

Lecture 23: Comb Drive

- Announcements:
- Reminder: 2nd project slide due tomorrow
- Module 13 now online

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- Reading: Senturia, Chpt. 5, Chpt. 6
 - Lecture Topics:

↳ Energy Conserving Transducers

- Charge Control
- Voltage Control

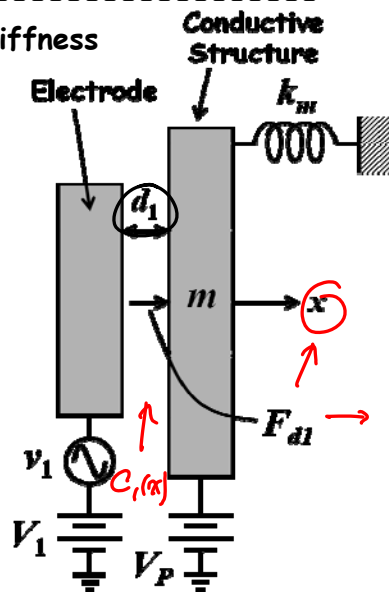
↳ Parallel-Plate Capacitive Transducers

- Linearizing Capacitive Actuators
- Electrical Stiffness

↳ Electrostatic Comb-Drive

- 1st Order Analysis
- 2nd Order Analysis

• Last Time: Electrical Stiffness



More Complete Expressions

$$C_1(x) = \frac{\epsilon A}{d_0 + x} = C_0 \left(1 + \frac{x}{d_1}\right)^{-1} \rightarrow \frac{\partial C_1}{\partial x} = -\frac{C_0}{d_1} \left(1 + \frac{x}{d_1}\right)^{-2}$$

(Expand the Taylor series further)

$$\frac{\partial C_1}{\partial x} = -\frac{C_0}{d_1} \left(1 + A_1 x + A_2 x^2 + A_3 x^3 + \dots\right)$$

where $A_1 = -\frac{2}{d_1}$, $A_2 = \frac{3}{d_1^2}$, $A_3 = -\frac{4}{d_1^3}$, ...

$$F_{d1} = \frac{1}{2} \frac{\partial C_1}{\partial x} (V_P - V_1 - N_1)^2 = \frac{1}{2} \frac{\partial C_1}{\partial x} (V_{P1} - N_1)^2$$

[small displacements: $x \ll d_1$] $V_{P1} = V_P - V_1$

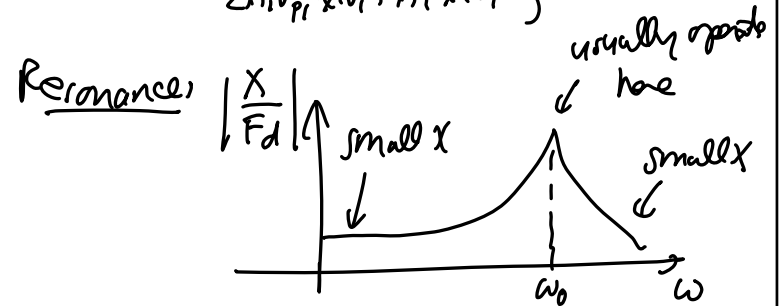
for d_1 !

$$F_{d1} = \frac{1}{2} \left(-\frac{C_0}{d_1}\right) \left(1 + A_1 x\right) \left(V_{P1}^2 - 2V_{P1}N_1 + N_1^2\right)$$

$$= \frac{1}{2} \left(-\frac{C_0}{d_1}\right) \left\{ V_{P1}^2 - 2V_{P1}N_1 + N_1^2 + A_1 V_{P1}^2 x - 2A_1 V_{P1} x N_1 + A_1 x N_1^2 \right\}$$

@ $\omega_0 \rightarrow N_1 = V_1 \cos \omega t$

@ ω_0



@ resonance

$$x = \frac{Q F_d}{jk} = \frac{Q}{jk} \frac{\partial C}{\partial x} V_{pi} N_i$$

\swarrow \nwarrow 90° phase shift \rightarrow displacement x is 90° phase-shifted wrt force, F_d

Thus:

$$N_i = N_i \cos \omega_0 t \rightarrow x = |x| \sin \omega_0 t$$

Force terms only @ ω_0 :

$$F_{d1}|_{\omega_0} = V_{pi} \frac{C_0}{d_i} N_i \cos \omega_0 t + V_{pi}^2 \frac{C_0}{d_i^2} |x| \sin \omega_0 t$$

forcing term

k_e + electrical stiffness

proportional to x

90°

in phase wrt displacement

Electrical stiffness

- ① A negative spring constant!
- ② Derives from V_p :

$$k_e = V_{pi}^2 \frac{C_0}{d_i^2} = V_{pi}^2 \frac{\epsilon A}{d_i^3}$$

overlap area of C

DC Bias

3rd power dependence on gap!

$k_e \rightarrow$ can affect resonance freq., f_0

$\omega_0 \triangleq$ radian resonance freq. w/ no V_p applied (i.e., $V_{pi} = 0V$)

$$\omega_0' = \sqrt{\frac{k}{m}} = \sqrt{\frac{k_m - k_e}{m}}$$

$$= \sqrt{\frac{k_m}{m}} \left(1 - \frac{k_e}{k_m}\right)^{1/2}$$

$$\omega_0' = \omega_0 \left[1 - \frac{V_{pi}^2 \epsilon A}{k_m d_i^2}\right]^{1/2}$$

now a fan of dc bias!
(voltage-controllable!)

- Go through Module 12 slides 26-35

Electrostatic Comb-Drive

Top View

Side View

Shuttle Finger

Drive Finger

V_p

V_i

d

x

L_f

h ← thickness

y

z

$F_d = \frac{\partial W^i}{\partial x} = \frac{L}{2} \frac{\partial C}{\partial x} (V_p - V_i)^2$ *Need C(x).*

$C(x) = \frac{2\epsilon_0 x h}{d} \rightarrow \frac{\partial C}{\partial x} = \frac{2\epsilon_0 h}{d}$ → not a fun of x !

$F_d = \frac{1}{2} \frac{\epsilon_0 h}{d} (V_p^2 - 2V_p V_i + V_i^2)$ (ideally)

can be done out by symmetrically placed electrodes

$V_i \ll V_p$

$F_d = -2V_p \frac{\epsilon_0 h}{d} V_i$

\therefore no electrical stiffness!
(no k_{ei})

- Go through the rest of Module 12, starting from slide 38
- Then start Module 13 (on Equivalent Circuits) through slide 8