to any value you want (using $\Delta \vec{u}[t]$)?

- ie, can you **control** $\Delta \vec{x}[t]$ completely?

- say $\Delta \vec{x}[0] = 0$ (w.l.o.g, see notes)

$$\Delta \vec{x}[t] = A^t \Delta \vec{x}[0] + \sum_{i=1}^t A^{t-i} \vec{b} \Delta u[i - 1]$$

$$\Delta \vec{x}[n] = \begin{bmatrix} A^{n-1} \vec{b}, A^{n-2} \vec{b}, \dots, A \vec{b}, \vec{b} \end{bmatrix} \begin{bmatrix} \Delta u[0] \\ \Delta u[1] \\ \vdots \\ \Delta u[n - 2] \\ \Delta u[n - 1] \end{bmatrix}$$

Controllability

nxn matrix **nx1 vector**

• Given (linearized) S.S.R: $\Delta \vec{x}[t + 1] = A \Delta \vec{x}[t] + \vec{b} \Delta u[t]$

• can you drive $\Delta \vec{x}[t]$ to any value you want (using $\Delta \vec{u}[t]$)?

→ ie, can you **control** $\Delta \vec{x}[t]$ completely?

• say $\Delta \vec{x}[0] = 0$ (w.l.o.g, see notes)

$$\Delta \vec{x}[t] = A^t \Delta \vec{x}[0] + \sum_{i=1}^t A^{t-i} \vec{b} \Delta u[i - 1]$$

$$\Delta \vec{x}[n] = \begin{bmatrix} A^{n-1} \vec{b}, A^{n-2} \vec{b}, \dots, A \vec{b}, \vec{b} \end{bmatrix} \begin{bmatrix} \Delta u[0] \\ \Delta u[1] \\ \vdots \\ \Delta u[n - 2] \\ \Delta u[n - 1] \end{bmatrix}$$

• would like to make $\Delta \vec{x}[n]$ anything we like in \mathcal{R}^n

Controllability

nxn matrix **nx1 vector**

• Given (linearized) S.S.R: $\Delta \vec{x}[t + 1] = A \Delta \vec{x}[t] + \vec{b} \Delta u[t]$

• can you drive $\Delta \vec{x}[t]$ to any value you want (using $\Delta \vec{u}[t]$)?

→ ie, can you **control** $\Delta \vec{x}[t]$ completely?

$$\Delta \vec{x}[t] = A^t \Delta \vec{x}[0] + \sum_{i=1}^t A^{t-i} \vec{b} \Delta u[i - 1]$$

• say $\Delta \vec{x}[0] = 0$ (w.l.o.g, see notes)

must be full rank (ie, rank = n)

$$\Delta \vec{x}[n] = \begin{bmatrix} A^{n-1} \vec{b}, A^{n-2} \vec{b}, \dots, A \vec{b}, \vec{b} \end{bmatrix} \begin{bmatrix} \Delta u[0] \\ \Delta u[1] \\ \vdots \\ \Delta u[n-2] \\ \Delta u[n-1] \end{bmatrix}$$

• would like to make $\Delta \vec{x}[n]$ anything we like in \mathcal{R}^n

• rank: number of lin. indep. columns (= # of lin. indep. rows)

Controllability

nxn matrix **nx1 vector**

• Given (linearized) S.S.R: $\Delta \vec{x}[t + 1] = A \Delta \vec{x}[t] + \vec{b} \Delta u[t]$

• can you drive $\Delta \vec{x}[t]$ to any value you want (using $\Delta \vec{u}[t]$)?

→ ie, can you **control** $\Delta \vec{x}[t]$ completely?

$$\Delta \vec{x}[t] = A^t \Delta \vec{x}[0] + \sum_{i=1}^t A^{t-i} \vec{b} \Delta u[i - 1]$$

• say $\Delta \vec{x}[0] = 0$ (w.l.o.g, see notes)

must be full rank (ie, rank = n)

$$\Delta \vec{x}[n] = \begin{bmatrix} A^{n-1} \vec{b}, & A^{n-2} \vec{b}, & \dots, & A \vec{b}, & \vec{b} \end{bmatrix} \begin{bmatrix} \Delta u[0] \\ \Delta u[1] \\ \vdots \\ \Delta u[n - 2] \\ \Delta u[n - 1] \end{bmatrix}$$

• would like to make $\Delta \vec{x}[n]$ anything we like in \mathcal{R}^n

• rank: number of lin. indep. columns (= # of lin. indep. rows)

Controllability: simple example

- $\text{span}(\left[A^{t-1}\vec{b} \mid A^{t-2}\vec{b} \mid \dots \mid A\vec{b} \mid \vec{b} \right]) = \text{span}(\left[\vec{b} \mid A\vec{b} \mid \dots \mid A^{t-1}\vec{b} \right])$

Controllability: simple example

- $\text{span}(\left[A^{t-1}\vec{b} \mid A^{t-2}\vec{b} \mid \dots \mid A\vec{b} \mid \vec{b} \right]) = \text{span}(\left[\vec{b} \mid A\vec{b} \mid \dots \mid A^{t-1}\vec{b} \right])$
- $A = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix}, \quad B = \vec{b} = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$

Controllability: simple example

- $\text{span}(\left[A^{t-1}\vec{b} \mid A^{t-2}\vec{b} \mid \dots \mid A\vec{b} \mid \vec{b} \right]) = \text{span}(\left[\vec{b} \mid A\vec{b} \mid \dots \mid A^{t-1}\vec{b} \right])$
- $A = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix}, \quad B = \vec{b} = \begin{bmatrix} 1 \\ 0 \end{bmatrix} \quad [B \mid AB \mid A^2B \mid \dots] = \begin{bmatrix} 1 & 1 & 1 & \dots \\ 0 & 0 & 0 & \dots \end{bmatrix}$

Controllability: simple example

- $\text{span}(\left[A^{t-1}\vec{b} \mid A^{t-2}\vec{b} \mid \dots \mid A\vec{b} \mid \vec{b} \right]) = \text{span}(\left[\vec{b} \mid A\vec{b} \mid \dots \mid A^{t-1}\vec{b} \right])$
- $A = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix}, \quad B = \vec{b} = \begin{bmatrix} 1 \\ 0 \end{bmatrix} \quad [B \mid AB \mid A^2B \mid \dots] = \begin{bmatrix} 1 & 1 & 1 & \dots \\ 0 & 0 & 0 & \dots \end{bmatrix}$
rank = 1 < n=2

Controllability: simple example

- $\text{span}(\left[A^{t-1}\vec{b} \mid A^{t-2}\vec{b} \mid \dots \mid A\vec{b} \mid \vec{b} \right]) = \text{span}(\left[\vec{b} \mid A\vec{b} \mid \dots \mid A^{t-1}\vec{b} \right])$

- $A = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix}, B = \vec{b} = \begin{bmatrix} 1 \\ 0 \end{bmatrix} \quad [B \mid AB \mid A^2B \mid \dots] = \begin{bmatrix} 1 & 1 & 1 & \dots \\ 0 & 0 & 0 & \dots \end{bmatrix}$
not controllable **rank = 1 < n=2**

Controllability: simple example

- $\text{span}(\left[A^{t-1}\vec{b} \mid A^{t-2}\vec{b} \mid \dots \mid A\vec{b} \mid \vec{b} \right]) = \text{span}(\left[\vec{b} \mid A\vec{b} \mid \dots \mid A^{t-1}\vec{b} \right])$
- $A = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix}$, $B = \vec{b} = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$ $[B \mid AB \mid A^2B \mid \dots] = \begin{bmatrix} 1 & 1 & 1 & \dots \\ 0 & 0 & 0 & \dots \end{bmatrix}$
not controllable rank = 1 < n=2
- **The system:** $\begin{bmatrix} \Delta x_1[t+1] \\ \Delta x_2[t+1] \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix} \begin{bmatrix} \Delta x_1[t] \\ \Delta x_2[t] \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(t)$

Controllability: simple example

- $\text{span}(\left[A^{t-1}\vec{b} \mid A^{t-2}\vec{b} \mid \dots \mid A\vec{b} \mid \vec{b} \right]) = \text{span}(\left[\vec{b} \mid A\vec{b} \mid \dots \mid A^{t-1}\vec{b} \right])$
- $A = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix}$, $B = \vec{b} = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$ $[B \mid AB \mid A^2B \mid \dots] = \begin{bmatrix} 1 & 1 & 1 & \dots \\ 0 & 0 & 0 & \dots \end{bmatrix}$
not controllable rank = 1 < n=2

- **The system:**
$$\begin{bmatrix} \Delta x_1[t+1] \\ \Delta x_2[t+1] \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix} \begin{bmatrix} \Delta x_1[t] \\ \Delta x_2[t] \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(t)$$

 $\Delta x_2[t+1] = 2\Delta x_2[t]$

Controllability: simple example

- $\text{span}(\left[A^{t-1}\vec{b} \mid A^{t-2}\vec{b} \mid \dots \mid A\vec{b} \mid \vec{b} \right]) = \text{span}(\left[\vec{b} \mid A\vec{b} \mid \dots \mid A^{t-1}\vec{b} \right])$
- $A = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix}$, $B = \vec{b} = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$ $[B \mid AB \mid A^2B \mid \dots] = \begin{bmatrix} 1 & 1 & 1 & \dots \\ 0 & 0 & 0 & \dots \end{bmatrix}$
not controllable rank = 1 < n=2

- **The system:** $\begin{bmatrix} \Delta x_1[t+1] \\ \Delta x_2[t+1] \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix} \begin{bmatrix} \Delta x_1[t] \\ \Delta x_2[t] \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(t)$
 $\Delta u(t)$ has no influence on $\Delta x_2[t]$ $\Delta x_2[t+1] = 2\Delta x_2[t]$

Controllability: simple example

- $\text{span}(\left[A^{t-1}\vec{b} \mid A^{t-2}\vec{b} \mid \dots \mid A\vec{b} \mid \vec{b} \right]) = \text{span}(\left[\vec{b} \mid A\vec{b} \mid \dots \mid A^{t-1}\vec{b} \right])$
- $A = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix}, B = \vec{b} = \begin{bmatrix} 1 \\ 0 \end{bmatrix} \quad [B \mid AB \mid A^2B \mid \dots] = \begin{bmatrix} 1 & 1 & 1 & \dots \\ 0 & 0 & 0 & \dots \end{bmatrix}$
not controllable rank = 1 < n=2

- **The system:** $\begin{bmatrix} \Delta x_1[t+1] \\ \Delta x_2[t+1] \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix} \begin{bmatrix} \Delta x_1[t] \\ \Delta x_2[t] \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(t)$
 $\Delta u(t)$ has no influence on $\Delta x_2[t]$ $\Delta x_2[t+1] = 2\Delta x_2[t]$

- When does $\vec{b}, A\vec{b}, \dots$ run out of lin. indep vectors?

Controllability: simple example

- $\text{span}(\left[A^{t-1}\vec{b} \mid A^{t-2}\vec{b} \mid \dots \mid A\vec{b} \mid \vec{b} \right]) = \text{span}(\left[\vec{b} \mid A\vec{b} \mid \dots \mid A^{t-1}\vec{b} \right])$
- $A = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix}$, $B = \vec{b} = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$ $[B \mid AB \mid A^2B \mid \dots] = \begin{bmatrix} 1 & 1 & 1 & \dots \\ 0 & 0 & 0 & \dots \end{bmatrix}$
not controllable rank = 1 < n=2

- **The system:** $\begin{bmatrix} \Delta x_1[t+1] \\ \Delta x_2[t+1] \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix} \begin{bmatrix} \Delta x_1[t] \\ \Delta x_2[t] \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(t)$
 $\Delta u(t)$ has no influence on $\Delta x_2[t]$ $\Delta x_2[t+1] = 2\Delta x_2[t]$

- When does $\vec{b}, A\vec{b}, \dots$ run out of lin. indep vectors?
- every A has a **minimal polynomial** (result from lin. alg.)
 \rightarrow ie, for some $k \leq n$, $A^k + c_{k-1}A^{k-1} + c_{k-2}A^{k-2} + \dots + c_1A + c_0I = 0$

Controllability: simple example

- $\text{span}(\left[A^{t-1}\vec{b} \mid A^{t-2}\vec{b} \mid \dots \mid A\vec{b} \mid \vec{b} \right]) = \text{span}(\left[\vec{b} \mid A\vec{b} \mid \dots \mid A^{t-1}\vec{b} \right])$
- $A = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix}, B = \vec{b} = \begin{bmatrix} 1 \\ 0 \end{bmatrix} \quad [B \mid AB \mid A^2B \mid \dots] = \begin{bmatrix} 1 & 1 & 1 & \dots \\ 0 & 0 & 0 & \dots \end{bmatrix}$
not controllable rank = 1 < n=2

- **The system:** $\begin{bmatrix} \Delta x_1[t+1] \\ \Delta x_2[t+1] \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix} \begin{bmatrix} \Delta x_1[t] \\ \Delta x_2[t] \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(t)$
 $\Delta u(t)$ has no influence on $\Delta x_2[t]$ $\Delta x_2[t+1] = 2\Delta x_2[t]$

- When does $\vec{b}, A\vec{b}, \dots$ run out of lin. indep vectors?
- every A has a **minimal polynomial** (result from lin. alg.)
 - ie, for some $k \leq n$, $A^k + c_{k-1}A^{k-1} + c_{k-2}A^{k-2} + \dots + c_1A + c_0I = 0$
 - ie, $A^k\vec{b} = \underbrace{-c_{k-1}A^{k-1}\vec{b} - c_{k-2}A^{k-2}\vec{b} - \dots - c_1A\vec{b} - c_0\vec{b}}_{\text{linear comb. of } [b, Ab, A^2b, \dots, A^{k-1}b]}$

Controllability: simple example

- $\text{span}(\left[A^{t-1}\vec{b} \mid A^{t-2}\vec{b} \mid \dots \mid A\vec{b} \mid \vec{b} \right]) = \text{span}(\left[\vec{b} \mid A\vec{b} \mid \dots \mid A^{t-1}\vec{b} \right])$
- $A = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix}, B = \vec{b} = \begin{bmatrix} 1 \\ 0 \end{bmatrix} \quad [B \mid AB \mid A^2B \mid \dots] = \begin{bmatrix} 1 & 1 & 1 & \dots \\ 0 & 0 & 0 & \dots \end{bmatrix}$
not controllable rank = 1 < n=2

- **The system:**
$$\begin{bmatrix} \Delta x_1[t+1] \\ \Delta x_2[t+1] \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix} \begin{bmatrix} \Delta x_1[t] \\ \Delta x_2[t] \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(t)$$

 $\Delta u(t)$ has no influence on $\Delta x_2[t]$ $\rightarrow \Delta x_2[t+1] = 2\Delta x_2[t]$

- When does $\vec{b}, A\vec{b}, \dots$ run out of lin. indep vectors?
- every A has a **minimal polynomial** (result from lin. alg.)
 - \rightarrow ie, for some $k \leq n$, $A^k + c_{k-1}A^{k-1} + c_{k-2}A^{k-2} + \dots + c_1A + c_0I = 0$
 - \rightarrow ie, $A^k\vec{b} = \underbrace{-c_{k-1}A^{k-1}\vec{b} - c_{k-2}A^{k-2}\vec{b} - \dots - c_1A\vec{b} - c_0\vec{b}}_{\text{linear comb. of } [b, Ab, A^2b, \dots, A^{k-1}b]}$
 - \rightarrow ie, $A^k\vec{b}, A^{k+1}\vec{b}, \dots$ will not contribute new linearly indep. columns

Cayley-Hamilton Theorem

- Every matrix A satisfies its own characteristic polynomial!


- char. poly.: $p_A(\lambda) \triangleq \det(A - \lambda I)$
$$= \lambda^n + a_{n-1}\lambda^{n-1} + \dots + a_1\lambda + a_0$$

Cayley-Hamilton Theorem

- Every matrix A satisfies its own characteristic polynomial!

- char. poly.: $p_A(\lambda) \triangleq \det(A - \lambda I)$
 $= \lambda^n + a_{n-1}\lambda^{n-1} + \dots + a_1\lambda + a_0$

scalar



Cayley-Hamilton Theorem

- Every matrix A satisfies its own characteristic polynomial!

- char. poly.: $p_A(\lambda) \triangleq \det(A - \lambda I)$
 $= \lambda^n + a_{n-1}\lambda^{n-1} + \dots + a_1\lambda + a_0$

scalar
↓

- C-H Thm: $p_A(A) = 0$

$$\Rightarrow A^n + a_{n-1}A^{n-1} + \dots + a_1A + a_0 = 0$$

Cayley-Hamilton Theorem

- Every matrix A satisfies its own characteristic polynomial!

- char. poly.: $p_A(\lambda) \triangleq \det(A - \lambda I)$
 $= \lambda^n + a_{n-1}\lambda^{n-1} + \dots + a_1\lambda + a_0$

- C-H Thm: $p_A(A) = 0$
 $\Rightarrow A^n + a_{n-1}A^{n-1} + \dots + a_1A + a_0 = 0$


scalar

matrix


Cayley-Hamilton Theorem

- Every matrix A satisfies its own characteristic polynomial!

- char. poly.: $p_A(\lambda) \triangleq \det(A - \lambda I)$
 $= \lambda^n + a_{n-1}\lambda^{n-1} + \dots + a_1\lambda + a_0$

scalar 

- C-H Thm: $p_A(A) = 0$
 $\Rightarrow A^n + a_{n-1}A^{n-1} + \dots + a_1A + a_0 = 0$

matrix 

- implication:

$\rightarrow A^n \vec{b} = \underbrace{-a_{n-1}A^{n-1}\vec{b} - a_{n-2}A^{n-2}\vec{b} - \dots - a_1A\vec{b} - a_0\vec{b}}_{\text{linear comb. of } [b, Ab, A^2b, \dots, A^{n-1}b]}$

Cayley-Hamilton Theorem

- Every matrix A satisfies its own characteristic polynomial!

- char. poly.: $p_A(\lambda) \triangleq \det(A - \lambda I)$
 $= \lambda^n + a_{n-1}\lambda^{n-1} + \dots + a_1\lambda + a_0$

scalar →

- C-H Thm: $p_A(A) = 0$
 $\Rightarrow A^n + a_{n-1}A^{n-1} + \dots + a_1A + a_0 = 0$

matrix →

- implication:

→ $A^n \vec{b} = \underbrace{-a_{n-1}A^{n-1}\vec{b} - a_{n-2}A^{n-2}\vec{b} - \dots - a_1A\vec{b} - a_0\vec{b}}_{\text{linear comb. of } [b, Ab, A^2b, \dots, A^{n-1}b]}$

- ie, $A^n b, A^{n+1} b, \dots$ will not contribute new linearly indep. columns

Cayley-Hamilton Theorem

- Every matrix A satisfies its own characteristic polynomial!

- char. poly.: $p_A(\lambda) \triangleq \det(A - \lambda I)$
 $= \lambda^n + a_{n-1}\lambda^{n-1} + \dots + a_1\lambda + a_0$

scalar →

- C-H Thm: $p_A(A) = 0$
 $\Rightarrow A^n + a_{n-1}A^{n-1} + \dots + a_1A + a_0 = 0$

matrix →

- implication:

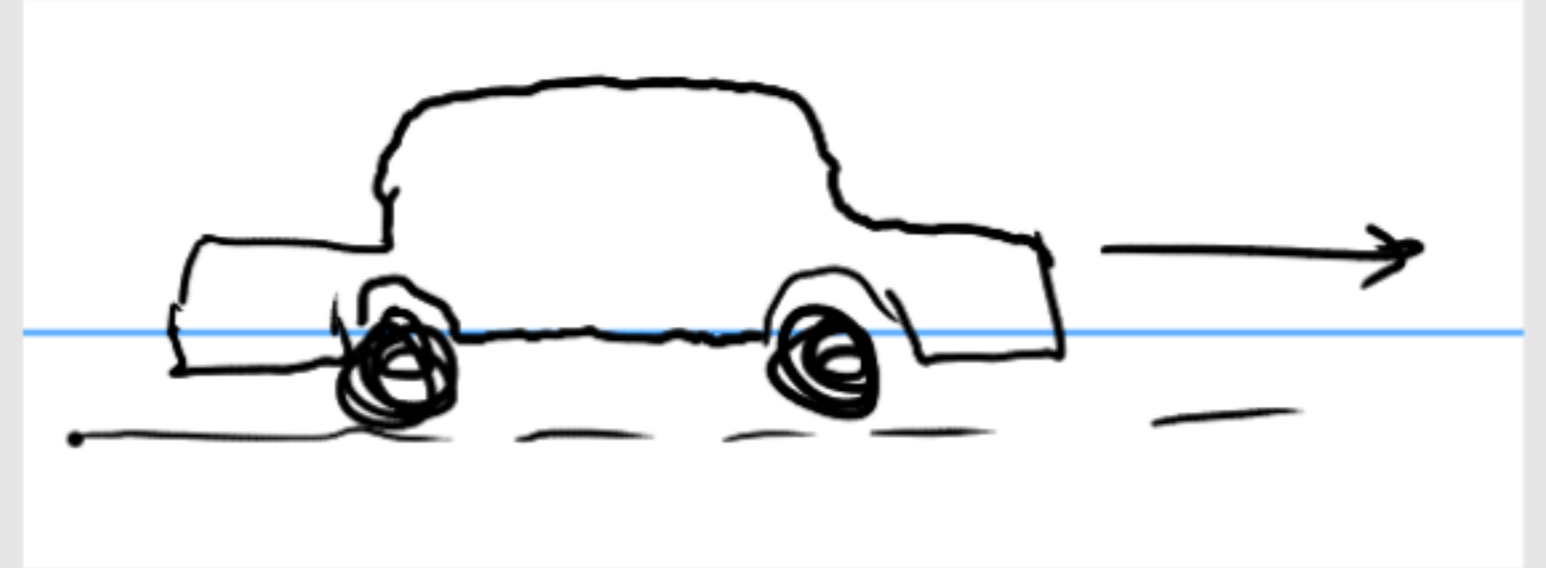
→ $A^n \vec{b} = \underbrace{-a_{n-1}A^{n-1}\vec{b} - a_{n-2}A^{n-2}\vec{b} - \dots - a_1A\vec{b} - a_0\vec{b}}_{\text{linear comb. of } [b, Ab, A^2b, \dots, A^{n-1}b]}$

- ie, $A^n b, A^{n+1} b, \dots$ will not contribute new linearly indep. columns

if no eigenvalues repeated, then n is the degree of the minimal polynomial (ie, $k=n$)

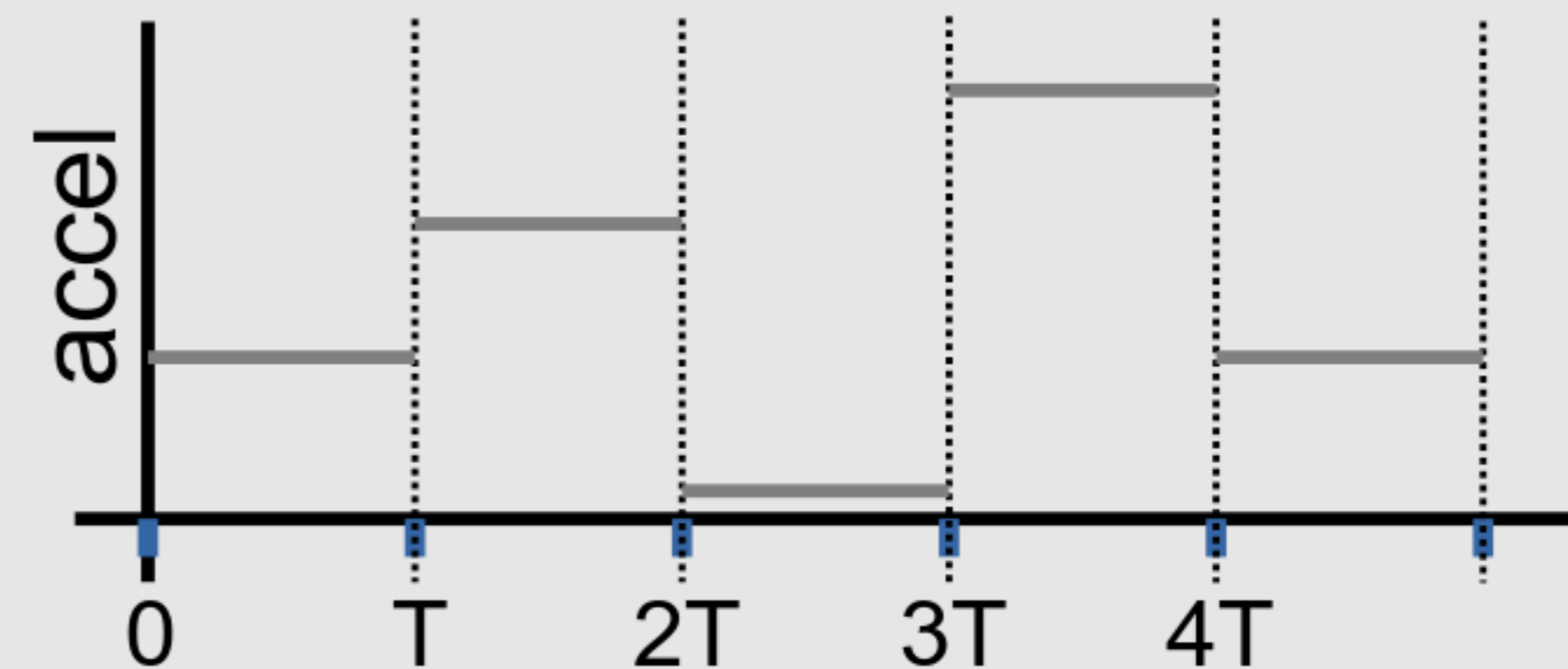
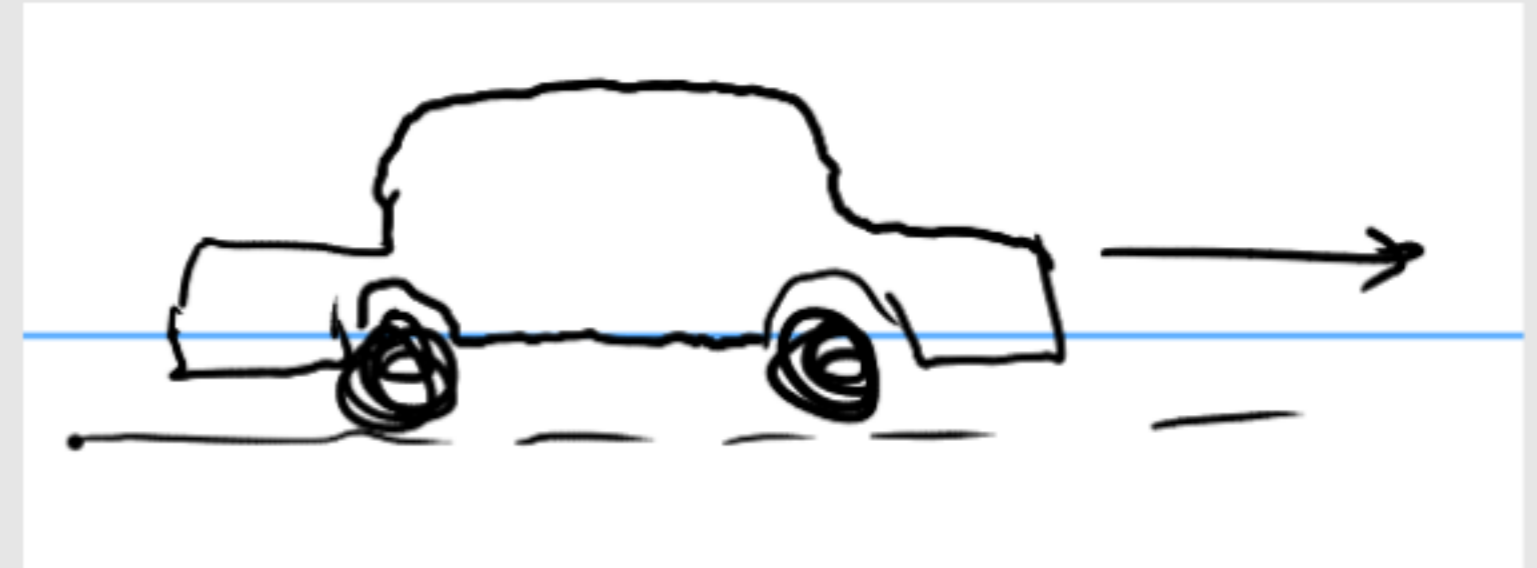
Example: Accelerating Car

- control input: **acceleration**



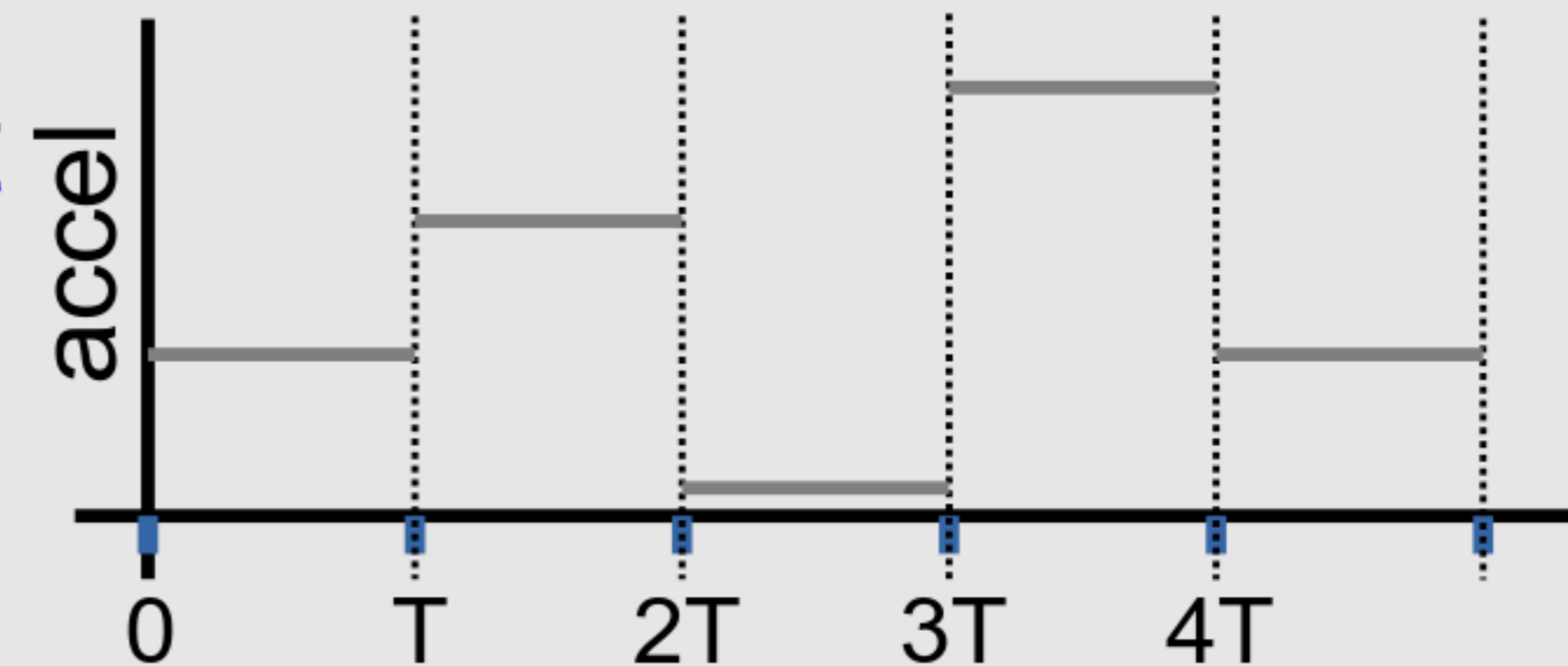
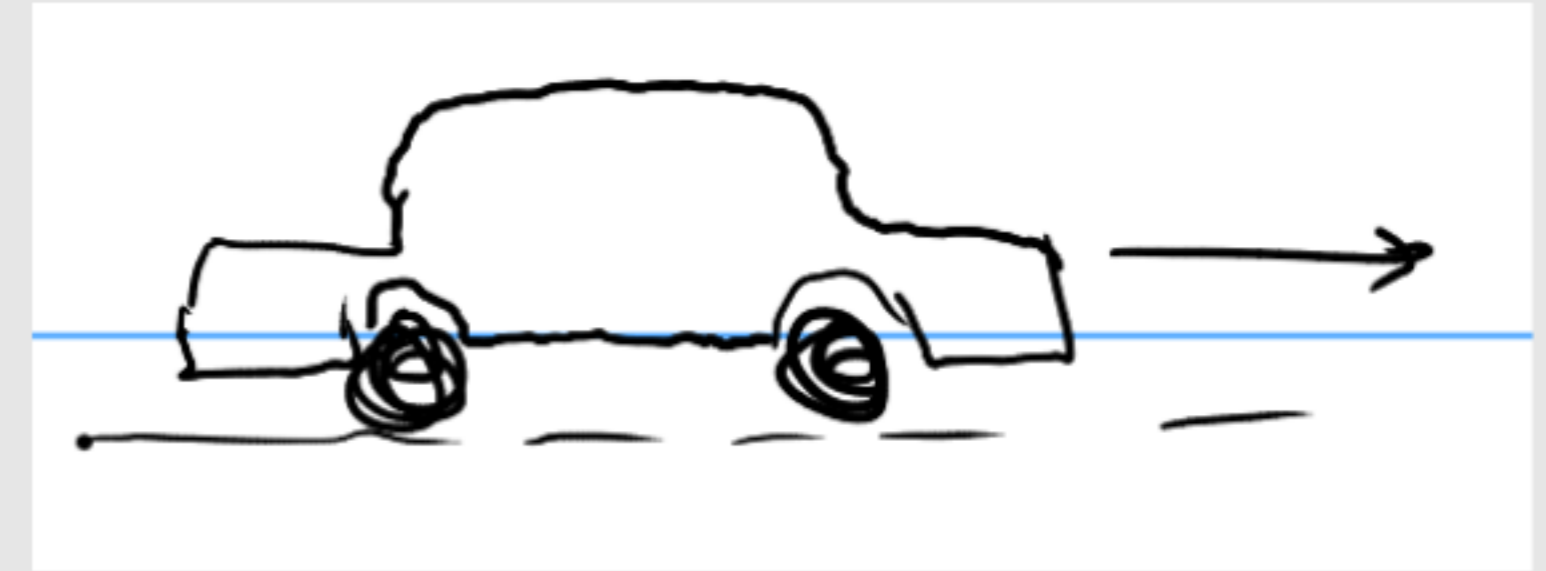
Example: Accelerating Car

- control input: **acceleration**
- can change only every T secs
 - stays constant in between



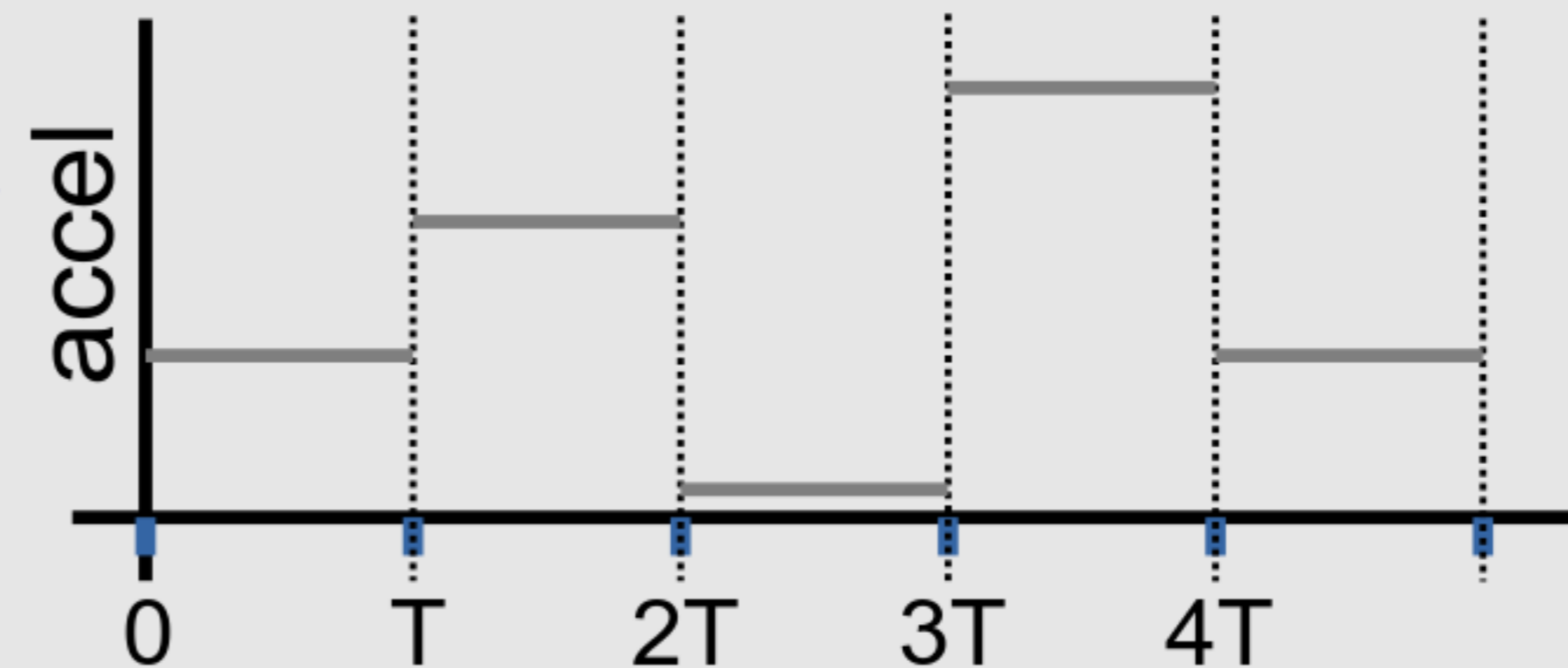
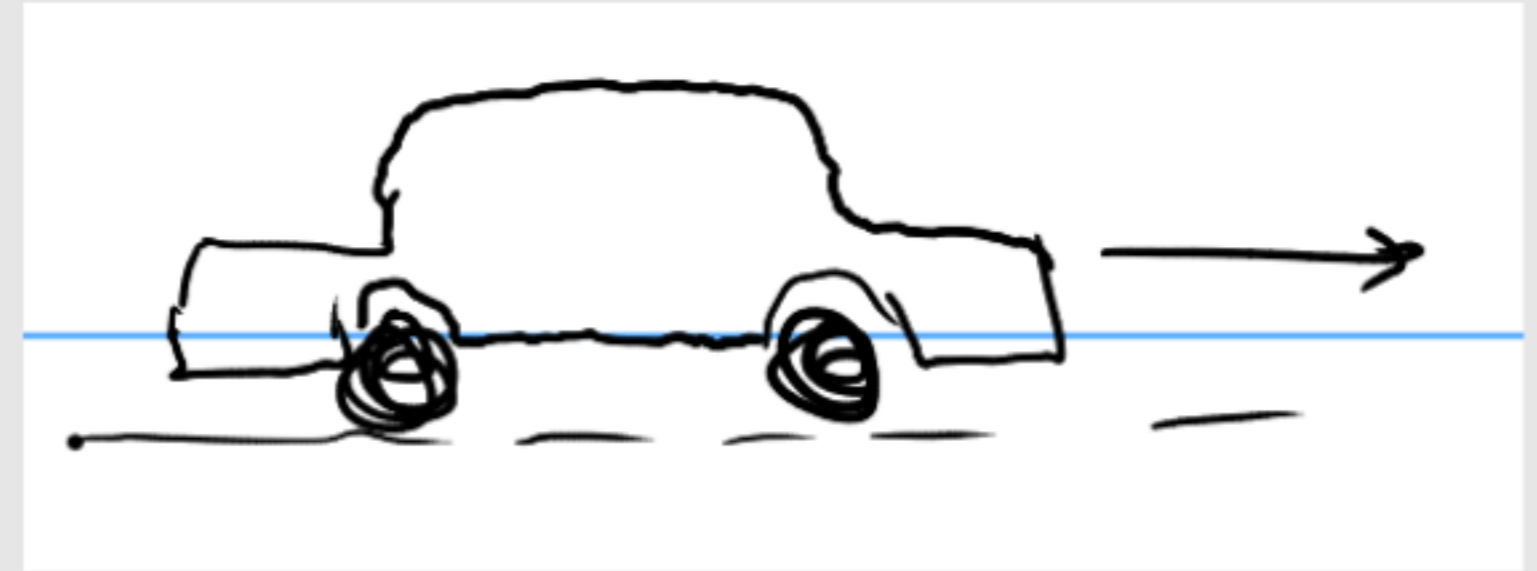
Example: Accelerating Car

- control input: **acceleration**
- can change only every T secs
→ stays constant in between
- Q: can we set its **position AND velocity** to whatever we want (at time = multiples of T)?



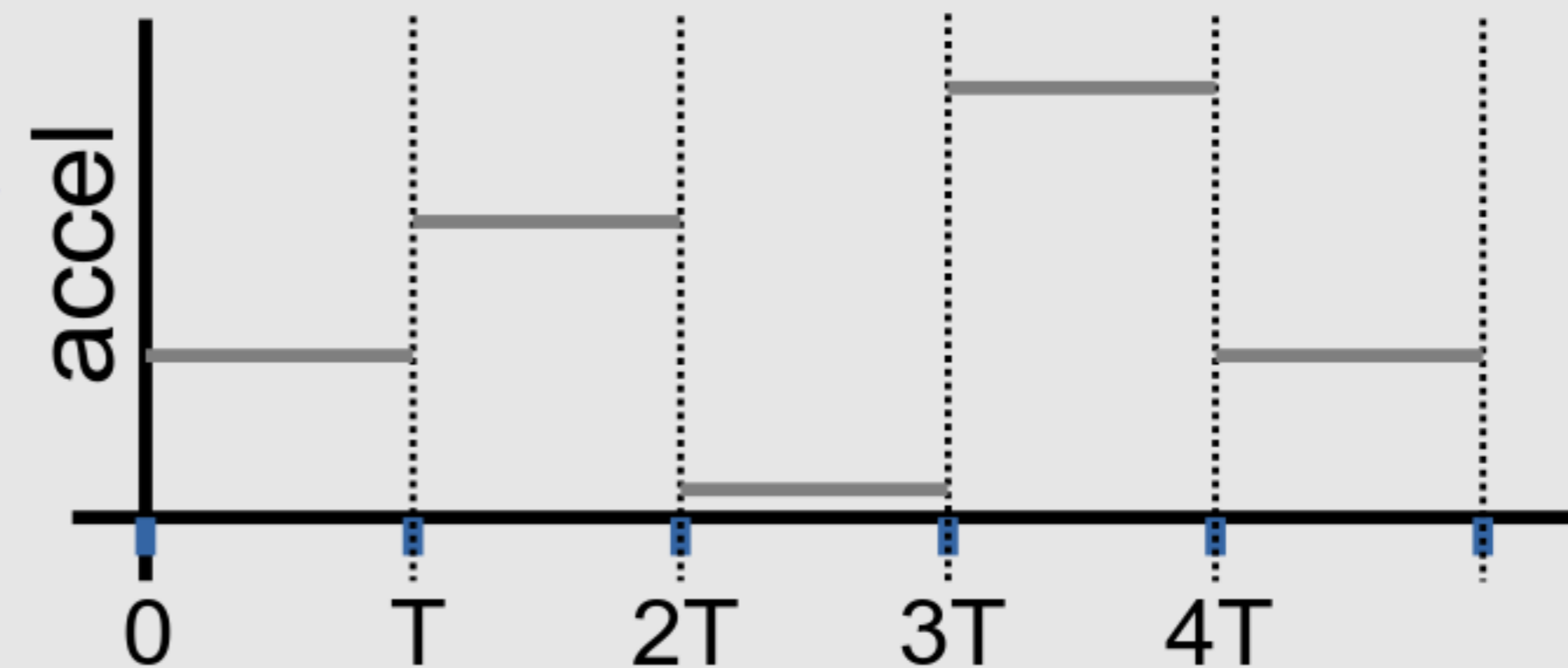
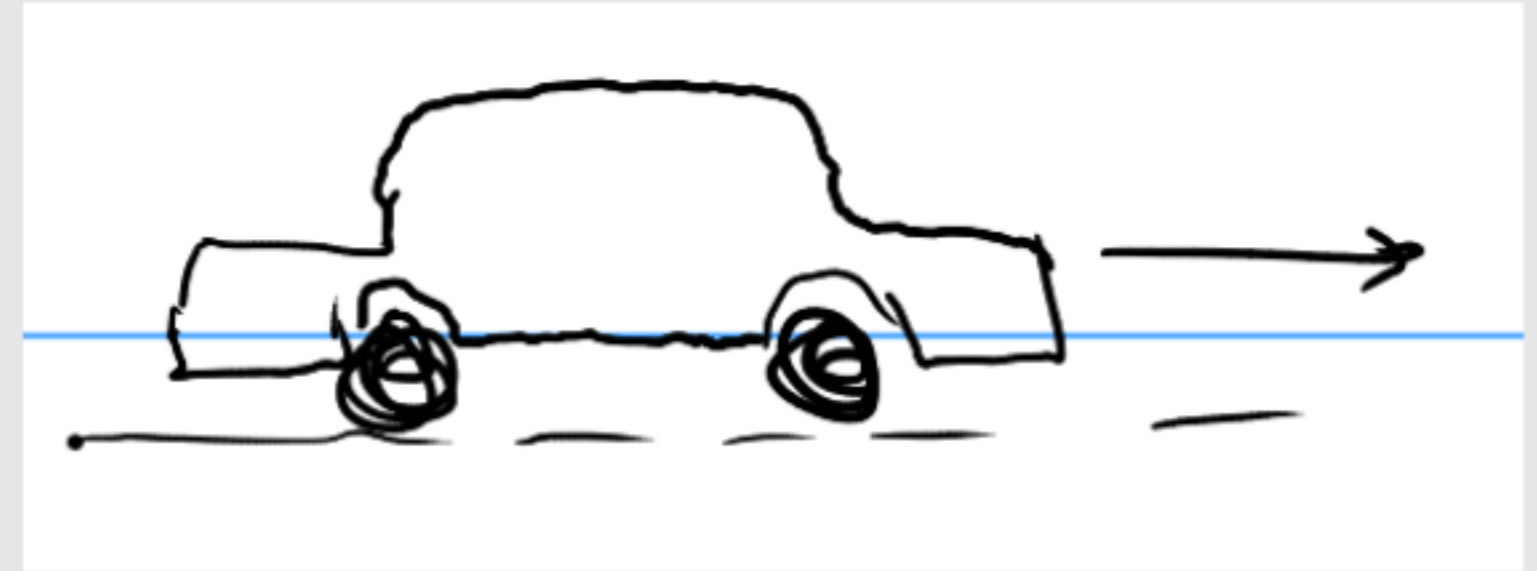
Example: Accelerating Car

- control input: **acceleration**
- can change only every T secs
→ stays constant in between
- Q: can we set its **position AND velocity** to whatever we want (at time = multiples of T)?
- analysis approach
 - find a discrete SSR for position/vel.
 - analyse its controllability



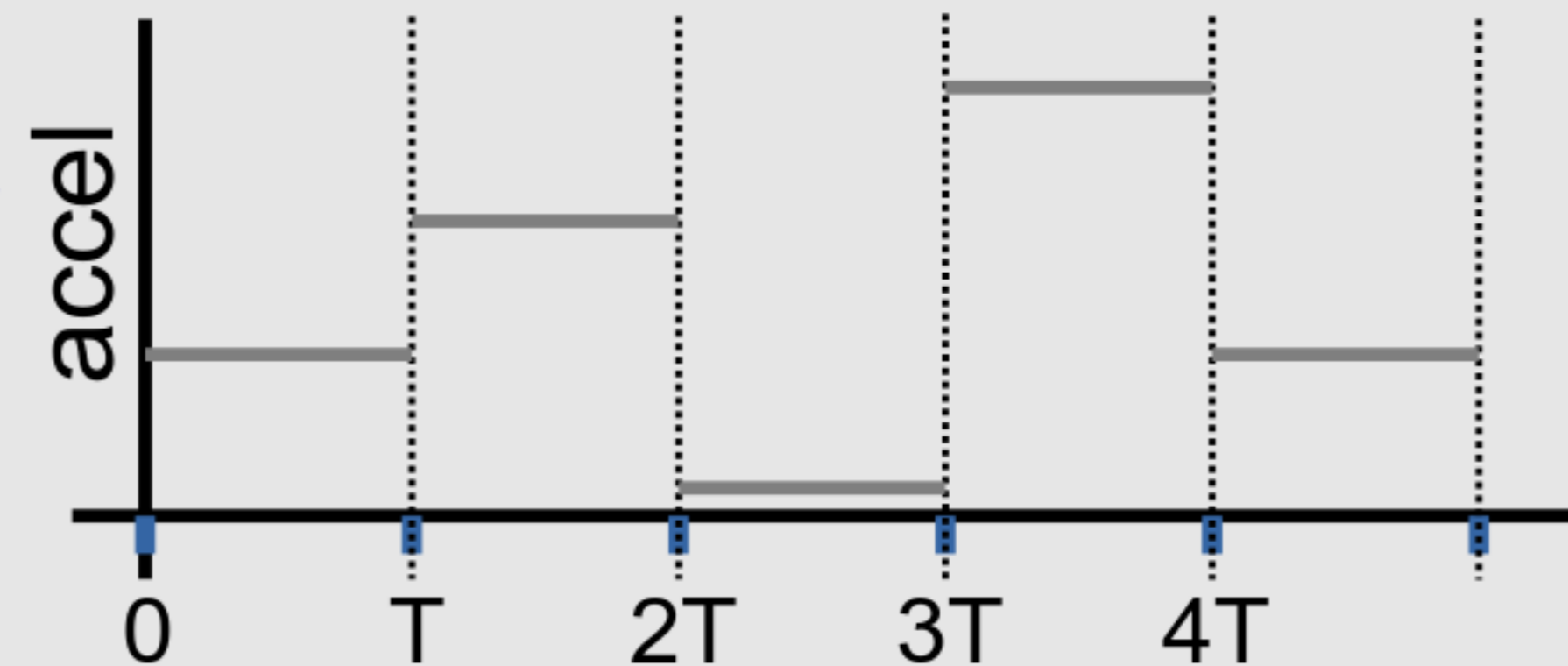
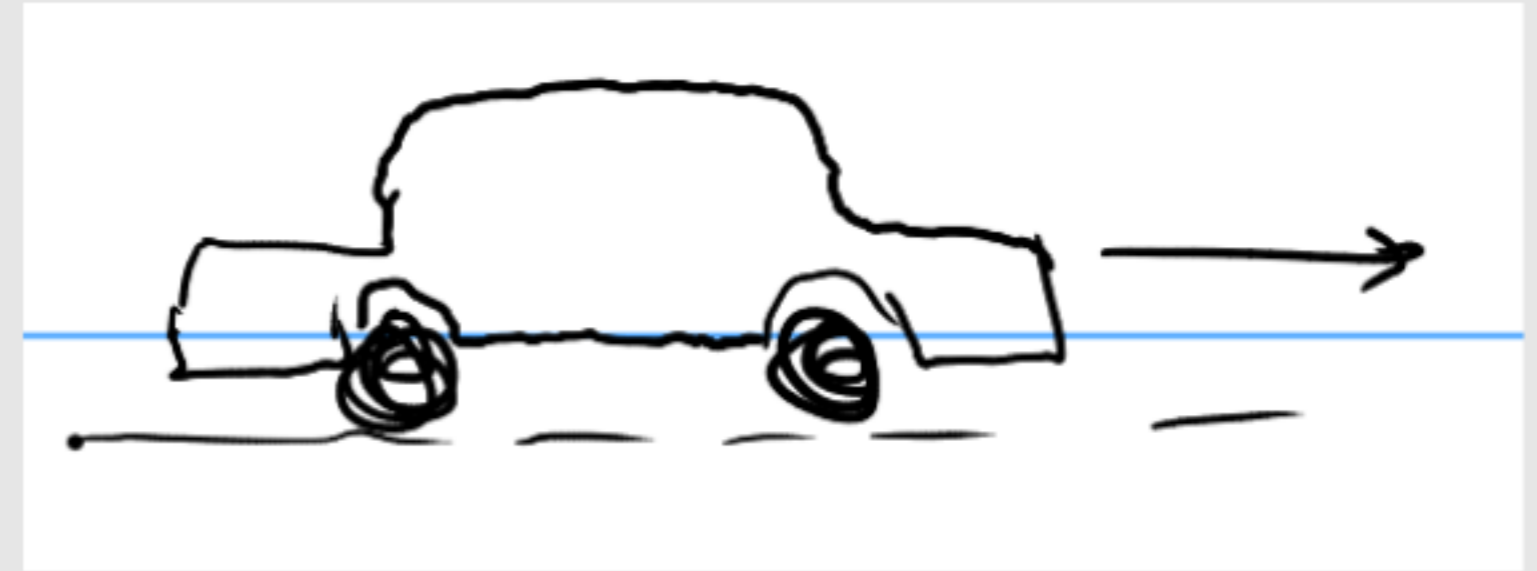
Example: Accelerating Car

- control input: **acceleration**
- can change only every T secs
→ stays constant in between
- Q: can we set its **position AND velocity** to whatever we want (at time = multiples of T)?
- analysis approach
 - find a discrete SSR for position/vel.
 - analyse its controllability
- acceleration: a ; velocity: v ; position: x



Example: Accelerating Car

- control input: **acceleration**
- can change only every T secs
→ stays constant in between
- Q: can we set its **position AND velocity** to whatever we want (at time = multiples of T)?
- analysis approach
 - find a discrete SSR for position/vel.
 - analyse its controllability

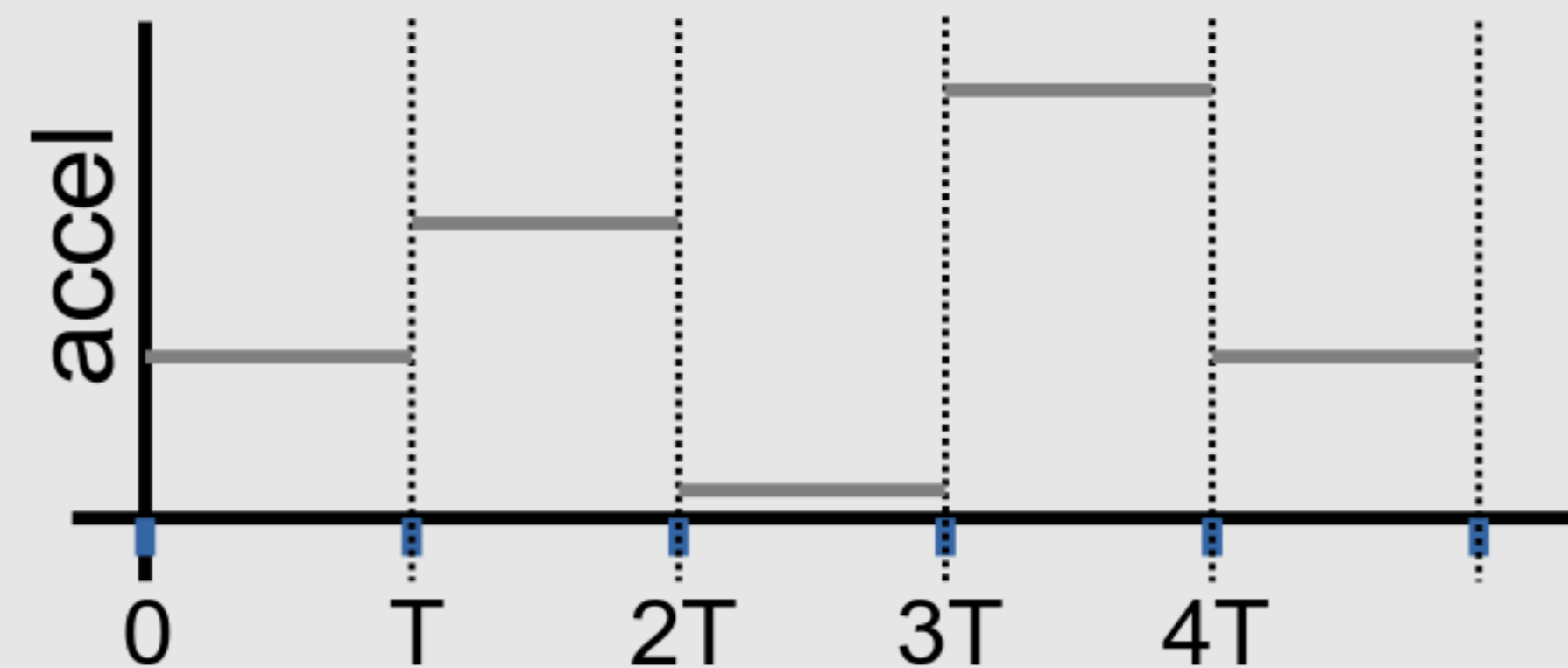
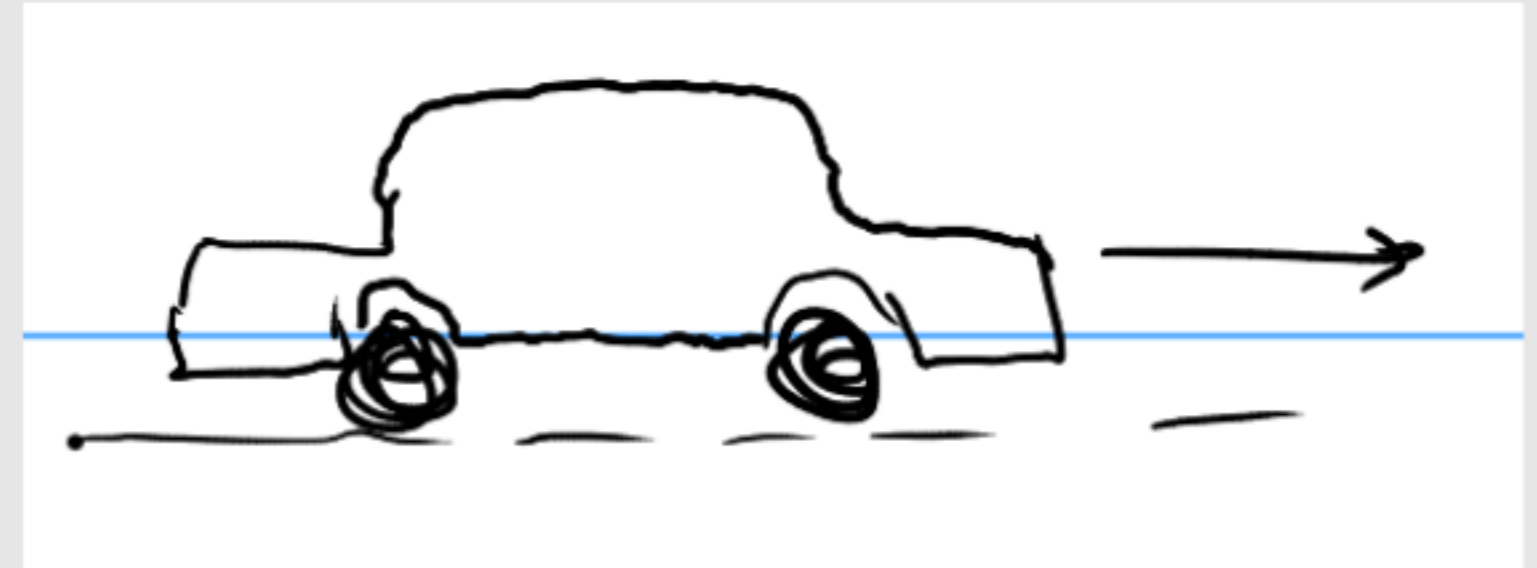


- acceleration: **a**; velocity: **v**; position: **x**

- $v(\tau) = \int_0^\tau a(\tau_2) d\tau_2, \quad x(\tau) = \int_0^\tau v(\tau_2) d\tau_2$

Example: Accelerating Car

- control input: **acceleration**
- can change only every T secs
→ stays constant in between
- Q: can we set its **position AND velocity** to whatever we want (at time = multiples of T)?
- analysis approach
 - find a discrete SSR for position/vel.
 - analyse its controllability



- acceleration: **a**; velocity: **v**; position: **x**

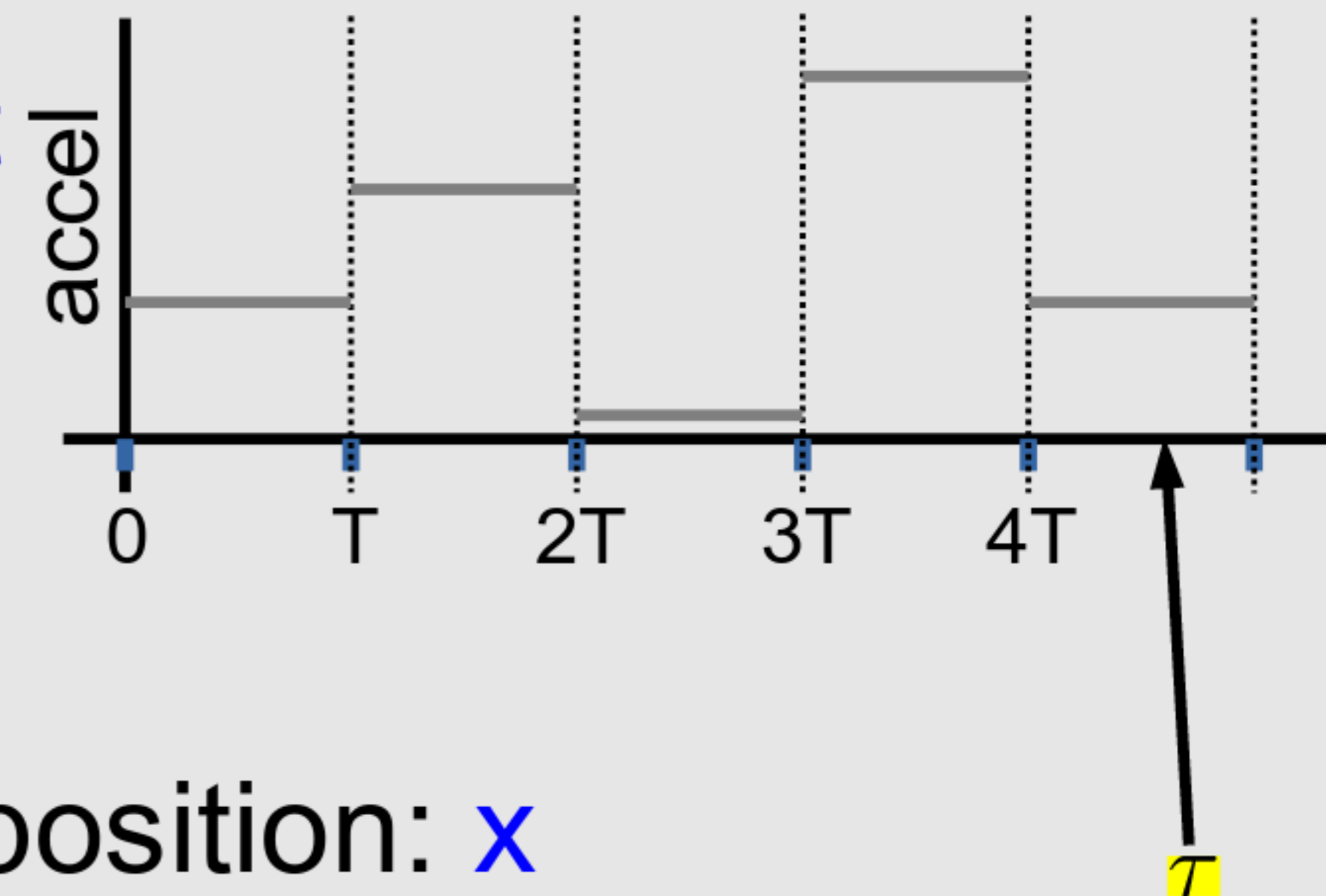
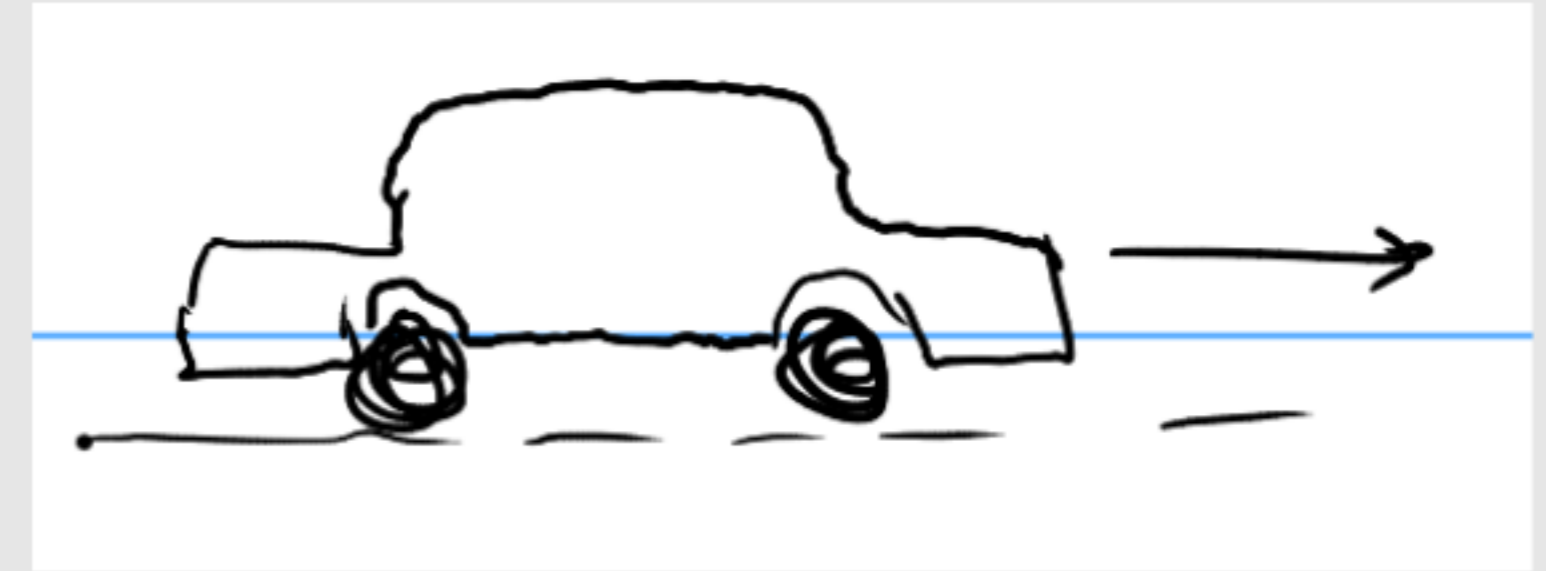
- $v(\tau) = \int_0^\tau a(\tau_2) d\tau_2, \quad x(\tau) = \int_0^\tau v(\tau_2) d\tau_2$

- $v(\tau) - v(tT) = \int_{tT}^\tau a(\tau_2) d\tau_2$

$$tT \leq \tau \leq (t+1)T$$

Example: Accelerating Car

- control input: **acceleration**
- can change only every T secs
→ stays constant in between
- Q: can we set its **position AND velocity** to whatever we want (at time = multiples of T)?
- analysis approach
 - find a discrete SSR for position/vel.
 - analyse its controllability



- acceleration: **a**; velocity: **v**; position: **x**

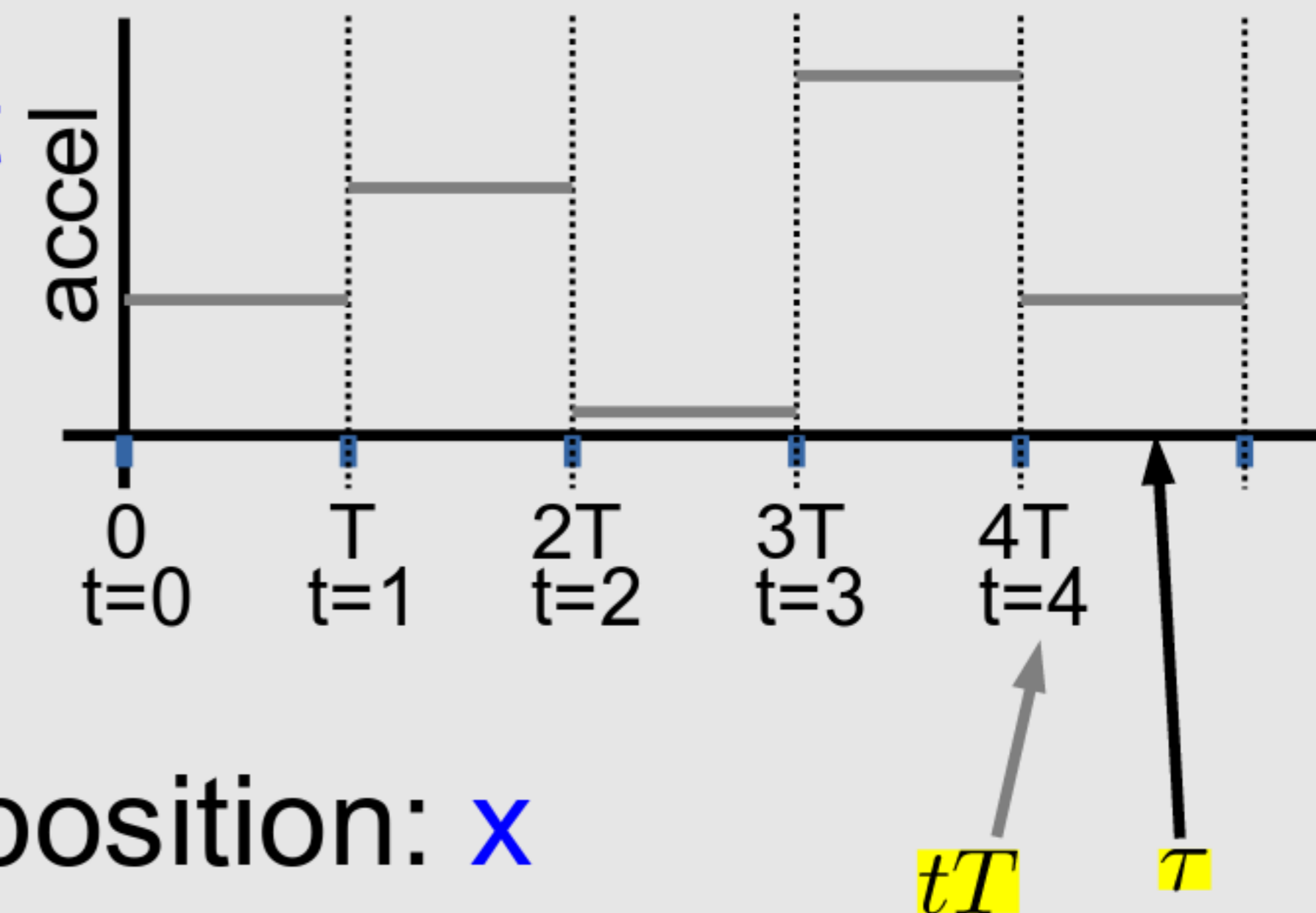
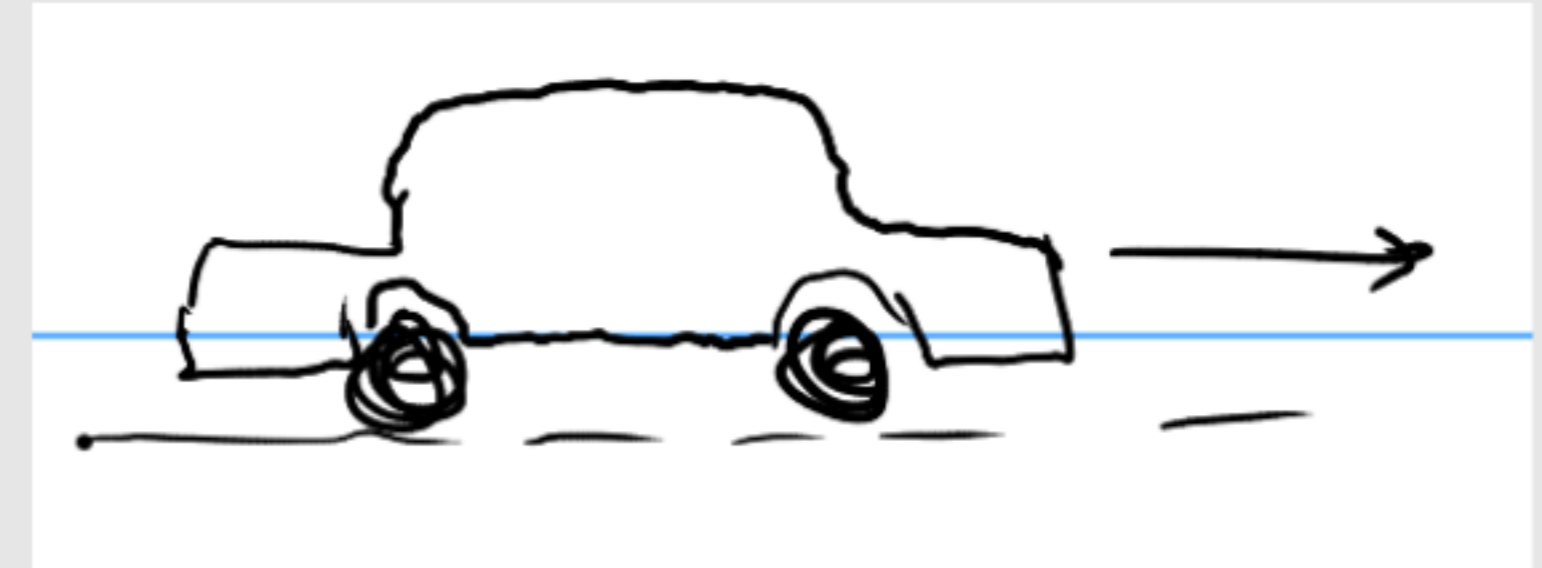
- $v(\tau) = \int_0^\tau a(\tau_2) d\tau_2, \quad x(\tau) = \int_0^\tau v(\tau_2) d\tau_2$

- $v(\tau) - v(tT) = \int_{tT}^\tau a(\tau_2) d\tau_2$

$$tT \leq \tau \leq (t+1)T$$

Example: Accelerating Car

- control input: **acceleration**
- can change only every T secs
→ stays constant in between
- Q: can we set its **position AND velocity** to whatever we want (at time = multiples of T)?
- analysis approach
 - find a discrete SSR for position/vel.
 - analyse its controllability



- acceleration: **a**; velocity: **v**; position: **x**

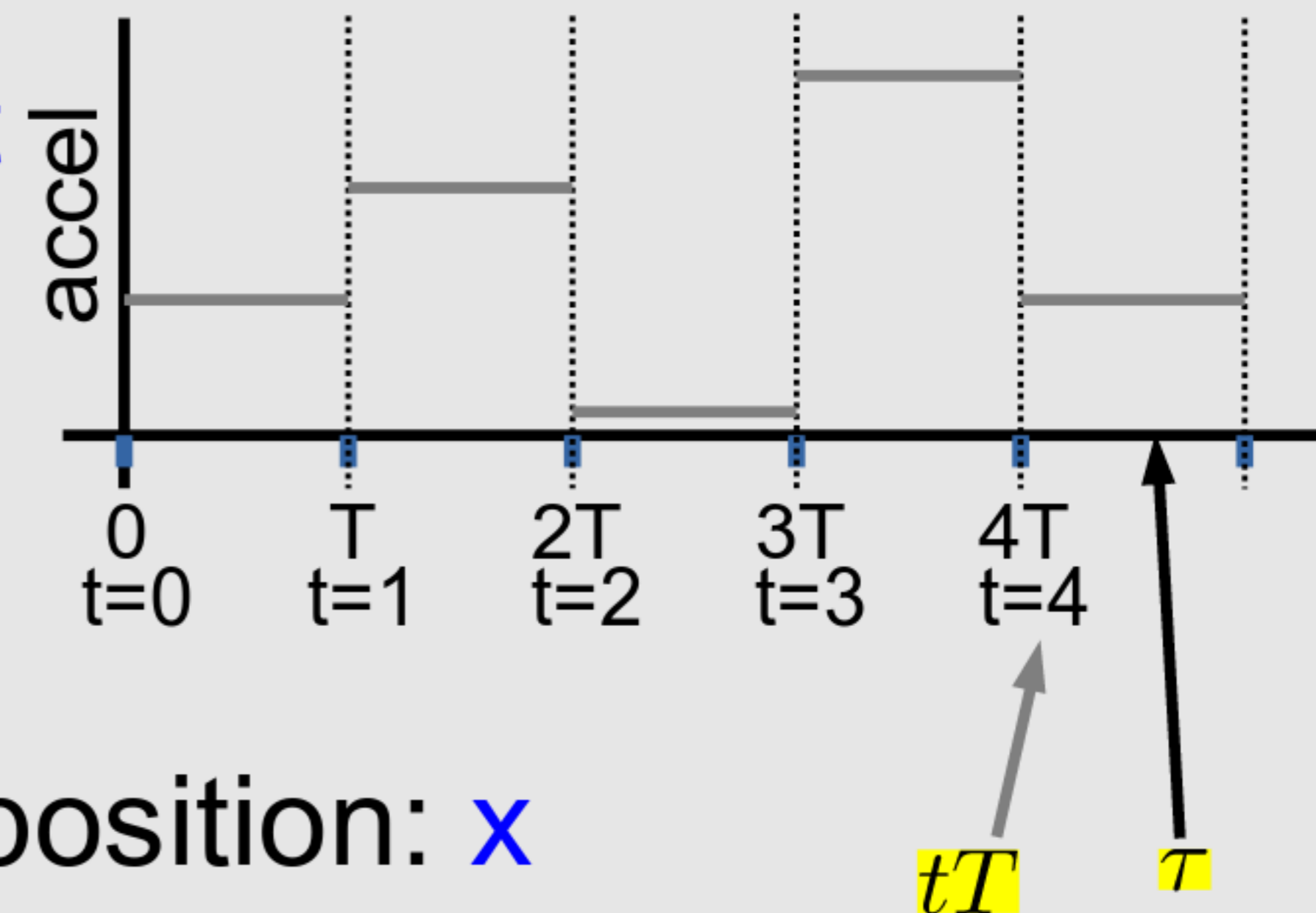
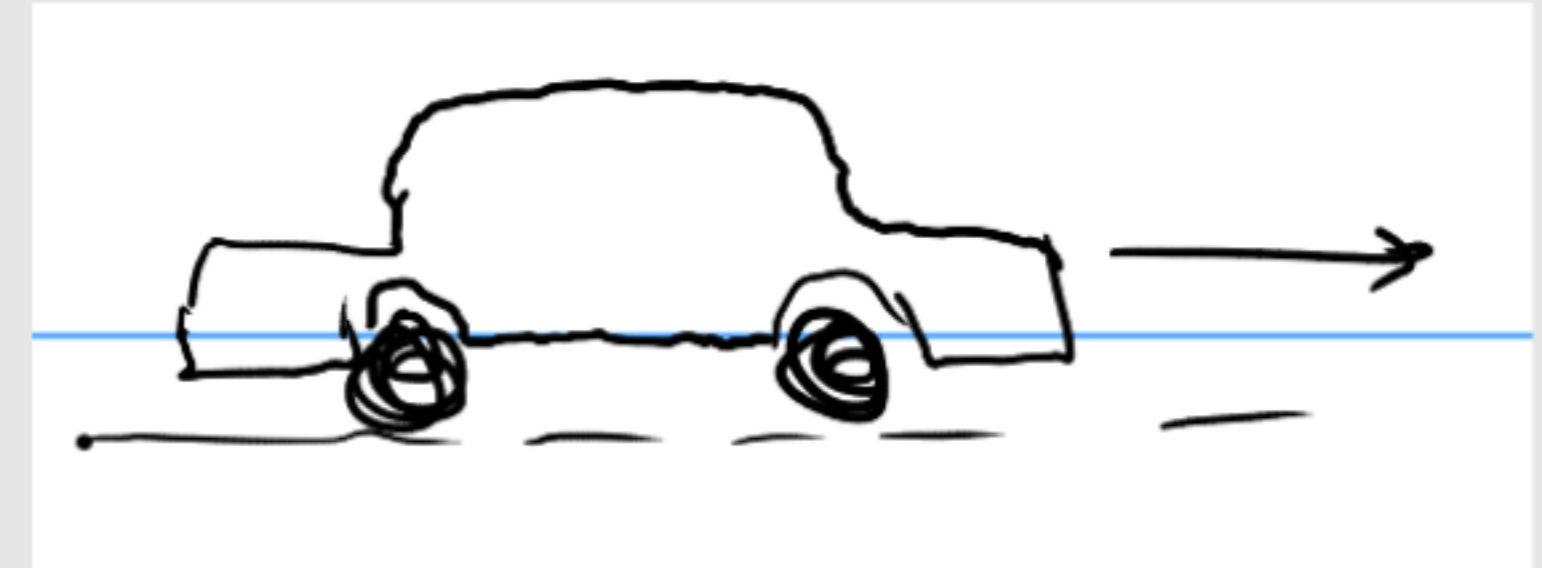
- $v(\tau) = \int_0^\tau a(\tau_2) d\tau_2, \quad x(\tau) = \int_0^\tau v(\tau_2) d\tau_2$

- $v(\tau) - v(tT) = \int_{tT}^\tau a(\tau_2) d\tau_2$

$$tT \leq \tau \leq (t+1)T$$

Example: Accelerating Car

- control input: **acceleration**
- can change only every T secs
→ stays constant in between
- Q: can we set its **position AND velocity** to whatever we want (at time = multiples of T)?
- analysis approach
 - find a discrete SSR for position/vel.
 - analyse its controllability



- acceleration: **a**; velocity: **v**; position: **x**

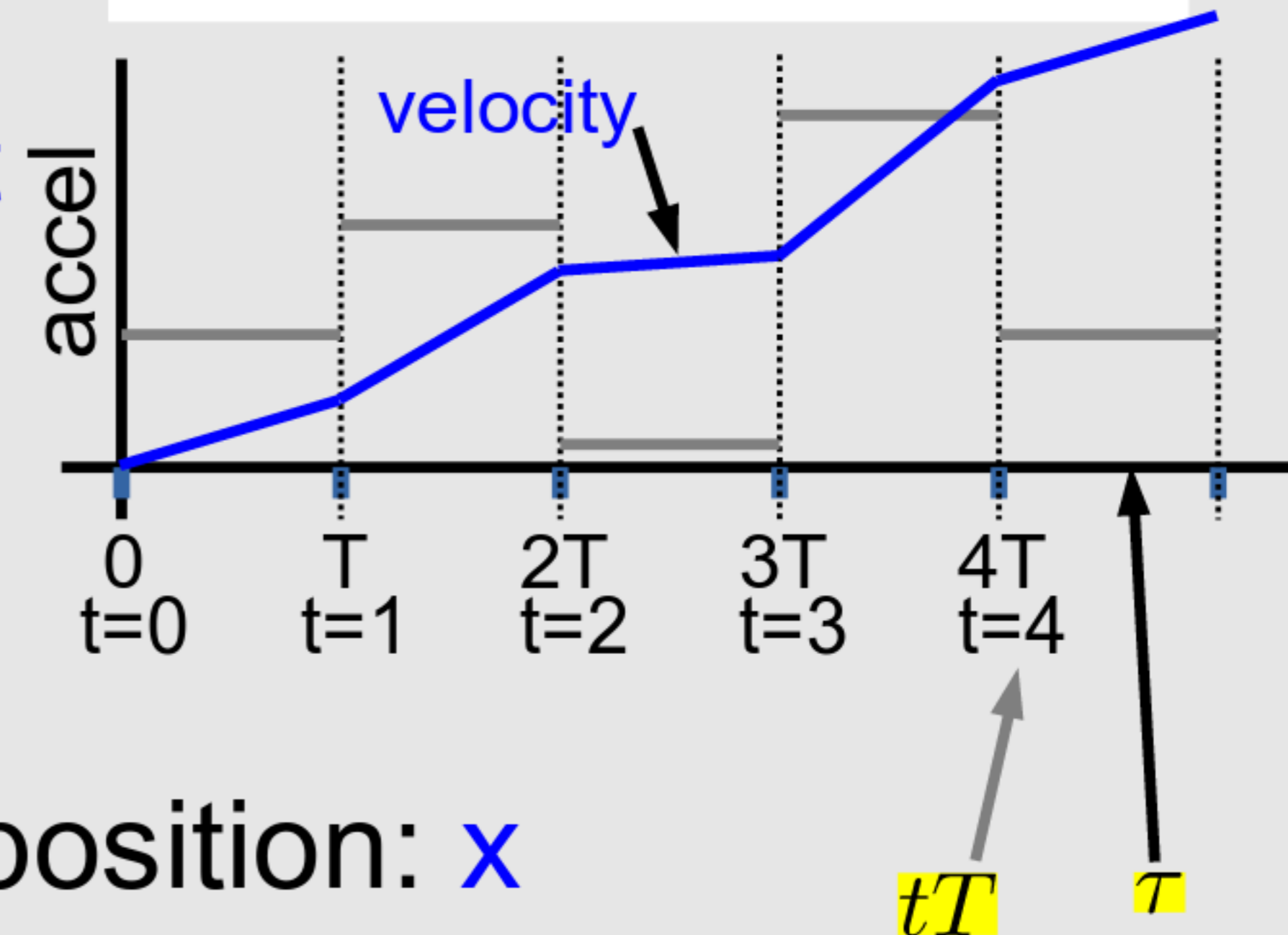
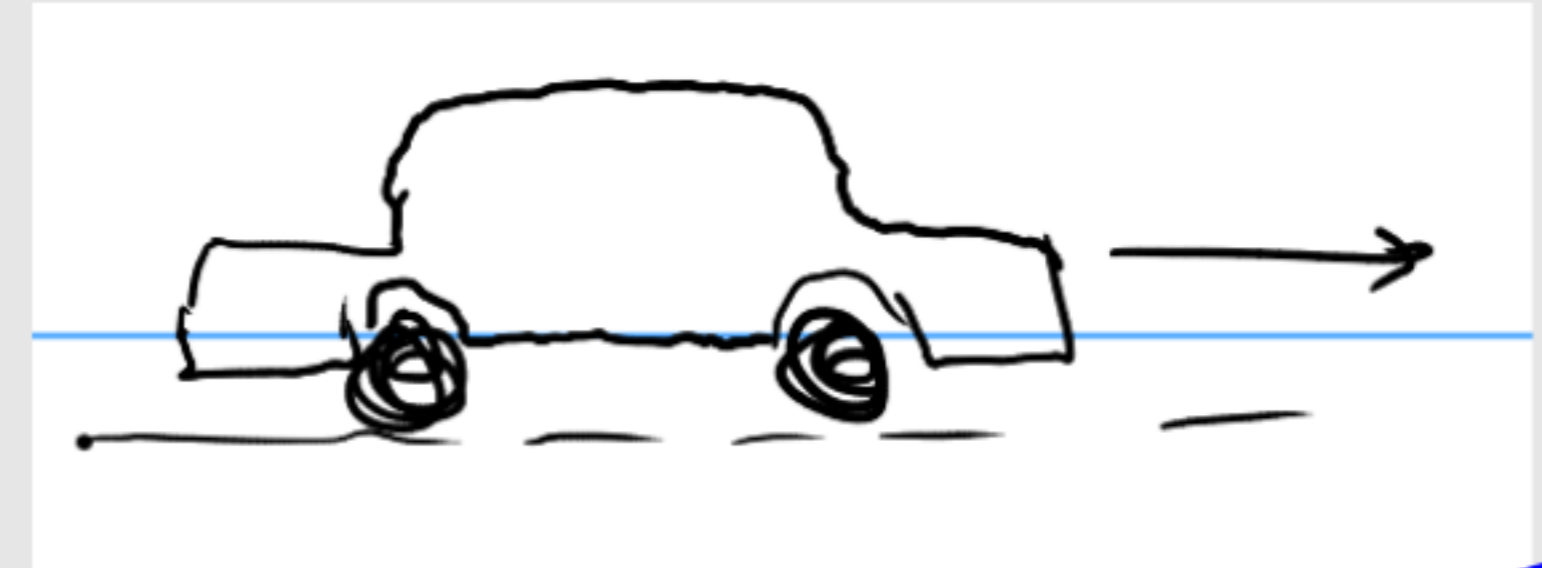
$$\bullet v(\tau) = \int_0^\tau a(\tau_2) d\tau_2, \quad x(\tau) = \int_0^\tau v(\tau_2) d\tau_2$$

$$\bullet v(\tau) - v(tT) = \int_{tT}^\tau a(\tau_2) d\tau_2 = a(tT) \int_{tT}^\tau d\tau_2 = (\tau - tT)a(tT)$$

$tT \leq \tau \leq (t+1)T$

Example: Accelerating Car

- control input: **acceleration**
- can change only every T secs
→ stays constant in between
- Q: can we set its **position AND velocity** to whatever we want (at time = multiples of T)?
- analysis approach
 - find a discrete SSR for position/vel.
 - analyse its controllability



- acceleration: **a**; velocity: **v**; position: **x**

$$\bullet v(\tau) = \int_0^\tau a(\tau_2) d\tau_2, \quad x(\tau) = \int_0^\tau v(\tau_2) d\tau_2$$

$$\bullet v(\tau) - v(tT) = \int_{tT}^\tau a(\tau_2) d\tau_2 = a(tT) \int_{tT}^\tau d\tau_2 = (\tau - tT)a(tT)$$

$tT \leq \tau \leq (t+1)T$

Accelerating car (contd.)

- acceleration: \mathbf{a} ; velocity: \mathbf{v} ; position: \mathbf{x}

- $v(\tau) = \int_0^\tau a(\tau_2) d\tau_2, \quad x(\tau) = \int_0^\tau v(\tau_2) d\tau_2$

- $v(\tau) - v(tT) = \int_{tT}^\tau a(\tau_2) d\tau_2 = a(tT) \int_{tT}^\tau d\tau_2 = (\tau - tT)a(tT)$
 $tT \leq \tau \leq (t+1)T$

Accelerating car (contd.)

- acceleration: \mathbf{a} ; velocity: \mathbf{v} ; position: \mathbf{x}

- $v(\tau) = \int_0^\tau a(\tau_2) d\tau_2, \quad x(\tau) = \int_0^\tau v(\tau_2) d\tau_2$

- $v(\tau) - v(tT) = \int_{tT}^\tau a(\tau_2) d\tau_2 = a(tT) \int_{tT}^\tau d\tau_2 = (\tau - tT)a(tT)$
 $tT \leq \tau \leq (t+1)T$

- $x(\tau) - x(tT) = \int_{tT}^\tau v(\tau_2) d\tau_2 = \int_{tT}^\tau [v(tT) + a(tT)(\tau_2 - tT)] d\tau_2$

$$tT \leq \tau \leq (t+1)T$$

Accelerating car (contd.)

- acceleration: \mathbf{a} ; velocity: \mathbf{v} ; position: \mathbf{x}

- $v(\tau) = \int_0^\tau a(\tau_2) d\tau_2, \quad x(\tau) = \int_0^\tau v(\tau_2) d\tau_2$

- $v(\tau) - v(tT) = \int_{tT}^\tau a(\tau_2) d\tau_2 = a(tT) \int_{tT}^\tau d\tau_2 = (\tau - tT)a(tT)$
 $tT \leq \tau \leq (t+1)T$

- $x(\tau) - x(tT) = \int_{tT}^\tau v(\tau_2) d\tau_2 = \int_{tT}^\tau [v(tT) + a(tT)(\tau_2 - tT)] d\tau_2$
 $= (\tau - tT)v(tT) + a(tT) \frac{(\tau - tT)^2}{2} \quad tT \leq \tau \leq (t+1)T$

Accelerating car (contd.)

- acceleration: \mathbf{a} ; velocity: \mathbf{v} ; position: \mathbf{x}

- $v(\tau) = \int_0^\tau a(\tau_2) d\tau_2, \quad x(\tau) = \int_0^\tau v(\tau_2) d\tau_2$

- $v(\tau) - v(tT) = \int_{tT}^\tau a(\tau_2) d\tau_2 = a(tT) \int_{tT}^\tau d\tau_2 = (\tau - tT)a(tT)$
 $tT \leq \tau \leq (t+1)T$

- $x(\tau) - x(tT) = \int_{tT}^\tau v(\tau_2) d\tau_2 = \int_{tT}^\tau [v(tT) + a(tT)(\tau_2 - tT)] d\tau_2$
 $= (\tau - tT)v(tT) + a(tT) \frac{(\tau - tT)^2}{2}$ $tT \leq \tau \leq (t+1)T$

- set $\tau = (t+1)T$; the above become:

- $x((t+1)T) = x(tT) + Tv(tT) + \frac{T^2 a(tT)}{2}$
 $v((t+1)T) = v(tT) + Ta(tT)$

Accelerating car (contd. - 2)

- $$x((t + 1)T) = x(tT) + Tv(tT) + \frac{T^2 a(tT)}{2}$$
$$v((t + 1)T) = v(tT) + Ta(tT)$$

Accelerating car (contd. - 2)

- $x((t + 1)T) = x(tT) + Tv(tT) + \frac{T^2 a(tT)}{2}$

$$v((t + 1)T) = v(tT) + Ta(tT)$$

- **S.S.R in matrix-vector form:**

- $$\begin{bmatrix} x((t + 1)T) \\ v((t + 1)T) \end{bmatrix} = \begin{bmatrix} 1 & T \\ 0 & 1 \end{bmatrix} \begin{bmatrix} x(tT) \\ v(tT) \end{bmatrix} + \begin{bmatrix} \frac{T^2}{2} \\ T \end{bmatrix} a(t)$$

Accelerating car (contd. - 2)

- $x((t + 1)T) = x(tT) + Tv(tT) + \frac{T^2 a(tT)}{2}$
 $v((t + 1)T) = v(tT) + Ta(tT)$
- **S.S.R in matrix-vector form:**
- $$\begin{bmatrix} x((t + 1)T) \\ v((t + 1)T) \end{bmatrix} = \begin{bmatrix} 1 & T \\ 0 & 1 \end{bmatrix} \begin{bmatrix} x(tT) \\ v(tT) \end{bmatrix} + \begin{bmatrix} \frac{T^2}{2} \\ T \end{bmatrix} a(t)$$
- **Controllability matrix:**
$$\begin{bmatrix} \vec{b} & | & A\vec{b} \end{bmatrix} = \begin{bmatrix} \frac{T^2}{2} & 3\frac{T^2}{2} \\ T & T \end{bmatrix}$$

Accelerating car (contd. - 2)

- $x((t + 1)T) = x(tT) + Tv(tT) + \frac{T^2 a(tT)}{2}$
 $v((t + 1)T) = v(tT) + Ta(tT)$
- **S.S.R in matrix-vector form:**
- $$\begin{bmatrix} x((t + 1)T) \\ v((t + 1)T) \end{bmatrix} = \begin{bmatrix} 1 & T \\ 0 & 1 \end{bmatrix} \begin{bmatrix} x(tT) \\ v(tT) \end{bmatrix} + \begin{bmatrix} \frac{T^2}{2} \\ T \end{bmatrix} a(t)$$
- **Controllability matrix:** $[\vec{b} \mid A\vec{b}] = \begin{bmatrix} \frac{T^2}{2} & 3\frac{T^2}{2} \\ T & T \end{bmatrix}$
- $$\det \left(\begin{bmatrix} \frac{T^2}{2} & 3\frac{T^2}{2} \\ T & T \end{bmatrix} \right) = \frac{T^3}{2} - 3\frac{T^3}{2} = -T^3$$

Accelerating car (contd. - 2)

- $x((t + 1)T) = x(tT) + Tv(tT) + \frac{T^2 a(tT)}{2}$
 $v((t + 1)T) = v(tT) + Ta(tT)$
- **S.S.R in matrix-vector form:**
- $$\begin{bmatrix} x((t + 1)T) \\ v((t + 1)T) \end{bmatrix} = \begin{bmatrix} 1 & T \\ 0 & 1 \end{bmatrix} \begin{bmatrix} x(tT) \\ v(tT) \end{bmatrix} + \begin{bmatrix} \frac{T^2}{2} \\ T \end{bmatrix} a(t)$$
- **Controllability matrix:** $[\vec{b} \mid A\vec{b}] = \begin{bmatrix} \frac{T^2}{2} & 3\frac{T^2}{2} \\ T & T \end{bmatrix}$
- $\det \left(\begin{bmatrix} \frac{T^2}{2} & 3\frac{T^2}{2} \\ T & T \end{bmatrix} \right) = \frac{T^3}{2} - 3\frac{T^3}{2} = -T^3 \leftarrow \text{always nonzero (for } T \neq 0)$

Accelerating car (contd. - 2)

- $x((t + 1)T) = x(tT) + Tv(tT) + \frac{T^2 a(tT)}{2}$
 $v((t + 1)T) = v(tT) + Ta(tT)$
- **S.S.R in matrix-vector form:**
- $$\begin{bmatrix} x((t + 1)T) \\ v((t + 1)T) \end{bmatrix} = \begin{bmatrix} 1 & T \\ 0 & 1 \end{bmatrix} \begin{bmatrix} x(tT) \\ v(tT) \end{bmatrix} + \begin{bmatrix} \frac{T^2}{2} \\ T \end{bmatrix} a(t)$$
- **Controllability matrix:** $[\vec{b} \mid A\vec{b}] = \begin{bmatrix} \frac{T^2}{2} & 3\frac{T^2}{2} \\ T & T \end{bmatrix}$
- $\det \left(\begin{bmatrix} \frac{T^2}{2} & 3\frac{T^2}{2} \\ T & T \end{bmatrix} \right) = \frac{T^3}{2} - 3\frac{T^3}{2} = -T^3 \leftarrow \text{always nonzero (for } T \neq 0)$
- **A: YES, we can drive the car's position AND velocity to whatever values we want (at every $\tau = tT$ for $t \geq 2$)**

Continuous Time Controllability

- **System:** $\frac{d}{dt}\Delta\vec{x}(t) = A\Delta\vec{x}(t) + B\Delta\vec{u}(t)$
 $\xrightarrow{\text{nxn matrix}} \quad \xrightarrow{\text{nxm matrix}}$

Continuous Time Controllability

- **System:** $\frac{d}{dt} \Delta \vec{x}(t) = \overset{\text{nxn matrix}}{A} \Delta \vec{x}(t) + \overset{\text{nxm matrix}}{B} \Delta \vec{u}(t)$
- **Controllability: same condition as for discrete**

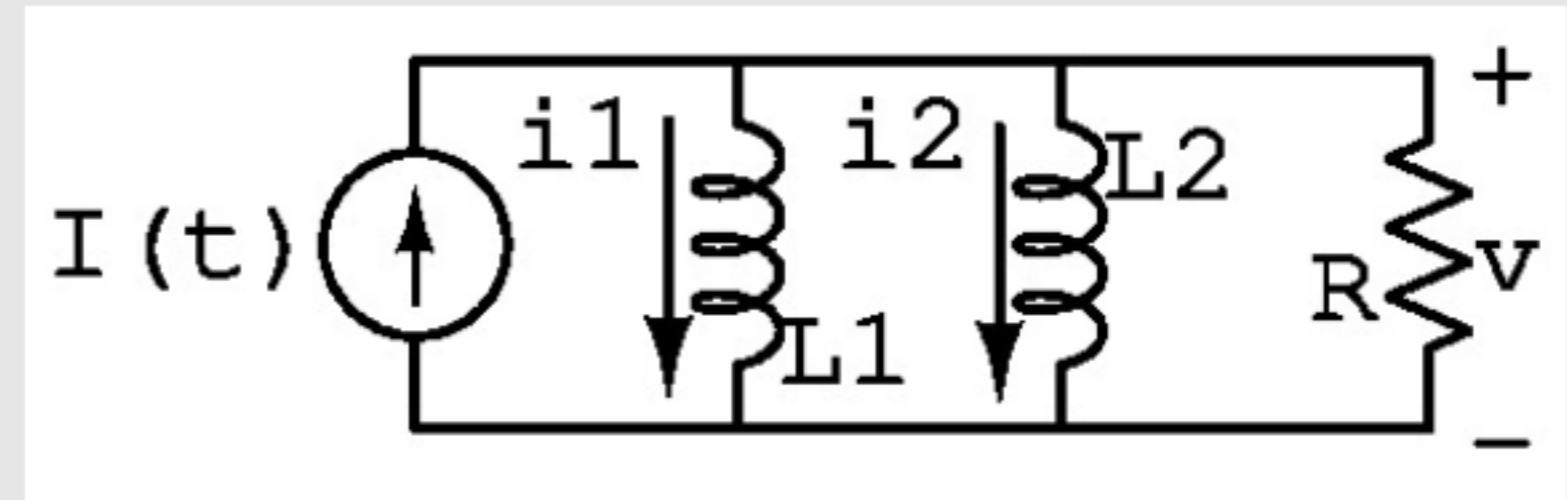
$$\text{rank} \left([B \mid AB \mid \cdots \mid A^{t-1}B] \right) = n$$

Continuous Time Controllability

- **System:** $\frac{d}{dt} \Delta \vec{x}(t) = \overset{\text{nxn matrix}}{A} \Delta \vec{x}(t) + \overset{\text{nxm matrix}}{B} \Delta \vec{u}(t)$
- **Controllability: same condition as for discrete**

$$\text{rank} \left([B \mid AB \mid \cdots \mid A^{t-1}B] \right) = n$$

- **Example: RL circuit**

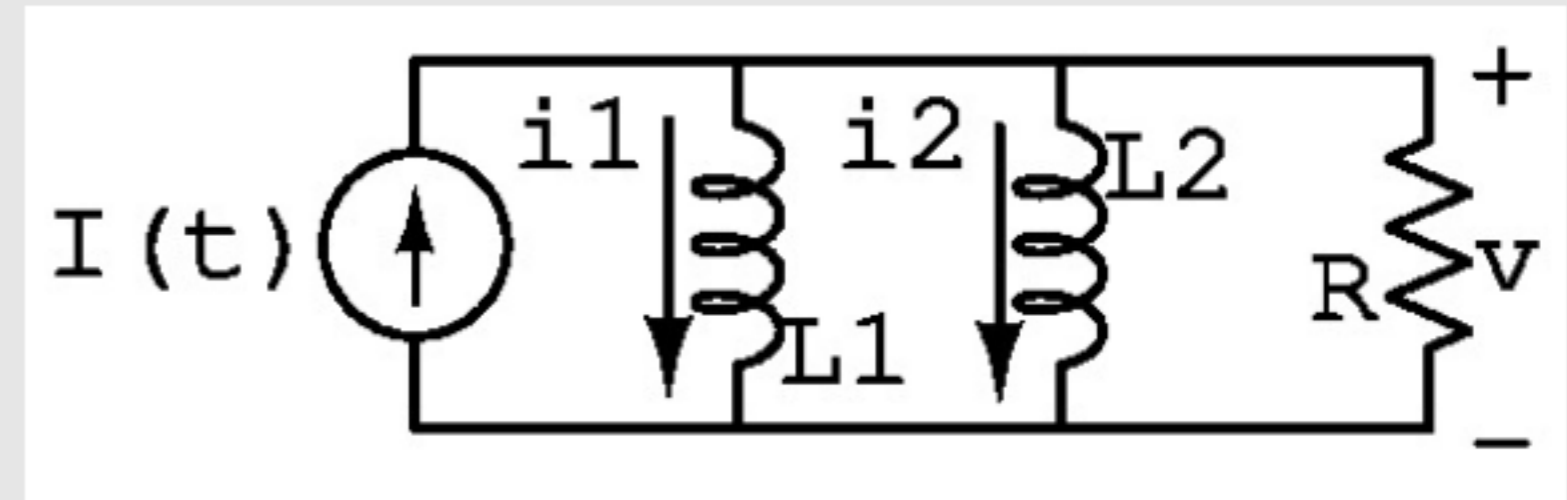


Continuous Time Controllability

- **System:** $\frac{d}{dt} \Delta \vec{x}(t) = \overset{\text{nxn matrix}}{A} \Delta \vec{x}(t) + \overset{\text{nxm matrix}}{B} \Delta \vec{u}(t)$
- **Controllability: same condition as for discrete**

$$\text{rank} \left([B \mid AB \mid \cdots \mid A^{t-1}B] \right) = n$$

- **Example: RL circuit**



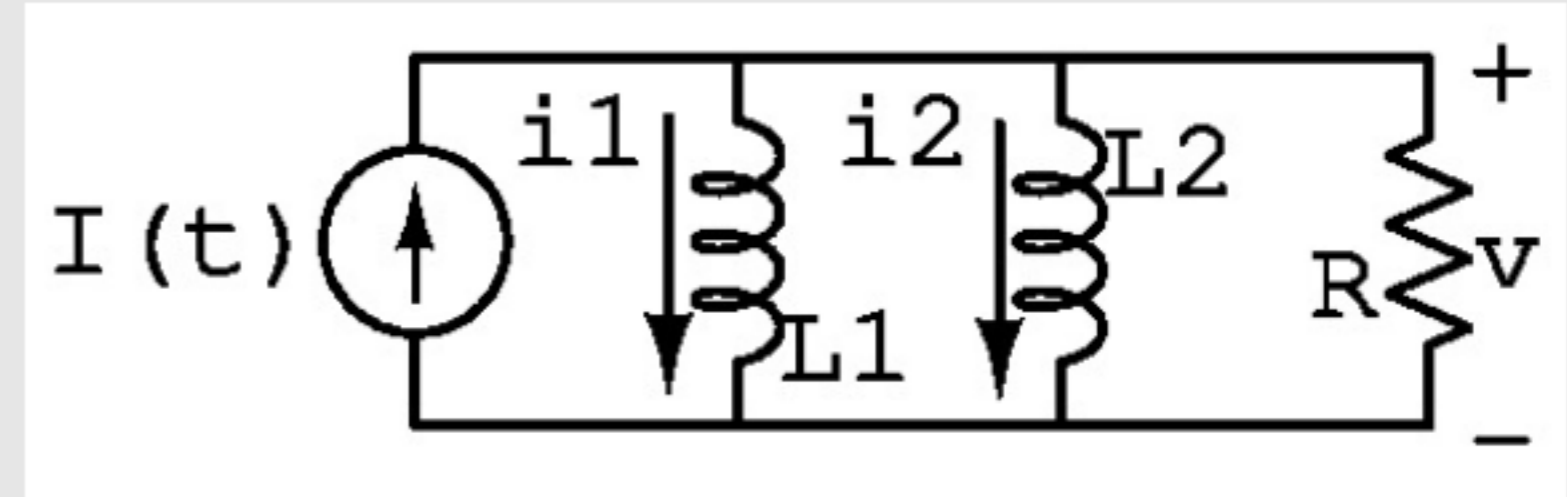
$$i_1 + i_2 + \frac{v}{R} = I_1(t), \quad \frac{di_1}{dt} = \frac{v}{L_1}, \quad \frac{di_2}{dt} = \frac{v}{L_2}$$

Continuous Time Controllability

- **System:** $\frac{d}{dt} \Delta \vec{x}(t) = \overset{\text{nxn matrix}}{A} \Delta \vec{x}(t) + \overset{\text{nxm matrix}}{B} \Delta \vec{u}(t)$
- **Controllability: same condition as for discrete**

$$\text{rank} \left([B \mid AB \mid \dots \mid A^{t-1}B] \right) = n$$

- **Example: RL circuit**



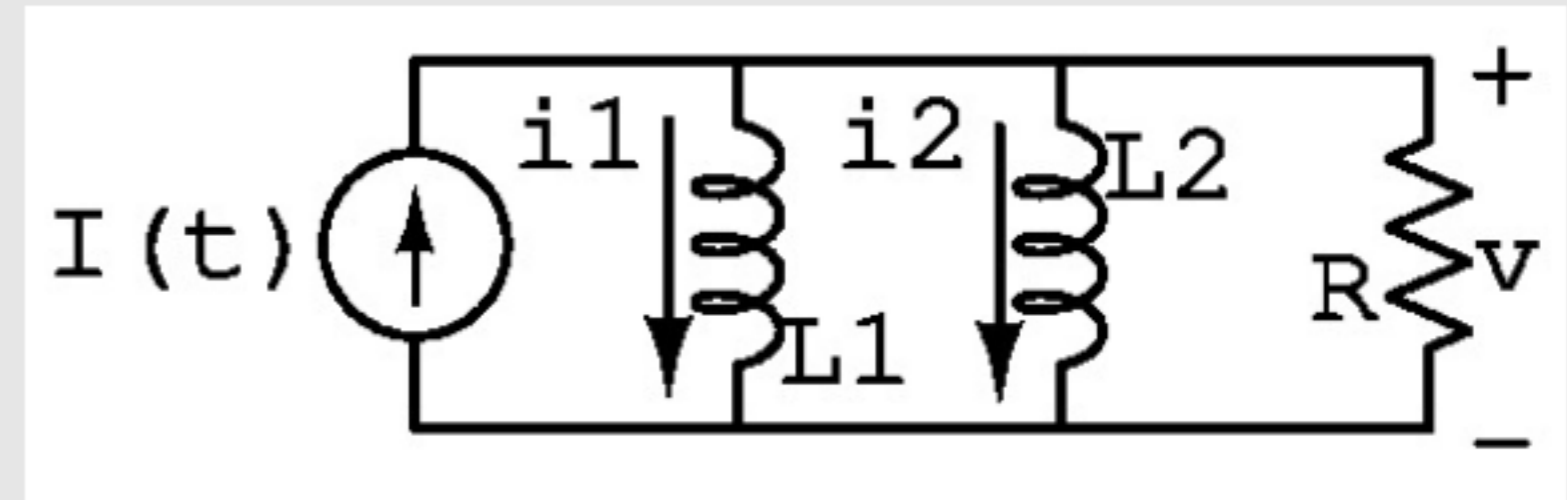
- $i_1 + i_2 + \frac{v}{R} = I_1(t), \quad \frac{di_1}{dt} = \frac{v}{L_1}, \quad \frac{di_2}{dt} = \frac{v}{L_2}$
- $\frac{di_1}{dt} = \frac{R(I_1(t) - i_1(t) - i_2(t))}{L_1}, \quad \frac{di_2}{dt} = \frac{R(I_1(t) - i_1(t) - i_2(t))}{L_2}$

Continuous Time Controllability

- **System:** $\frac{d}{dt} \Delta \vec{x}(t) = \overset{\text{nxn matrix}}{A} \Delta \vec{x}(t) + \overset{\text{nxm matrix}}{B} \Delta \vec{u}(t)$
- **Controllability: same condition as for discrete**

$$\text{rank} \left([B \mid AB \mid \dots \mid A^{t-1}B] \right) = n$$

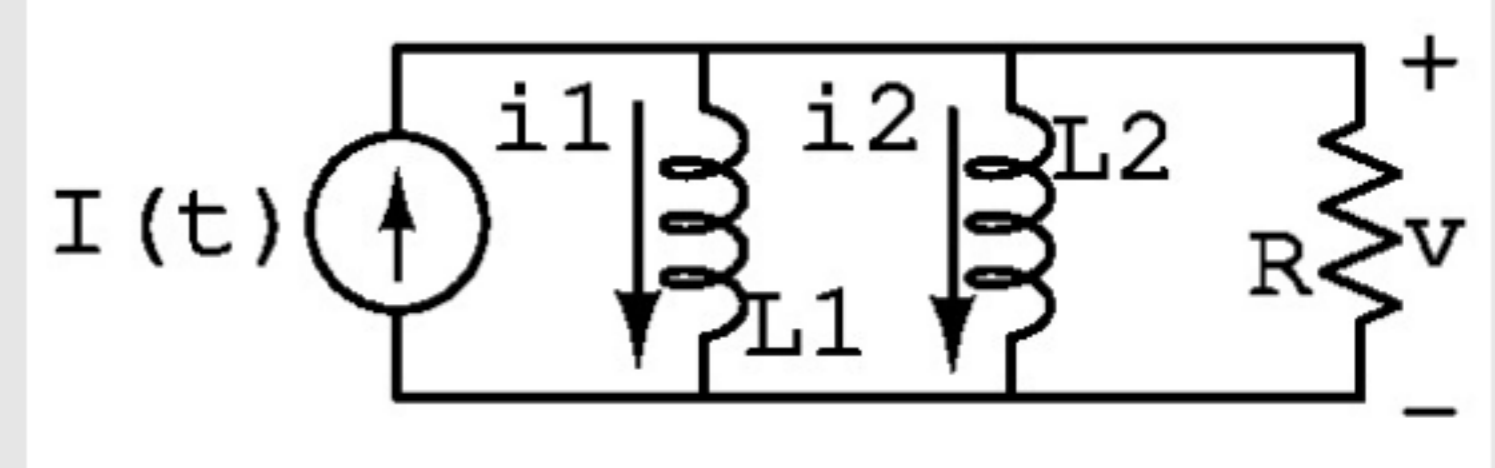
- **Example: RL circuit**



- $i_1 + i_2 + \frac{v}{R} = I_1(t), \quad \frac{di_1}{dt} = \frac{v}{L_1}, \quad \frac{di_2}{dt} = \frac{v}{L_2}$
- $\frac{di_1}{dt} = \frac{R(I_1(t) - i_1(t) - i_2(t))}{L_1}, \quad \frac{di_2}{dt} = \frac{R(I_1(t) - i_1(t) - i_2(t))}{L_2}$
- $\frac{d}{dt} \begin{bmatrix} i_1(t) \\ i_2(t) \end{bmatrix} = \begin{bmatrix} -\frac{R}{L_1} & -\frac{R}{L_1} \\ -\frac{R}{L_2} & -\frac{R}{L_2} \end{bmatrix} \begin{bmatrix} i_1(t) \\ i_2(t) \end{bmatrix} + \begin{bmatrix} \frac{R}{L_1} \\ \frac{R}{L_2} \end{bmatrix} I(t)$

Continuous Controllability (contd.)

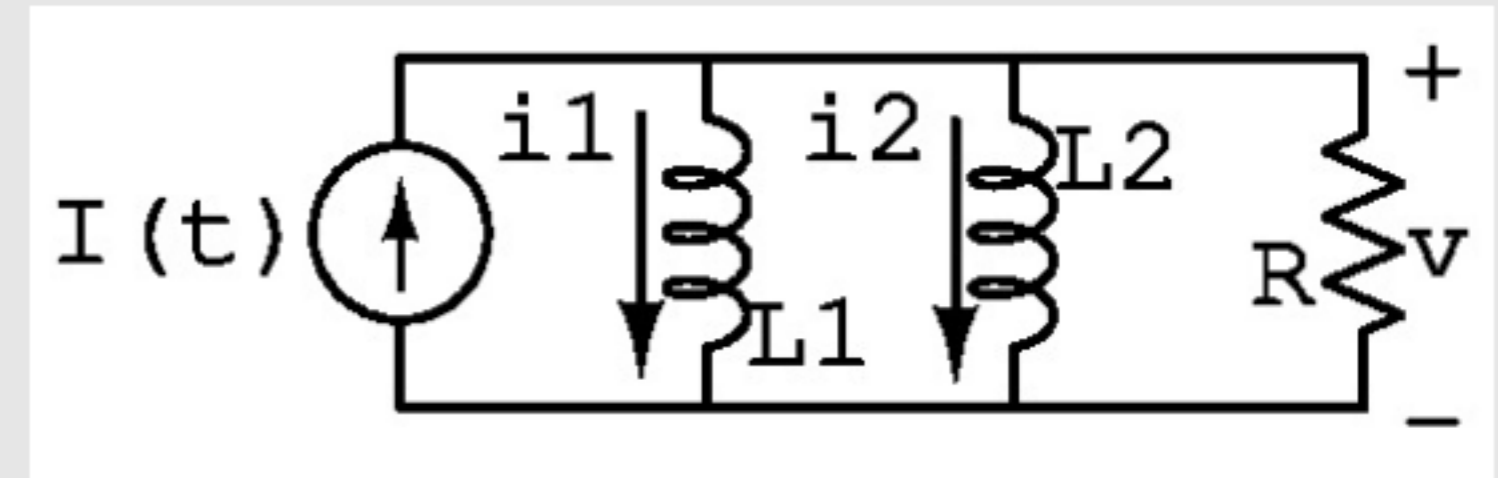
- $$\frac{d}{dt} \begin{bmatrix} i_1(t) \\ i_2(t) \end{bmatrix} = \begin{bmatrix} -\frac{R}{L_1} & -\frac{R}{L_1} \\ -\frac{R}{L_2} & -\frac{R}{L_2} \end{bmatrix} \begin{bmatrix} i_1(t) \\ i_2(t) \end{bmatrix} + \begin{bmatrix} \frac{R}{L_1} \\ \frac{R}{L_2} \end{bmatrix} I(t)$$



Continuous Controllability (contd.)

- $$\frac{d}{dt} \begin{bmatrix} i_1(t) \\ i_2(t) \end{bmatrix} = \begin{bmatrix} -\frac{R}{L_1} & -\frac{R}{L_1} \\ -\frac{R}{L_2} & -\frac{R}{L_2} \end{bmatrix} \begin{bmatrix} i_1(t) \\ i_2(t) \end{bmatrix} + \begin{bmatrix} \frac{R}{L_1} \\ \frac{R}{L_2} \end{bmatrix} I(t)$$

- Controllability matrix:

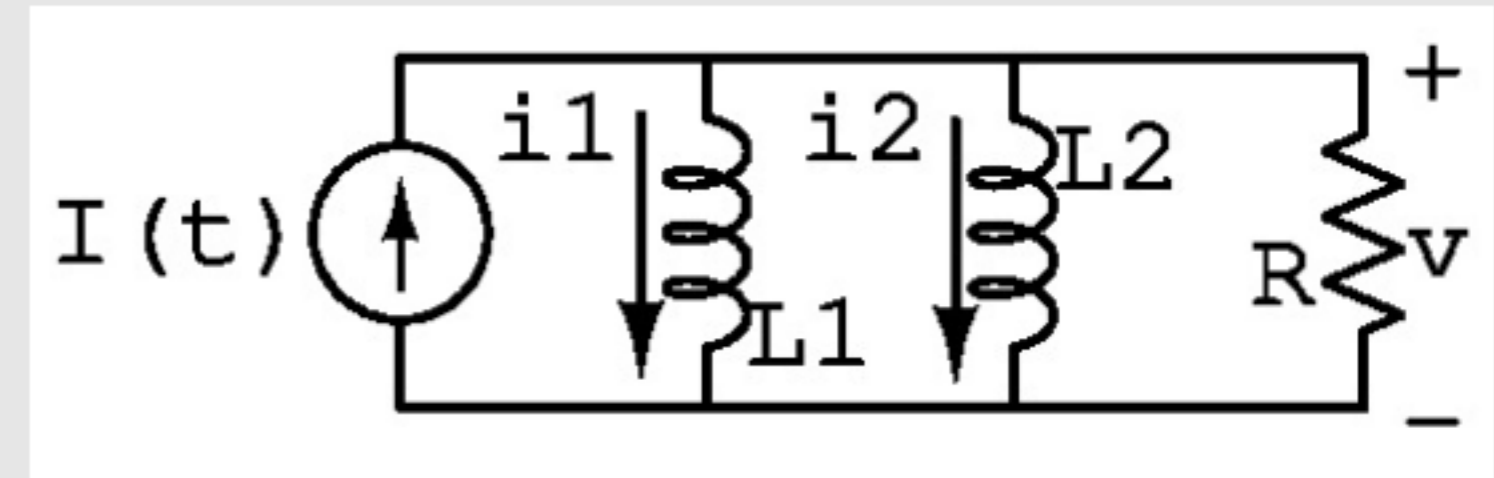


- $$\begin{bmatrix} \vec{b} & | & A\vec{b} \end{bmatrix} = \begin{bmatrix} \frac{R}{L_1} & -\frac{R^2}{L_1^2} & -\frac{R^2}{L_1 L_2} \\ \frac{R}{L_2} & -\frac{R^2}{L_1 L_2} & -\frac{R^2}{L_2^2} \end{bmatrix} = \begin{bmatrix} 1 & -\frac{R}{L_1} & -\frac{R}{L_2} \\ 1 & -\frac{R}{L_1} & -\frac{R}{L_2} \end{bmatrix} \begin{bmatrix} \frac{R}{L_1} & \\ 0 & \frac{R}{L_2} \end{bmatrix}$$

Continuous Controllability (contd.)

- $$\frac{d}{dt} \begin{bmatrix} i_1(t) \\ i_2(t) \end{bmatrix} = \begin{bmatrix} -\frac{R}{L_1} & -\frac{R}{L_1} \\ -\frac{R}{L_2} & -\frac{R}{L_2} \end{bmatrix} \begin{bmatrix} i_1(t) \\ i_2(t) \end{bmatrix} + \begin{bmatrix} \frac{R}{L_1} \\ \frac{R}{L_2} \end{bmatrix} I(t)$$

- Controllability matrix:



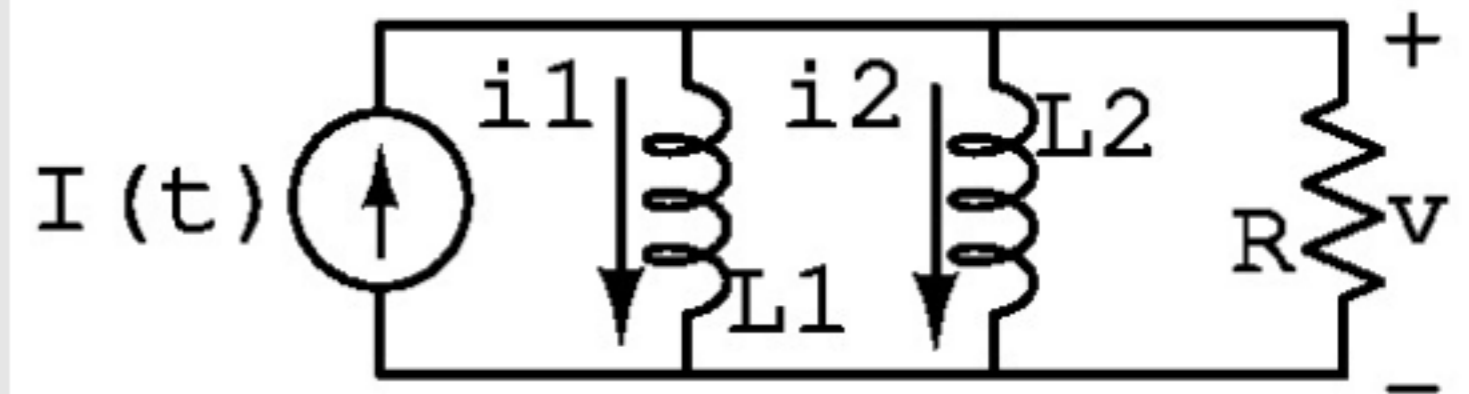
- $$\begin{bmatrix} \vec{b} & | & A\vec{b} \end{bmatrix} = \begin{bmatrix} \frac{R}{L_1} & -\frac{R^2}{L_1^2} & -\frac{R^2}{L_1 L_2} \\ \frac{R}{L_2} & -\frac{R^2}{L_1 L_2} & -\frac{R^2}{L_2^2} \end{bmatrix} = \begin{bmatrix} 1 & -\frac{R}{L_1} & -\frac{R}{L_2} \\ 1 & -\frac{R}{L_1} & -\frac{R}{L_2} \end{bmatrix} \begin{bmatrix} \frac{R}{L_1} & \\ 0 & \frac{R}{L_2} \end{bmatrix}$$

rank = 1 < n=2

Continuous Controllability (contd.)

- $$\frac{d}{dt} \begin{bmatrix} i_1(t) \\ i_2(t) \end{bmatrix} = \begin{bmatrix} -\frac{R}{L_1} & -\frac{R}{L_1} \\ -\frac{R}{L_2} & -\frac{R}{L_2} \end{bmatrix} \begin{bmatrix} i_1(t) \\ i_2(t) \end{bmatrix} + \begin{bmatrix} \frac{R}{L_1} \\ \frac{R}{L_2} \end{bmatrix} I(t)$$

not controllable



- Controllability matrix:

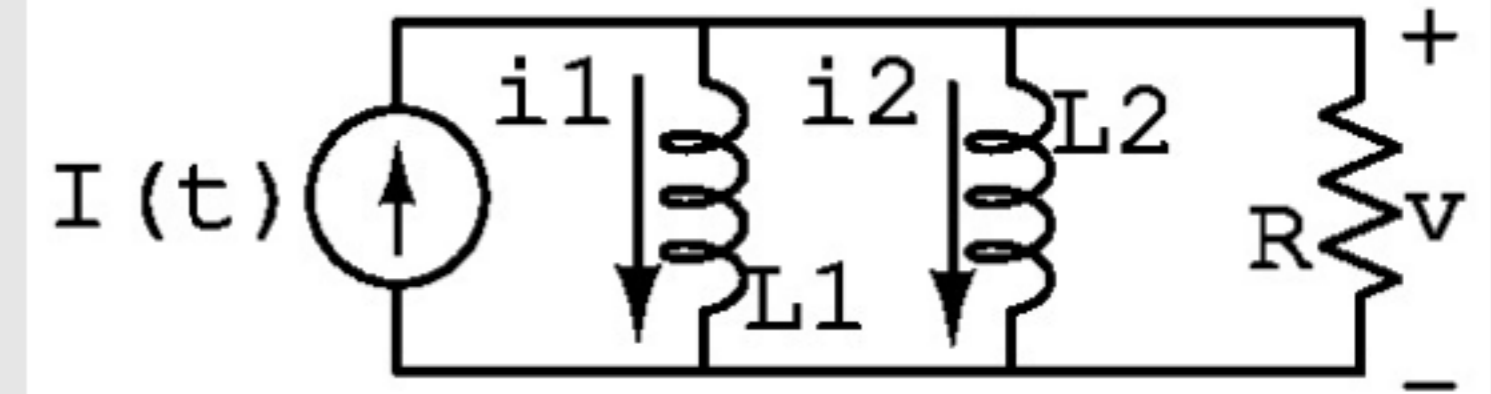
- $$\begin{bmatrix} \vec{b} & | & A\vec{b} \end{bmatrix} = \begin{bmatrix} \frac{R}{L_1} & -\frac{R^2}{L_1^2} & -\frac{R^2}{L_1 L_2} \\ \frac{R}{L_2} & -\frac{R^2}{L_1 L_2} & -\frac{R^2}{L_2^2} \end{bmatrix} = \begin{bmatrix} 1 & -\frac{R}{L_1} & -\frac{R}{L_2} \\ 1 & -\frac{R}{L_1} & -\frac{R}{L_2} \end{bmatrix} \begin{bmatrix} \frac{R}{L_1} & \\ 0 & \frac{R}{L_2} \end{bmatrix}$$

rank = 1 < n=2

Continuous Controllability (contd.)

- $$\frac{d}{dt} \begin{bmatrix} i_1(t) \\ i_2(t) \end{bmatrix} = \begin{bmatrix} -\frac{R}{L_1} & -\frac{R}{L_1} \\ -\frac{R}{L_2} & -\frac{R}{L_2} \end{bmatrix} \begin{bmatrix} i_1(t) \\ i_2(t) \end{bmatrix} + \begin{bmatrix} \frac{R}{L_1} \\ \frac{R}{L_2} \end{bmatrix} I(t)$$

not controllable



- Controllability matrix:

- $$\begin{bmatrix} \vec{b} & | & A\vec{b} \end{bmatrix} = \begin{bmatrix} \frac{R}{L_1} & -\frac{R^2}{L_1^2} & -\frac{R^2}{L_1 L_2} \\ \frac{R}{L_2} & -\frac{R^2}{L_1 L_2} & -\frac{R^2}{L_2^2} \end{bmatrix} = \begin{bmatrix} 1 & -\frac{R}{L_1} & -\frac{R}{L_2} \\ 1 & -\frac{R}{L_1} & -\frac{R}{L_2} \end{bmatrix} \begin{bmatrix} \frac{R}{L_1} & \\ 0 & \frac{R}{L_2} \end{bmatrix}$$

rank = 1 < n=2

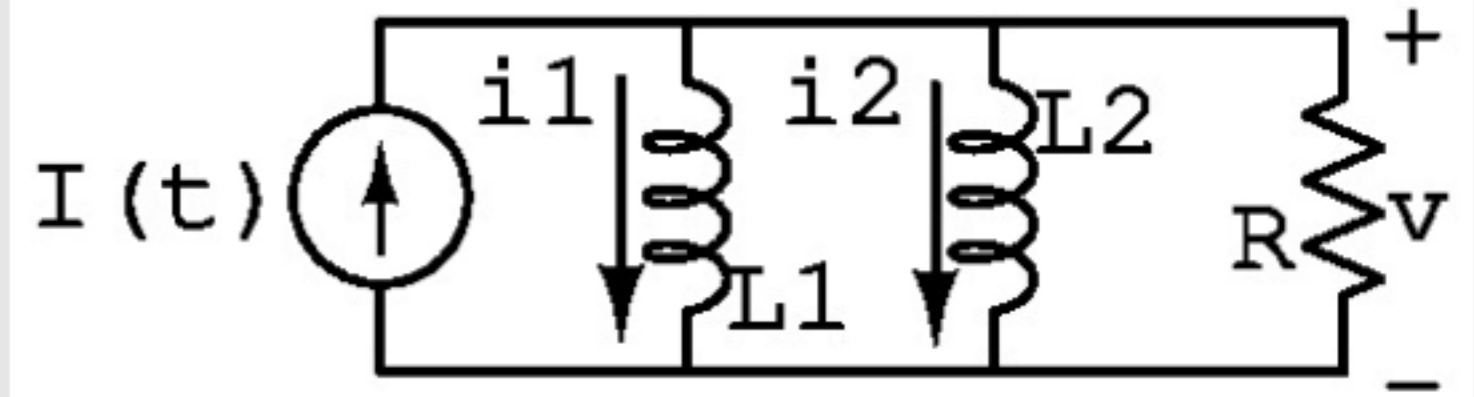
- Intuitive/“physical” way to see it:

- i_1 and i_2 both **directly determined** by the same $v(t)$

Continuous Controllability (contd.)

- $$\frac{d}{dt} \begin{bmatrix} i_1(t) \\ i_2(t) \end{bmatrix} = \begin{bmatrix} -\frac{R}{L_1} & -\frac{R}{L_1} \\ -\frac{R}{L_2} & -\frac{R}{L_2} \end{bmatrix} \begin{bmatrix} i_1(t) \\ i_2(t) \end{bmatrix} + \begin{bmatrix} \frac{R}{L_1} \\ \frac{R}{L_2} \end{bmatrix} I(t)$$

not controllable



- Controllability matrix:

- $$\begin{bmatrix} \vec{b} & | & A\vec{b} \end{bmatrix} = \begin{bmatrix} \frac{R}{L_1} & -\frac{R^2}{L_1^2} & -\frac{R^2}{L_1 L_2} \\ \frac{R}{L_2} & -\frac{R^2}{L_1 L_2} & -\frac{R^2}{L_2^2} \end{bmatrix} = \begin{bmatrix} 1 & -\frac{R}{L_1} & -\frac{R}{L_2} \\ 1 & -\frac{R}{L_1} & -\frac{R}{L_2} \end{bmatrix} \begin{bmatrix} \frac{R}{L_1} & \\ 0 & \frac{R}{L_2} \end{bmatrix}$$

rank = 1 < n=2

- Intuitive/“physical” way to see it:

- i_1 and i_2 both **directly determined** by the same $v(t)$

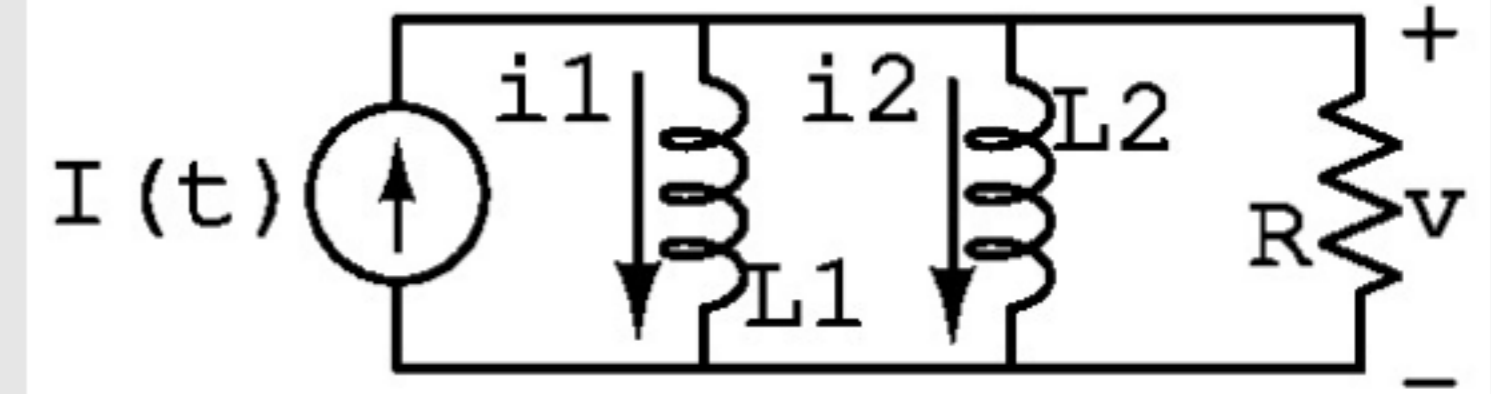
- $$\frac{di_1}{dt} = \frac{v}{L_1}, \quad \frac{di_2}{dt} = \frac{v}{L_2}$$

- $$\frac{d}{dt} (L_1 i_1(t) - L_2 i_2(t)) = 0 \rightarrow L_1 i_1(t) - L_2 i_2(t) = \text{constant}$$

Continuous Controllability (contd.)

- $$\frac{d}{dt} \begin{bmatrix} i_1(t) \\ i_2(t) \end{bmatrix} = \begin{bmatrix} -\frac{R}{L_1} & -\frac{R}{L_1} \\ -\frac{R}{L_2} & -\frac{R}{L_2} \end{bmatrix} \begin{bmatrix} i_1(t) \\ i_2(t) \end{bmatrix} + \begin{bmatrix} \frac{R}{L_1} \\ \frac{R}{L_2} \end{bmatrix} I(t)$$

not controllable



- Controllability matrix:

- $$\begin{bmatrix} \vec{b} & | & A\vec{b} \end{bmatrix} = \begin{bmatrix} \frac{R}{L_1} & -\frac{R^2}{L_1^2} & -\frac{R^2}{L_1 L_2} \\ \frac{R}{L_2} & -\frac{R^2}{L_1 L_2} & -\frac{R^2}{L_2^2} \end{bmatrix} = \begin{bmatrix} 1 & -\frac{R}{L_1} & -\frac{R}{L_2} \\ 1 & -\frac{R}{L_1} & -\frac{R}{L_2} \end{bmatrix} \begin{bmatrix} \frac{R}{L_1} & \\ 0 & \frac{R}{L_2} \end{bmatrix}$$

rank = 1 < n=2

- Intuitive/“physical” way to see it:

- i_1 and i_2 both **directly determined** by the same $v(t)$

- $$\frac{di_1}{dt} = \frac{v}{L_1}, \quad \frac{di_2}{dt} = \frac{v}{L_2}$$

cannot be set independently

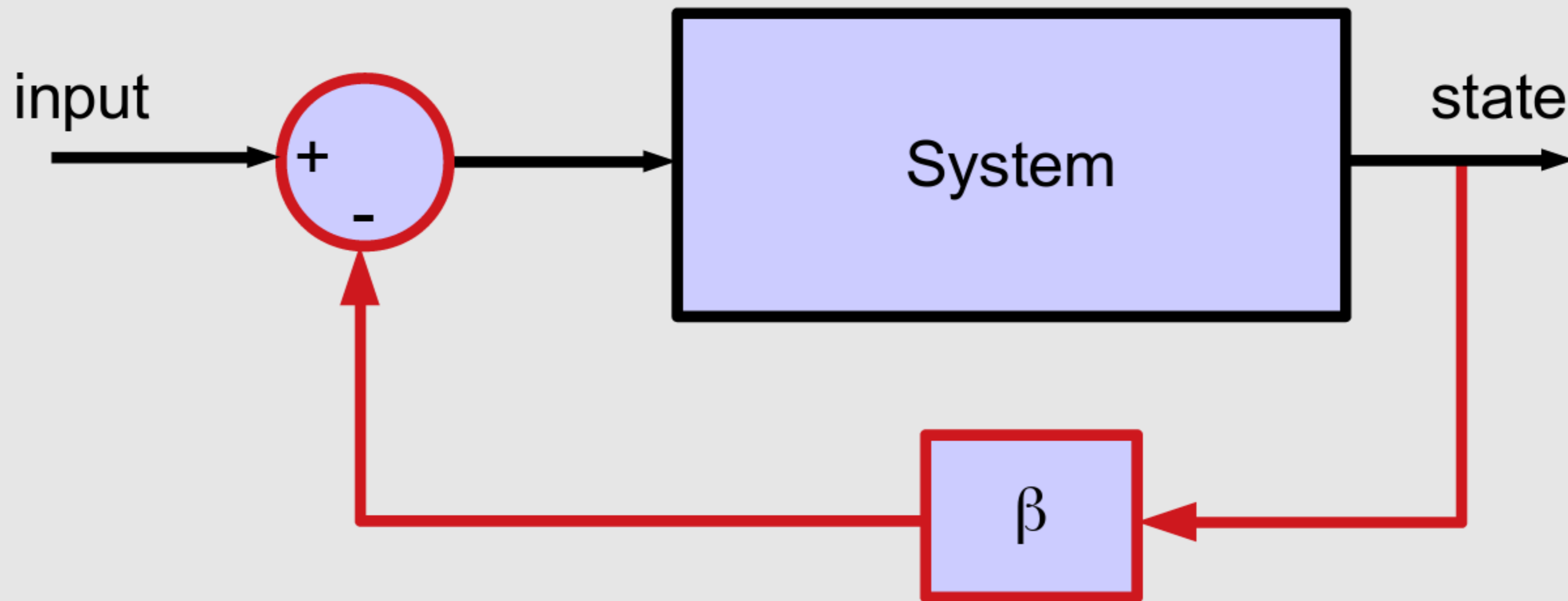
- $$\frac{d}{dt} (L_1 i_1(t) - L_2 i_2(t)) = 0 \rightarrow L_1 i_1(t) - L_2 i_2(t) = \text{constant}$$

Feedback

- The concept of **feedback**
 - add/subtract some of the output/state from the input

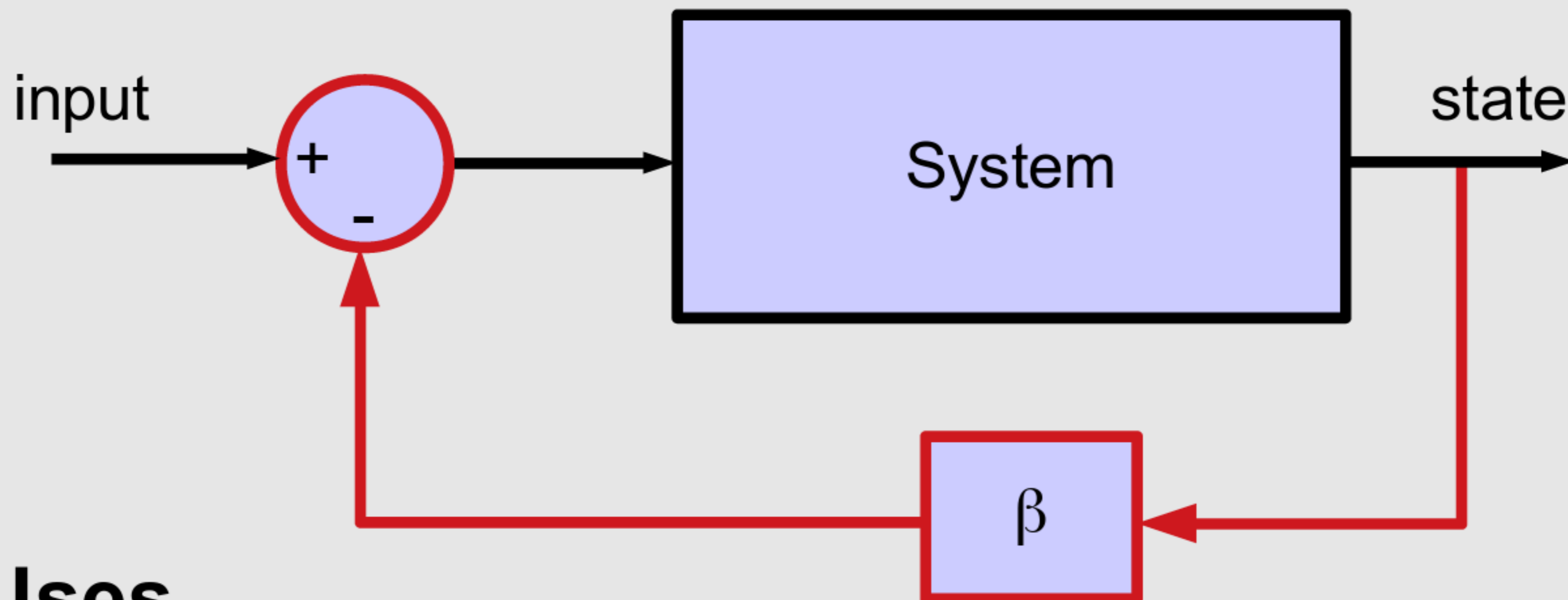
Feedback

- The concept of **feedback**
 - add/subtract some of the output/state from the input



Feedback

- The concept of **feedback**
 - add/subtract some of the output/state from the input



- **Uses**

- making systems **less sensitive** to undesired noise and uncertainties (ALWAYS PRESENT in practical systems)
- **stabilizing unstable systems** (if they are controllable)
 - thus making them practically usable

The Problem with Open Loop Control

- “open loop” means: no feedback
- “closed loop” means a system with feedback

The Problem with Open Loop Control

- “open loop” means: no feedback
- “closed loop” means a system with feedback
- example: $\dot{x}(t) = ax(t) + u(t), \quad a = 1 > 0$

* dropping Δ from Δx and Δu (for convenience)

The Problem with Open Loop Control

- “open loop” means: no feedback
- “closed loop” means a system with feedback
- example: $\dot{x}(t) = ax(t) + u(t)$, $a = 1 > 0$ ← **unstable**
* dropping Δ from Δx and Δu (for convenience)

The Problem with Open Loop Control

- “open loop” means: no feedback
- “closed loop” means a system with feedback
- example: $\dot{x}(t) = ax(t) + u(t)$, $a = 1 > 0$ ← unstable
- but controllable (why?) * dropping Δ from Δx and Δu (for convenience)

The Problem with Open Loop Control

- “open loop” means: no feedback
- “closed loop” means a system with feedback
- example: $\dot{x}(t) = ax(t) + u(t)$, $a = 1 > 0$ ← unstable
- but controllable (why?) * dropping Δ from Δx and Δu (for convenience)
- goal: make $x(t=10) = 1$, starting with I.C. $x(0) = 1$

The Problem with Open Loop Control

- “open loop” means: no feedback
- “closed loop” means a system with feedback
- example: $\dot{x}(t) = ax(t) + u(t)$, $a = 1 > 0$ ← **unstable**
- but controllable (why?) * dropping Δ from Δx and Δu (for convenience)
- goal: make $x(t=10) = 1$, starting with I.C. $x(0) = 1$
- $$x(10) = 1 \cdot e^{10} + \int_0^{10} e^{10-\tau} u(\tau) d\tau$$

The Problem with Open Loop Control

- “open loop” means: no feedback
- “closed loop” means a system with feedback
- example: $\dot{x}(t) = ax(t) + u(t)$, $a = 1 > 0$ ← **unstable**
- but controllable (why?) *** dropping Δ from Δx and Δu (for convenience)**
- goal: **make $x(t=10) = 1$, starting with I.C. $x(0) = 1$**
- $x(10) = 1 \cdot e^{10} + \int_0^{10} e^{10-\tau} u(\tau) d\tau$ **try $u(\tau) = \text{constant}$**

The Problem with Open Loop Control

- “open loop” means: no feedback
- “closed loop” means a system with feedback
- example: $\dot{x}(t) = ax(t) + u(t)$, $a = 1 > 0$ ← **unstable**
- but controllable (why?) *** dropping Δ from Δx and Δu (for convenience)**
- goal: **make $x(t=10) = 1$, starting with I.C. $x(0) = 1$**
- $x(10) = 1 \cdot e^{10} + \int_0^{10} e^{10-\tau} u(\tau) d\tau = 1 \cdot e^{10} + u(e^{10} - 1)$ **try $u(\tau) = \text{constant}$**

The Problem with Open Loop Control

- “open loop” means: no feedback
- “closed loop” means a system with feedback
- example: $\dot{x}(t) = ax(t) + u(t)$, $a = 1 > 0$ ← **unstable**
- but controllable (why?) *** dropping Δ from Δx and Δu (for convenience)**
- goal: **make $x(t=10) = 1$, starting with I.C. $x(0) = 1$**
- $x(10) = 1 \cdot e^{10} + \int_0^{10} e^{10-\tau} u(\tau) d\tau = 1 \cdot e^{10} + u(e^{10} - 1)$ **try $u(\tau) = \text{constant}$**
- want: $1 = x(10) = 1 \cdot e^{10} + u(e^{10} - 1)$

The Problem with Open Loop Control

- “open loop” means: no feedback
- “closed loop” means a system with feedback
- example: $\dot{x}(t) = ax(t) + u(t)$, $a = 1 > 0$ ← **unstable**
- but controllable (why?) *** dropping Δ from Δx and Δu (for convenience)**
- goal: **make $x(t=10) = 1$, starting with I.C. $x(0) = 1$**
- $x(10) = 1 \cdot e^{10} + \int_0^{10} e^{10-\tau} u(\tau) d\tau = 1 \cdot e^{10} + u(e^{10} - 1)$
try $u(\tau) = \text{constant}$
- want: $1 = x(10) = 1 \cdot e^{10} + u(e^{10} - 1)$
-1

The Problem with Open Loop Control

- “open loop” means: no feedback
- “closed loop” means a system with feedback
- example: $\dot{x}(t) = ax(t) + u(t)$, $a = 1 > 0$ ← **unstable**
- but controllable (why?) *** dropping Δ from Δx and Δu (for convenience)**
- goal: **make $x(t=10) = 1$, starting with I.C. $x(0) = 1$**
- $x(10) = 1 \cdot e^{10} + \int_0^{10} e^{10-\tau} u(\tau) d\tau = 1 \cdot e^{10} + u(e^{10} - 1)$
try $u(\tau) = \text{constant}$
- want: $1 = x(10) = 1 \cdot e^{10} + u(e^{10} - 1)$
-1
- suppose there's a **0.1% error in the IC: $1 \rightarrow 1.001$**

The Problem with Open Loop Control

- “open loop” means: no feedback
- “closed loop” means a system with feedback
- example: $\dot{x}(t) = ax(t) + u(t)$, $a = 1 > 0$ ← **unstable**
- but controllable (why?) *** dropping Δ from Δx and Δu (for convenience)**
- goal: **make $x(t=10) = 1$, starting with I.C. $x(0) = 1$**
- $x(10) = 1 \cdot e^{10} + \int_0^{10} e^{10-\tau} u(\tau) d\tau = 1 \cdot e^{10} + u(e^{10} - 1)$
try $u(\tau) = \text{constant}$
- want: $1 = x(10) = 1 \cdot e^{10} + u(e^{10} - 1)$
-1
- suppose there's a **0.1% error in the IC: $1 \rightarrow 1.001$**
- new $x(10) = 1.001 \cdot e^{10} + u(e^{10} - 1) = 1 + 10^{-3} e^{10}$

The Problem with Open Loop Control

- “open loop” means: no feedback
- “closed loop” means a system with feedback
- example: $\dot{x}(t) = ax(t) + u(t)$, $a = 1 > 0$ ← **unstable**
- but controllable (why?) *** dropping Δ from Δx and Δu (for convenience)**
- goal: **make $x(t=10) = 1$, starting with I.C. $x(0) = 1$**
- $x(10) = 1 \cdot e^{10} + \int_0^{10} e^{10-\tau} u(\tau) d\tau = 1 \cdot e^{10} + u(e^{10} - 1)$
try $u(\tau) = \text{constant}$
- want: $1 = x(10) = 1 \cdot e^{10} + u(e^{10} - 1)$
-1
- suppose there's a **0.1% error in the IC: $1 \rightarrow 1.001$**
- new $x(10) = 1.001 \cdot e^{10} + u(e^{10} - 1) = 1 + 10^{-3} e^{10}$

$$e^{10} \simeq 22026$$

The Problem with Open Loop Control

- “open loop” means: no feedback
- “closed loop” means a system with feedback
- example: $\dot{x}(t) = ax(t) + u(t)$, $a = 1 > 0$ ← **unstable**
- but controllable (why?) * dropping Δ from Δx and Δu (for convenience)
- goal: make $x(t=10) = 1$, starting with I.C. $x(0) = 1$
- $x(10) = 1 \cdot e^{10} + \int_0^{10} e^{10-\tau} u(\tau) d\tau = 1 \cdot e^{10} + u(e^{10} - 1)$

try $u(\tau) = \text{constant}$
- want: $1 = x(10) = 1 \cdot e^{10} + u(e^{10} - 1)$

-1
- suppose there's a 0.1% error in the IC: $1 \rightarrow 1.001$
- new $x(10) = 1.001 \cdot e^{10} + u(e^{10} - 1) = 1 + 10^{-3} e^{10} \rightarrow \sim 22$

$e^{10} \simeq 22026$

The Problem with Open Loop Control

- “open loop” means: no feedback
- “closed loop” means a system with feedback
- example: $\dot{x}(t) = ax(t) + u(t)$, $a = 1 > 0$ ← **unstable**
- but controllable (why?) * dropping Δ from Δx and Δu (for convenience)
- goal: make $x(t=10) = 1$, starting with I.C. $x(0) = 1$
- $x(10) = 1 \cdot e^{10} + \int_0^{10} e^{10-\tau} u(\tau) d\tau = 1 \cdot e^{10} + u(e^{10} - 1)$

try $u(\tau) = \text{constant}$
- want: $1 = x(10) = 1 \cdot e^{10} + u(e^{10} - 1)$

-1
- suppose there's a 0.1% error in the IC: $1 \rightarrow 1.001$
- new $x(10) = 1.001 \cdot e^{10} + u(e^{10} - 1) = 1 + 10^{-3} e^{10} \rightarrow \sim 22$
- **0.1% error in IC \rightarrow 2200% error in $x(10)$** $e^{10} \simeq 22026$

The Problem with Open Loop Control

- “open loop” means: no feedback
- “closed loop” means a system with feedback
- example: $\dot{x}(t) = ax(t) + u(t)$, $a = 1 > 0$ ← **unstable**
- but controllable (why?) * dropping Δ from Δx and Δu (for convenience)
- goal: make $x(t=10) = 1$, starting with I.C. $x(0) = 1$
- $x(10) = 1 \cdot e^{10} + \int_0^{10} e^{10-\tau} u(\tau) d\tau = 1 \cdot e^{10} + u(e^{10} - 1)$

try $u(\tau) = \text{constant}$
- want: $1 = x(10) = 1 \cdot e^{10} + u(e^{10} - 1)$

-1
- suppose there's a 0.1% error in the IC: $1 \rightarrow 1.001$
- new $x(10) = 1.001 \cdot e^{10} + u(e^{10} - 1) = 1 + 10^{-3} e^{10} \rightarrow \sim 22$
- **0.1% error in IC \rightarrow 2200% error in $x(10)$** $e^{10} \simeq 22026$

if system unstable, control in the presence of errors/noise is impossible in practice

The Problem with Open Loop Control

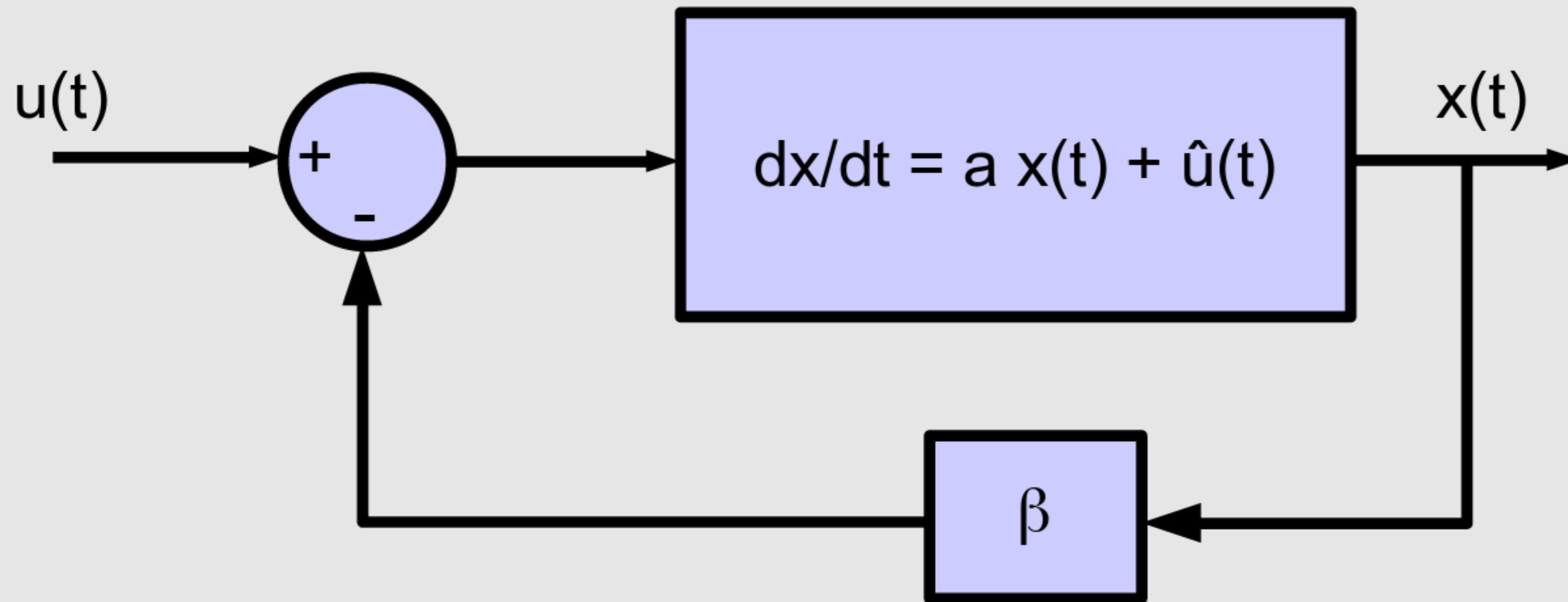
- “open loop” means: no feedback
- “closed loop” means a system with feedback
- example: $\dot{x}(t) = ax(t) + u(t)$, $a = 1 > 0$ ← **unstable**
- but controllable (why?) * dropping Δ from Δx and Δu (for convenience)
- goal: **make $x(t=10) = 1$, starting with I.C. $x(0) = 1$**
- $x(10) = 1 \cdot e^{10} + \int_0^{10} e^{10-\tau} u(\tau) d\tau = 1 \cdot e^{10} + u(e^{10} - 1)$

try $u(\tau) = \text{constant}$
- want: $1 = x(10) = 1 \cdot e^{10} + u(e^{10} - 1)$

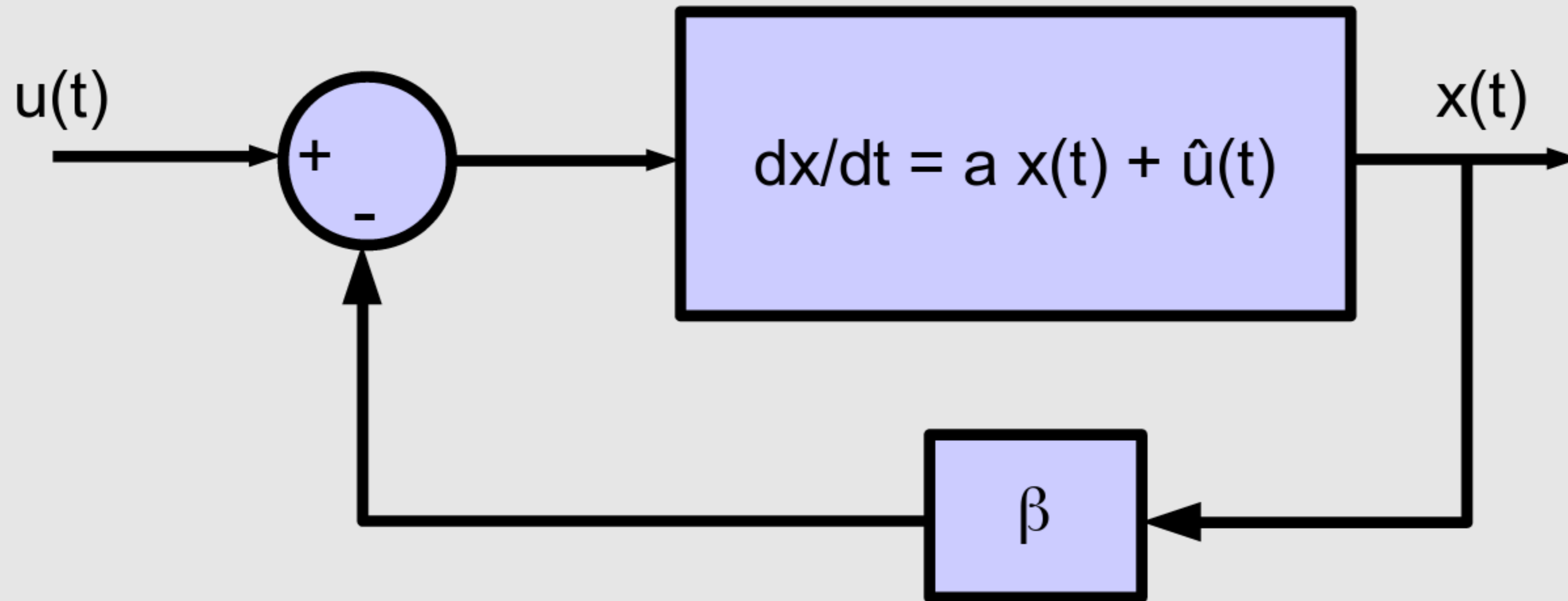
-1
- suppose there's a **0.1% error in the IC: $1 \rightarrow 1.001$**
- new $x(10) = 1.001 \cdot e^{10} + u(e^{10} - 1) = 1 + \boxed{10^{-3} e^{10}} \rightarrow \sim 22$
- **0.1% error in IC \rightarrow 2200% error in $x(10)$** $e^{10} \simeq 22026$
- How will this change if $a = -1$?

if system unstable, control in the presence of errors/noise is impossible in practice

Stabilization via Feedback (Scalar)

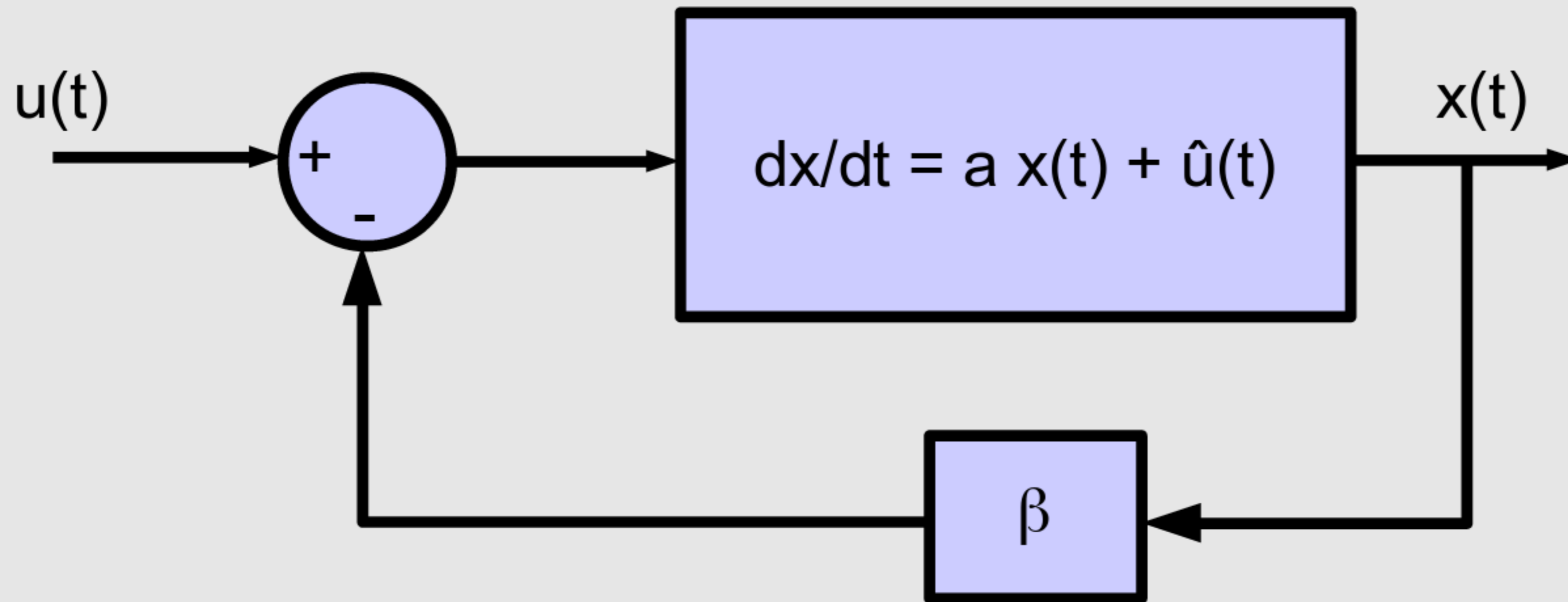


Stabilization via Feedback (Scalar)



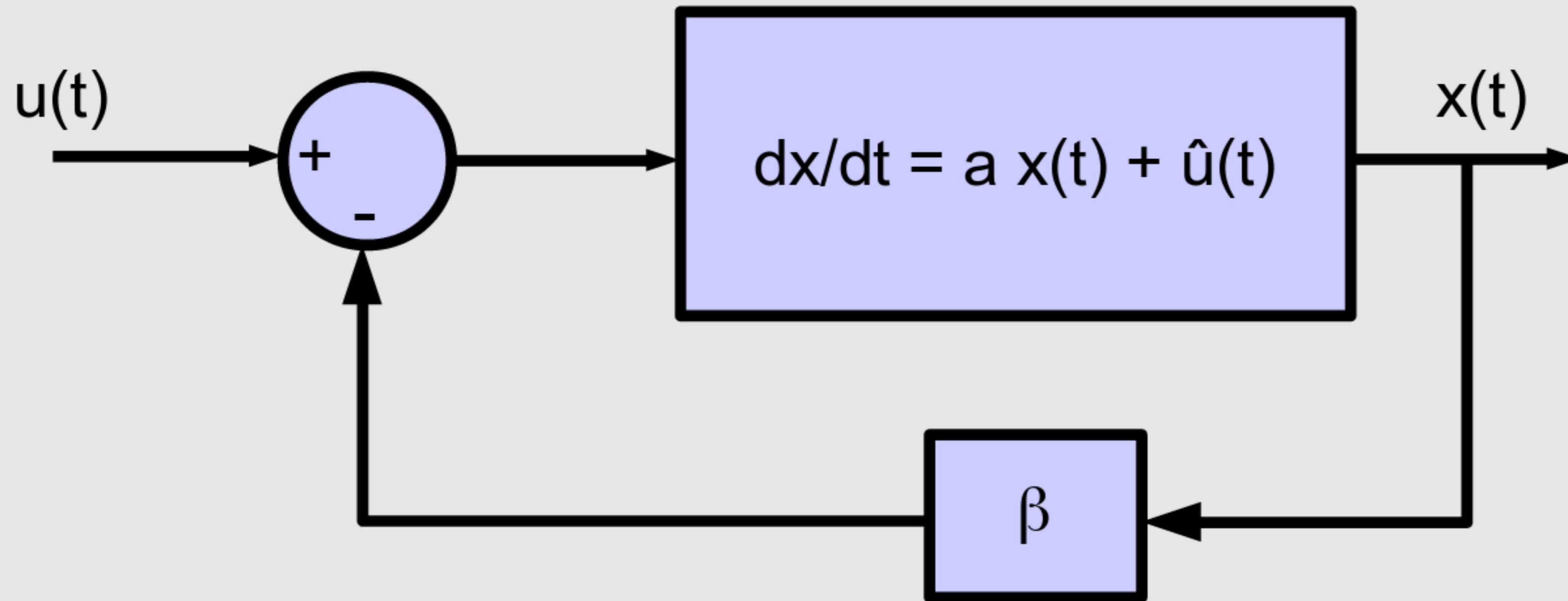
- apply feedback: $\hat{u}(t) \mapsto u(t) - \beta x(t)$

Stabilization via Feedback (Scalar)



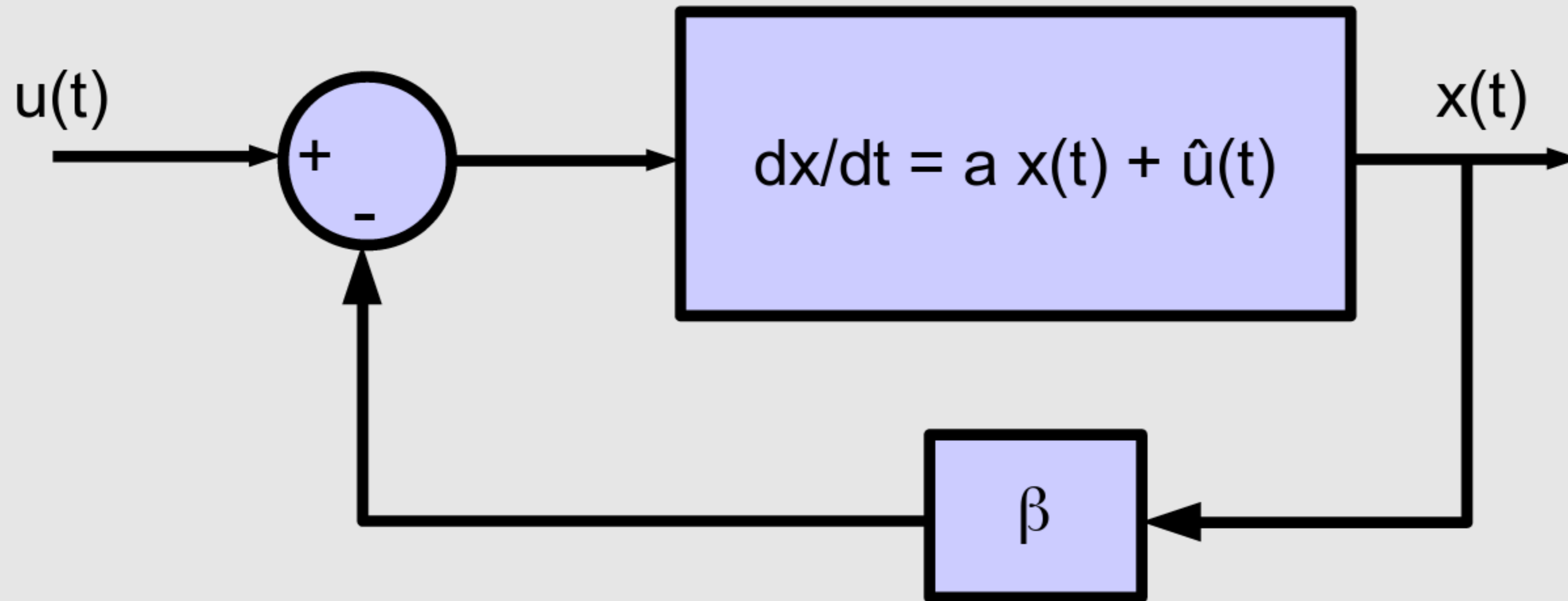
- apply feedback: $\hat{u}(t) \mapsto u(t) - \beta x(t)$
- $\dot{x}(t) = ax(t) + \hat{u}(t) \mapsto \dot{x}(t) = ax(t) + u(t) - \beta x(t), \quad a = 1 > 0$

Stabilization via Feedback (Scalar)



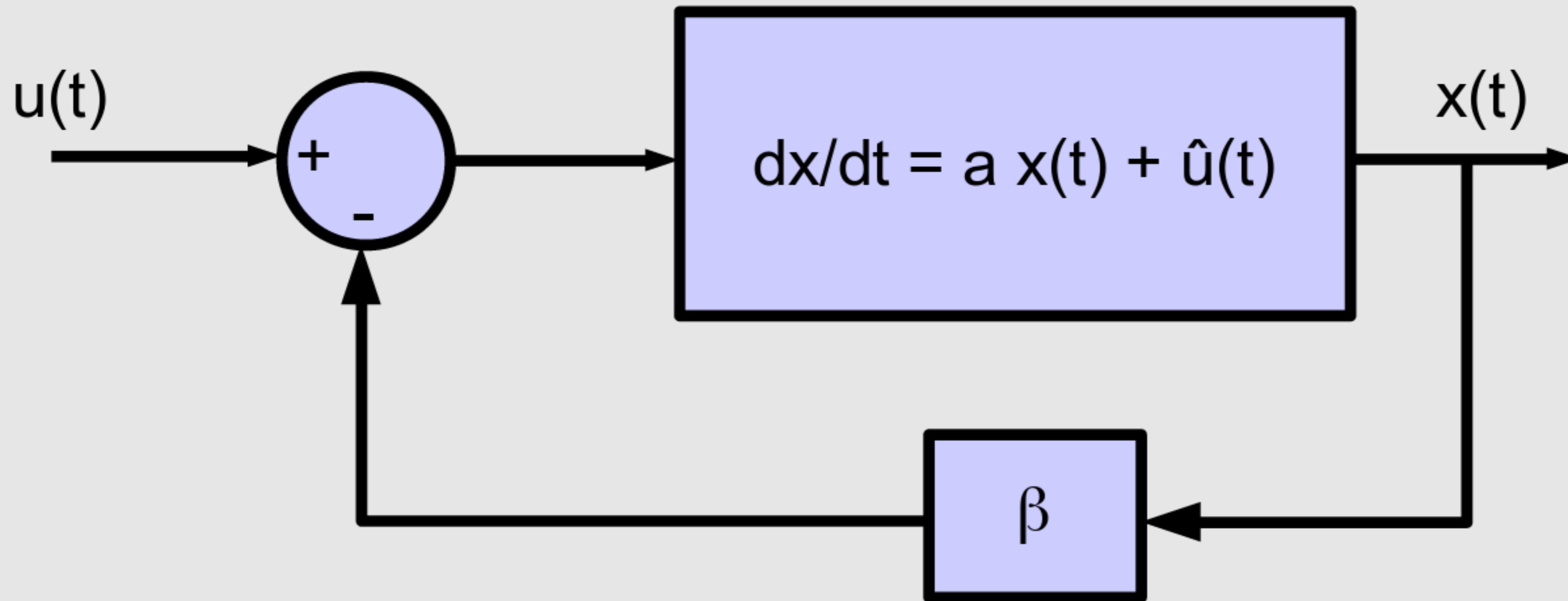
- apply feedback: $\hat{u}(t) \mapsto u(t) - \beta x(t)$
- $\dot{x}(t) = ax(t) + \hat{u}(t) \mapsto \dot{x}(t) = ax(t) + u(t) - \beta x(t), \quad a = 1 > 0$

Stabilization via Feedback (Scalar)



- apply feedback: $\hat{u}(t) \mapsto u(t) - \beta x(t)$
- $\dot{x}(t) = ax(t) + \hat{u}(t) \mapsto \dot{x}(t) = ax(t) + u(t) - \beta x(t), \quad a = 1 > 0$
- $\dot{x}(t) = (a - \beta)x(t) + u(t), \quad a = 1 > 0$

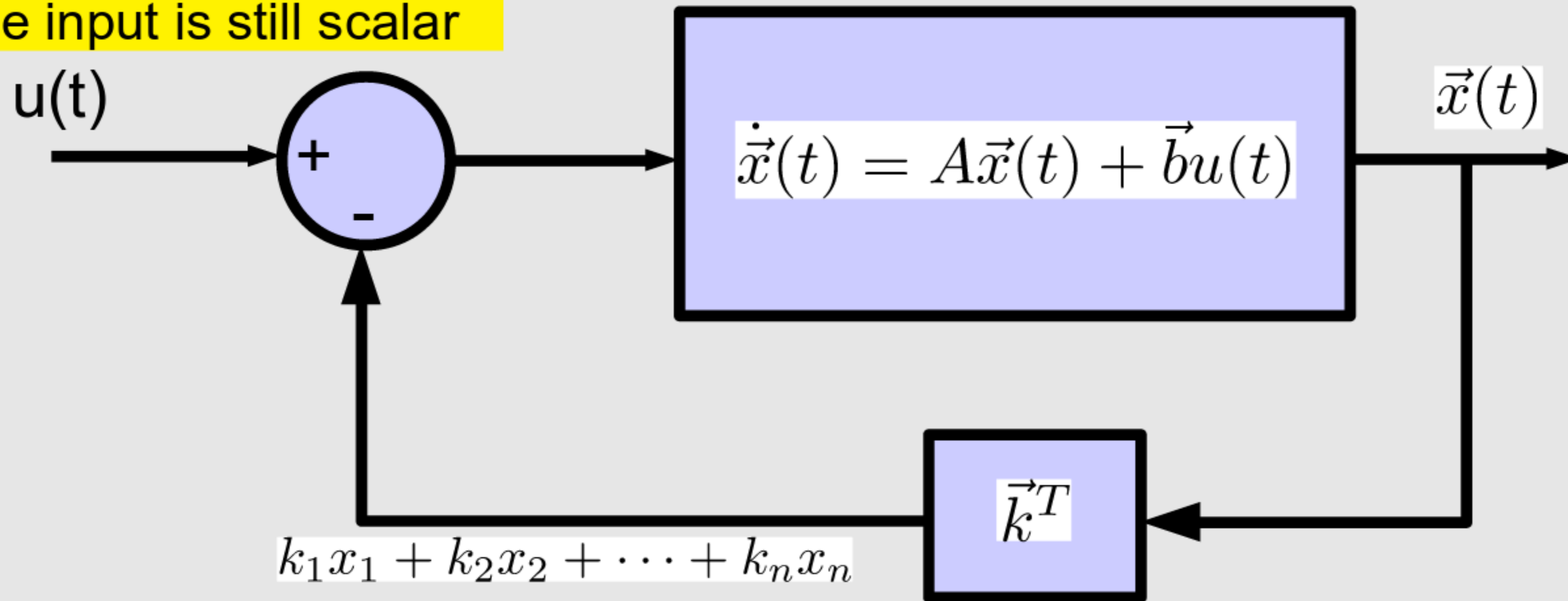
Stabilization via Feedback (Scalar)



- apply feedback: $\hat{u}(t) \mapsto u(t) - \beta x(t)$
- $\dot{x}(t) = ax(t) + \hat{u}(t) \mapsto \dot{x}(t) = ax(t) + u(t) - \beta x(t), \quad a = 1 > 0$
- $\dot{x}(t) = (a - \beta)x(t) + u(t), \quad a = 1 > 0$
choose $\beta > a \rightarrow$ **system is stabilized**

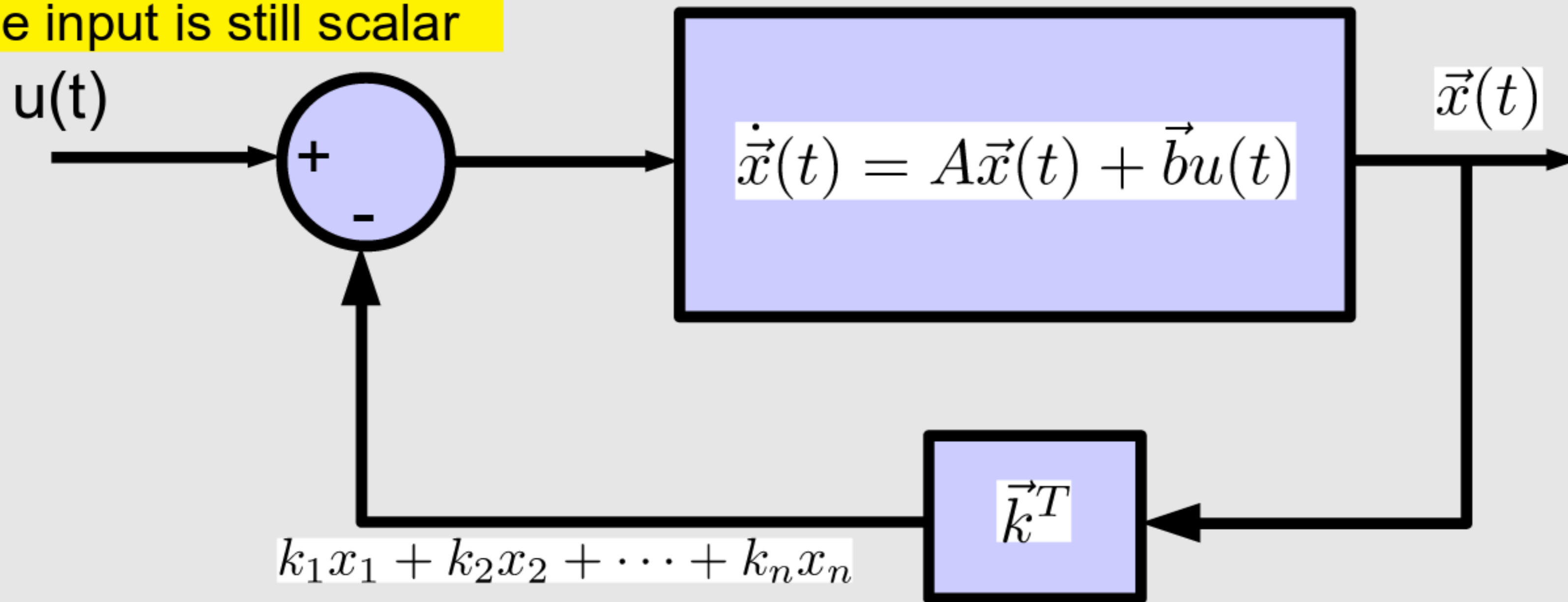
Feedback for Vector S.S. Systems

assumption for simplicity:
the input is still scalar



Feedback for Vector S.S. Systems

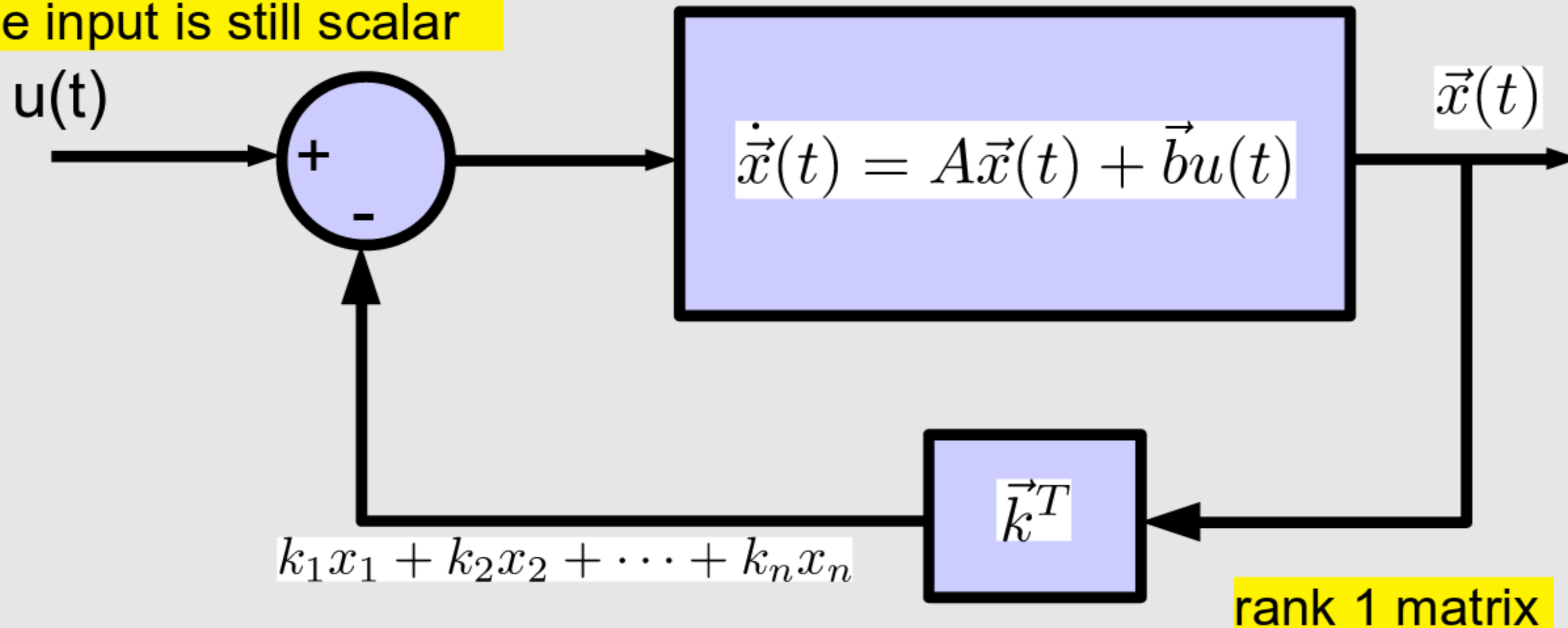
assumption for simplicity:
the input is still scalar



- system w feedback: $\dot{\vec{x}}(t) = (A - \vec{b}\vec{k}^T)\vec{x}(t) + \vec{b}u(t)$

Feedback for Vector S.S. Systems

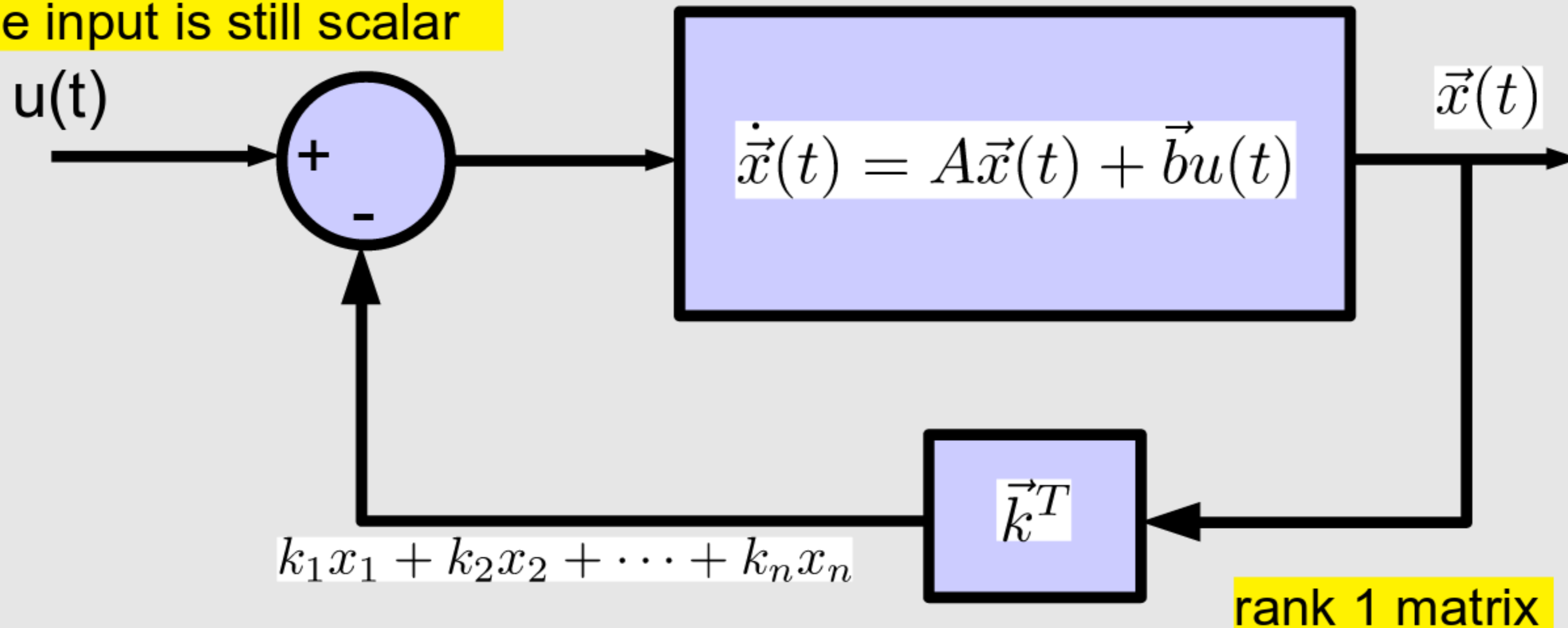
assumption for simplicity:
the input is still scalar



- system w feedback: $\dot{\vec{x}}(t) = (A - \vec{b}\vec{k}^T)\vec{x}(t) + \vec{b}u(t)$

Feedback for Vector S.S. Systems

assumption for simplicity:
the input is still scalar

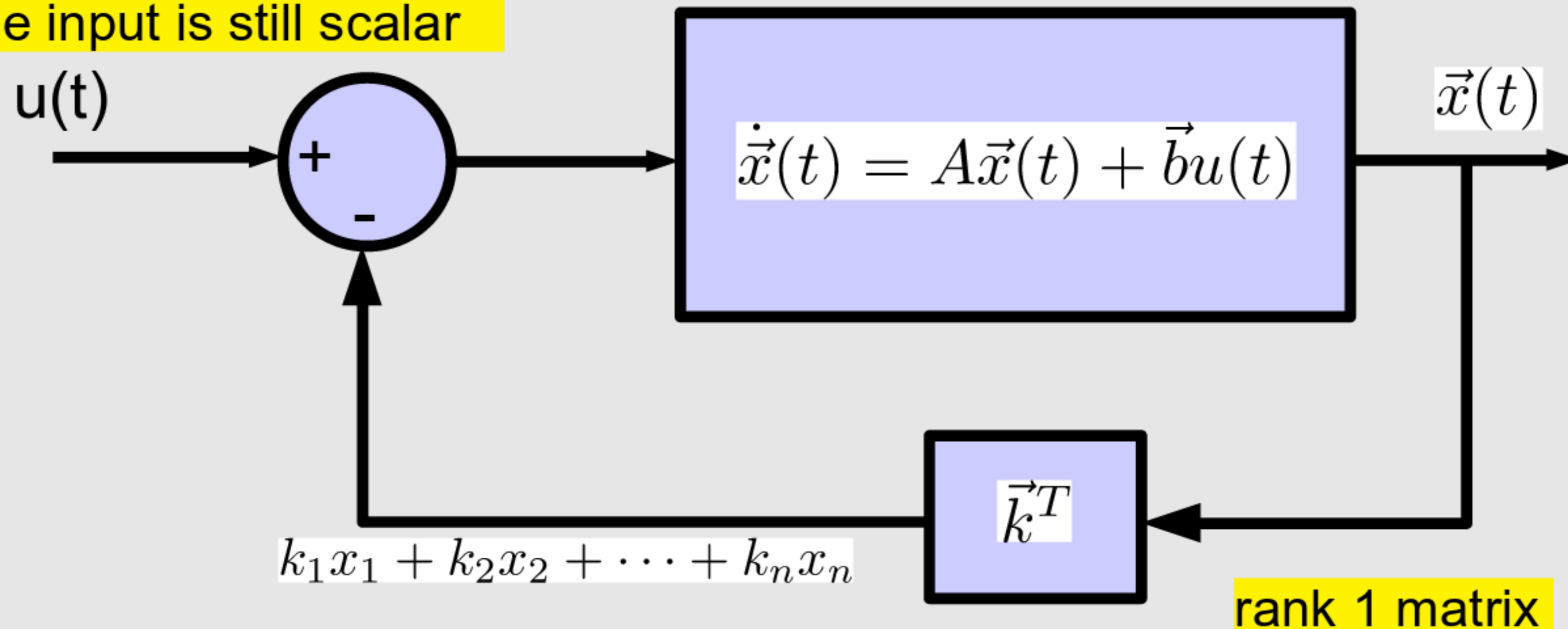


rank 1 matrix

- system w feedback: $\dot{\vec{x}}(t) = (A - \vec{b}\vec{k}^T)\vec{x}(t) + \vec{b}u(t)$
- stability governed by eigenvalues of $A - \vec{b}\vec{k}^T$

Feedback for Vector S.S. Systems

assumption for simplicity:
the input is still scalar

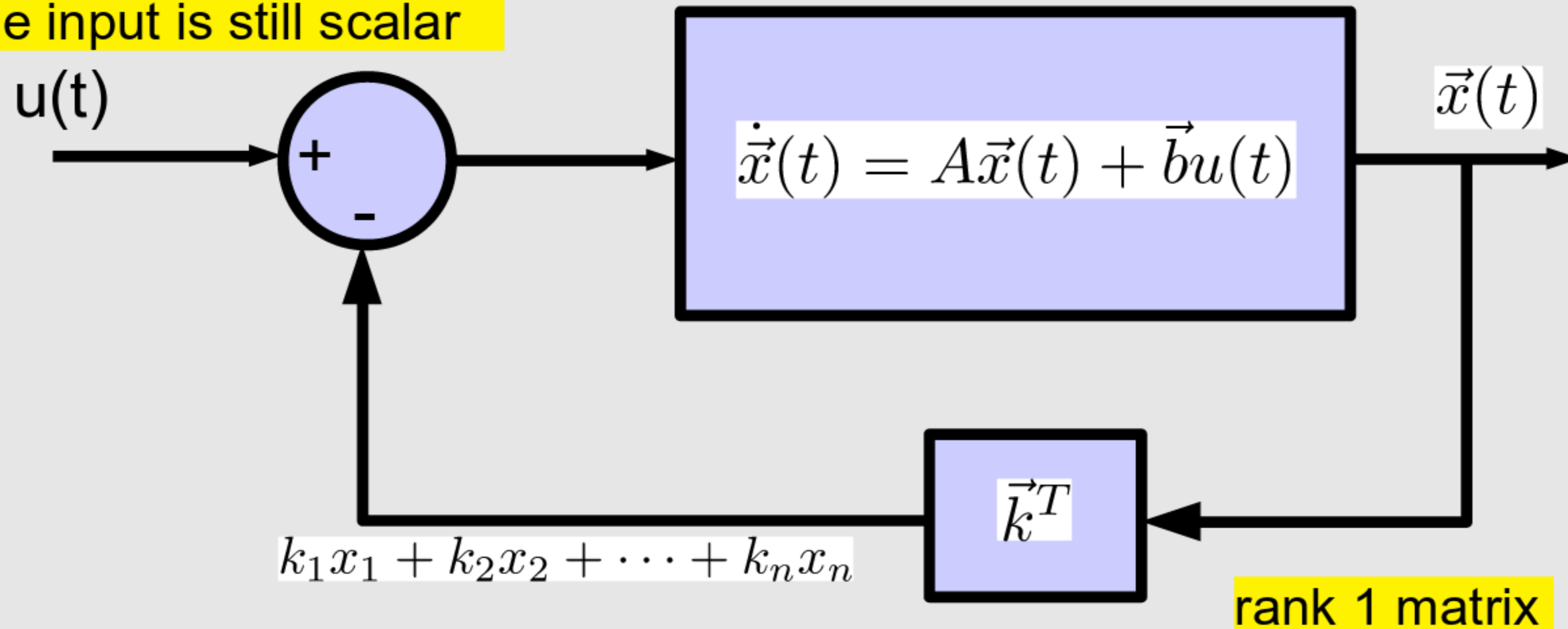


rank 1 matrix

- system w feedback: $\dot{\vec{x}}(t) = (A - \vec{b}\vec{k}^T)\vec{x}(t) + \vec{b}u(t)$
- stability governed by eigenvalues of $A - \vec{b}\vec{k}^T$
- Q: how do the e.values of A change due to $\vec{b}\vec{k}^T$

Feedback for Vector S.S. Systems

assumption for simplicity:
the input is still scalar

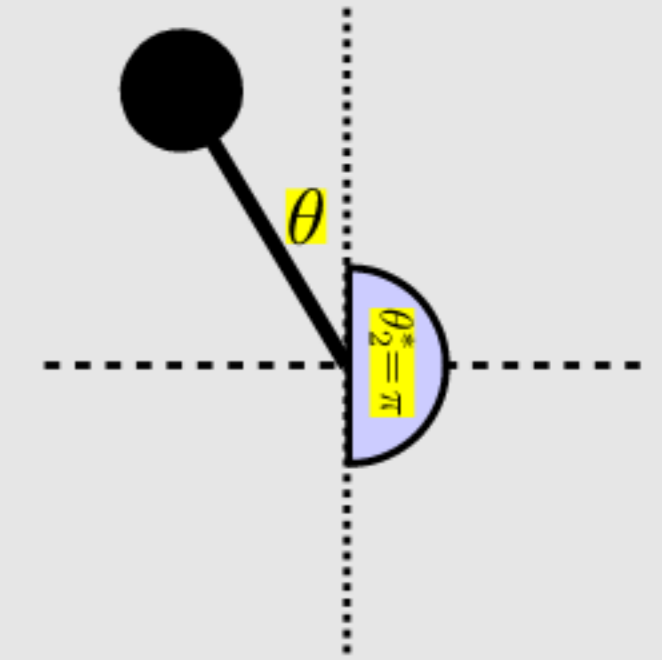


rank 1 matrix

- system w feedback: $\dot{\vec{x}}(t) = (A - \vec{b}\vec{k}^T)\vec{x}(t) + \vec{b}u(t)$
- stability governed by eigenvalues of $A - \vec{b}\vec{k}^T$
- Q: how do the e.values of A change due to
- **very difficult to figure out analytically!**
 - can do simple examples; otherwise, numerically

Example: stabilizing an inverted pendulum using feedback

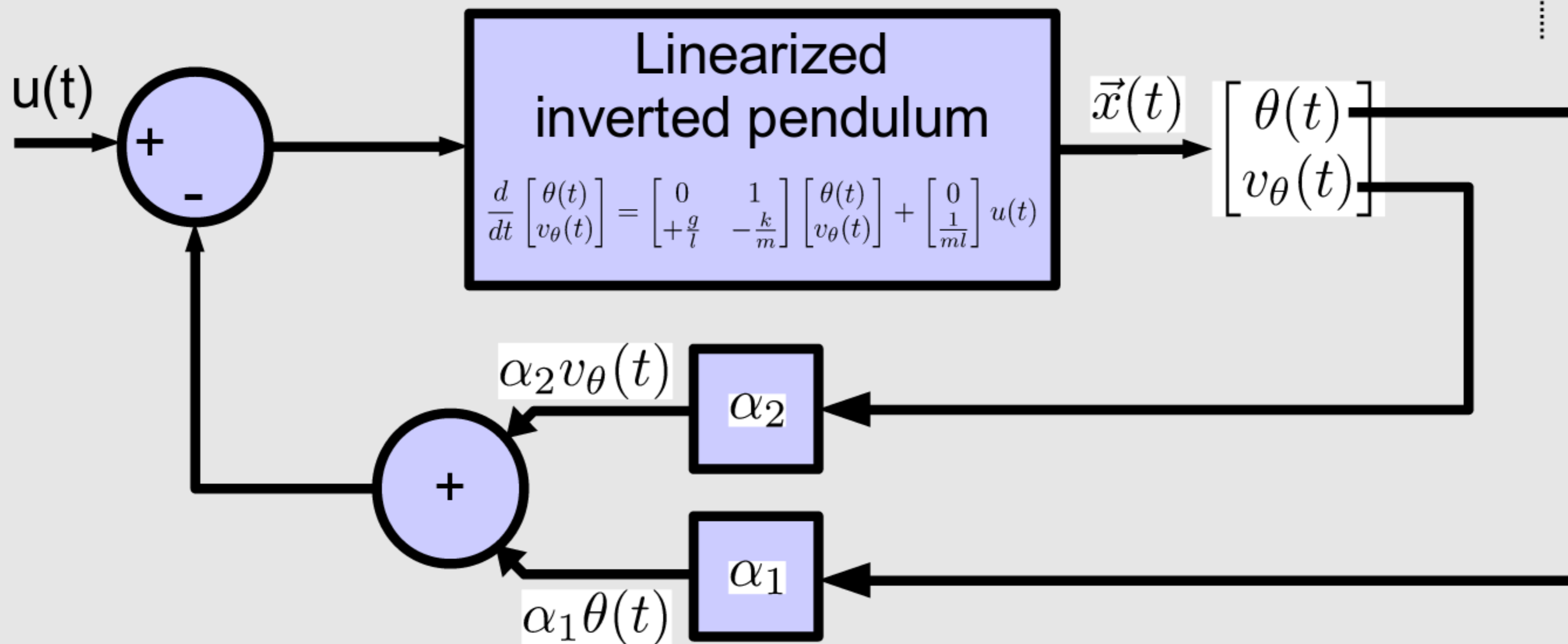
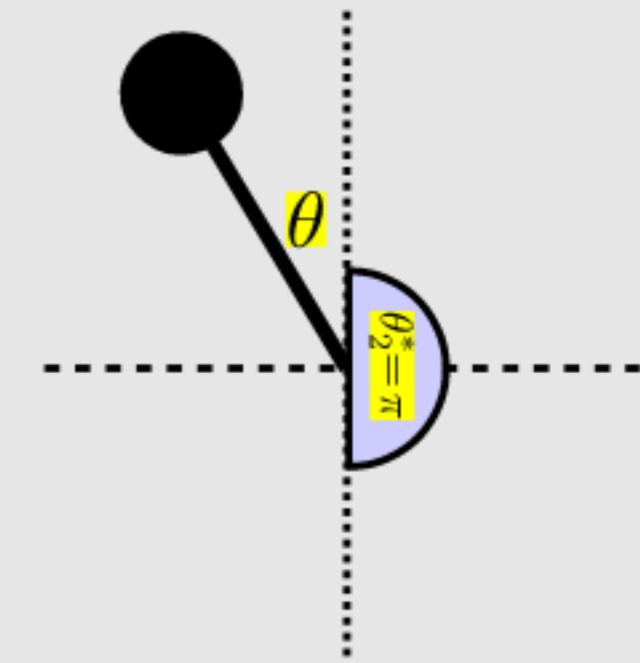
- i.p.:
$$\frac{d}{dt} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ +\frac{g}{l} & -\frac{k}{m} \end{bmatrix} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{ml} \end{bmatrix} u(t)$$



Example: stabilizing an inverted pendulum using feedback

- i.p.: $\frac{d}{dt} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ +\frac{g}{l} & -\frac{k}{m} \end{bmatrix} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{ml} \end{bmatrix} u(t)$

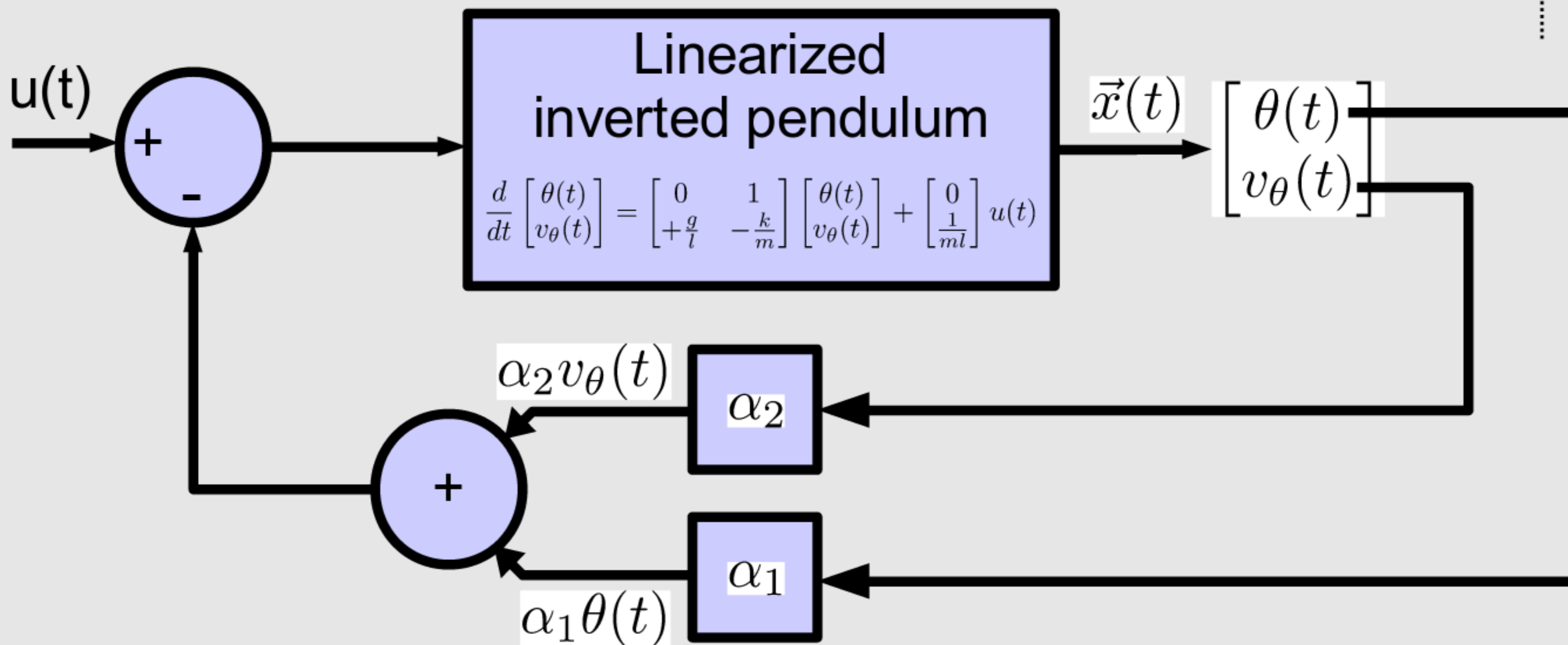
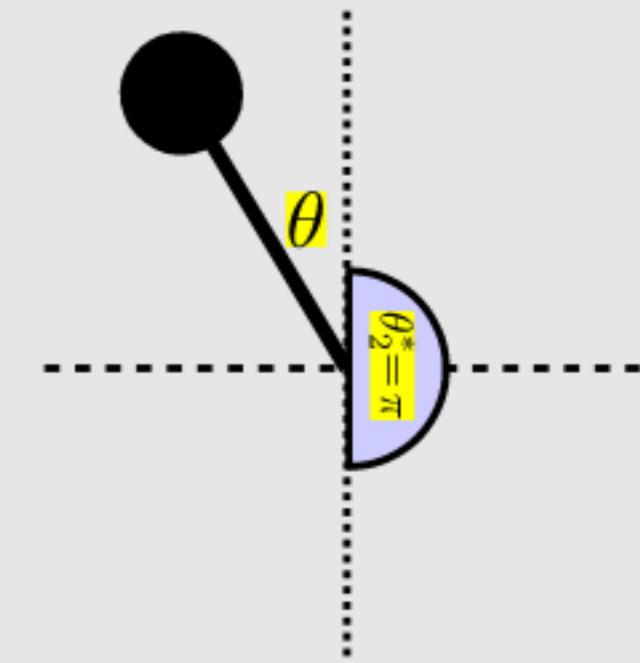
- Closed loop system (ie, with feedback)



Example: stabilizing an inverted pendulum using feedback

- i.p.:
$$\frac{d}{dt} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ +\frac{g}{l} & -\frac{k}{m} \end{bmatrix} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{ml} \end{bmatrix} u(t)$$

- Closed loop system (ie, with feedback)



- $$\frac{d}{dt} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \frac{mg - \alpha_1}{ml} & \frac{-kl - \alpha_2}{ml} \end{bmatrix} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{ml} \end{bmatrix} u(t)$$

Stabilizing I.P. via feedback (contd.)

- I.P. w F.:
$$\frac{d}{dt} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \frac{mg - \alpha_1}{ml} & \frac{-kl - \alpha_2}{ml} \end{bmatrix} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{ml} \end{bmatrix} u(t)$$

Stabilizing I.P. via feedback (contd.)

- I.P. w F.: $\frac{d}{dt} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \frac{mg - \alpha_1}{ml} & \frac{-kl - \alpha_2}{ml} \end{bmatrix} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{ml} \end{bmatrix} u(t)$
- eigenvalues of this determine stability

Stabilizing I.P. via feedback (contd.)

- I.P. w F.: $\frac{d}{dt} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \frac{mg-\alpha_1}{ml} & \frac{-kl-\alpha_2}{ml} \end{bmatrix} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{ml} \end{bmatrix} u(t)$

- eigenvalues of this determine stability

$$\rightarrow \det \left(\begin{bmatrix} -\lambda & 1 \\ \frac{mg-\alpha_1}{ml} & \frac{-kl-\alpha_2-m\lambda}{ml} \end{bmatrix} \right) = 0$$

Stabilizing I.P. via feedback (contd.)

- I.P. w F.: $\frac{d}{dt} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \frac{mg - \alpha_1}{ml} & \frac{-kl - \alpha_2}{ml} \end{bmatrix} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{ml} \end{bmatrix} u(t)$
 - eigenvalues of this determine stability
- $\det \left(\begin{bmatrix} -\lambda & 1 \\ \frac{mg - \alpha_1}{ml} & \frac{-kl - \alpha_2 - ml\lambda}{ml} \end{bmatrix} \right) = 0 \Rightarrow ml\lambda^2 + (kl + \alpha_2)\lambda - (mg - \alpha_1) = 0$

Stabilizing I.P. via feedback (contd.)

- I.P. w F.: $\frac{d}{dt} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \frac{mg - \alpha_1}{ml} & \frac{-kl - \alpha_2}{ml} \end{bmatrix} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{ml} \end{bmatrix} u(t)$
- eigenvalues of this determine stability
 - $\det \left(\begin{bmatrix} -\lambda & 1 \\ \frac{mg - \alpha_1}{ml} & \frac{-kl - \alpha_2 - ml\lambda}{ml} \end{bmatrix} \right) = 0 \Rightarrow ml\lambda^2 + (kl + \alpha_2)\lambda - (mg - \alpha_1) = 0$
 - $\lambda_{1,2} = \frac{-(kl + \alpha_2)}{2ml} \pm \frac{\sqrt{(kl + \alpha_2)^2 + 4ml(mg - \alpha_1)}}{2ml}$

Stabilizing I.P. via feedback (contd.)

- I.P. w F.: $\frac{d}{dt} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \frac{mg - \alpha_1}{ml} & \frac{-kl - \alpha_2}{ml} \end{bmatrix} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{ml} \end{bmatrix} u(t)$
- eigenvalues of this determine stability
 - $\det \left(\begin{bmatrix} -\lambda & 1 \\ \frac{mg - \alpha_1}{ml} & \frac{-kl - \alpha_2 - ml\lambda}{ml} \end{bmatrix} \right) = 0 \Rightarrow ml\lambda^2 + (kl + \alpha_2)\lambda - (mg - \alpha_1) = 0$
 - $\lambda_{1,2} = \frac{-(kl + \alpha_2)}{2ml} \pm \frac{\sqrt{(kl + \alpha_2)^2 + 4ml(mg - \alpha_1)}}{2ml}$
- to stabilize: make both evs -ve (real part)

Stabilizing I.P. via feedback (contd.)

- I.P. w F.: $\frac{d}{dt} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \frac{mg - \alpha_1}{ml} & \frac{-kl - \alpha_2}{ml} \end{bmatrix} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{ml} \end{bmatrix} u(t)$
- eigenvalues of this determine stability
 - $\det \left(\begin{bmatrix} -\lambda & 1 \\ \frac{mg - \alpha_1}{ml} & \frac{-kl - \alpha_2 - ml\lambda}{ml} \end{bmatrix} \right) = 0 \Rightarrow ml\lambda^2 + (kl + \alpha_2)\lambda - (mg - \alpha_1) = 0$
 - make this negative
 - $\lambda_{1,2} = \frac{-(kl + \alpha_2)}{2ml} \pm \frac{\sqrt{(kl + \alpha_2)^2 + 4ml(mg - \alpha_1)}}{2ml}$
- to stabilize: make both evs -ve (real part)

Stabilizing I.P. via feedback (contd.)

- I.P. w F.: $\frac{d}{dt} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \frac{mg-\alpha_1}{ml} & \frac{-kl-\alpha_2}{ml} \end{bmatrix} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{ml} \end{bmatrix} u(t)$

- eigenvalues of this determine stability

→ $\det \left(\begin{bmatrix} -\lambda & 1 \\ \frac{mg-\alpha_1}{ml} & \frac{-kl-\alpha_2}{ml} - ml\lambda \end{bmatrix} \right) = 0 \Rightarrow ml\lambda^2 + (kl + \alpha_2)\lambda - (mg - \alpha_1) = 0$

make this negative

make this smaller than $|kl + \alpha_2|$

→ $\lambda_{1,2} = \frac{-(kl + \alpha_2)}{2ml} \pm \frac{\sqrt{(kl + \alpha_2)^2 + 4ml(mg - \alpha_1)}}{2ml}$

- to stabilize: make both evs -ve (real part)

Stabilizing I.P. via feedback (contd.)

- I.P. w F.: $\frac{d}{dt} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \frac{mg-\alpha_1}{ml} & \frac{-kl-\alpha_2}{ml} \end{bmatrix} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{ml} \end{bmatrix} u(t)$

- eigenvalues of this determine stability

→ $\det \left(\begin{bmatrix} -\lambda & 1 \\ \frac{mg-\alpha_1}{ml} & \frac{-kl-\alpha_2}{ml} - ml\lambda \end{bmatrix} \right) = 0 \Rightarrow ml\lambda^2 + (kl + \alpha_2)\lambda - (mg - \alpha_1) = 0$

make this negative

make this smaller than $|kl + \alpha_2|$

→ $\lambda_{1,2} = \frac{-(kl + \alpha_2)}{2ml} \pm \frac{\sqrt{(kl + \alpha_2)^2 + 4ml(mg - \alpha_1)}}{2ml}$ ← make this negative

- to stabilize: make both evs -ve (real part)

Stabilizing I.P. via feedback (contd.)

- I.P. w F.: $\frac{d}{dt} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \frac{mg-\alpha_1}{ml} & \frac{-kl-\alpha_2}{ml} \end{bmatrix} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{ml} \end{bmatrix} u(t)$

- eigenvalues of this determine stability

→ $\det \left(\begin{bmatrix} -\lambda & 1 \\ \frac{mg-\alpha_1}{ml} & \frac{-kl-\alpha_2}{ml} - ml\lambda \end{bmatrix} \right) = 0 \Rightarrow ml\lambda^2 + (kl + \alpha_2)\lambda - (mg - \alpha_1) = 0$

make this negative

make this smaller than $|kl + \alpha_2|$

→ $\lambda_{1,2} = \frac{-(kl + \alpha_2)}{2ml} \pm \frac{\sqrt{(kl + \alpha_2)^2 + 4ml(mg - \alpha_1)}}{2ml}$ ← make this negative

- to stabilize: make both evs -ve (real part)

→ choose any $\alpha_2 > -kl, \quad \alpha_1 > mg$

Stabilizing I.P. via feedback (contd.)

- I.P. w F.: $\frac{d}{dt} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \frac{mg-\alpha_1}{ml} & \frac{-kl-\alpha_2}{ml} \end{bmatrix} \begin{bmatrix} \theta(t) \\ v_\theta(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{ml} \end{bmatrix} u(t)$

- eigenvalues of this determine stability

→ $\det \left(\begin{bmatrix} -\lambda & 1 \\ \frac{mg-\alpha_1}{ml} & \frac{-kl-\alpha_2-m\lambda}{ml} \end{bmatrix} \right) = 0 \Rightarrow ml\lambda^2 + (kl + \alpha_2)\lambda - (mg - \alpha_1) = 0$

make this negative

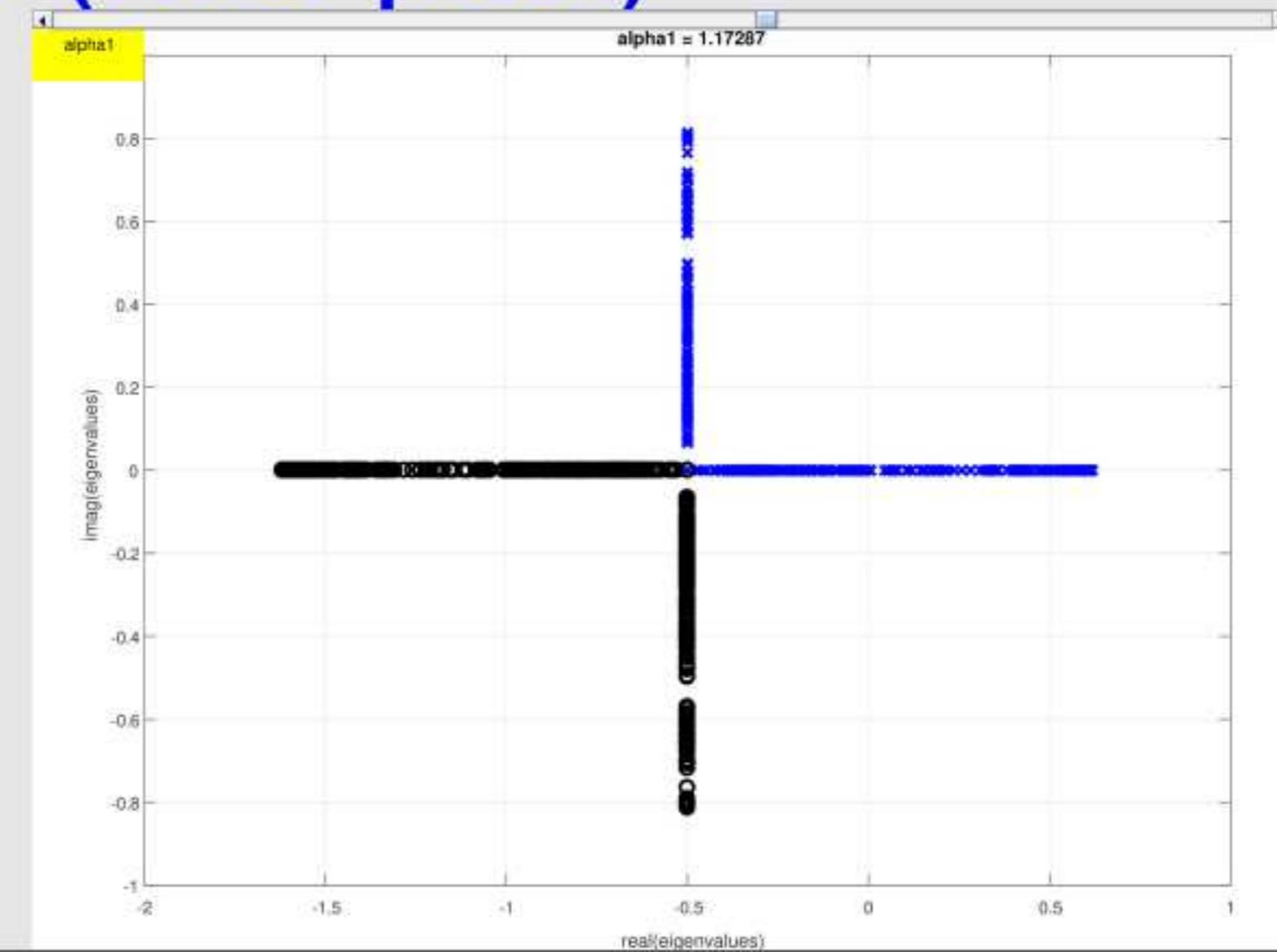
make this smaller than $|kl+\alpha_2|$

→ $\lambda_{1,2} = \frac{-(kl + \alpha_2)}{2ml} \pm \frac{\sqrt{(kl + \alpha_2)^2 + 4ml(mg - \alpha_1)}}{2ml}$ ← make this negative

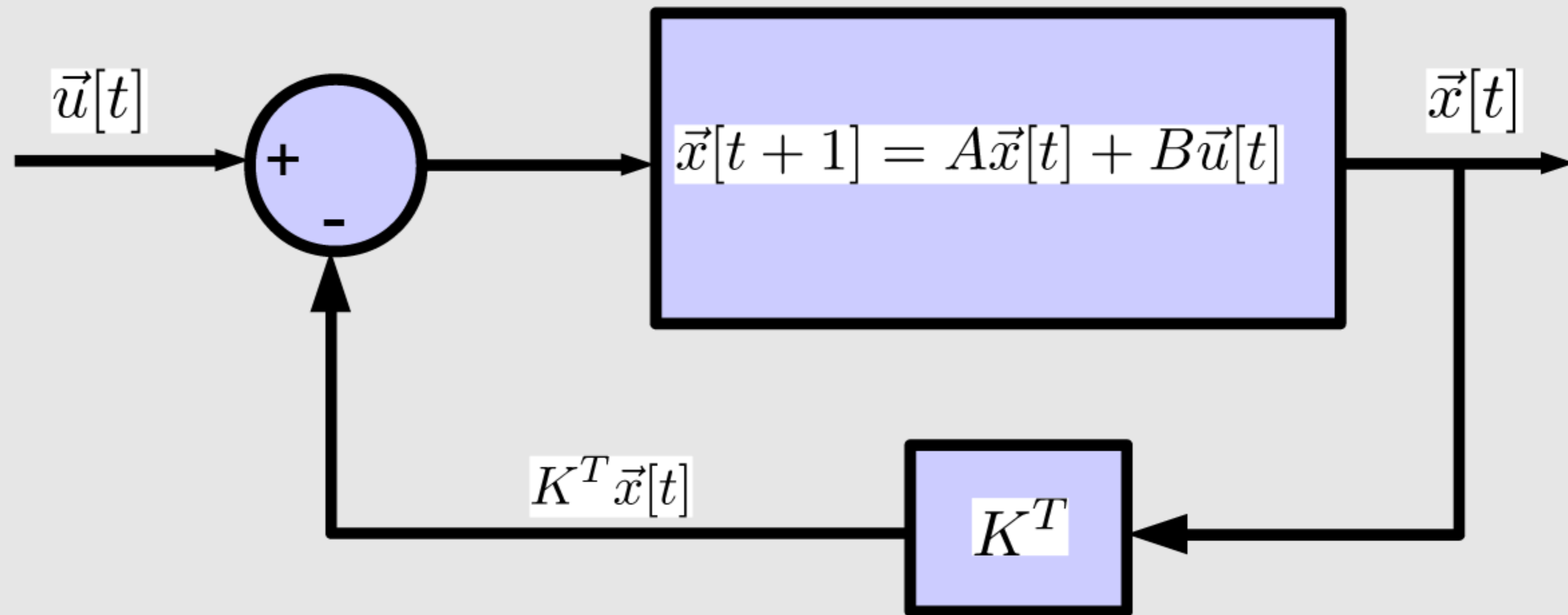
- to stabilize: make both evs -ve (real part)

→ choose any $\alpha_2 > -kl$, $\alpha_1 > mg$

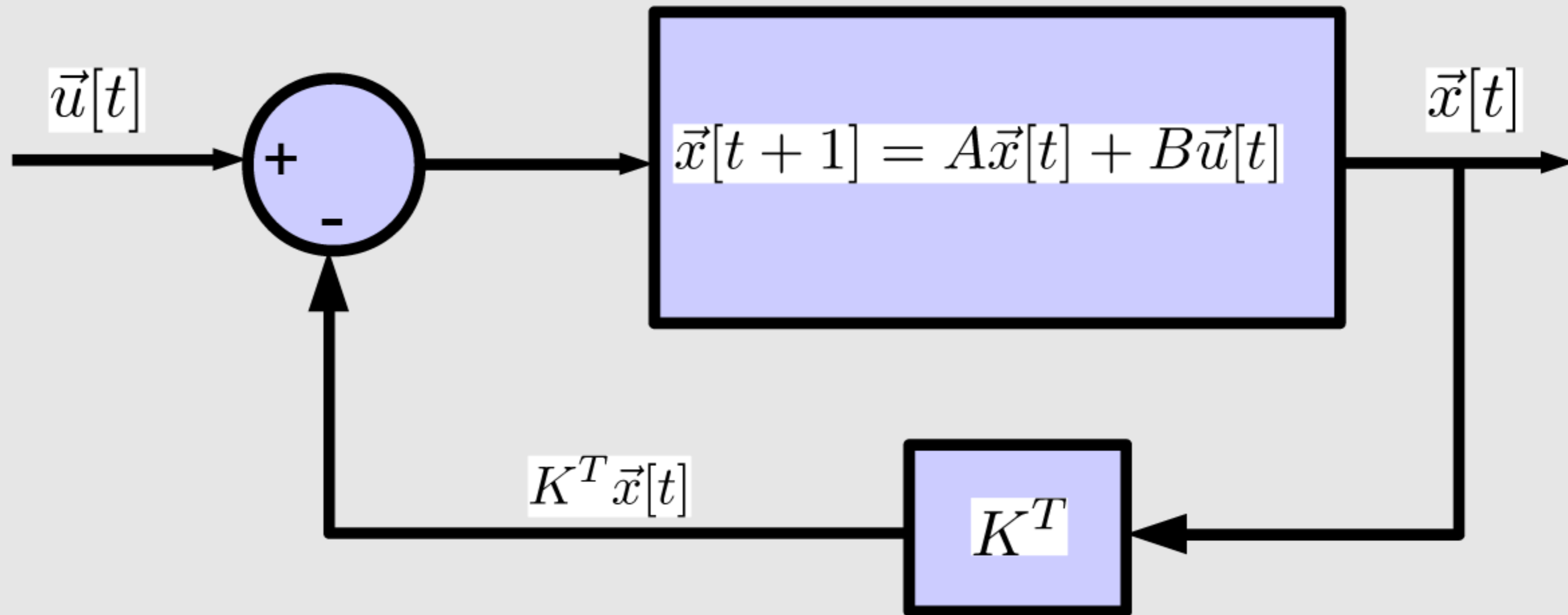
run MATLAB demo
inverted_pendulum_w_feedback_root_locus.m



Feedback for Discrete-Time S.S.Rs

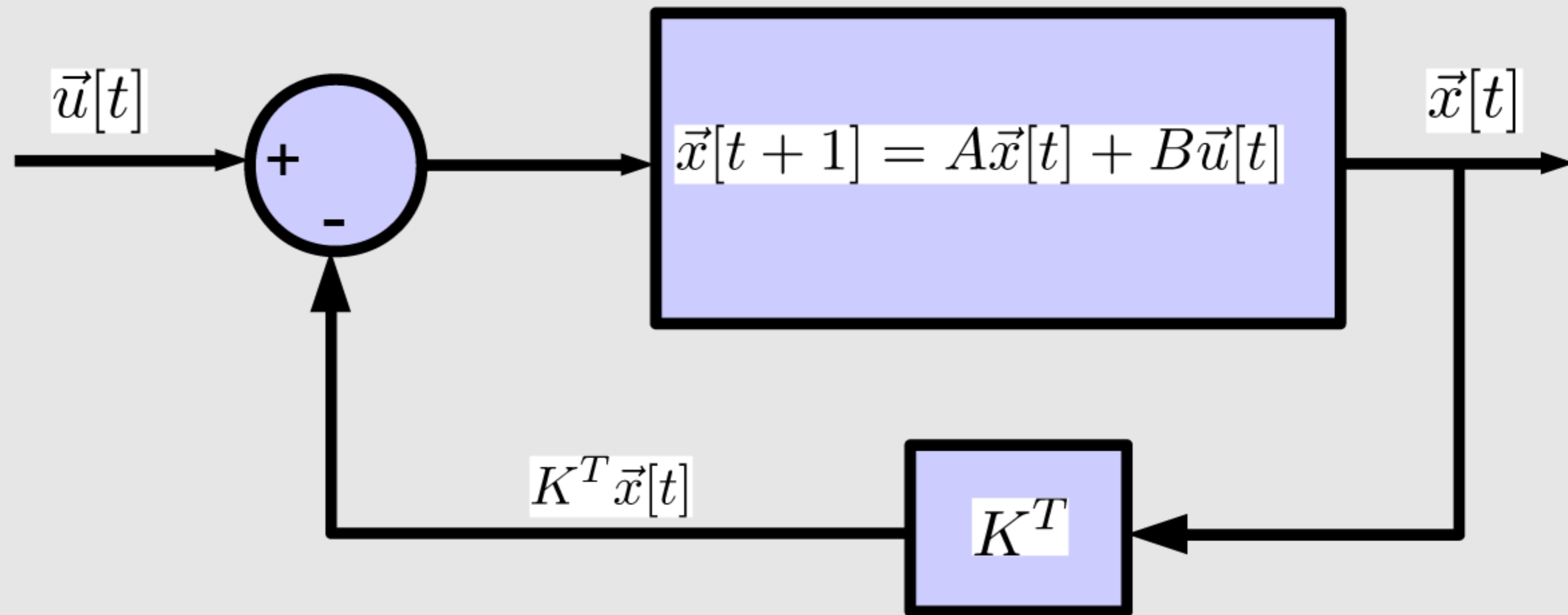


Feedback for Discrete-Time S.S.Rs



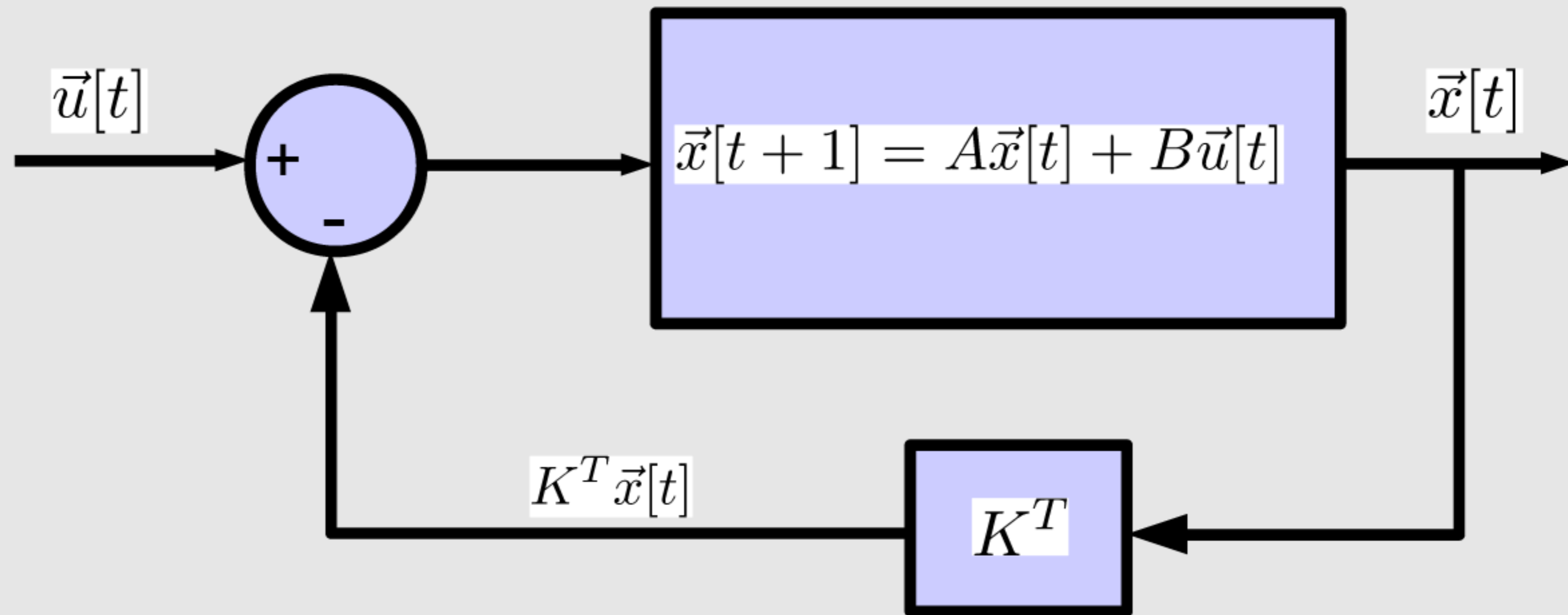
- system w feedback: $\vec{x}[t+1] = (A - BK^T)\vec{x}[t] + B\vec{u}[t]$

Feedback for Discrete-Time S.S.Rs



- system w feedback: $\vec{x}[t+1] = (A - BK^T)\vec{x}[t] + B\vec{u}[t]$
- stability still governed by the eigenvalues of $A - BK^T$

Feedback for Discrete-Time S.S.Rs



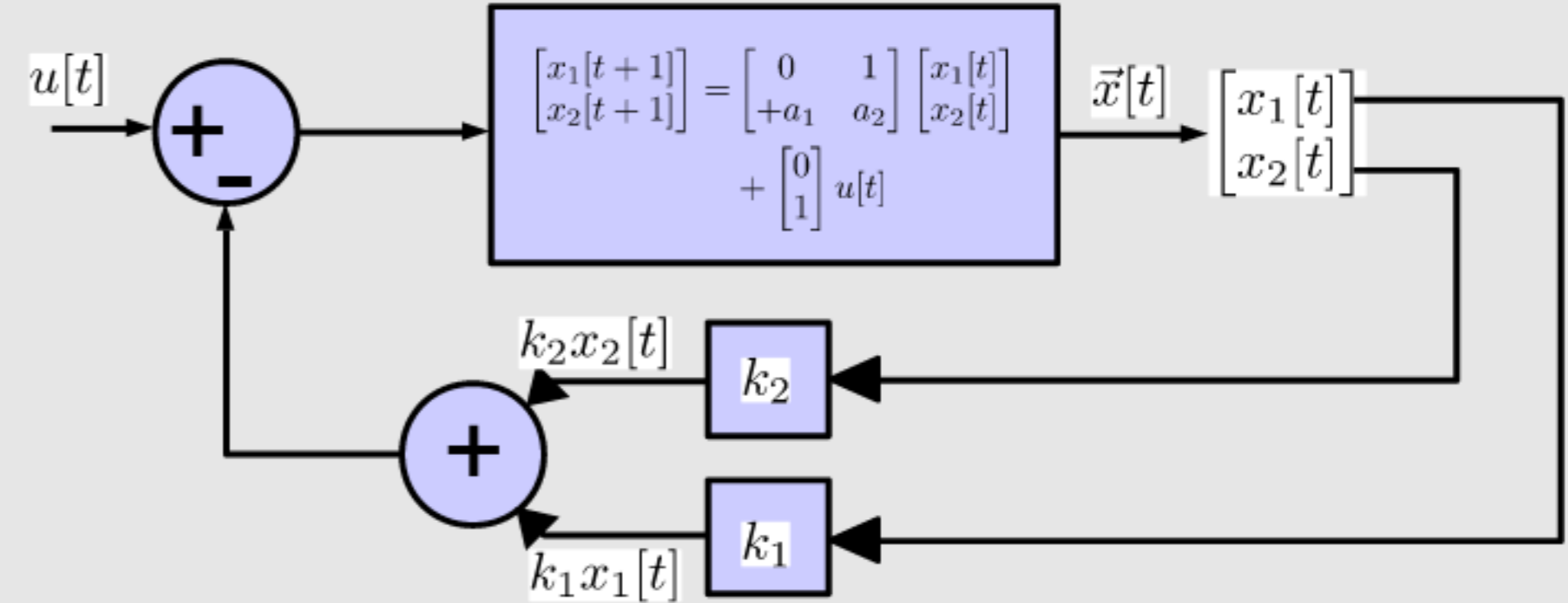
- system w feedback: $\vec{x}[t+1] = (A - BK^T)\vec{x}[t] + B\vec{u}[t]$
- stability still governed by the eigenvalues of $A - BK^T$
- stability (discr.) \rightarrow magnitude of eigenvalues < 1
- different from the continuous case

Example: Discrete-Time Feedback

- $$\vec{x}[t + 1] = \begin{bmatrix} 0 & 1 \\ a_1 & a_2 \end{bmatrix} \vec{x}[t] + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u[t]$$

Example: Discrete-Time Feedback

- $\vec{x}[t+1] = \begin{bmatrix} 0 & 1 \\ a_1 & a_2 \end{bmatrix} \vec{x}[t] + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u[t]$

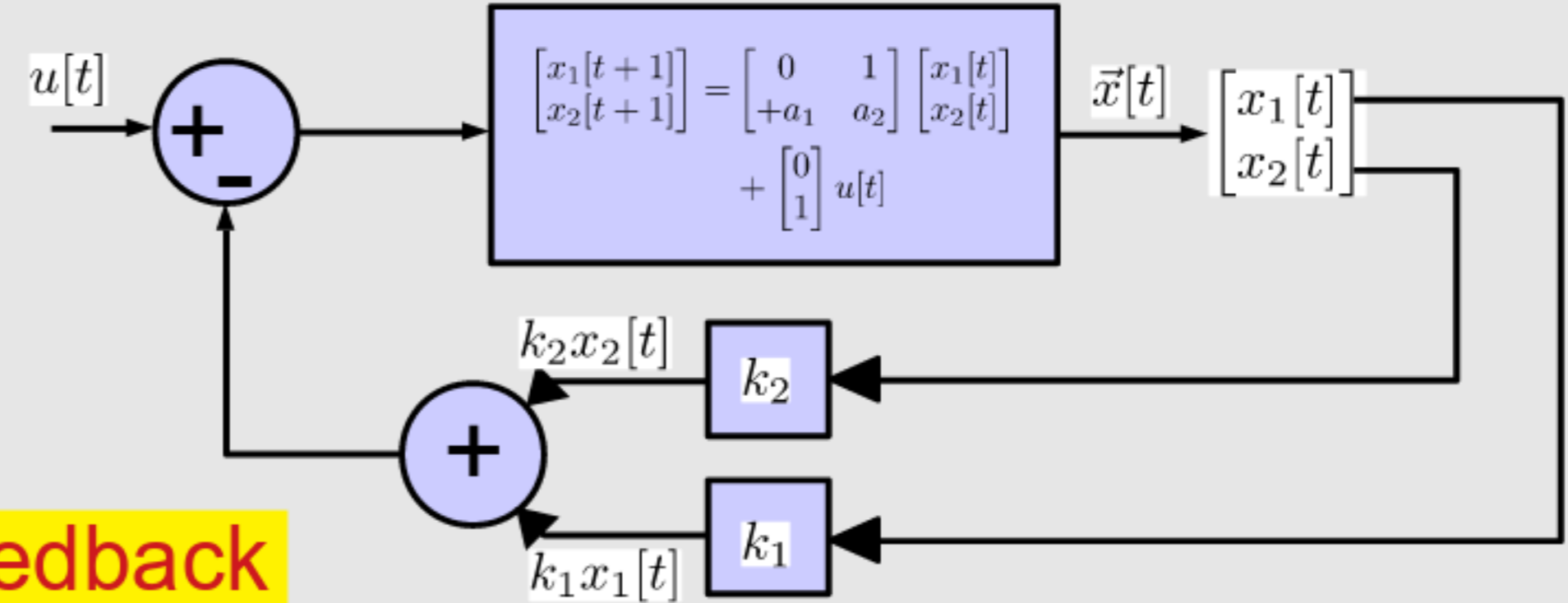


Example: Discrete-Time Feedback

- $$\vec{x}[t+1] = \begin{bmatrix} 0 & 1 \\ a_1 & a_2 \end{bmatrix} \vec{x}[t] + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u[t]$$

$$\begin{bmatrix} 0 & 1 \\ a_1 - k_1 & a_2 - k_2 \end{bmatrix}$$

w feedback

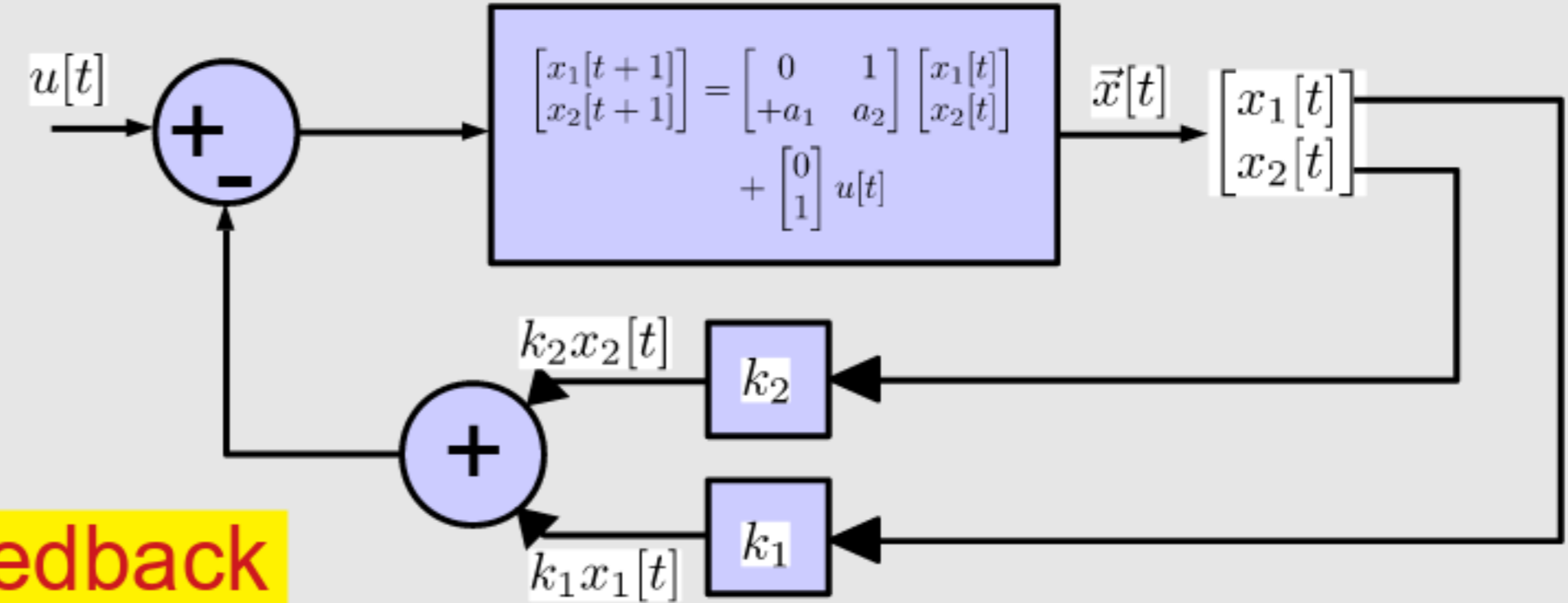


Example: Discrete-Time Feedback

- $$\vec{x}[t+1] = \begin{bmatrix} 0 & 1 \\ a_1 & a_2 \end{bmatrix} \vec{x}[t] + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u[t]$$

$$\begin{bmatrix} 0 & 1 \\ a_1 - k_1 & a_2 - k_2 \end{bmatrix}$$

w feedback



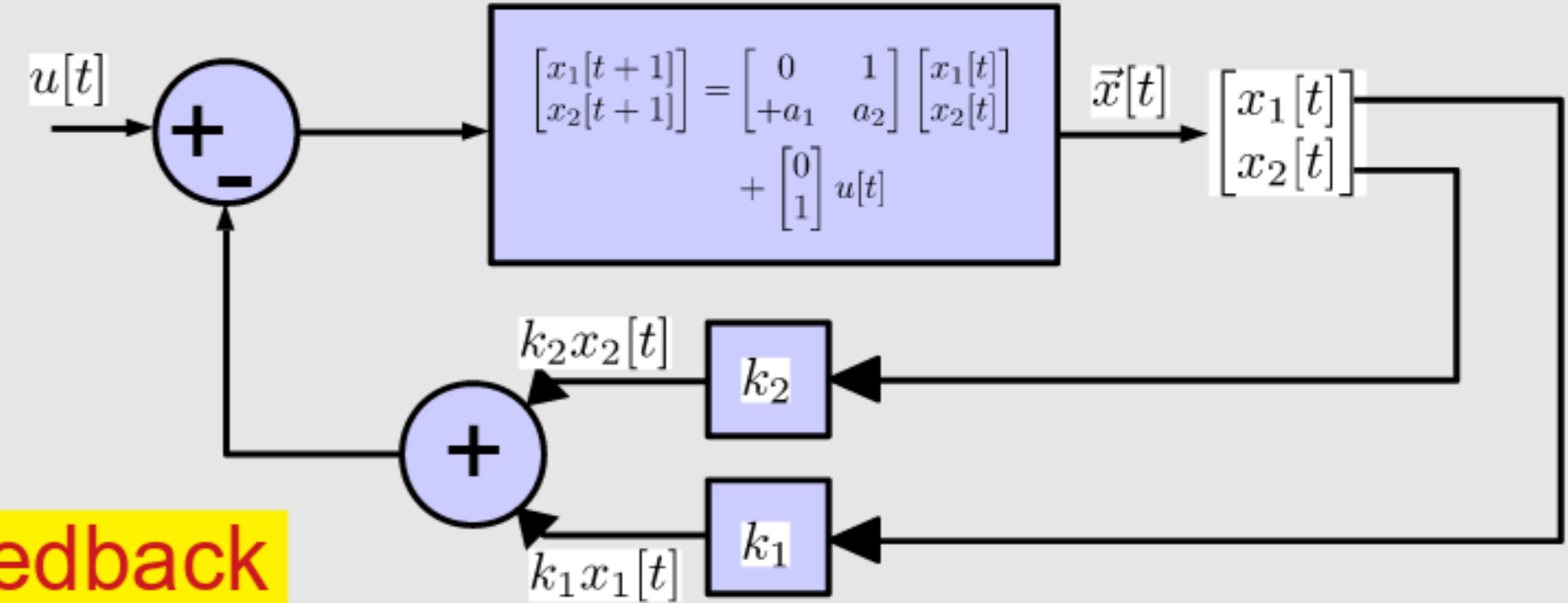
- char. poly.:** $\lambda^2 - (a_2 - k_2)\lambda - (a_1 - k_1) = 0$

Example: Discrete-Time Feedback

- $$\vec{x}[t + 1] = \begin{bmatrix} 0 & 1 \\ a_1 & a_2 \end{bmatrix} \vec{x}[t] + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u[t]$$

$$\begin{bmatrix} 0 & 1 \\ a_1 - k_1 & a_2 - k_2 \end{bmatrix}$$

w feedback



- char. poly.:** $\lambda^2 - (a_2 - k_2)\lambda - (a_1 - k_1) = 0$

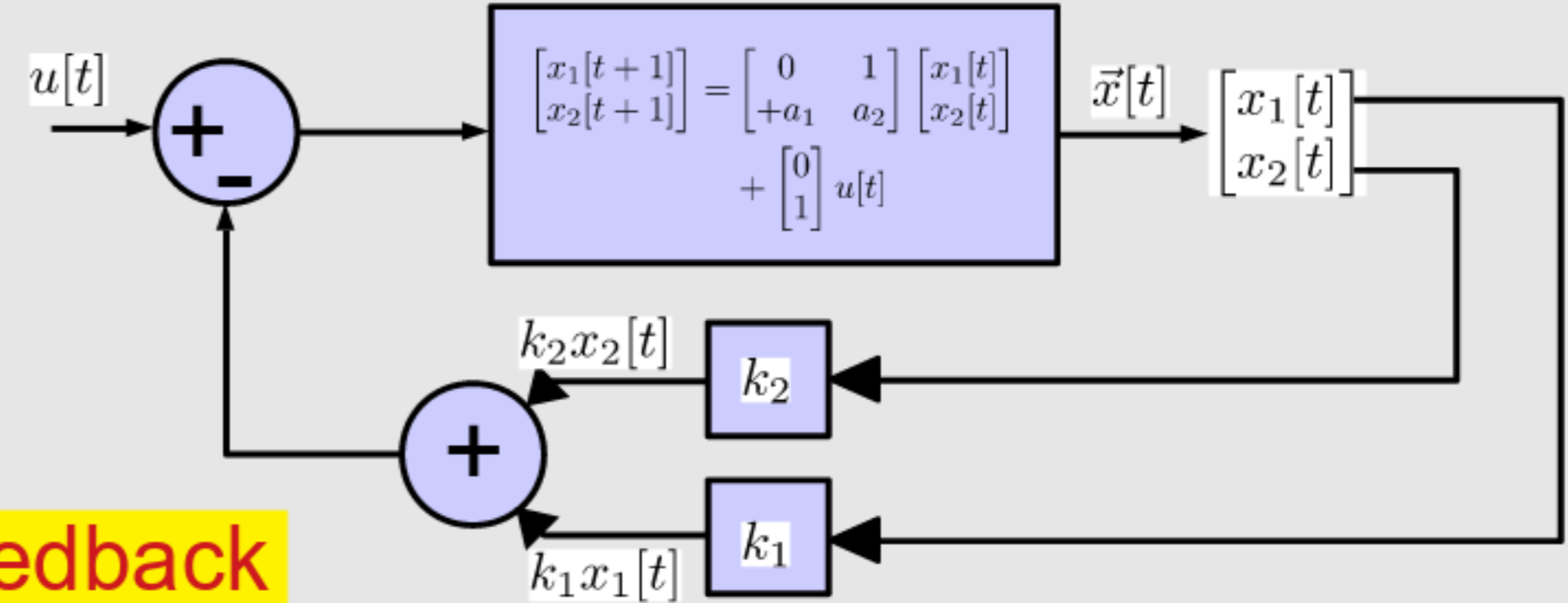
- roots:** $\lambda_{1,2} = \frac{a_2 - k_2}{2} \pm \frac{1}{2} \sqrt{(a_2 - k_2)^2 + 4(a_1 - k_1)}$

Example: Discrete-Time Feedback

- $$\vec{x}[t+1] = \begin{bmatrix} 0 & 1 \\ a_1 & a_2 \end{bmatrix} \vec{x}[t] + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u[t]$$

$$\begin{bmatrix} 0 & 1 \\ a_1 - k_1 & a_2 - k_2 \end{bmatrix}$$

w feedback



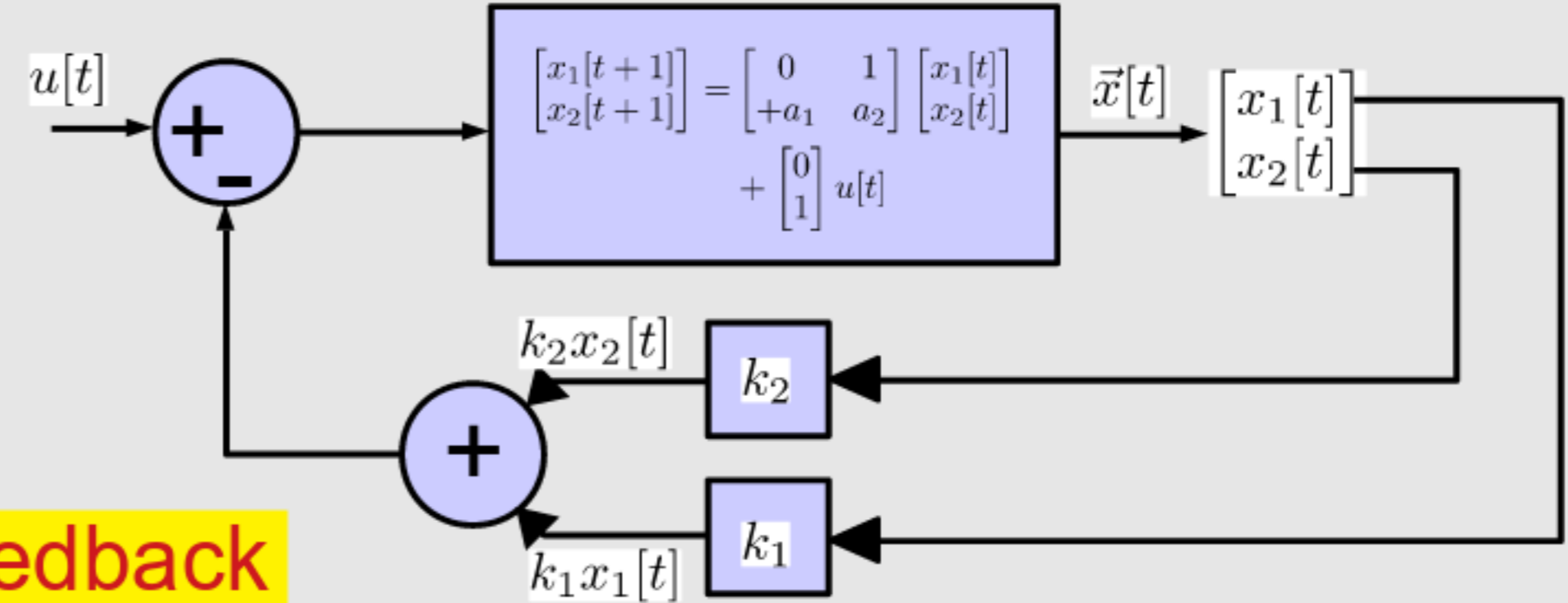
- char. poly.:** $\lambda^2 - (a_2 - k_2)\lambda - (a_1 - k_1) = 0$
- roots:** $\lambda_{1,2} = \frac{a_2 - k_2}{2} \pm \frac{1}{2} \sqrt{(a_2 - k_2)^2 + 4(a_1 - k_1)}$
- easy to express k_1, k_2 in terms of λ_1, λ_2 :**
 - $k_1 = \lambda_1 \lambda_2 - a_1$
 - $k_2 = a_2 - \lambda_1 - \lambda_2$

Example: Discrete-Time Feedback

- $$\vec{x}[t + 1] = \begin{bmatrix} 0 & 1 \\ a_1 & a_2 \end{bmatrix} \vec{x}[t] + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u[t]$$

$$\begin{bmatrix} 0 & 1 \\ a_1 - k_1 & a_2 - k_2 \end{bmatrix}$$

w feedback



- char. poly.:** $\lambda^2 - (a_2 - k_2)\lambda - (a_1 - k_1) = 0$

- roots:** $\lambda_{1,2} = \frac{a_2 - k_2}{2} \pm \frac{1}{2} \sqrt{(a_2 - k_2)^2 + 4(a_1 - k_1)}$

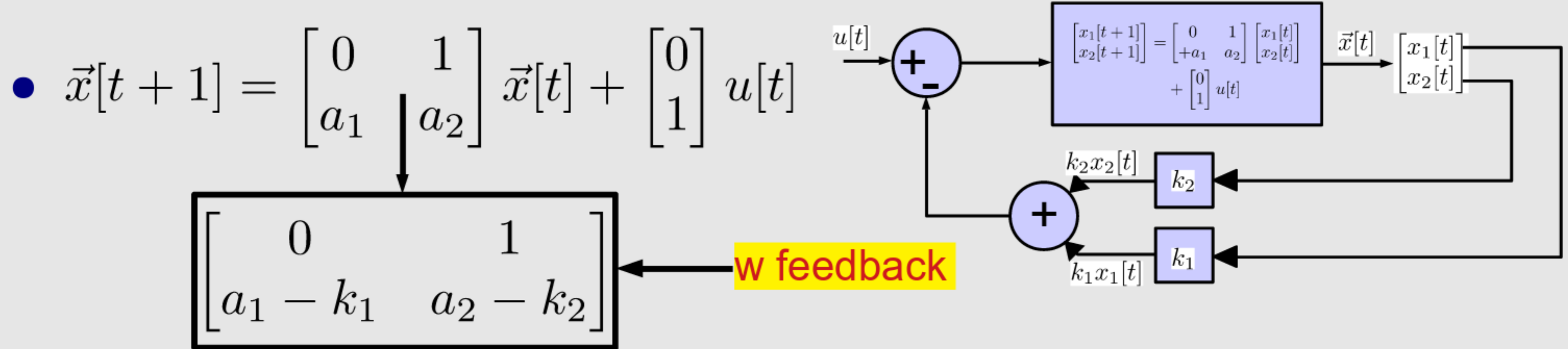
- easy to express k_1, k_2 in terms of λ_1, λ_2 :**

- $k_1 = \lambda_1 \lambda_2 - a_1$

- $k_2 = a_2 - \lambda_1 - \lambda_2$

choose any λ_1 and λ_2 (eg, stable ones); set k_1 and k_2

Example: Discrete-Time Feedback



- char. poly.: $\lambda^2 - (a_2 - k_2)\lambda - (a_1 - k_1) = 0$

- roots: $\lambda_{1,2} = \frac{a_2 - k_2}{2} \pm \frac{1}{2} \sqrt{(a_2 - k_2)^2 + 4(a_1 - k_1)}$

- easy to express k_1, k_2 in terms of λ_1, λ_2 :

- $k_1 = \lambda_1 \lambda_2 - a_1$
 $k_2 = a_2 - \lambda_1 - \lambda_2$

choose any λ_1 and λ_2 (eg, stable ones); set k_1 and k_2

- if** λ_1 is complex: **make sure** λ_2 is the conjugate of λ_1 !

- otherwise, $k_1/k_2/x_1/x_2$ will have imaginary components
 - which would be physically meaningless

Another D-T. Feedback Example

- $\vec{x}[t + 1] = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix} \vec{x}[t] + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u[t]$

Another D-T. Feedback Example

- $\vec{x}[t + 1] = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix} \vec{x}[t] + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u[t]$

$$\begin{bmatrix} 1 - k_1 & 1 - k_2 \\ 0 & 2 \end{bmatrix}$$

w feedback

Another D-T. Feedback Example

- $\vec{x}[t + 1] = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix} \vec{x}[t] + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u[t]$

The diagram shows the feedback matrix $\begin{bmatrix} 1 - k_1 & 1 - k_2 \\ 0 & 2 \end{bmatrix}$ enclosed in a black box. A yellow highlight is placed over the text "w feedback" to the right of the box. A black arrow points from the text "w feedback" to the right side of the matrix box. A black arrow also points from the matrix box back up to the matrix in the equation above.

- **char. poly.:** $(1 - k_1 - \lambda)(2 - \lambda) = 0$

Another D-T. Feedback Example

- $\vec{x}[t + 1] = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix} \vec{x}[t] + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u[t]$

$$\begin{bmatrix} 1 - k_1 & 1 - k_2 \\ 0 & 2 \end{bmatrix} \leftarrow \text{w feedback}$$

- **char. poly.:** $(1 - k_1 - \lambda)(2 - \lambda) = 0$

- **roots:** $\lambda_1 = 1 - k_1, \quad \lambda_2 = 2$

Another D-T. Feedback Example

- $\vec{x}[t + 1] = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix} \vec{x}[t] + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u[t]$

$$\begin{bmatrix} 1 - k_1 & 1 - k_2 \\ 0 & 2 \end{bmatrix}$$

← w feedback

- **char. poly.:** $(1 - k_1 - \lambda)(2 - \lambda) = 0$

- **roots:** $\lambda_1 = 1 - k_1,$

$$\lambda_2 = 2$$

← does not depend on k_1 or k_2 ; ie, cannot be altered via feedback

Another D-T. Feedback Example

- $\vec{x}[t + 1] = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix} \vec{x}[t] + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u[t]$ ← not controllable

$$\begin{bmatrix} 1 - k_1 & 1 - k_2 \\ 0 & 2 \end{bmatrix} \leftarrow \text{w feedback}$$

- char. poly.: $(1 - k_1 - \lambda)(2 - \lambda) = 0$

- roots: $\lambda_1 = 1 - k_1$, $\lambda_2 = 2$ ← does not depend on k_1 or k_2 ; ie, cannot be altered via feedback

Another D-T. Feedback Example

- $\vec{x}[t + 1] = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix} \vec{x}[t] + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u[t]$ ← not controllable

$$\begin{bmatrix} 1 - k_1 & 1 - k_2 \\ 0 & 2 \end{bmatrix}$$

← w feedback

- char. poly.: $(1 - k_1 - \lambda)(2 - \lambda) = 0$

- roots: $\lambda_1 = 1 - k_1$, $\lambda_2 = 2$ ← does not depend on k_1 or k_2 ; ie, cannot be altered via feedback

- suspicions (based on a few examples)

- controllable → can place all eigenvalues via careful feedback
- not controllable → might not be able to place all evs

Summary

- **Controllability**
 - controllability matrix must be full rank
 - C-H Theorem
 - examples: accelerating car (discrete), R-L1-L2 ckt
- **Feedback**
 - **controllable + unstable = useless**
 - **uncontrollable + unstable = REALLY useless?**
 - **feedback (from state to input) can stabilize (eigs moved)**
 - inverted pendulum and other examples